

Signal Processing Techniques and Its Applications
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Lecture - 12
Tutorial for Final Examination

So, I will use the two lectures I have given on spectral estimation, which will be the content of this week 12, week 12. Right now, I am discussing what kind of problem you will face in that exam and what kind of question patterns will be there. So, those things I am discussing right now are ok. So, as you expect, signal processing is a lot of mathematical things.

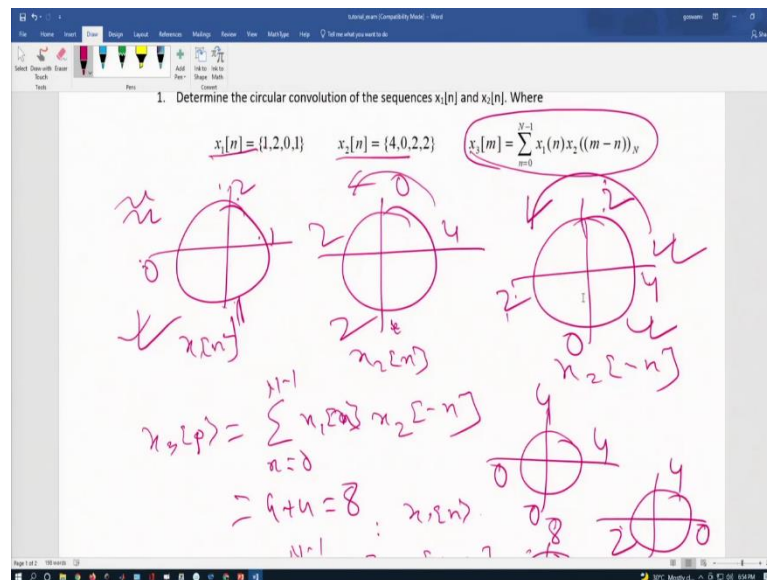
So, you can expect that all those questions will be related to numeric. So, there will be a 3 section, section a, section b and section c. So, section a contains 20 short questions, and 10 short questions carry 2 marks. So, 20 marks section b contains 10 short questions that carry 2 marks, that is 20 marks. So, 40 marks, and then section c contains 20 questions, and every question is 3 marks, which contains 60 marks; so, there is a total of 100 marks.

Section A has 20 marks, Section B has 20 marks, Section C has 60 marks, and Section 3 is divided into 20, so there will be 3 marks for each question and 20 questions. Section A and Section B require very low computation, maybe 1-line or 2-line computation, but Section C requires a computation. So, you may be required to compute, or you have to do that problem in pen and paper, and then you have to find out the answer, and then you have to give that correct answer.

Although it is a multiple-choice question, you must calculate those things; I can guarantee 90 per cent of cases. The question is numerical, and only a few questions are there where some basic importance theories are tested. I am also not asking for any formula for theory. So, the question does not say you derive this formula or what will be the formula for this no; many times, I have included the formula inside the question paper itself.

So, the formula is given in the question paper itself, but you have to feed the numerical things in that question in that formula, and then you have to solve it, and then you have to compute that thing.

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For example, let us say see that question number 1. So, determine the circular convolution of the sequence of $x_1[n]$ $x_2[n]$, where I can say $x_1[n]$ is given, $x_2[n]$ is given, and the formula is given. So, how do you compute? As you know, I can use circular things. So, you know the forward when I say $x_1[n]$; this is 0. So, this is 1 2 0 1, and when I say $x_2[n]$, $x_2[n]$, this is nothing but a 4 0 2 2.

$$x_3[p] = \sum_{n=0}^{N-1} x_1[n] \cdot x_2[-n]$$

Now, when I say, let us say I want to draw the sample value for $x_2[-n]$, so it is reversed. So, I can say if it is anticlockwise, then it will be clockwise. So, if it is clockwise, then there will be 4 0 2 2, then what do you know that if I say what is x_3 ? $x_3[m]$ so, $x_3[0]$ is nothing but a n equal to 0 to N minus 1 $x_1[0]$ multiply by x_2 $x_1[n]$ multiply by $x_2[-n]$, n equal to 0. So, if you see that, what is $x_1[n]$? This one. What is $x_2[-n]$? I have to take the product and add it to this one.

So, I can say 4 multiplied by 1. So, the product is 4, 2 multiplied by the 2 4, 0 multiplied by the 2 0, then it is 1 2 0, there is 1, here is 1. So, 1 multiplied by 0 0, then what is the sum? It is nothing, but a 4 plus 4 is equal to 8; now, when I say x_3 , $x_3[1]$.

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$$x_3[p] = \sum_{n=0}^{N-1} x_1[n] x_2[-n]$$

= 4+4=8 : $x_1[n]$

$$x_3[1] = \sum_{n=0}^{N-1} x_1[n] x_2[1-n]$$

1. Determine the value of $h[0]$, $h[1]$, $h[2]$ and $h[3]$ of a linear-phase FIR filter of length $M = 4$ which has a symmetric unit sample response and a frequency response that satisfies the conditions as in equation-2

Then $x_3[1]$ is nothing but a $x_1[n]$ plus multiplied by so, $x_3[n]$ when I say $x_3[1]$,

$$x_3[1] = \sum_{n=0}^{N-1} x_1[n] \cdot x_2[1-n]$$

So, 1 minus means shifting by one sample, which you are shifting, which will be rotated by one sample. This $x[-n]$ rotated anti-clockwise by one sample. So, I can say if it is rotated, then that 4 will come here, these 2 will come here, this 2 will come here, and this 0 will come here.

Then, it is the product, so 0 is multiplied by 1. So, what is the product sequence? 0 multiply by the 1[0], 4 multiply by the 2 8, 2 multiply by the 0[0], 2 multiply by the 1 2. So, it is nothing but an 8 plus 2 10. So, you have to compute it; then, you have to choose the correct option. So, in the question paper, you have to compute it and then choose the correct option.

(Refer Slide Time: 06:30)

1. Determine the value of $h[0]$, $h[1]$, $h[2]$ and $h[3]$ of a linear-phase FIR filter of length $M = 4$ which has a symmetric unit sample response and a frequency response that satisfies the conditions as in equation-2

$$H_r\left(\frac{2\pi k}{4}\right) = \{2, -1, 0, 8, 0\}$$

Where

$$G[k] = (-1)^k H_r\left(\frac{2\pi k}{N}\right)$$

$$h[n] = \frac{1}{N} \left[G[0] + 2 \sum_{k=0}^{U-1} G[k] \cos\left(\frac{2\pi k}{N} \left(n + \frac{1}{2}\right)\right) \right]$$

$$U = \begin{cases} (N-1)/2 & \text{for } N \text{ is odd} \\ N/2 - 1 & \text{for } N \text{ is Even} \end{cases}$$

Handwritten notes:

$$G[0] = (-1)^0 \cdot H_r\left(\frac{2\pi \cdot 0}{4}\right) = 1 \cdot 2 = 2$$

Similarly, now this question, let us say this question. I will come to the next page because that question and answer can be seen in the whole question, and the whole answer can be seen. So, let us say I said to determine the value of $h[0]$, $h[1]$, $h[2]$, $h[3]$ for a linear phase FIR filter M equal to 4, which has a symmetric unit impulse response frequency response is given; the formula is given.

So, this formula is given, the formula is given, and all formula is given. So, M is equal to 4; that means M is even. So, u is equal to N by 2 minus 1. So, 4 by 2 minus 1 is equal to 2. Sorry, 2 minus 1. Now, how do I calculate $h[0]$? $H[0]$ is nothing but a ; so, first, you say what $G[0]$ is? Let us say that

$$G(0) = (-1)^0 \cdot H_r\left(\frac{2\pi \cdot 0}{N}\right)$$

So, it is nothing but a 1 into $H_r(0)$. So, what is the value of $H_r(0)$? It is nothing but a 2, so 1 into 2 is equal to 2.

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Handwritten notes on the slide:

$$G(1) = (-1)^1 H_r(1)$$

$$= -1 \cdot -1 = 1$$

$$h[0] = \frac{1}{N} \left[G(0) + 2 \sum_{k=0}^{N/2-1} G(k) \cos\left(\frac{2\pi k}{N} \cdot \frac{1}{2}\right) \right]$$

$$= \frac{1}{4} \left[G(0) + 2 \cdot [G(0) + G(1) \cos(\frac{2\pi \cdot 1}{4} \cdot \frac{1}{2})] \right]$$

$$= \frac{1}{4} [2 + 2 \cdot [2 + 1 \cdot \cos(\frac{\pi}{2})]]$$

2. Convert the analog filter with system function in equation (2) into a digital IIR filter by means of the impulse invariance method where sampling frequency is 10kHz.

$$H_a(s) = \frac{s-2}{s^2-4s+5}$$

3. Consider the simple signal processing system as in figure.

Then what is $G(1)$? If it is nothing but a minus 1 to the power 1, then $H_r(1)$; so, k equals 2 k equal to 1, so $H_r(1)$. So, $H_r(1)$ is nothing but a minus 1. So, I can say it is nothing but a minus 1 into minus 1. Now, if you see if I want to calculate $h[0]$, what is the value of $h[0]$? So,

$$h[0] = \frac{1}{N} \left[G(0) + 2 \sum_{k=0}^{N/2-1} G(k) \cos\left(\frac{2\pi k}{N}\right) \cdot \frac{1}{2} \right]$$

so, I can say into half. Now, if you see k equal to 0 to 1, k equal to 0 means two parts will be there.

So, one part will be $\frac{1}{4} G(0)$ into plus 2 into k equal to 0 $G(0)$ plus $G(0) \cos k$ equal to 0, which means $\cos 0 = 1$ plus k equal to 1. So, $G(1)$ into $\cos 2\pi$ into 1 into half divided by 4. So, it is $\frac{1}{4}$. So, I can say it is nothing but a $\frac{1}{4}$; what is the value of $G(0)$? Is $G(0)$ value? What is the $G(0)$ value? 0 value is 2. So, I can say 2 plus 2 into 2 plus $G(1)$ is 1, 1 into $\cos \pi$ by 4. So I can calculate the value.

So, you have to compute the value and then give that answer. So, that is very much required; you have to compute the value, and then you have to give the answer.

(Refer Slide Time: 10:23)

2. Convert the analog filter with system function in equation (2) into a digital IIR filter by means of the impulse invariance method where sampling frequency is 10kHz.

$$H_a(s) = \frac{s-2}{s^2-4s+5}$$

Handwritten notes:

$$s^2-4s+5=0$$
$$s^2-2.s.2+4+1=0$$
$$(s-2)^2=-1$$
$$s=+2\pm\sqrt{-1}=2\pm j$$

Poles: $p_1=2+j$, $p_2=2-j$

So, all the questions in section c are numerical, which is related to the value. Let us say the next question; the next question is to determine how to convert the filter system function equation 2 using a digital IIR filter by means of the impulse invariance method where the sampling frequency is 10 kilohertz. So, what do I have to do? First, I have to calculate the pole of the system. What is the pole value? $s^2 - 4s + 5 = 0$, $s^2 - 2 \cdot s \cdot 2 + 4 + 1 = 0$, ok or not.

So, $(s-2)^2 = -1$. So, I can say s is equal to $2 \pm \sqrt{-1}$. So, it is nothing but a $2 \pm j$. So, I can say this system has a 2 pole: p_1 is equal to $2 + j$, and p_2 is equal to $2 - j$. Now, what is required? I have to represent the transfer function in a fractional sum.

(Refer Slide Time: 11:42)

The image shows a handwritten derivation on a whiteboard. At the top, the sum of residues is given as $\sum_{k=1}^N \frac{c_k}{s-p_k}$. The transfer function is $H(s) = \frac{c_1}{(s-p_1)(s-p_2)} + \frac{c_2}{(s-p_1)(s-p_2)}$. The residue c_1 is calculated as $c_1 = \lim_{s \rightarrow p_1} (s-p_1) H(s) = \frac{(s-p_2)}{(s-p_1)(s-p_2)} \bigg|_{s=p_1} = \frac{1}{p_1 - p_2}$. Similarly, $c_2 = \frac{1}{p_2 - p_1}$. The final expression for $H(s)$ is $H(s) = \frac{1/2}{(s-p_1)(s-p_2)} + \frac{1/2}{(s-p_1)(s-p_2)} = \frac{1}{(s-p_1)(s-p_2)}$. The z-domain transfer function is then given as $H(z) = \frac{1}{(z-p_1)(z-p_2)}$.

So, I can say that it is nothing but a k equal to 0 or k equal to 1 to N minus or N minus k divided by k into H_k . I can say that. So, or I can say this $H(s)$ is nothing but a H of a s is equal to c_1 by c_1 into I can say that c_1 by s minus p_1 into s minus p_2 plus c_2 by s minus p_1 by s . So, it is nothing but a c_k divided by s minus p_k ok. So, c_1 c_1 p_1 divided by s minus p_2 . Now, how do I get the value of c_1 ? I know that c_1 by s 1.

So, I can say that H of s into s minus p_1 at s equal to p_1 gives me the c_1 value. So, this is nothing but the H of s . So, it is nothing but a c_1 equal to H of s is nothing but the s plus s minus 2, s minus 2 into s minus p_1 divided by s minus p_1 into s minus p_2 at s is equal to p_1 . So, I can say this cancel. So, it is nothing but a s is s minus 2 s is p_1 .

So, it is nothing but a 2 plus j minus 2 divided by s minus 2, s is 2 plus j , and minus 2 is there, s minus p_2 . So, s is nothing but a 2 plus j . What is p_2 ? p_2 is nothing but a minus 2 minus j minus 2 plus sorry plus j . So, I can say that these 2, 2 cancel 2, 2 cancels. So, it is nothing but a j by 2 j and nothing but a half. So, I can say c_1 equals half, and c_2 is half.

So, I can say H of s is nothing but a half of H of s is nothing but a half of H of s . It is nothing but a half of s minus p_1 into s minus p_2 plus half of s minus p_1 into s minus p_2 . Now, I know what the value of s is. What is the value of z ? So, I put that s and z I get that value. So, that way, you have to determine that $H(z)$ because I have to find out if the $H(z)$ is ok.

(Refer Slide Time: 14:41)

3. Consider the simple signal processing system as in figure.

Figure 1

If the signal $x_a(t)$ is applied at the input, determine the expression of output $y_d(t)$.

$x_a(t) = 15 \cos 120\pi t + 6 \cos 500\pi t$

$F_s = \frac{1}{T} = \frac{1}{5 \text{ ms}} = \frac{1000}{5} \text{ Hz} = 200 \text{ Hz}$

The next problem is, let us say this problem, this problem is related to the chapter number. I think 2 or 1, 1, not 2, or I can say the lecture number week number 1. So, this will be here, week number 1. So, what I said? I said that some simple signal processing systems are in the figure when considering a signal processing process. This one is the single processing system; if the signal $x_a(t)$ is applied at the input, it determines the expression of the output $y_a(t)$.

So, there is an analogue-to-digital converter and a digital-to-analogue converter. So, what is that F_s in? Let us say this is F_1 and this is F_2 . So, what is the value of F_1 here, 1 minus 5 milliseconds, which is nothing but a tell me 1000 divided by 5 hertz, so 200 hertz? So, the sampling frequency of the signal is 200 hertz. So, what is the signal? This one is my signal.

(Refer Slide Time: 15:52)

Figure 1

If the signal $x_a(t)$ is applied at the input, determine the expression of output $y_a(t)$.

$$x_a(t) = 15 \cos 120\pi t + 6 \cos 500\pi t$$

$$f = \frac{1}{5 \text{ ms}} = \frac{1000}{5} \text{ Hz} = 200 \text{ Hz}$$

$$x[n] = 15 \cos 2\pi \cdot \frac{60}{200} n + 6 \cos 2\pi \cdot \frac{250}{200} n$$

$$= 15 \cos \frac{3\pi}{5} n + 6 \cos \frac{5\pi}{2} n$$

$$= 15 \cos \frac{3\pi}{5} n + 6 \cos (2\pi + \frac{\pi}{2}) n$$

So, I can say $x_a(t)$ is nothing but a $15 \cos 2\pi$. What is the analogue signal? Analogue frequency is 60 hertz divided by 200 n plus 6 cos 2π ; what is 500 mean? 250 divided by 200 n. So, that is nothing but an $x[n]$. So, I get the expression of $x[n]$ 15 cos. So, I can say 3π by 5, ok or not 3π by 5 n plus 6 cos; so, $5\pi/2$ n. Now, I know if the signal is separated by 2π , the signals are identical. So, I can say it is nothing but a $15 \cos 3\pi$ by 5 n plus 6 cos 2π plus tell me; $\pi/2$ n 2π plus 5 by 2 n. So, I can say the signal is separated by 2π are identical.

(Refer Slide Time: 17:17)

$$= 15 \cos \frac{3\pi}{5} n + 6 \cos \frac{\pi}{2} n$$

$$= 15 \cos 2\pi \left(\frac{3}{10} \right) n + 6 \cos 2\pi \left(\frac{1}{4} \right) n$$

$$f_1 = \frac{3}{10} = \frac{f}{f_s} \quad f_2 = \frac{1}{2} = \frac{1000}{2}$$

$$f_2 = \frac{1}{4} = \frac{f}{f_s} \quad f_2 = \frac{1}{4} \times 500 = 125 \text{ Hz}$$

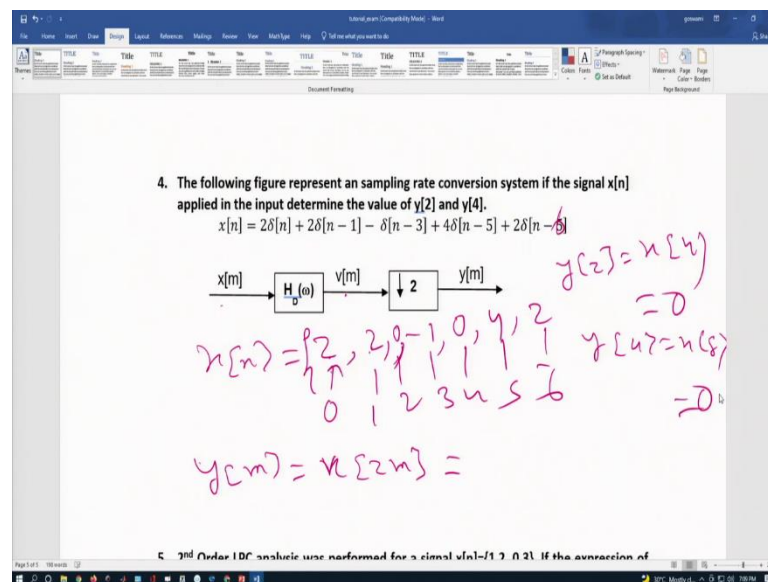
$$f_1 = \frac{3}{10} \times 500 = 150 \text{ Hz}$$

So, it is nothing but a $15 \cos 3\pi$ by $5n$ plus $6 \cos \pi/2 n$. Now, what is the discrete frequency F_1 ? Discrete frequency is nothing but a $\cos 2\pi$ into 3 by 10 into n plus $6 \cos 2\pi$ into 1 by $4n$. So, I can say that discrete frequency F_1 is equal to 3 by 10 ; F_1 is nothing but an F by F_s .

F is the analogue frequency, and F_2 is equal to 1 by 4 , which is nothing but an F by F_s ; F_s is the sampling frequency, and F is the analogue frequency. So, what is the sampling frequency in analogue to digital to analogue conversion? It is 2 milliseconds. So, I can say if let us this is F_s , this is F_s or F_2 , I can say F_2 is equal to 1 by 2 milliseconds. So, a 1000 divided by 2 is nothing but 500 hertz.

So, I can say F is equal to so, this is F_1 . Let us say this is F_2 ; F_1 is equal to 3 by 10 into F_s 150 hertz, and F_2 will be 1 by 4 into F_s . So, I know it is nothing but a 125 hertz. So, if you see that the input analogue frequency is my 60 hertz and 250 hertz, here I get 150 hertz and 125 -hertz analogue signal. So, I can get that ok. So, what is the expression? I can write down the expression easily.

(Refer Slide Time: 19:15)



Now, let us look at question number 2 4. What did I say? The following figure represents the sampling rate conversion system. If the signal x_n is applied in the input, determine the value of $y[2]$ and $y[4]$. So, what is $x[n]$? $x[n]$ What is the sampling repetition? 2 is the 0 th sample, so I know 2 is the 1 sample, and -1 is the third sample, so that means 0 1 . So, it is -3 ; that means 0 1 , then 1 0 will be there, and then -1 is okay.

And, then minus 5, then 4 and 5; this is 0, this is 1, this is 2, this is 3, this is 4, this is the 5th sample. Then so, there will be another 5th. Sorry, this will be 6; let 6, then 2. That is the 6th sample. Now, what I said is dissemination by 2, so every alternative signal will be said. So, I can say $y[m]$ is nothing but a , so $x[m]$ is equal to $v[m]$. So, $y[m]$ is nothing but $a x[2m]$.

So, I can say that. So, y of 0. So, what I said is that $y[2]$ is nothing but an equal to $x[4]$. So, what is $x[4]$? $x[4]$ is equal to 0, $y[4]$ is equal to $x[8]$ which is equal to again 0. So, that way, I can do it ok.

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5. 2nd Order LPC analysis was performed for a signal $x[n] = \{1, 2, 0, 3\}$. If the expression of forward prediction error is given in equation below determine value of the forward PARCOR coefficients k_1 and k_2 .

$$e^i[m] = e^{i-1}[m] - k_1 b^{i-1}[m-1]$$

Handwritten notes on the slide:

- $x[n] = \{1, 2, 0, 3\}$
- $x[0] = 1$
- $x[1] = 2$
- $x[2] = 0$
- $x[3] = 3$
- $\frac{d}{dk_i} \{e^i[m]\}$
- $= \frac{d}{dk_i} \{e^{i-1}[m] - k_1 b^{i-1}[m-1]\}$
- Handwritten k_1 and k_2 above the equation.

Similarly, question number 5. This is a good question; you can see that this required a lot of calculation. That is why I went to that page first and then did it.

So, 2nd order LPC analysis was performed using a signal $x[n]$ is given 1, 2, 0, 3; that means $x[0]$ is equal to 1 $x[1]$ is equal to 2 $x[2]$ is equal to 0 and $x[3]$ is equal to 3 and partial reflection coefficient is k_1 and k_2 because the order of the analysis is 2. So, what do I have to find out? Determine the value of the forward prediction coefficients k_1 and k_2 .

So, how do I get that? So, as you know, how do I get the minimum value of k_1 mean square error? So, I can say

$$\frac{d e^2}{d k_i} [m]$$

So, I know that this is nothing but a

$$\frac{d [e^{i-1}[m] - k b^{i-1}[n-1]}{d k_i}$$

So, this is why I have to derivate it square.

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Handwritten derivation on a whiteboard:

$$e^2 = [e^{i-1}[m] - k b^{i-1}[n-1]]^2$$

$$\frac{d}{d k_i} [e^{i-1}[m] - k b^{i-1}[n-1]]^2 = 2 \cdot [e^{i-1}[m] - k b^{i-1}[n-1]] \cdot \frac{d}{d k_i} [e^{i-1}[m] - k b^{i-1}[n-1]]$$

$$= 2 \cdot [e^{i-1}[m] - k b^{i-1}[n-1]] \cdot (-b^{i-1}[n-1])$$

$$k_i = \frac{\sum_{m=0}^{N-1} e^{i-1}[m] \cdot b^{i-1}[m-1]}{\sum_{m=0}^{N-1} (b^{i-1}[m-1])^2}$$

So, the square means it is nothing but a 2 into $e^{i-1} m$ minus $k b^{i-1} m$ minus 1 multiplied by $b^{i-1} n$ minus 1. So, I have to find out that the mean square error is equal to 0. So, this is equal to 0. So, I can say that k_i that k_i this multiply by this and this multiply by this.

So, I can say.

$$k_i = \frac{\sum_{m=0}^{N-1} e^{i-1}[m] \cdot b^{i-1}[m-1]}{\sum_{m=0}^{N-1} (b^{i-1}[m-1])^2}$$

So, now, m varies, so there will be a sum m range from 0 to N minus 1. So, here also, m varies from 0 to N minus 1. So, what is N ? N is the length of the signal. So, what is the length of the signal?

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Handwritten derivation of the autocorrelation function k_1 for a signal $x[n]$ with length 4. The signal values are $x[0]=1, x[1]=2, x[2]=0, x[3]=3$. The derivation starts with the general formula for the autocorrelation function:

$$e^j[m] = e^{j-1}[m] \cdot k b^{j-1}[m-1]$$

Then, the derivative with respect to k_1 is taken, and the result is set to zero:

$$\frac{d}{dk_1} [e^{j-1}[m] \cdot k b^{j-1}[m-1]] = 0$$

This leads to the equation:

$$2 \cdot [e^{j-1}[m] \cdot k b^{j-1}[m-1]] = 0$$

Finally, the autocorrelation function k_1 is calculated as the ratio of the sum of the product of the signal and its shifted version to the sum of the squared magnitude of the shifted version:

$$k_1 = \frac{\sum_{m=0}^{N-1} e^{j-1}[m] b^{j-1}[m-1]}{\sum_{m=0}^{N-1} (b^{j-1}[m-1])^2}$$

The signal length is 4. So, I can say that this is k was k , and this is m n equal to 4; 4 minus 1 means 3.

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Handwritten derivation of the autocorrelation function k_1 for a signal $x[n]$ with length 4. The signal values are $x[0]=1, x[1]=2, x[2]=0, x[3]=3$. The derivation starts with the general formula for the autocorrelation function:

$$k_1 = \frac{\sum_{m=0}^{N-1} e^{j-1}[m] b^{j-1}[m-1]}{\sum_{m=0}^{N-1} (b^{j-1}[m-1])^2}$$

Then, the derivative with respect to k_1 is taken, and the result is set to zero:

$$\frac{d}{dk_1} [e^{j-1}[m] \cdot k b^{j-1}[m-1]] = 0$$

This leads to the equation:

$$2 \cdot [e^{j-1}[m] \cdot k b^{j-1}[m-1]] = 0$$

Finally, the autocorrelation function k_1 is calculated as the ratio of the sum of the product of the signal and its shifted version to the sum of the squared magnitude of the shifted version:

$$k_1 = \frac{\sum_{m=0}^3 e^0[m] \cdot b^0[m-1]}{\sum_{m=0}^3 (b^0[m-1])^2}$$

Substituting the signal values, we get:

$$k_1 = \frac{1 \cdot 1 + 2 \cdot 2 + 0 \cdot 0 + 3 \cdot 3}{1^2 + 2^2 + 0^2 + 3^2} = \frac{14}{14} = 1$$

So, I can say

$$k_1 = \frac{\sum_{m=0}^3 e^0[m] \cdot b^0[m-1]}{\sum_{m=0}^3 (b^0[m-1])^2}$$

So, Now, as you know $e^0[m]$ is equal to $b^0[m]$ is equal to x^m . So, I can say that $e^0[0]$ is nothing but a x^0 or equal to $b^0[0]$. So, if my x is 1 2 0 3; so, x is 1 2 0 3. So, I can say $e^0[0]$ is nothing but a 1.

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Handwritten mathematical derivation on a whiteboard:

$$K_1 = \sum_{m=0}^{n-1} e^{i-1}[m] \cdot b^{i-1}[m-1]$$

For $m=0$, $e^{i-1}[0] = b^0[0] = 1$

For $m=1$, $e^{i-1}[1] = b^0[1] = 2$

For $m=2$, $e^{i-1}[2] = b^0[2] = 0$

For $m=3$, $e^{i-1}[3] = b^0[3] = 3$

So, $K_1 = 1 + 2 + 0 + 3 = 6$

Now, $K_2 = \sum_{m=0}^{n-1} b^0[m-1]$

For $m=0$, $b^0[-1] = 0$

For $m=1$, $b^0[0] = 1$

For $m=2$, $b^0[1] = 2$

For $m=3$, $b^0[2] = 0$

For $m=4$, $b^0[3] = 3$

So, $K_2 = 0 + 1 + 2 + 0 + 3 = 6$

Final result: $K_1 = 2$ (Note: This appears to be a typo in the original image, as the calculation shows 6).

Multiply by $e^0[b^0[m] - 1]$, minus 1 means 0 plus $u^{1/2}$ multiply by the m equal to 1. So, $b^0[0]$ which is nothing but a 1 plus; so, 1 after 1. So, I can say 0 multiply by 2 plus 3 multiply by 0 plus 0 multiply by 3. So, I can say that it is nothing but this term 0, this term 0, this term 0, this term 0, it is nothing but a 2 2.

So, this term is 2. Now, what is this term? This term is nothing but a square. So, whether 0 will be not there, so, 0 square plus 1 square plus 2 square plus 0 square plus. So, sorry, this will be $m - 1$ plus p , so p is equal to 2. So, that is why $m - 1$ plus p .

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Handwritten mathematical derivation for K_1 in a video player window. The derivation shows the simplification of a double summation formula. It starts with $K_1 = \frac{\sum_{m=0}^{N-1} e^{j\omega m} b^{*-1}[m-1]}{\left[\sum_{m=0}^{N-1} b^{*-1}[m-1] \right]^2}$. This is simplified to $K_1 = \frac{\sum_{m=0}^{N-1} e^{j\omega m} b^{*-1}[m-1]}{\sum_{m=0}^{N-1} [b^{*-1}[m-1]]^2}$. The denominator is then calculated for $N=5$ with values $1, 2, 0, 3, 1$ (labeled as $1, 2, 0, 3, 1$ in the video), resulting in $1^2 + 2^2 + 0^2 + 3^2 + 1^2 = 14$. The numerator is calculated as $1 \cdot 0 + 2 \cdot 1 + 0 \cdot 2 + 3 \cdot 0 + 1 \cdot 3 = 4$. The final result is $K_1 = \frac{4}{14} = \frac{2}{7}$.

So, 4 minus 1 plus 2; so, this will be 3 plus 2 5. So, it is m varies from 0 to 5 because, 2nd order analysis is 0 to 5 N minus 1 plus p ok. So, now, I get this one will be plus 3 square.

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Handwritten calculation for K_1 in a video player window. It shows the calculation of the denominator $\sum_{m=0}^{N-1} [b^{*-1}[m-1]]^2$ for $N=5$ with values $1, 2, 0, 3, 1$. The calculation is $1^2 + 2^2 + 0^2 + 3^2 + 1^2 = 14$. The numerator is calculated as $1 \cdot 0 + 2 \cdot 1 + 0 \cdot 2 + 3 \cdot 0 + 1 \cdot 3 = 4$. The final result is $K_1 = \frac{4}{14} = \frac{2}{7}$.

So, I will get how much? 1 plus 4 plus 9. So, it is nothing but a 14. So, I can say the K_1 is nothing but a 2 by 14. So, it is nothing but a 1 by 7.

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Handwritten notes on a whiteboard:

$$k_2 = \frac{\sum_{m=0}^5 e^1[m] b^1[m]}{\sum_{m=0}^5 [b^1[m-1]]^2}$$

$$e^i[m] = e^{i-1}[m] - k_1 b^{i-1}[m-1]$$

$$e^1[m] = e^0[m] - k_1 b^0[m-1]$$

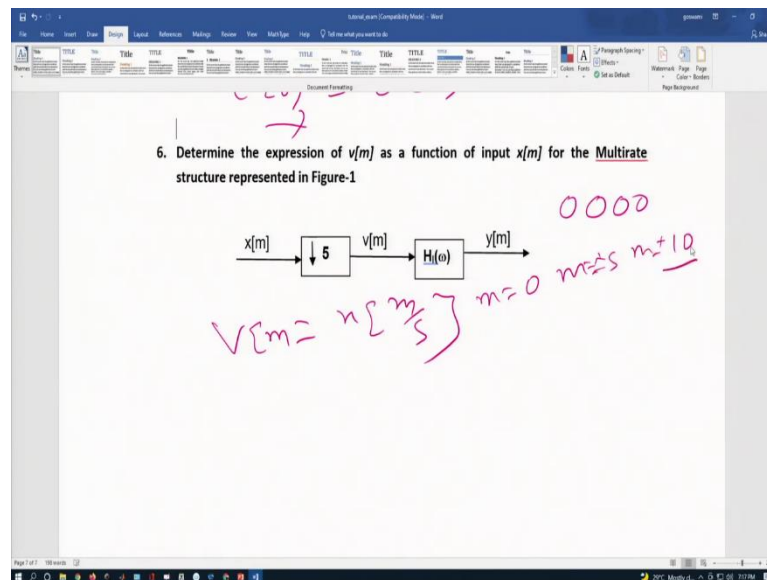
$$e^1[0] = e^0[0] - k_1 b^0[-1]$$

6. Determine the expression of $v[m]$ as a function of input $x[m]$ for the M structure represented in Figure-1

Then, I have to calculate k_2 . So, when I say calculate k_2 , k_2 is nothing but a I have to set m equal to 0 to 5 $e^1[m] b^1[m]$ divided by 0 to 5 $b^1[m-1]$ whole square. So, I have to determine $e^1[m]$. So, what is $e^1[m]$? So, how do I get the $e^1[m]$? I know $e^i[m]$ is equal to $e^{i-1}[m] - k_1 b^{i-1}[m-1]$.

So, I can say that $e^1[m]$ is equal to $e^0[m]$ minus k_1 into $b^0[m-1]$. From there, I can calculate that $e^1[0]$ is equal to $e^0[0]$ minus k_1 and $b^0[-1]$. So, from there, I can calculate $e^1[0]$, $e^1[1]$, and $e^1[2]$. Similarly, I can calculate $b^0[0]$, $b^1[1]$, $b^1[2]$; then, once I get that $b^1[n-1]$, I can calculate the value of k_2 . So, that required a calculation.

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Then, I will determine the expression of a $v[m]$ as a function of input $x[m]$ in multi-rate signal processing. So, what is $v[m]$? $v[m]$ is nothing but a $x[m]$ m by 5. So, what is required? So, I have to 4 0 is inserted between 1 to 5th sample 1 1 to 2nd sample, where m is equal to 0, m is equal to plus minus 5, m is equal to plus minus 10 like that ok.

Then the so, this is the signal, this kind of mathematics you can expect from that question paper. So, I will say, please carefully look into the assignment and try to practice all the mathematics I have given in the assignment; that is very important. If you practice all the mathematics, I am sure you will be able to solve all the questions in the question paper.

So, thank you very much for attending the entire course. So, there is a review, and the feedback that the question is whether the course is too much mathematics; yes, signal processing is mathematics. But, I try to explain it as is your practical sense and give you some assignments that can use your brain to see what is happening practically because signal processing is the application of, or you can say, the physical phenomena expressed in a mathematical expression.

So, that is why this is too much mathematics, ok? So, thank you for your patience in attending the entire course. I hope that all of you will perform better in the exam. also if you have any questions or any queries regarding any point, my email id is available; you can send me an email, or you can drop a message in your NPTEL forum; so, that I will be able to answer you. So, thank you very much for your kind patience.

Thank you.