

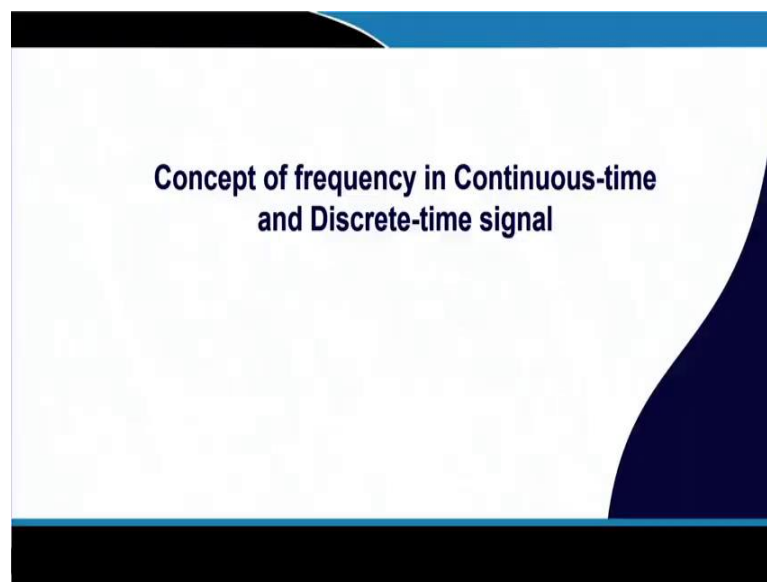
Signal Processing Techniques and Its Applications
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Lecture - 06

Concept of frequency in Continuous-time and Discrete-time signal

So, in the last class, we discussed analog-to-digital conversion and digital-to-analog conversion. Basically, I am interested in the concept of sampling frequency and quantization. We also discussed the memory requirement for storing a digital signal, which may be a one-dimensional signal or maybe a multidimensional signal. So, in both cases, we have done that.

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Today we will discuss the Concept of frequency in Continuous time and Discrete time signal. So, continuous time and discrete time signals basically talk about sinusoidal signals. So, what are the concepts of frequency, discrete frequency, and analogue frequency of all kinds of things we will talk about.

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Concept of frequency in continuous-time


Continuous sinusoidal Time Signal $x_a(t) = A \cos(\Omega t + \theta)$

$\Omega \rightarrow 2\pi F$

- For every fixed value of the frequency F , $x(t)$ is periodic.
- Continuous time sinusoidal signal with distinct frequencies are themselves distinct.
- Increase the frequency result in increase in the rate of oscillation of the signal \rightarrow more period are included.

Complex exponent form $x_a(t) = A e^{j(\Omega t + \theta)}$

$F = 200 \text{ Hz}$
 $400 \text{ } T = \frac{1}{400} \text{ sec}$ $T = \frac{1}{200} \text{ sec}$ $x_a(t) = A \cos(2\pi \times 200 t + \theta)$



Let us consider the analogue sinusoidal. So, I can write down continuous sinusoidal time signal $x_a(t)$. So, $x_a(t)$ is nothing but $A \cos(\Omega t + \theta)$. This ω is analogue frequency; that means the radian per second, θ is the phase, and t is the continuous time. So, if this is my signal, then I can say that for every fixed value of the frequency F , $x_a(t)$ is periodic. So, what is Ω ? Ω is nothing but a $2\pi f$. This F is analog frequency, which is radian per second. So, F $x(t)$ is periodic for the fixed value of the frequency. What is the meaning?

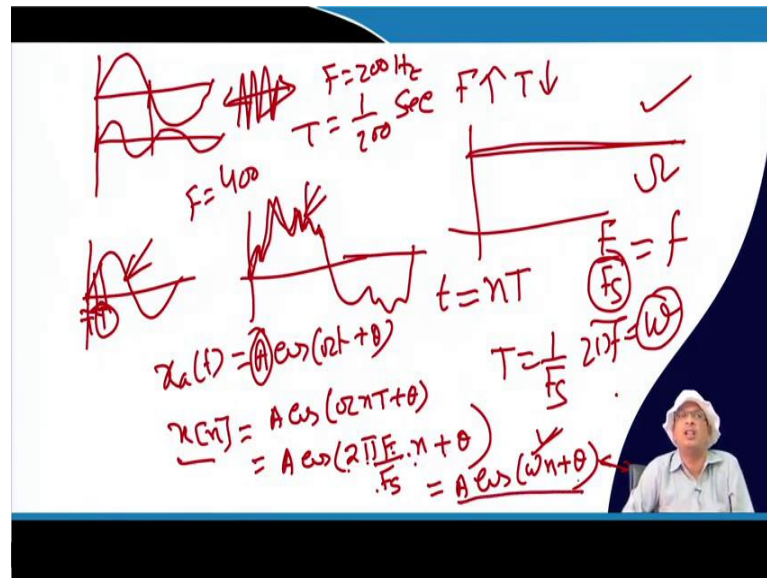
Suppose I said I want to generate a 200 Hz sinusoidal signal. So, small f is equal to 200 Hz. Now, what should be the analog frequency? What should the signal be? So, $x_a(t)$ should be $A \cos(2\pi \times 200t + \theta)$. So, I can say in that case, if the frequency F is 200, then what is the period T ? T is equal to 1 by 200 seconds.

Now, if I change the frequency F to 400, the period will also be 1 by 400 second. So, for every fixed value of the frequency F , the $x(t)$ is periodic, whose period is 1 by F . So, continuous time sinusoidal signals with distinct frequencies are themselves distinct. So, if I said a $\cos 200$ Hz cosine wave and a 400 Hz cosine wave, they are two distinct.

So, they are distinct. So, if I say you have already learned about the Fourier transform. So, when I say $\cos 200$ Hz and turn the frequency transform, if this axis is my F , then I get a peak at 200 Hz. When I converted the 400 Hz, I got a peak at 400 Hz. And those two peaks are distinct themselves. They do not overlap in nature. So, continuous time sinusoidal with distinct frequencies are themselves distinct.

So, $x_a(t)$ is 200 Hz, and let's say $x_{1a}(t)$ is 400 Hz. They are not equal. They are distinct. Increasing the frequency results in an increased rate of oscillation.

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So, if I say I take a 200 Hz sinusoidal signal, it will look like this. Now, when I take the 400 Hz sinusoidal signal, then within a half period, there will be a complete sinusoidal signal. So, if I see, if I increase the frequency, the number of oscillations will increase per second because T decreases.

So, how many oscillations will be in 1 second if the frequency F equals 200 Hz? So, 200 Hz frequency means T equals 1 by 200 Hz second. So, in 1 second, how many oscillations will be there? I can calculate the number of frequencies, which is 200. Similarly, when I say F equals 400, then 400 complete periods will be there within 1 second.

So, if I increase the frequency, the rate of oscillation will increase, and the number of periods will increase within 1 second, or I can say the period will be less. T will be decreasing. F is increasing, and T is decreasing. So, there are three properties for a continuous time sinusoidal signal, the frequency concept. One is that every fixed value of the frequency $x(t)$ is periodic.

Yes, I can calculate. Continuous time sinusoidal signals with distinct frequencies are themselves distinct, yes. 200 Hz signal does not overlap with 400 Hz signals. They are distinct. An increase in the frequency means the rate of oscillation will increase. So, that

is the concept of frequency in a sinusoidal signal. When I say these two signals, one signal looks like this.

Another signal looks like this. If I ask you which one contains a high-frequency component? So, what is a frequency? With more variation, more oscillation will increase. So, if I see more oscillations here, that is why these contain the high frequency component compared to this signal. So, that is the concept of frequency in a sinusoidal signal.

Similarly, if I told you the concept of frequency in an image. Suppose I say black whiteboard like this PPT white portion. If this is completely white, there is no change in the pixel; then I said it is totally DC, with no frequency component. So, when I say no oscillation, it means my signal will look like this. No oscillation signal is constant. So, it is a DC signal.

Once I include variation, only oscillation will come. So, in the case of an image, if it is completely white or completely black, there is no only DC. Now, once I oscillate, suppose here when there is a boundary change. So, there is a change in pixel colours. So, here, some frequency is introduced. So, high frequency means more variation and low frequency means less variation. So, that is the concept of frequency in the analogue signals. Is it clear?

So, now I go for the digital signal or discrete signal. So, what is a discrete signal? Suppose I said I have a signal $x_a(t)$; let the same example, which is $A \cos(\Omega t + \theta)$. I want to discretize. So, say discrete time. Discrete value means A will be quantized. Discrete-time means t is quantized. So, what should be the discrete signal of $x_a(t)$? Which is nothing but a represented by a $x[n]$.

So, once I want to do the discretization what parameters I required? I required a sampling frequency. So, what is sampling frequency? This means after what interval I am collecting the evidence of the analogue signal, or I can say what is the resolution of the time axis that I have to know.

So, I can say continuous time is represented by nT . Capital T is the difference or distance between the two samples. So, suppose I take this sample and this sample. So, this distance is T , and the next distance is also T . So, who determines the T ? Sampling frequency

So, when I write down nT , what is capital T? It is nothing but an A cos what is ω ? ω is nothing but the $2\pi F$. So, F denotes the analogue frequency in T.

So, this F by F_s is called normalized discrete frequency denoted by small f .

Now, this $2\pi f$ we write down by this ω , not this. This ω is analogue; this ω is normalized discrete frequency. So, when I say normalized discrete frequency, that means radian per sample. Since it is divided by F_s , it is called radian per sample. So, I write this as $A\cos(\omega n + \theta)$. There will be no T ; n is the sample index.

So, this ω is a discrete normalized discrete frequency, which is nothing but an F by Fs. Is it clear? So, due to the sampling, when I write down a discrete signal, it is nothing but a $(2\pi F/F_s) * n + \theta$. So, it is nothing but $A \cos(\omega n + \theta)$.


Concept of frequency in Discrete time Sinusoidal

$x[n] = A \cos(\omega n + \theta)$
 $\omega = 2\pi f$
 $f = \frac{F}{T}$
 $T = 2\pi F$

☐ A discrete time signal is periodic if its frequency f is a rational number

$x[n + N] = x[n]$ for all n
 $A \cos(2\pi f_0(n + N) + \theta) = A \cos(2\pi f_0 n + \theta)$
This relation is true if and only if there exists an integer k such that

$2\pi f_0 N = 2\pi k$
 $f_0 = \frac{k}{N}$
 $\cos(2\pi f_0(n + N) + \theta) = \cos(2\pi f_0 n + \theta)$
 $\cos(2\pi f_0 n + 2\pi f_0 N + \theta) = \cos(2\pi f_0 n + \theta)$
 $2\pi f_0 N = 2\pi k$



So, when I say discrete frequency, $x[n]$ discrete signal. So, discrete sinusoidal signal. So, I said $x[n] = A \cos(\omega n + \theta)$, where A is the amplitude, ω is nothing but a $2\pi f$. And this ω is

nothing but a radian per sample. Is it clear? So, this small f is nothing but a F divided by F_s . So, it is a ratio ok.

And when I write capital ω , it is nothing but a $2\pi F$. Now, what is the concept of frequency in discrete sinusoidal signals? It said a discrete-time signal is periodic if the frequency f is a rational number. So, if it is periodic, f must be a rational number. So, how do I prove it? Let us say what a period in a digital signal is.

So, if I say a sinusoidal signal, I say that two points to this point is a period. In terms of samples, let's say there is an N number of samples, and T is time. Now, time is converted to index in the discrete signal. So, let us say there are N samples within this period. So, instead of period T , I write capital N , which is the number of samples after which signals repeat themselves.

So, if N is the period, then $x[n+N] = x[n]$, then only N is a period because the signal has to repeat itself for all n . Now if it is that if I put that, if you change the index n to n plus capital N , it should be equal to $A \cos 2\pi f_0 n + \theta$.

So, from this equation, I put the value of this one, and I get this one. Now, when will this equation be true? So, this relation is true if and only if there exists an integer k such that because this is nothing but A . So, I have to prove

$$A \cos (2\pi f_0 (n+N) + \theta) = A \cos (2\pi f_0 n + \theta)$$

When is it possible?

Now, when will this be true? When this part must be equal to the integer multiple of 2π . You know that $\cos (2\pi + \theta) = \cos \theta$. So, only when $2\pi f_0 N$ should equal the integer multiple of 2π . So, I can say $2\pi f_0 N = 2\pi * k$, and k is an integer. It may be 2π , it may be 4π , it may be 6π does not matter. So, in that case, I can say 2π , 2π cancel. So, f_0 is equal to k/N , What is k ? k is an integer? What is N ? N is also an integer.

Because that is the period index of that period. So, integer by integer is nothing but a rational number. So, I said a discrete time sinusoidal signal is periodic if f is a rational number, clear.

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Discrete time sinusoidal whose frequency are separated by an integer multiple of 2π are identical

$\omega = 0, 1, 2, \dots$

$\cos((\omega_0 + 2\pi)n + \theta) = \cos(\omega_0 n + 2\pi n + \theta) = \cos(\omega_0 n + \theta)$

All sinusoidal sequences

$x_k = A \cos(\omega_k n + \theta)$

$\omega_k = \omega_0 + 2\pi k$

$-\pi \leq \omega_0 \leq \pi$

where

Sinusoid with a frequency $\omega > \pi$ or $f > \frac{1}{2}$ is identical to sequence obtained from sinusoidal signal with frequency $\omega < \pi$ or $f < \frac{1}{2}$

$x_1 = \cos(2\pi n)$
 $x_2 = \cos(4\pi n)$
 $x_3 = \cos(6\pi n)$

$x_5 = \cos(10\pi n)$

$x_7 = \cos(14\pi n)$

$x_9 = \cos(18\pi n)$

$x_{11} = \cos(22\pi n)$

$x_{13} = \cos(26\pi n)$

$x_{15} = \cos(30\pi n)$

$x_{17} = \cos(34\pi n)$

$x_{19} = \cos(38\pi n)$

$x_{21} = \cos(42\pi n)$

$x_{23} = \cos(46\pi n)$

$x_{25} = \cos(50\pi n)$

$x_{27} = \cos(54\pi n)$

$x_{29} = \cos(58\pi n)$

$x_{31} = \cos(62\pi n)$

$x_{33} = \cos(66\pi n)$

$x_{35} = \cos(70\pi n)$

$x_{37} = \cos(74\pi n)$

$x_{39} = \cos(78\pi n)$

$x_{41} = \cos(82\pi n)$

$x_{43} = \cos(86\pi n)$

$x_{45} = \cos(90\pi n)$

$x_{47} = \cos(94\pi n)$

$x_{49} = \cos(98\pi n)$

$x_{51} = \cos(102\pi n)$

$x_{53} = \cos(106\pi n)$

$x_{55} = \cos(110\pi n)$

$x_{57} = \cos(114\pi n)$

$x_{59} = \cos(118\pi n)$

$x_{61} = \cos(122\pi n)$

$x_{63} = \cos(126\pi n)$

$x_{65} = \cos(130\pi n)$

$x_{67} = \cos(134\pi n)$

$x_{69} = \cos(138\pi n)$

$x_{71} = \cos(142\pi n)$

$x_{73} = \cos(146\pi n)$

$x_{75} = \cos(150\pi n)$

$x_{77} = \cos(154\pi n)$

$x_{79} = \cos(158\pi n)$

$x_{81} = \cos(162\pi n)$

$x_{83} = \cos(166\pi n)$

$x_{85} = \cos(170\pi n)$

$x_{87} = \cos(174\pi n)$

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$x_{121} = \cos(242\pi n)$

$x_{123} = \cos(246\pi n)$

$x_{125} = \cos(250\pi n)$

$x_{127} = \cos(254\pi n)$

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$x_{259} = \cos(518\pi n)$

$x_{261} = \cos(522\pi n)$

$x_{263} = \cos(526\pi n)$

$x_{265} = \cos(530\pi n)$

The next one is a discrete-time sinusoidal whose frequencies are identical and separated by an integer multiple of 2π . Before that, let me give you the back concept in the sampling; what is the concept? Suppose I have a baseband signal like fm, this kind of structure. If I sample this signal with sampling frequency F_s . So, what is the frequency response we have drawn for the frequency?

So, this is Fs. So, this is fm, and this is fm. This is 0 to fm and this is fm to Fs. So, that is why Fs is equal to twice fm. If it is not, there will be the aliasing effect that we have discussed during the sampling. So, I said a discrete time sinusoidal whose frequencies are separated by integer multiple 2π are identical. Why? Because I know $\cos(2\pi + \theta) = \cos \theta$. So, whatever is written so for $\cos(4\pi n + \theta)$ that can be written as $\cos((2\pi + 2\pi)n + \theta)$.

So, I can say if it is $\cos 2\pi$ separated by 2π , then the \cos signal will be again returned. So, how do I prove it? Let us say $\cos(\omega_0 + 2\pi)$. So, this signal is 2π is separation is 2π . So, I added the separation 2π here. So, I changed the ω_0 and added 2π . So, $\cos((\omega_0 + 2\pi)n + \theta)$.
so

$$\cos ((\omega_0 + 2\pi)n + \theta) = \cos (\omega_0 n + 2\pi n + \theta) = \cos (\omega_0 n + \theta)$$

n is an integer, and n varies from 0, 1, 2, 3 dot dot dot number of sample.

So, this product is nothing but an integer multiple of 2π . So, I know if the \cos signal is separated by an integer multiple of 2π , the \cos signal again comes back. So, this is nothing, but again $\cos(\omega_0 n + \theta)$. Because I can write down this one $\cos(\omega_0 n + 2\pi n + \theta)$, which equals nothing but a $\cos(\omega_0 n + \theta)$.

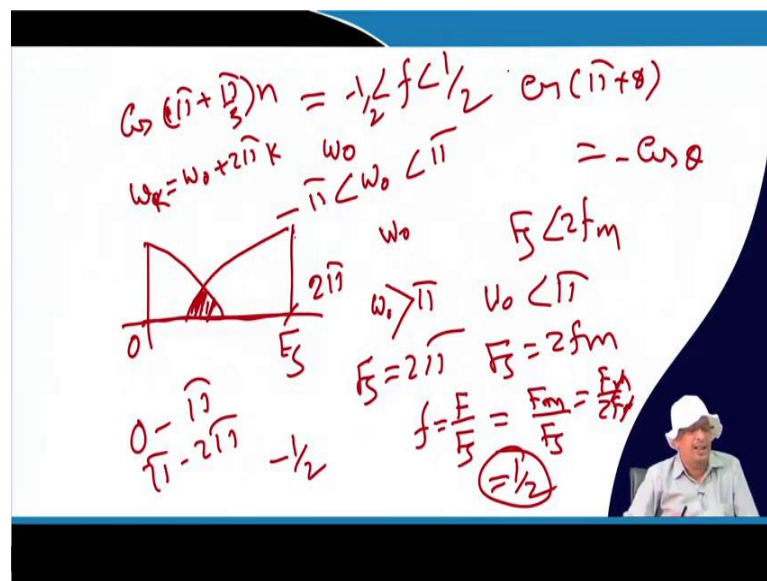
So, a sinusoidal whose frequency is separated by 2π , ω value is separated by 2π . So, I have ω_1 and ω_2 . If the difference is 2π , then they are identical signals. So, I give an example. Let us say

$$x_1 = \cos\left(\frac{2\pi}{3}\right)n,$$

$$x_2 = \cos\left(2\pi + \left(\frac{2\pi}{3}\right)n\right) = \cos\left(\frac{8\pi}{3}\right)n$$

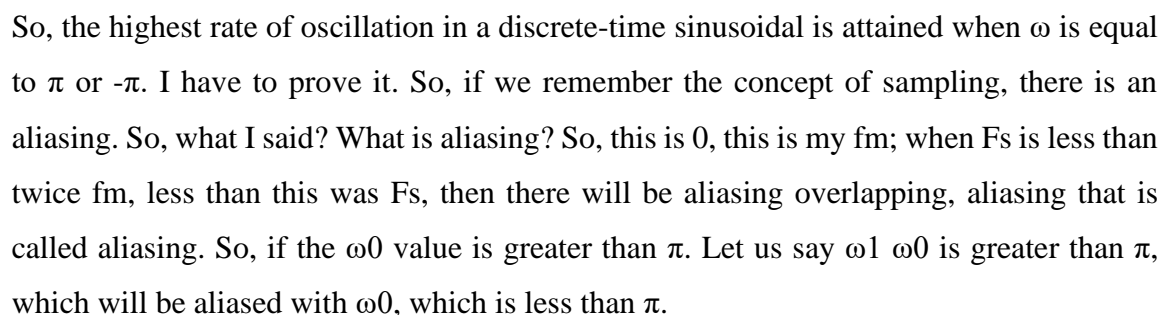
So, this signal and this signal are identical because they can be written like this. So, if it is written like this, then $\cos(2\pi + \theta)$ is nothing but a $\cos\theta$, again $\cos\left(\frac{2\pi}{3}\right)n$. So, I am saying if signals are separated by 2π , the signal will repeat itself ok. So, if the signal is separated by 2π , then the integer multiple of 2π , then the signal will repeat itself. Then what is the variation of the ω_0 ? ω_0 varies from $-\pi$ to $+\pi$, how? How is it possible?

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If I told you, let say $\cos(\pi + (\pi/3))n$ is equal to how much?. So, $\cos \pi + \theta$ is equal to how much? $-\cos\theta$. So, I am saying that ω_0 , if it is separated by an integer multiple of ω_k , is

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So, why is it possible? As you know, the highest rate of oscillation is 2π because, after 2π , the signal will repeat itself separated by 2π . So, 2π is nothing but an Fs. So, I said Fs is equal to 2π . According to the sampling theory, the Fs must equal twice the fm. So, without aliasing, what should be the discrete frequency normalized discrete frequency f is equal to analog frequency divided by Fs.

So, I can say F_m divided by F_s which is nothing but a F_m divided by twice F_m . So, it is nothing but a half. So, if I consider from 0 to π , it is half; if I consider π to 2π . So, 0 to π and π to 2π , if I consider so, that means it will start from 0 to $-\pi$. So, max, then it will come at minus half. So, ω_0 varies between $-\pi$ to π , or I can say small f varies from minus half to half ok.

So, now let us say that it says the highest rate of oscillation in a discrete-time sinusoidal is attained when ω is equal to π . So, I said the highest rate of oscillation would only be attained when ω is equal to π , or ω is equal to $-\pi$. If the ω is greater than π , then it will be aliased with ω less than π . How would it be possible? So, at $\omega = 0$, there is no oscillation then which means it is a uniform signal. Then let's say ω is equal to $\pi/8$.

So, what is the period? So, ω is $2\pi f = \pi/8$. So, I can say the f is equal to $1/16$. So, $T = 16$, 16 samples in one period. Now if $\omega = \pi/4$, then $2\pi f = \pi/4$. So π, π cancel $f = 1/8$, so $T = 8$. So, the value of ω_0 can be 0, $\pi/8$, $\pi/16$, I can say the value of small f is 0, $1/16$, $1/8$, and if it is $\pi/2$, it will be $1/4$ like that or T will be 16, 8, 4.

So, if I say if I increase ω , ω_0 increases within π 0 to π , then the rate of oscillation is increasing because T is decreasing. So, the rate of oscillation is increasing; when ω_0 is equal to 0, the T is equal to infinite because t is nothing but a 1 by f , equal to 0. So, 1 by 0 is infinite. So, it is an infinite period.

So, when the ω is increasing from 0 to π the oscillation rate of oscillation is increasing. What is the highest rate of oscillation when it is possible? So, when ω_0 is equal to π ω_0 is equal to π this case, ω_0 is equal to π ; that means $2\pi f$ 0 is equal to π . So, I can say the f 0 is equal to half.

Let us say I am increasing further; I am increasing ω_0 further. So, I am increasing 0 to π is ok. Let us say ω_0 is greater than π ; that means, let us say I take a sinusoidal signal. So, $\omega \omega_1$, so ω_0 is greater than π and less than 2π . So, I am increasing the ω_0 , not 0 to π from π to 2π .

So, I am increasing ω_0 from π to 2π ok. So, I am taking a signal. So, ω_1 is nothing but ω_0 . So, ω_0 varies, taking from π to 2π . Let us say ω_2 is another frequency of 2π minus ω_0 . So, I can say that ω_2 varies from π to 0 when ω_0 is equal to π , that is π , when ω_0 is equal to 2π ω_2 will be 0. I am saying I am taking ω_2 . Let us take another slide.

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Handwritten notes on a whiteboard:

$$w_0 \rightarrow \pi + 2\pi$$

$$w_1 = w_0$$

$$w_2 = 2\pi - w_0$$

$$w_0 = \pi, w_2 = 0$$

$$x_1[n] = A \cos(w_0 n + \theta) = A \cos((2\pi - w_0)n + \theta)$$

$$= A \cos(-w_0 n + \theta) = A \cos(w_0 n + \theta) = x_1[n]$$

So, what I am saying is that I am saying ω_0 varies from 2π . I take ω_1 as a frequency that is equal to ω_0 , and I take ω_2 as nothing but a $2\pi - \omega_0$. Because after $2\pi + \omega_0$ will be ω_0 , that will be this thing that we already proved. So, within 2π . So, it varies from $2\pi - \omega_0$. So, when $\omega_0 = \pi$, then $\omega_2 = \pi$. When ω_0 is equal to 2π , then ω_2 is equal to 0. So, I can say ω_2 varies from π to 0.

Now, if I write $x_1[n]$ equals $A \cos \omega_1 n$ let us say θ is you can write or not write. So, it is nothing but $A \cos (\omega_0 n + \theta)$. Because ω_1 is equal to ω_0 . Now I say x_2 is another signal: $A \cos (\omega_2 n + \theta)$. So, I can say it is nothing but an $A \cos ((2\pi - \omega_0)n + \theta)$, which is nothing but an $A \cos (-\omega_0 n + \theta)$, which is nothing but an $A \cos (\omega_0 n + \theta)$, which is nothing but a $x_1[n]$.

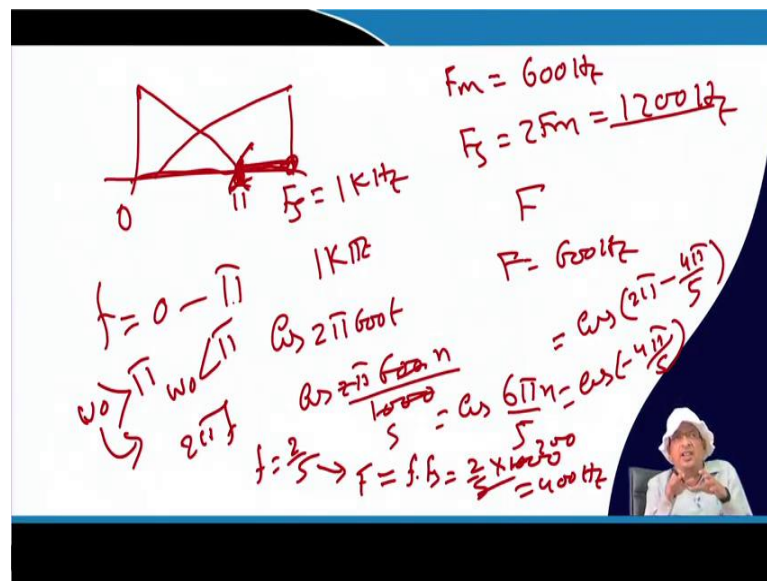
So, I am saying if my ω_2 varies from π to 0. So, if it is ω_0 , it is greater than π and less than 2π . So, when I increase ω_0 above the π , I see that it again comes back to the x_1 . So, I am saying $x_1[n]$ and $x_2[n]$ are aliasing. They are overlapping; they are the same signal, and they are overlapping; because they are the same, they are not distinct signals.

So, I cannot say that ω_0 can vary from above 2π because they are not distinct; they come back again within the x_1 . So, the value of sinusoidal, which varies from 0 to π oscillation, is increasing. Now, once I say that it is greater than π . So, when it is π , then I know the oscillation is maximum; when it is 0, it again comes back to 0. So, as at ω_0 is equal to 0

and ω_0 is equal to that $2\omega_2$ when ω_2 is equal to ω_0 is equal to 2π then also ω_2 is equal to 0.

So, the concept is that a signal oscillation is increasing up to π , then it starts decreasing, and it will become again 0 when ω_0 is equal to 2π . Now, if I can represent it in a graphically. Let us say this one. Again, I draw the same graph.

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So, what I said is this is my 0, this is my F_s and if the signal is aliased. Let us say this is F_s is equal to 1 kHz. Let us say, my baseband signal is F_m is equal to, let us say, 600 Hz. So, what did the sampling frequency say? If F_s is equal to twice F_m , that means Hz, then there will be no aliasing.

So, I can say this is 0 to 600 Hz. Normalized frequency f varies from 0 to π because F_s is when it is π ; that means, it is 600 Hz. If it is F_s is equal to 2 kHz, 1.2 kHz, then this will be the value of π . Now, if you see this value, it will be aliased with this value. So, this value is 600 Hz.

So, if I sample it. So, if the signal is sampled by 1 kHz, what will happen? Let us consider $\cos 2\pi 600 t$ as my analog signal. That is, F is equal to 600 Hz. This is sampled with the sampling frequency 1 kHz. This is nothing but a $\cos 6\pi n$ by 5. Which I can write; how can I write?

Cos 2π minus how much? 6π by 5, I have to do. So, 2π minus 4π by 5. Then, it is nothing but a 6π by 5. So, that is nothing but a cos minus 4π by 5. So, what is the f ? If I say cos $2\pi f$, then my f will be 2 by 5, and my f will be 2 by 5, ok or not. So, what is the analogue frequency if I convert? So, f is equal to f into F_s . So, it is 2 by 5 into 1 kHz 400 Hz.

So, instead of 600 Hz, I get 400 Hz up. So, this is due to the aliasing. So, this is nothing but 400 Hz from this point to this point. So, instead of 600 Hz, I get a 400 Hz signal because it is aliased. So, that is why we said if ω_0 is greater than π , that will be aliased with ω_0 less than π ; is it clear? So, this is the concept of frequency in discrete time sinusoidal.

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Continuous time signal ✓	Discrete time signal ✓
$\Omega = 2\pi f$ Ω is in Radians/sec and F is in Hz $-\infty < \Omega < \infty$ $-\infty < F < \infty$	$\omega = 2\pi f$ ω is in Radians/sample and f is in cycles/sample $-\pi < \omega < \pi$ $-1/2 < f < 1/2$
$\Omega = \omega / T_s$ $F = f \cdot F_s$	$\omega = \Omega \cdot T_s$ $f = F / F_s$
Where F_s is the sampling frequency and $T_s = 1/F_s$	

$F_s = 2f_m$ $f = \frac{F}{F_s} = \frac{2f_m}{2f_m} = 1$

So, please remember this part: continuous-time signal, discrete time signal. Here, F is the analogue F in Hz; that means radian per second. Here is normalized discrete normal this normalized discrete frequency. This F is in radian per sample radian per sample. So, that F is nothing but a F divided by F_s . So, the frequency can vary from minus infinity to infinity, but once it is sampled, it is restricted by to $-\pi$ to π , or normalized frequency can vary from minus half to plus half because I know that F_s is equal to twice f_m .

So, I know f equals F by f_m , F by F_s . So, it is nothing but a twice f_m divided by F_s sorry f_m divided by twice f_m , f_m f_m cancel it is half. Either it will be minus half, or it will be plus half, ok. So, please remember this part because we always use normalized discrete

frequency in signal processing. So, even if it is given in Hz, you have to convert it into radians per sample.

So, this will be used in many ways; when you implement it, if you just consider the Hz, it will not come. You have to modify it to normalize discrete frequency. So, that concept will again take forward because this concept will be used in many places of signal processing.

Thank you.