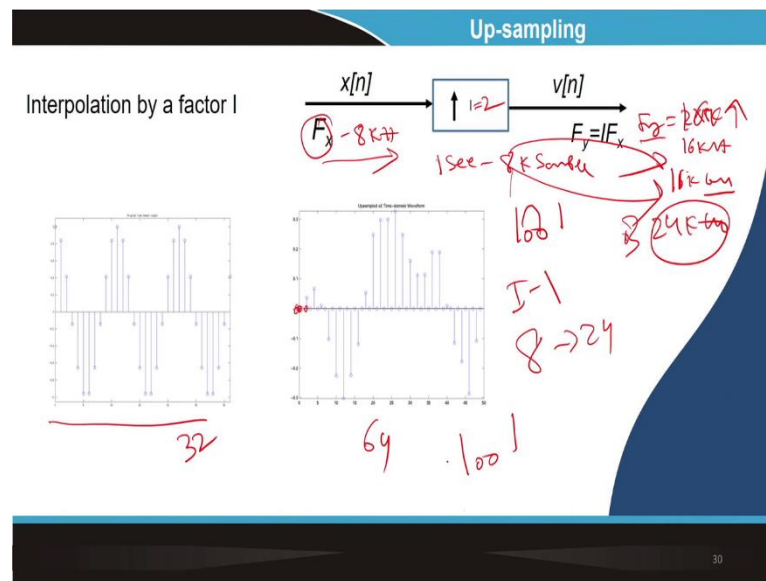


**Signal Processing Techniques and its Applications**  
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**Advanced Technology Development Centre**  
**Indian Institute of Technology, Kharagpur**

**Lecture - 52**  
**Fractional Rate Conversion**

Now, I come to sampling or interpolation. So, what is upsampling?

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.So, upsampling means that  $F_x$ , let us say  $F_x$  is equal to 8-kilo hertz. I want to make  $F_y$  equal to  $I$  is equal to 2, then  $F_y$  will be 2 into 8-kilo hertz; that means 16 kilohertz. So, that factor  $I$  is equal to 2. So, let us say I have a signal  $F_x$  is equal to 8 kilohertz. So, within 1 second of the signal, there is an 8 k sample.

Now, once I make it 2,  $F_y$  equals 16 kilohertz. So, within 1 second, I have a 16 k sample, 16 kilo sample. So, I have to double that number of samples. In the same example, there is a 32 sample. So, if I want to make it double frequency, the number of samples should be 64. So, what will we do? In between 2 samples, I put a 0.

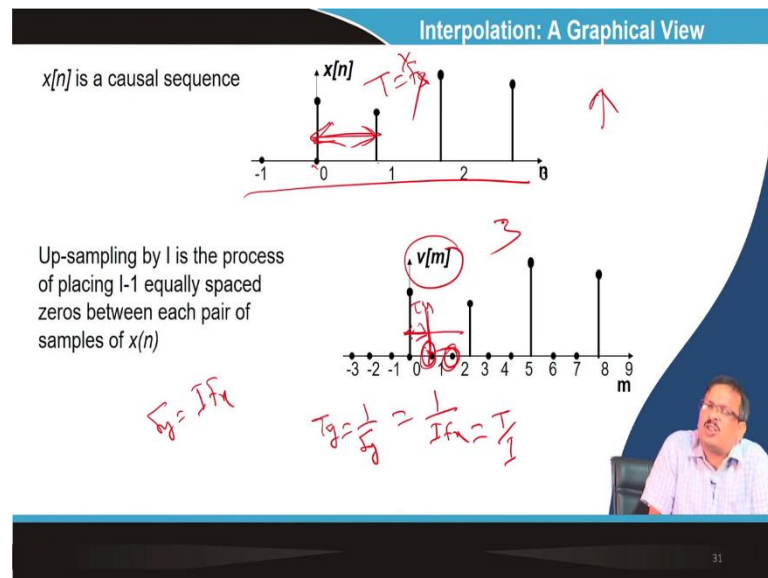
Now, if I want to make it 3 times, then, in between 2 samples, I have to put two 0. So, a number of 0 will be  $I$  minus 1. If it is factor 2, then 1 sample one 0. Suppose it is factor 3, then two 0. If it is factor 4, then three 0. I have to put three 0s because factor 2 means I

have to double the number of samples. Factor 3 means I have to 3 times that; when it is 3, then it will be 8 into 3 24 k samples in 1 second.

So, I have to make it 24K. So, 8k to 24k. So, in between 2 samples, I have to put two 0 OKs. So, if I minus 1 number of 0, I can put 0. So, there is a sample value, and there is a sample value I put 0. So, what I am doing is interpolating the curve.

So, upsampling is nothing, but an interpolation down sampling is nothing but a discard. I have discarded the sample. So, dissemination or this disseminated the sample and discarded the sample. Here, I am inserting the sample. So, that is why it is called interpolation ok.

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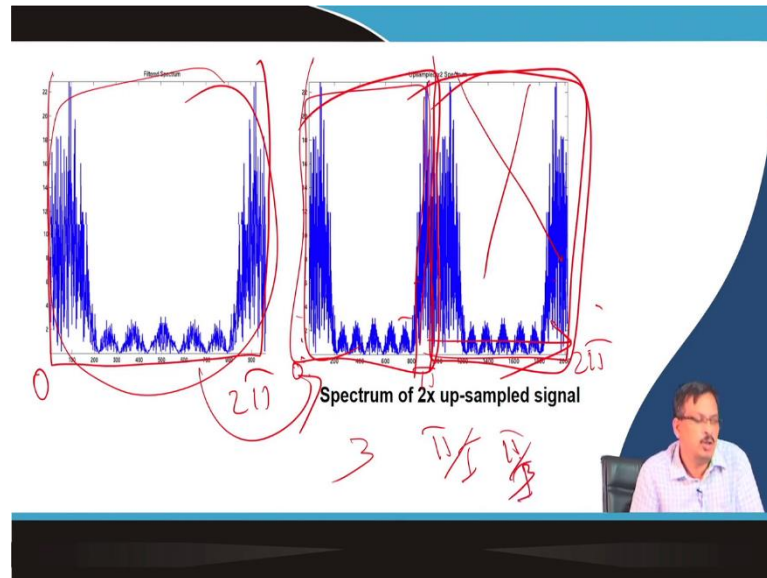


So, I know that up-sampling means interpolation. So, if the  $x[n]$  is the sequence, then the  $v[m]$  is the up sample by 2, 3 times, then I know in between 2 samples, I have to put to another sample. Because I know the distance between the 2 samples here,  $T$  is equal to  $1$  by  $F_x$ .

But here, the distance between the 2 sample this is also a sample is also a sample. So, this is my  $T_y$ . So,  $T_y$  is equal to  $1$  by  $F_y$ . I know the  $F_y$  is equal to  $I$  into  $F_x$ . So, I put  $1$  by  $I$  into  $F_x$ . So, I can say  $T_y$  is equal to  $T$  by  $I$ . So, I have divided the duration, but factor  $I$ .

So, that is why you know that if the sampling frequency increases, then the distance between the 2 samples decreases, which is why it decreases by  $T$ . That is why I got a larger sample.

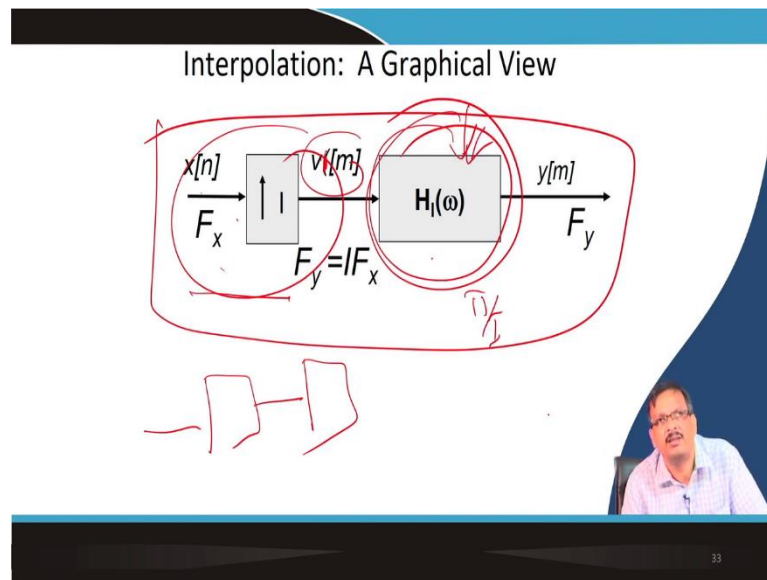
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So, if I say that, what will happen in the frequency domain? Now, if this is my, let us say, spectrum of a signal. So, this is 0, and this is  $2\pi$ . Let us say when I up the sample by factor 2, and I get 2 mirror images. This is 0, and this is  $2\pi$ .

So, a number of mirror images will be. So, if it is, if it is factor 3, I get 3 mirror images using 0 to  $2\pi$ . If it is factor 4, I get 4 mirror images between 0 to  $2\pi$ .

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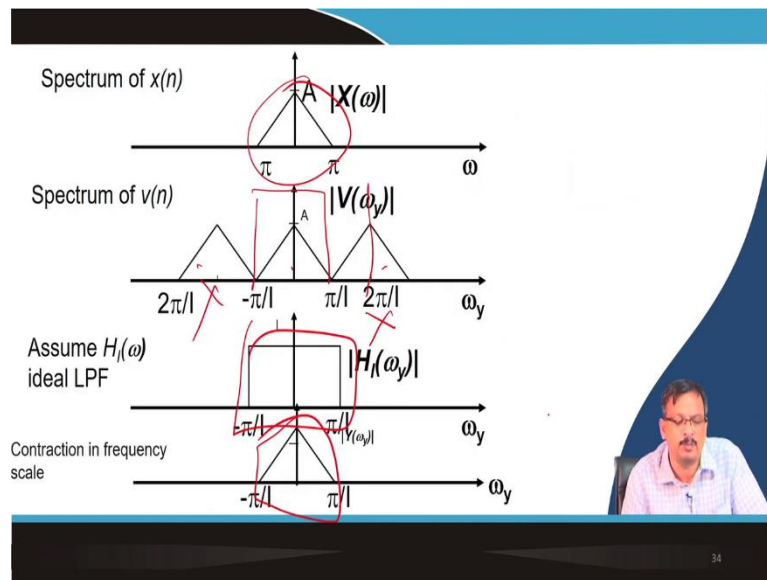


So, I can say I have to analyze in the frequency domain and what? I have to restrict it. So, that means I have a lot of images I have given. I am getting 2 images. So, when I say the 2 images, I am getting, but what do I want? I want a single image.

So, somehow, after sampling, I have to remove these multiple images. So, what do we have to do after half-sampling? I have to pass through a low pass filter again. Because 0 to  $\pi$  I am interested, not this portion I am interested.

So, again, I required a filter low pass filter whose cut-off frequency is  $\pi$  by I. Because it is 2 times, it is  $\pi$  by I  $\pi$  by 2. If it is 3 times, that will be  $\pi$  by 3. So, that means  $\pi$  by I. So, after that sampling, I required a filter whose cut-off frequency is  $\pi$  by I. So, I discarded all the other images and only kept the single image. So, that is an up-sampling diagram, okay?

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Now, let us say what will be given. So, the spectrum of  $x[n]$  is equal to this one. So, the spectrum of  $v[m]$  will be multiple. So, this line will be, you know,  $v[m]$  and will be a multiple of  $x[n]$ . So, a mirror image of  $x[n]$  will be there. So, I only want this portion. So, I designed a low pass filter and kept this portion's rest discarded.

So, that is my requirement. That is why upsampling is also followed by a low pass filter. In the case of downsampling, first low pass filtering then downsampling. In the case of upsampling, after upsampling, I have to reduce that mirror image. That is why I have to put a low pass filter to get the signal ok. So, that is the up-sampling diagram.


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The spectrum of the up-sampled signal  $v[m]$  contains not only the primary band of interest, i.e., from  $-\pi/I$  to  $\pi/I$  but also its images centered around the integer multiples of  $2\pi/I$  radians

This means that signal in the desired frequency range

$$0 \leq |\omega_x| \leq \pi \text{ maps to } 0 \leq |\omega_y| \leq \frac{\pi}{I}$$

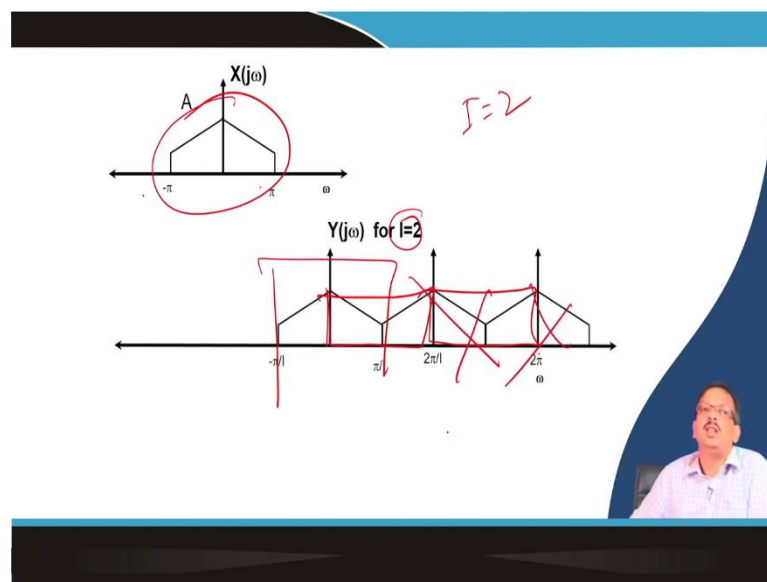
where  $\omega_y = 2\pi f / f_y = 2\pi f / (If_x) = \omega_x / I$



35

Now, the spectrum of the up-sample signal contains not only the primary band of interest but also the image. So, I have to reduce those images by a low pass filter 0 to the mod of  $x$  is equal to  $\pi$  0 to the mod of  $y$  equal to  $\pi$  by  $n$ ; that means minus  $\pi$  by  $I$   $2\pi$  by  $I$  ok.

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Now, let us say here let us say this is my  $j\omega x$  minus  $\pi$  to  $\pi$ . So, after  $I$  equal to 2, I will get 2 that, what I said? If  $I$  equal to 2, I get 2 mirror images.


So, this to this is 1, and this to this is another one. So, within  $2\pi$ , I get 2 mirror images. So, instead of that, I want only a single one. So, I will keep this portion, and this portion will be discarded using a low pass filter.

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The anti-imaging filter  $H_I(\omega)$  (interpolation filter) recovers the primary band of interest.

$$H_I(\omega_y) = \begin{cases} 1 & |\omega_y| \leq \frac{\pi}{I} \\ 0 & \text{else} \end{cases} \quad H_I(\omega_y) = \begin{cases} 1 & |\omega_y| \leq \frac{\pi}{I} \\ 0 & \text{else} \end{cases}$$

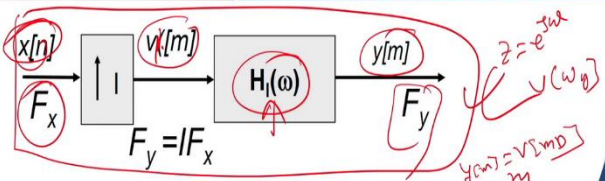
A compression in frequency of the primary image with respect to the original spectrum of  $x(n)$  corresponds to an expansion in the time scale of the signal.



So, again, it is called an anti-aliasing filter or interpolation filter to recover the primary band of interest instead of all other bands being discarded. So, I know it is equal to 1. So, it is 1. So, this is equal to  $\pi$  by 1. So, how is this coming?

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Interpolation by a factor of  $I$



Up-sampling:

$$v[m] = \begin{cases} x\left[\frac{m}{I}\right] & m=0, \pm I, \pm 2I, \dots \\ 0 & \text{else} \end{cases}$$

$$V(z) = \sum_{m=-\infty}^{\infty} v[m] z^{-m} = \sum_{m=-\infty}^{\infty} x\left[\frac{m}{I}\right] z^{-m} = \sum_{k=-\infty}^{\infty} x[k] z^{-Ik} = X(z^I)$$


$$V(\omega) = X(\omega/I)$$

Lowpass filtering:

$$y[m] = \sum_{k=-\infty}^{\infty} h_I(m-k) v[k]$$

$$= \sum_{k=-\infty}^{\infty} h_I(m-kI) x[k]$$

Handwritten notes include:  $\omega/I = 0$ ,  $\omega/I = \pi$ ,  $\omega/I = 2\pi$ ,  $\omega/I = 3\pi$ ,  $\omega/I = 4\pi$ ,  $\omega/I = 5\pi$ ,  $\omega/I = 6\pi$ ,  $\omega/I = 7\pi$ ,  $\omega/I = 8\pi$ ,  $\omega/I = 9\pi$ ,  $\omega/I = 10\pi$ ,  $\omega/I = 11\pi$ ,  $\omega/I = 12\pi$ ,  $\omega/I = 13\pi$ ,  $\omega/I = 14\pi$ ,  $\omega/I = 15\pi$ ,  $\omega/I = 16\pi$ ,  $\omega/I = 17\pi$ ,  $\omega/I = 18\pi$ ,  $\omega/I = 19\pi$ ,  $\omega/I = 20\pi$ ,  $\omega/I = 21\pi$ ,  $\omega/I = 22\pi$ ,  $\omega/I = 23\pi$ ,  $\omega/I = 24\pi$ ,  $\omega/I = 25\pi$ ,  $\omega/I = 26\pi$ ,  $\omega/I = 27\pi$ ,  $\omega/I = 28\pi$ ,  $\omega/I = 29\pi$ ,  $\omega/I = 30\pi$ ,  $\omega/I = 31\pi$ ,  $\omega/I = 32\pi$ ,  $\omega/I = 33\pi$ ,  $\omega/I = 34\pi$ ,  $\omega/I = 35\pi$ ,  $\omega/I = 36\pi$ ,  $\omega/I = 37\pi$ ,  $\omega/I = 38\pi$ ,  $\omega/I = 39\pi$ ,  $\omega/I = 40\pi$ ,  $\omega/I = 41\pi$ ,  $\omega/I = 42\pi$ ,  $\omega/I = 43\pi$ ,  $\omega/I = 44\pi$ ,  $\omega/I = 45\pi$ ,  $\omega/I = 46\pi$ ,  $\omega/I = 47\pi$ ,  $\omega/I = 48\pi$ ,  $\omega/I = 49\pi$ ,  $\omega/I = 50\pi$ ,  $\omega/I = 51\pi$ ,  $\omega/I = 52\pi$ ,  $\omega/I = 53\pi$ ,  $\omega/I = 54\pi$ ,  $\omega/I = 55\pi$ ,  $\omega/I = 56\pi$ ,  $\omega/I = 57\pi$ ,  $\omega/I = 58\pi$ ,  $\omega/I = 59\pi$ ,  $\omega/I = 60\pi$ ,  $\omega/I = 61\pi$ ,  $\omega/I = 62\pi$ ,  $\omega/I = 63\pi$ ,  $\omega/I = 64\pi$ ,  $\omega/I = 65\pi$ ,  $\omega/I = 66\pi$ ,  $\omega/I = 67\pi$ ,  $\omega/I = 68\pi$ ,  $\omega/I = 69\pi$ ,  $\omega/I = 70\pi$ ,  $\omega/I = 71\pi$ ,  $\omega/I = 72\pi$ ,  $\omega/I = 73\pi$ ,  $\omega/I = 74\pi$ ,  $\omega/I = 75\pi$ ,  $\omega/I = 76\pi$ ,  $\omega/I = 77\pi$ ,  $\omega/I = 78\pi$ ,  $\omega/I = 79\pi$ ,  $\omega/I = 80\pi$ ,  $\omega/I = 81\pi$ ,  $\omega/I = 82\pi$ ,  $\omega/I = 83\pi$ ,  $\omega/I = 84\pi$ ,  $\omega/I = 85\pi$ ,  $\omega/I = 86\pi$ ,  $\omega/I = 87\pi$ ,  $\omega/I = 88\pi$ ,  $\omega/I = 89\pi$ ,  $\omega/I = 90\pi$ ,  $\omega/I = 91\pi$ ,  $\omega/I = 92\pi$ ,  $\omega/I = 93\pi$ ,  $\omega/I = 94\pi$ ,  $\omega/I = 95\pi$ ,  $\omega/I = 96\pi$ ,  $\omega/I = 97\pi$ ,  $\omega/I = 98\pi$ ,  $\omega/I = 99\pi$ ,  $\omega/I = 100\pi$ .



Let us say. So, I know that my interpolation filter in the upsampling block diagram looks like this  $x[n]$   $F_x$  is the input sampling frequency  $v[m]$  and then,  $H(\omega)$  and  $y[m]$   $F_y$ . This is my upsampling block diagram.

So,  $x[n]$  is the input. So, I know  $v[m]$  is nothing but a  $x[m]$  by  $I$ . Because if  $x[n]$  is equal to 1, 2, 3, 4, 5, then I know if  $d$  is,  $I$  is the upsampling factor, then I have to know  $I$  minus 1 number of the samples I have to insert in between that. That means that you know that in the case of  $x[n]$ , the distance between the 2 samples is  $T$ . So, it is nothing, but I can say that when I write  $x[n]$ , it is. Basically, it is  $x[n]$  into  $T$ . Now, my  $T_y$  is nothing but a  $T$  by  $I$ .

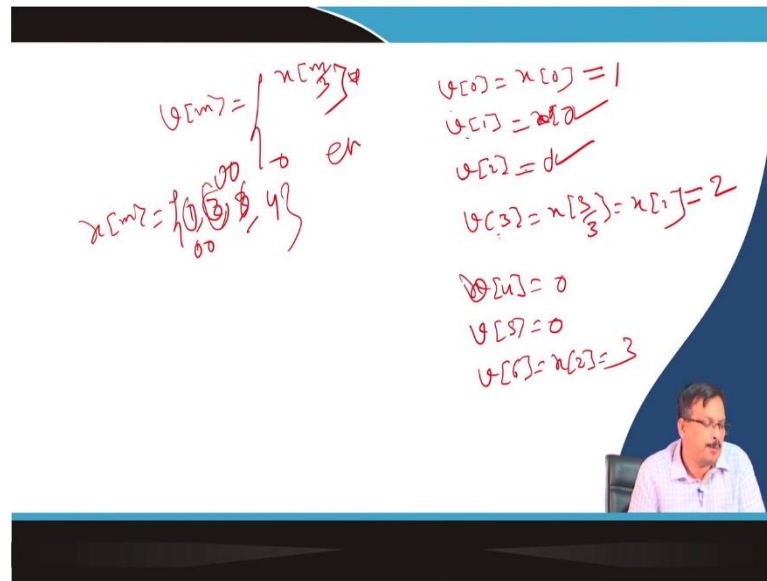
So, my  $y[m]$ , when I say it is nothing, but a  $y[m]$  into  $T$  by  $I$ . So,  $t$  I am not writing; when we write discrete signals, we do not write this  $T$ . But I know that the index is divided by  $I$ . So, I know  $v[m]$  is nothing but a  $x[m]$  by  $I$ . So,  $m$  equals 0; I know it is  $x[0]$  then  $x[1]$ , I cannot get, but 1 by let us say  $I$  equal to 2.

So, can I get it in  $x[1/2]$ ? That is why  $x[1/2]$  is nothing but a 0. Then I get the sample at  $x[2]$  by 2, which is nothing but a  $x[1]$ . So, that is why I said in between 2 samples, I put a 0. So, how do I put a 0? Using this function.

But in the case of downsampling, we said  $y[m]$  is equal to  $v[m]$  into  $d$ . We have discarded that; we have not taken it up. I am not discarding; I am putting 0 between 2 samples. So, if I say  $I$  equal to 3. So, I am putting two 0,  $I$  equal to 3. So, what will happen? So, if I say  $I$  equal to 3, I will take another slide here.



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So, when I say I equal to 3. So, I said,  $v[m]$  is equal to  $x[m]$  by 3 else 0. So, that means I am selecting  $v[0]$  is equal to  $x[0]$ ,  $v[1]$  is equal to 3 is equal to  $x[1]$  by 3. So, this is not else. So, this is 0.

Then,  $v[2]$  2 by  $x[2]$  by 3 0  $v[3]$   $x[3]$  by 3 is equal to  $x[1]$ . So, let us say I have a signal  $x[m]$  is equal to 1, 2, 3, 4. So, I can say,  $v[0]$  is equal to  $x[0]$  is equal to 1,  $v[1]$  is 0,  $v[2]$  is 0,  $v[3]$  is equal to 2. So, in between the sample 1 and 2, I put two 0 understand. Similarly,  $x[4]$   $v[4]$  is 4 by 3 is 0,  $v[5]$  5 by 3 0,  $v[6]$  6 by 3,  $x[2]$  is equal to 3.

So, in between 3 and 3 samples, I put another two 0. So, upsampling means the meaning of the upsampling is  $v[m]$  sequence is generated from  $x[n]$  selecting the signal index, which is an integer multiple of  $I$ , and elsewhere it is 0. Now I can say, let us say, what is  $V(z)$ ?  $V(z) = v[m] * z^{-m}$ .

So, what is  $v[m]$   $x[m]$  by  $I$  elsewhere it is 0,  $z^{-m}$  into 0 is nothing, but a 0. So,  $x[m]$  by  $I$  into  $z^{-m}$ . So, again, if I change the index. So, it will be  $x[m]$ , and this will be  $z^{-m}$   $I$ . So, I can say this is nothing but an  $X[z^I]$ . Because  $z$   $I$  whole  $-m$ . So, I act as a  $z$ .

Now,  $x[z]$   $I$ . Now, I can know the  $V(z)$ .  $V(z)$  is equal to  $x$  instead of  $X[z]$ ; it is  $X[z^I]$ . So, if I say  $z$  is equal to  $\omega$ . So,  $V(\omega y)$ . So, I know  $z$  is equal to  $e^{j\omega}$  ok. So,  $V(z)$  is  $V(\omega y)$  is equal to  $x$ ,  $x(\omega x)$ .

So,  $\omega x$  into  $I$ . So, it is  $e^{j\omega y}$  into  $I$ . So, I said  $\omega y$  into  $I$   $z$  is equal to  $e^{j\omega y}$ . So, this side is  $\omega y$ , which is  $z^I$ . So, it is  $\omega y$  into  $I$ .

Now, this signal will be low pass filter by a filter. So, the filter impulse response is  $m$   $k$  dash. So,  $k$  dash is equal to  $k$  again. It is nothing but an index changing  $k$   $l$ . So, it is nothing but a  $h$   $l$   $m$  minus  $k$   $l$  into  $x$   $k$ . Because this is  $\omega$ , this is  $x[z^I]$  or I can say,  $v[m]$  is equal to  $x[m]$  by  $I$  ok.

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Handwritten notes on the slide:

- $H_I(\omega_y) = \begin{cases} 1 & |\omega_y| \leq \frac{\pi}{I} \\ 0 & \text{otherwise} \end{cases}$
- $Y(\omega_y) = \begin{cases} X(\omega_y I) & |\omega_y| \leq \frac{\pi}{I} \\ 0 & \text{otherwise} \end{cases}$
- $y(m) = \begin{cases} x\left(\frac{m}{I}\right) & m = 0, \pm I, \pm 2I, \dots \\ 0 & \text{else} \end{cases}$
- The output  $Y(z)$  of the Interpolator:  $Y(z) = H_I(z) V(z) = X(z^I)$  where  $H_I(z)$  is the ideal LPF.
- Relationships:  $z = e^{j\omega_y I}$ ,  $\omega_y = \frac{\omega_x}{I}$ ,  $F_y = \frac{F_x}{I}$ ,  $T_y = I T_x$ .
- Diagram of a rectangular pulse from  $-\pi/I$  to  $\pi/I$  with height 1.
- Diagram of a discrete-time signal  $x[n]$  with a period of  $I$ .

Now, I can say what the value of this constant is. So, when I say that, what is the output of this filter? So, what is the filter? So,  $H_I(\omega)$ . So, I require a filter which is instead of 1  $\omega y$  less than equal to  $\pi$  by  $I$  equal to instead of 1; I want this to be  $C$ ;  $C$  is a constant; otherwise, it will be 0. So, I can say I required a low pass filter, which will be  $C$  up to  $\pi$  by  $I$ . After that, it will be 0.

So, minus  $\pi$  by  $I$  to minus  $\pi$  by  $I$  to plus  $\pi$  by  $I$ , it will be  $C$  that I want. So, how do I get the value of the  $C$ ? I know  $y$   $y$  if you see that  $y$   $y$  is nothing but a  $C$  into  $x$  of. So, what is  $y$   $y$ ?  $y$   $y(z)$  I can say  $y(z)$  is nothing but a  $H(z)$  multiplied by  $x[z]$   $v[z]$ . So,  $v[z]$  is nothing but an  $X[z^I]$ . So, I know that is nothing but an  $x[\omega y]$  into  $I$  and  $H(z)$ . I want a constant. So,  $H(z)$  is constant and  $x(\omega_i)$  I know.

Now, what is the initial value theorem? So, I can say  $y[0]$  power of the output. So, output power and input power will be the same. I have a sampling rate conversion system input

power, but I cannot generate the power. So, power will be the same. So, I can say  $y[0]$ . So, the power  $y[0]$  is nothing but a  $1$  by  $2\pi$  summation of all the spectra. So, I summing all the spectra. So, it is nothing, but the  $C$  by  $2\pi$  x of  $y$  i I by else are  $0$  into  $d y$ .

So, I can say that  $C$  by  $I$ , because you know  $d y$ . So, it is  $x(\omega_i)$   $\omega_y$  into  $I$ . So,  $\omega_x$  is equal to  $\omega_y$  into  $I$ . Because you know  $\omega_y$  is equal to  $2\pi f$  by  $F_y$ ,  $F_y$  is equal to  $I$  into  $x$ . So, I know  $\omega_y$  is equal to  $2\pi f$  by  $I$  into  $F_x$ . So, it is nothing but a  $\omega_x$  by  $I$ . So, I know  $\omega_y$  into  $I$  is equal to  $\omega_x$  and  $\omega_y$  into  $I$ ; that means divided by  $I$ . So,  $\omega_x$  divided by  $I$  will become here. So, it is  $C$  by  $I$  x  $0$ .

So, I know that the constant in this  $C$  is nothing but a  $C$  by  $I$  into  $x 0$ , which must be equal to  $I$ . So, for better, I can I can say that  $I$  is equal, or  $C$  is equal to  $I$  into  $y[0]$  divided by  $x 0$ . So, in normalized cases, the  $C$  must be equal to  $I$ . So, if the  $C$  is  $I$ , then  $H_I(z)$  is nothing but an  $I$ . So,  $y(z)$  is equal to  $I$  into  $X[z^1]$  or I can say the frequency response  $Y(\omega_y)$  is equal to  $I$  into  $x(\omega_y)$ . So, is the interpolation gain I am called interpolation gain? Is it clear? So, it is called interpolation gain.

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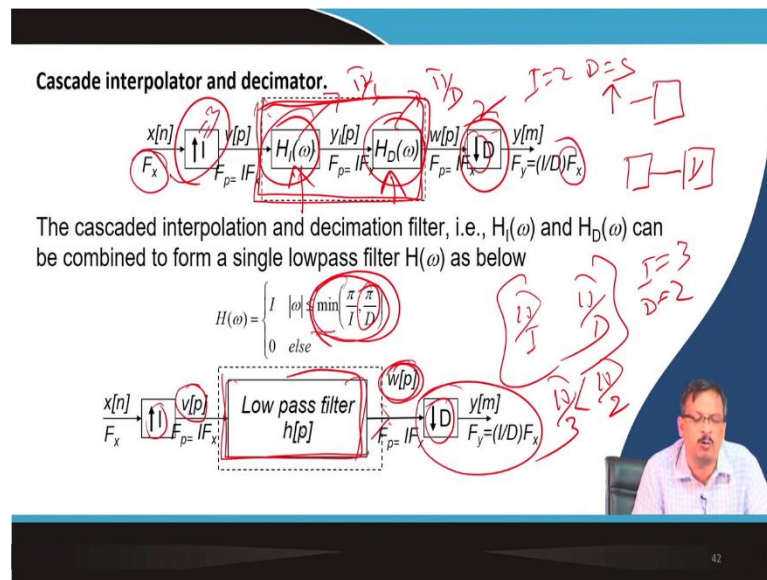
The slide is titled "Fractional Rate Conversion by I/D". It contains handwritten calculations in red ink:

- $F_x = 8 \text{ KHz}$
- $F_y = 12 \text{ KHz}$
- $\frac{F_y}{F_x} = \frac{12}{8} = \frac{3}{2}$  (with a box around the fraction  $\frac{3}{2}$ )
- An arrow pointing up with the number  $3$  (representing interpolation).
- An arrow pointing down with the number  $2$  (representing decimation).

A small video inset in the bottom right corner shows a man speaking.

Now, I said up sampling down sampling I know. Now I said fractional rate conversion. That means, suppose I have an  $F_x$  is equal to  $8$  kilohertz I want  $F_y$  is equal to  $12$  kilo hertz. That means, I know that  $F_y$  by  $F_x$  is equal to  $12$  by  $8$  is equal to nothing, but a  $3$  by  $2$ . So, I have to up sample by  $3$  times and down sample by  $2$  times. So, this is called fractional rate converter.

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So, how do I do the fractional rate conversion? So, in this case, the sample was increased by 3 times. So, here I equal to 3 and downsample by 2 times. So, you know that upsampling means upsampling followed by a low pass filter, and down-sampling means a low pass filter followed by a down sampler. So, upsampling followed by a low pass filter down-sampling means a low pass filter followed by a down-sampling.

Now, this is also a low-pass filter. This low pass filter cut-off frequency is  $\pi$  by I. This low pass filter cut-off frequency is  $\pi$  by D. So, can I combine these two low pass filters? Yes, I can combine these two low-pass filters. So, let us say I combine this low pass filter, then. What should be the cutoff frequency?

So, whichever will be the minima,  $\pi$  by I and  $\pi$  by D. So. Because I have to rest. So, let us say I is equal to 3 and D is equal to 2. So, if it is  $\pi$  by 3 then, then it supports  $\pi$  by 2. Because  $\pi$  by 3 is less than  $\pi$  by 2. So, I can say if I restricted the output signal of the up sampler to  $\pi$  by 3, it would support the restriction of  $\pi$  by 2. That is why I said a minimum of this is required. in case of, let us say, and I equal to 2 and D equal to 5, then the output minima will  $\pi$  by 5 only. Because I have to the input, which the down sampler can accept, is only  $\pi$  by 5. So, I cannot produce an output that is  $\pi$  by 2.

So, the cut-off frequency is the minimum, whichever is the minimum, but yes, I can combine these two low-pass filters in a single low-pass filter. So, the input is  $v_p$ , and the

output is  $w(p)$ . And then it passes through a disseminator. Let us analyze the system's first-time domain then, the frequency domain

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Analysis of Fractional Rate conversion

**In Time Domain:**

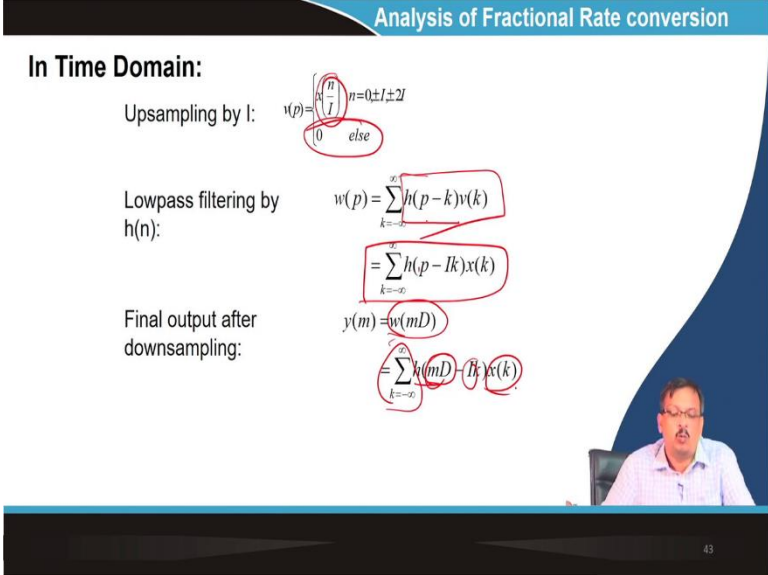
Upsampling by  $I$ : 
$$v(p) = \begin{cases} x\left(\frac{p}{I}\right) & n=0, \pm I, \pm 2I \\ 0 & \text{else} \end{cases}$$

Lowpass filtering by  $h(n)$ : 
$$w(p) = \sum_{k=-\infty}^{\infty} h(p-k)v(k)$$

$$= \sum_{k=-\infty}^{\infty} h(p-Ik)x(k)$$

Final output after downsampling: 
$$y(m) = w(mD)$$

$$= \sum_{k=-\infty}^{\infty} h\left(\frac{mD}{I} - Ik\right)x(k)$$



So, time domain analysis. So, I know what  $v(p)$  is? What is  $v(p)$ ?  $v(p)$  is nothing but an up sample by  $I$ . So, I can say  $v(p)$  is equal to  $x[n]$  by  $I$  else it is 0 unless the low pass filtering is  $h(p)$ .

So, once the  $h(p)$  is, what is the output of the low pass filter  $v(p)$  as an input and the low pass filter and output as a convolution? So, I convoluted the  $v(p)$  with low pass filter  $w(p)$ . So, this is the convolution, ok? Now, I have the final output after downsampling. So then,  $w(p)$  is down sample by  $D$ .

So, now that  $w(p)$  is downsampling, I know  $y[m]$  is equal to  $w(mD)$ . So, I can say it is nothing  $k$  equal to infinity  $h(mD)$ . So,  $p$  is changed to  $mD$  I  $k$  into  $x(k)$ . So, if I explain in pictorially, what will be the time domain factors? Let us say pictorial, and I will explain it. So, I will take a slide here.

(Refer Slide Time: 24:52)

Handwritten notes on a whiteboard showing signal processing steps:

- Top left:  $x[n]$  &  $F_x = 8 \text{ KHz}$
- Top right:  $F_y = 12 \text{ KHz}$
- Below  $F_y$ :  $I=3$  and  $D=2$
- Calculation:  $8 \times 3 = 24$  and  $24 / 2 = 12$
- Signal definition:  $x[n] = \{1, 2, 3, 4, 5, 6, 7, 8\}$
- Upsampled signal:  $V[m] = \{1, 0, 2, 0, 3, 0, 4, 0, 5, 0, 6, 0, 7, 0, 8, 0\}$
- Filtered signal:  $y[m] = \{1, 0, 2, 0, 3, 0, 4, 0, 5, 0, 6, 0, 7, 0, 8, 0\}$
- Final output:  $y[n] = \{1, 2, 3, 4, 5, 6, 7, 8\}$

So, I have a signal  $x[n]$  whose sampling frequency is  $F_x$  is equal to 8 kilohertz. Now, I want to make it  $F_y$ , which is equal to 12 kilohertz. That means the up sample is 3 and the down sample is 2, with  $I$  equal to 3 and  $D$  equal to 2. An up sample is measured by 3; that means 8 are divided into 3, and a down sample is measured by 2. So, it is 24, and downsampling by 2 means 12. Ok, done.

What is the or suppose now, not less  $x[n]$ , equal to this? Let us say 1,2,3,4,5,6,7,8 are all the sample values. So,  $x[0]$  is equal to 1,  $x[1]$  is equal to 2,  $x[2]$  is equal to 3. So, I know this.

Now, what is up? Sampling upsampling by 3. That means, in between 1 and 2, I put two 0, minus 1. So, 3 minus 1 means two 0. So, if I say  $v(p)$  is my up sample signal, then  $V$  is my up sample signal. So, I know  $V[0]$  is equal to 1,  $V[1]$  is equal to 0,  $V[2]$  is equal to 0, and  $V[3]$  is equal to 2. Because  $V[m]$  is equal to  $x[m]$  by  $I$  else 0,  $V[4]$  is equal to 0,  $V[5]$  is equal to 0, and  $V[6]$  is equal to 3.

So, in between 2 samples, I put 0, two 0. Then, I did the filtering. So, filtering does not change the index of the signal. So, I get the same  $V$  b signal  $v(p)$   $V[m]$  is passed through a filter, filter only changes the characteristics of the signal the length-wise there will change no change.

Then, after the filtering  $V[m]$  plus the  $V[m]$  after filtering, I get  $W[m]$ . So, this is equal to, let us say,  $W[m]$  filter is equal to 1 let us say. So, I get  $W[m]$ . I will get  $W[m]$ . Once I get the  $W[m]$  from the  $v[m]$ , then I have to get  $y[m]$  downsampled by 2. So, that means I am not taking all the samples.

So, let us say  $y[m]$  is equal to  $V[m]$ . Let us say that  $y[m]$   $v[m]$  or  $W[m]$  is equal to  $V[m]$ . Let us say that  $W[m]$  is equal to  $V[m]$ . Let us consider that the filter is equal to 1. Let us say the filter basically multiplies the  $v[m]$  with 1. So, now, I can say  $y[m]$  equals  $W$  into 2  $m$  because  $v$  equals 2. So, I am selecting  $y$ . So,  $y[0]$  is equal to  $W[0]$ , is equal to  $V[0]$ ,  $y[1]$  is equal to  $w[2]$ , is equal to  $v[2]$ , which is equal to 0.

So, I am selecting this sample then  $y[2]$  is equal to  $w[4]$  is equal to  $v[4]$ , and 4 is equal to 0. So, this sample is this sample. So, while I am doing multi-rate or fractional up-sampling downsampling, in that case, I am not required to compute all the interpolation. I can only compute this signal this  $V$  this  $V$  this  $V$  and this  $V$  others  $V$ s are not important to me.

So, now, if I give you another signal, let us say I give you  $x[n]$ , and then, I give you that  $I$  equal to 2 and  $D$  equal to 3; then, if the  $v[m]$  is equal to  $W[m]$  compute the sequence of  $y[m]$  you can do that ok.

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**In Frequency Domain**

$$W(\omega_y) = H(\omega_y) V(\omega_y)$$

where  $\omega_y = 2\pi f / f_s = 2\pi f / (f_x / I) = \omega_x / I$

Using definition of  $H(\omega)$ : 
$$W(\omega_y) = \begin{cases} IX(\omega_y I) & |\omega_y| \leq \min\left(\frac{\pi}{I}, \frac{\pi}{D}\right) \\ 0 & \text{else} \end{cases}$$

Down-sampling of  $w(n)$  by  $D$ : 
$$Y(\omega_y) = \sum_{k=0}^{D-1} W\left(\omega_y - \frac{2\pi k}{D}\right)$$

where  $\omega_y = 2\pi f / f_y = 2\pi f / (f_x / D) = (I/D)\omega_x$

Final output of the fractional rate converter:

$$Y(\omega_y) = \begin{cases} \frac{I}{D} X\left(\frac{\omega_y}{D}\right) & |\omega_y| \leq \min\left(\pi, \frac{\pi D}{I}\right) \\ 0 & \text{Otherwise} \end{cases}$$

*Handwritten notes on the slide:*

- $V(\omega) \propto W(\omega)$
- $Y(\omega_y) = \frac{1}{D} W(\omega_y)$
- $1 \times W(\omega_y) \rightarrow 12 \text{ kHz}$
- $\frac{1}{D} W(\omega_y) \rightarrow 24 \text{ kHz}$
- $\frac{1}{D} W(\omega_y) \rightarrow 6 \text{ kHz}$
- $8 \text{ kHz}$
- $I = 3/2$

Now, what is happening in the frequency domain? So, in the frequency domain, I know  $w(p)$  that if you see the diagram, what is  $w(p)$  that it is nothing but a  $v(p)$  multiplied by the



filter. So,  $w(p)$  is nothing, but a  $h(p)$ .  $w(p)w(\omega p)$  is nothing but a filter frequency response multiplied by the  $v(p)$  frequency response. I know the  $v$  frequency response  $\omega p$  is nothing, but an  $x(\omega p)$  because it is up the sample by  $I$  ok.

So,  $w(p)$  is equal to  $X$  into  $Z$   $w(p)$   $I$ , which I know is used for up-sampling the filter. Now, what do I require for the downsampling filter?  $Y$   $w$  is nothing, but a  $1$  by  $D$  into  $k$  equal to  $0$  to  $D$  minus  $1$   $W$  into  $D$   $y$  minus  $2\pi k$  divided by  $D$ . Now, for imaging to reduce the image after filtering, what do I want? I want to, only  $w$   $y$  by  $D$  and  $1$  by  $D$ . So,  $1$  by  $I$  can say  $(\omega y)$  is equal to  $1$  is  $1$  by  $D$  into  $W$   $\omega y$  by  $D$ , where  $W$   $\omega y$  by  $D$  is nothing, but an  $I$  into  $X$   $p$  the  $\omega p$  by  $I$   $\omega p$   $I$ .

So, I can say that it is nothing, but the  $I$  by  $D$  into  $x[\omega y]$  divided by  $D$ , or if I say  $\omega p$ , then  $\omega y$   $\omega y$  is equal to  $\omega p$  into  $\omega y$  is equal to  $I$  by  $D$  into  $\omega x$ . So, if I said the  $\omega y$  instead of  $\omega y$   $\omega x$ , then I have to put is  $I$  by  $D$  into  $\omega x$  here. So, it will be  $I$  by  $D$  square.

So, the frequency response  $I$  by  $D$  is the amplification factor. So,  $I$  is equal to  $3$ ,  $D$  is equal to  $2s$ , and is  $3$  by  $2$ . Then,  $\omega y$  by  $d$  is the output. So, in terms of frequency, let us say I have an 8-kilohertz signal. So, the maximum frequency content is 4 kilohertz. Now, once I say up the sample by 3 times, that means 24-kilo hertz. So, the maximum frequency content is 12 kilo hertz, then down the sample by 2 times; that means the 12 kilo hertz maximum frequency constant is 6 kilo hertz, understand or not.


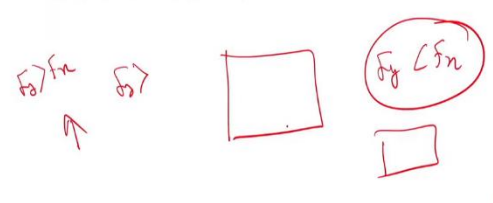
So, it is nothing but a  $3$  by  $2$  into  $4$ , 4 kilohertz is the input. So,  $3$  by  $2$  into  $4$ , that is 6 kilo hertz ok. So, this is the frequency domain representation of the up sampler, up simpler and down simpler. Both combine ok.



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**Summary Fractional Rate Conversion by I/D**

- Interpolation precedes decimation as the spectrum of the intermediate signal  $y(p)$  should, at least, contain all the components of the original spectrum.
- The composite filter  $H(\omega)$  has gain 1 as required due to interpolation by 1.
- If  $F_y > F_x$ , then low-pass filter acts as an anti-imaging filter and for  $F_y < F_x$  then low-pass filter acts as an anti-aliasing filter.



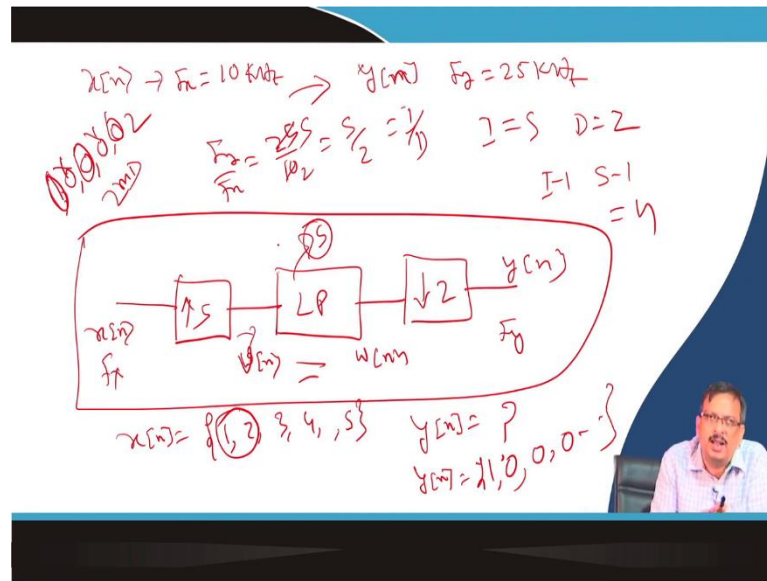
46

Now, so, the summary of the fractional rate conversion by I by D. Interpolation process, decimation, interpolation proceeds, decimation as the spectrum of the intermediate signal  $y(p)$  should at least contain all the components of the original spectrum.

The composite filter has a gain due to interpolation. Now, if you see that  $F_y$  is greater than  $F_x$ , then the low-pass filter acts as an anti-emerging imaging filter. So,  $F_y$  is greater than  $F_x$ , then the input is  $F_x$ , the input is  $F_x$ , and the output is  $F_y$ . So, if  $F_y$  is greater than  $F_x$ , that means it is up-sampling. So, in the case of upsampling, ultimately, it is upsampling. So, in the case of upsampling, the job of the filter is to remove multiple images.

Now, when  $F_y$  is less than  $F_x$ , that is downsampling. The job of the filter is to act as an anti-aliasing filter. So, now, suppose I give you a problem developed. Write down the fractional sample, such as the conversion signal flow diagram or the block diagram, for converting sampling frequency 2.

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Let us say I have a signal  $x[n]$ . I have a signal  $x[n]$  sampling frequency of  $x[n]$  is equal to, let us say, 10-kilo hertz. Write down the functional block diagram of a fractional signal sampling rate converter to convert to  $y[n]$  with the sampling frequency  $F_y$  is equal to, let us say, 25 kilohertz.

So, I want 10 kilo hertz input of 25 kilo hertz. So, in that case, what will I require? So, I know  $F_y$  by  $F_x$  is equal to 25 divided by 10, so, which is nothing but a 5, 2. So, 5 by 2. So, I by D. So, I is equal to 5, and D is equal to 2. So, I know this is my  $x[n]$  up sample upward array by 5 times then, I know low pass filter LP then, I know it has to be downsampled by 2 times, and I get  $y[n]$ . So, this is  $F_x$ , this is  $F_y$  up sample by 5 times, down sample by 2 times. Is it clear?

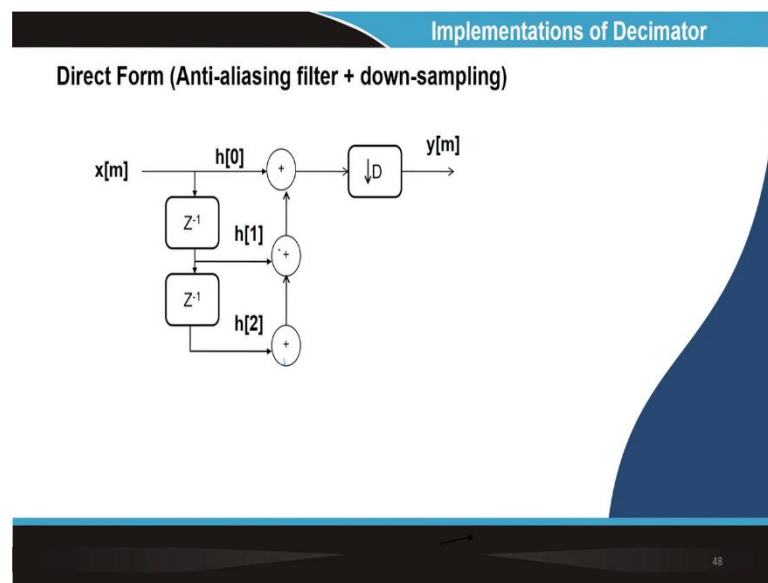
Similarly, I can give any arbitrary number and tell you to draw that signal flow diagram or block diagram of the fractional rate sampling rate converter. Now, if I told you that suppose  $x[n]$  is equal to 1, 2, 3, 4, 5, then what will be the  $y[n]$ ? So, upsampling is 5 means I am equal to. So, minus one 5 minus 1 means four 0 between 2 samples. So, there will be a 1, 0, 0, 1, 0, 0, 0, 0, 2 four 0 in between 2 samples.

Now, down sample by 2, down sample by 2 means,  $m \times D \times m$  into  $2 \times m$ . So, this one, skip on this one. So, the  $y_n$  will be 1 0; then this skips this one 0, then 2 will be skipped 0, so like that, ok. Let us say LP is less if it is  $V[m]$ , if it is  $V[m]$ , and if it is  $W[m]$ . Let us say that  $V[m]$  is equal to  $W[m]$ .  $V[m]$  LP does not change that sampling that  $V[m]$   $W[m]$

sample value, but basically, it will be changed because low pass filter interpolation will happen, and the samples will be changed.

So, I can say that it is nothing but a summation of 2 or 3 fields. Let us say LP is nothing but a summation of successive 5 samples. So, you can do it that way, ok. So, that is up sampling, down sampling and fractional sampling rate conversion.

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The next topic will be the implementation of that up sampler and down sampler in a digital scenario. How do I implement it? How do I implement the low pass filter? Ok. So, in the next class, I will talk about the implementation of those things.

Thank you.