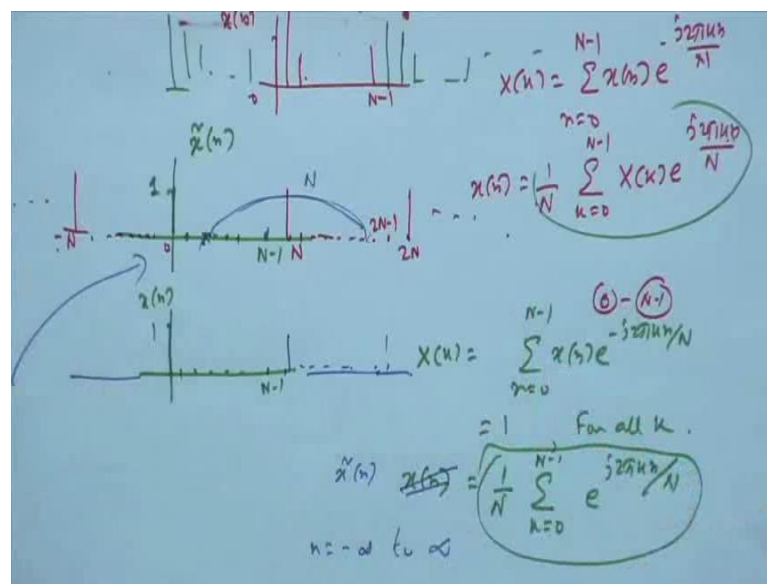


Discrete Time Signal Processing
Prof. Mrityunjay Chakraborty
Department of Electrics and Electrical Communication Engineering
Indian Institute of Science, Kharagpur

Lecture – 22
Decimation and DFT of Decimated Sequences

Okay just recall what we did in the last class towards the end.

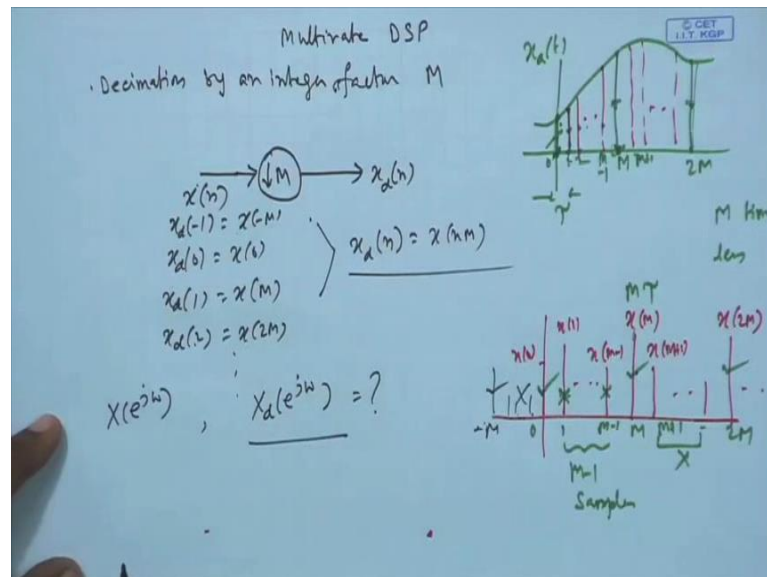
(Refer Slide Time: 00:22)



We considered what periodic sequence, for when one period you have value 1 at n equal to 0 and otherwise all 0s and that got we did it. Then I took out just one period separately called it x_n , so 1 here and all zeros up to N minus 1, we carried out this d f t, that d f t is 1 because if we take the summation $x_n e^{-j2\pi kn/N}$ this x_n is 1, only for n equal to 0, when e to the power 0 is 1 and all other x values are 0s, so 1 into 1 is the result, which is true for all k . Now in the inverse d f t, if you put that back; x_k is 1, so 1 into $e^{-j2\pi kn/N}$ by capital noise, this is x_n ; small n you choose from here, fix here and that n goes summation is over k .

And now in this formula, if you allow small n to go from minus infinity to infinity; this formula then what you get as we have seen is nothing, but periodic repetition of that one block which is nothing, but this.

(Refer Slide Time: 01:39)



So, that x tilde n which I called sampling function, sampling sequence that is actually can be written by this formula. Now I go to, I start another topic which is just, which is a big topic; I will just give some introduction to that only, very basic things I will discuss because we have very common place to; this topic is called Multirate D S P, I will just consider very basic things; Decimation by an integer factor. So, capital N will be an integer, either at least 2 or more than 2 and there is a decimator, this is the first block. One first important component in multirate d s p, as you going to it will understand why you need multirate D S P.

But first let us understand some of the basic blocks, this is one block; we denoted like this that the suppose there is a sequence x_n and we denoted this way this is called decimation and output you can give any name, I call it x_d ; d for decimated $x_d n$. It means, it is like this suppose x_n was obtained, suppose as an example x_n was obtained by sampling an analog function, where is an x a t your sampled here you are here, here here dot dot dot dot here and then again at this point. So, this is suppose 0, 1, 2 dot dot

dot and they this is m , this is $m - 1$, then again dot dot dot dot dot dot dot dot dot then again at $2m$, solve this 0 is sample, first sample, second sample, $m - 1$ sample m th, $m + 1$ th dot dot dot dot this also how you sampled.

But suppose I had sampled, I use a sampling rate which is m times less; that is sampling period m times larger, that is after taking one sample here, next sample would have would be here all right. Suppose this much is your one sampling period τ , so earlier sampling period was τ now it will be $m\tau$. So, 1τ , 2τ , 3τ , $m\tau$ will be here 1τ brings to 1, 2τ brings to 2 $m\tau$ brings 2. So, now, sampling period has become $m\tau$ because sampling speed has gone down, so instead of sampling at all these intermediate points, you sampling here and the next sample you will take here next will take. So, now, it will be zero th sample, it will be first sample, it will be second sample like that which means; even if it is not obtained by sampling and analog signal you have got some x_0 then x_1 dot dot dot may be x_{m-1} then x_m , then again x_{m+1} dot dot dot, then again x_{2m} dot dot dot dot. What we do? This is 1, this $m - 1$ this is m ; $m + 1$. What we do in this process, that is here I have drawn a general sequence, it has got nothing to with time and all that may be you generated the data yourself on a piece of paper or may be it is where computed.

So, just zeroth data, first data, second data but in decimation means if you are decimating by a factor name where after taking one sample you discard this samples $m - 1$ samples, then again take the next, then again these (Refer Time: 05:30) then again take the next like that, which is what you are doing here as though if you are doing sampling in the real time case, earlier you had zeroth sample, first sample, second sample, $m - 1$ th sample, (Refer Time: 05:43) sample dot dot dot dot. Now you are saying no my sampling rate I am bringing down, so I have speed, I am bringing down which means sampling period is very wider by a factor capital m . So, if τ was the sampling period earlier now it is m into τ , so after this zero m into τ will bring you to this point because τ brought you to number 1 then 2τ brought you to 2; $m\tau$ will give you $2m$ that will be m th sample, to be as m th sample in is now be the first sample because $m\tau$ is the new period, so after one period I will get sample first sample, after two period you get second sample.

So, this zeroth sample, first sample, second sample like that, so intermediate for us go, you rename this sample zeroth, m th. Earlier m th now becomes first, earlier 2 m th now become second, earlier 3 m th now become start same here, $x[0]$ you keep, m minus 1 sample just throw away, then this m th you retain. So, this will be a first guy now, then again next m minus 1 sample you throw away, $x[2m]$ you retain, but will be the second sample now dot dot dot dot and this will be called a decimated sequence. Now there is a danger of course, you see here you are bringing down sampling rate by a factor m , so there may be a possibility of (Refer Time: 06:55) because if the function is band limited and you must sample it at more than the (Refer Time: 07:00) rate that is twice the band limiting frequency.

Suppose you had been doing it, but now you are beginning down the sampling frequencies, sampling rate here, so there is a possibility that if you by this bringing down the sampling rate you go below (Refer Time: 07:15) rate there will be might in danger there will be (Refer Time: 07:18), so this you have to be careful. May be your IQ state is less, but your actual sampling rate here you know that at this rate, the τ ; τ period at that rate was very high. So, you could effort to bring it down by some factor m , still remaining higher than the (Refer Time: 07:33) state that is okay, but by bringing down do not bring it below (Refer Time: 07:36), then you are inviting danger in form of earlier scene, it is just a war of (Refer Time: 07:41). So, now $x_d[n]$ means x_d zeroth sample will be original zeroth, x_d first guy will be original m th, x_d second guy will be original 2 m th, x_d minus oneth guy will be original x minus m th because in that this second sample I will throw away, throw away at the I will take from minus m th right. So, which means dot dot dot dot, which means in general x_d if you take n th, it will be if it is 2, it is 2 m which is 1; 1 m , which 0, 0 m ; minus 1 minus m . So, $x_d[n]$ means $x[nm]$, n times m a in m th will be m th sample m th first sample, two m th second sample, three m th third sample, a in m th will be m th sample. So, this is a mathematical description all right $x_d[n]$ and this is the notation, then question is given this d t f t ; what is the d t f t of this into what.

(Refer Slide Time: 08:55)

$$X_d(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x_d(n) e^{-j\omega n}$$

$$= \dots + x_d(-1) e^{j\omega} + x_d(0) + x_d(1) e^{-j\omega} + x_d(2) e^{-j2\omega} + \dots$$

$$= \dots + x_d(-m) e^{j\omega m} + x_d(0) + x_d(m) e^{-j\omega m} + x_d(2m) e^{-j2\omega m} + \dots$$

$$= \dots + x_d(-m) e^{j\omega_1 m} + x_d(0) + x_d(m) e^{-j\omega_1 m} + x_d(2m) e^{-j2\omega_1 m} + \dots$$

$\frac{\omega}{m} = \omega_1$

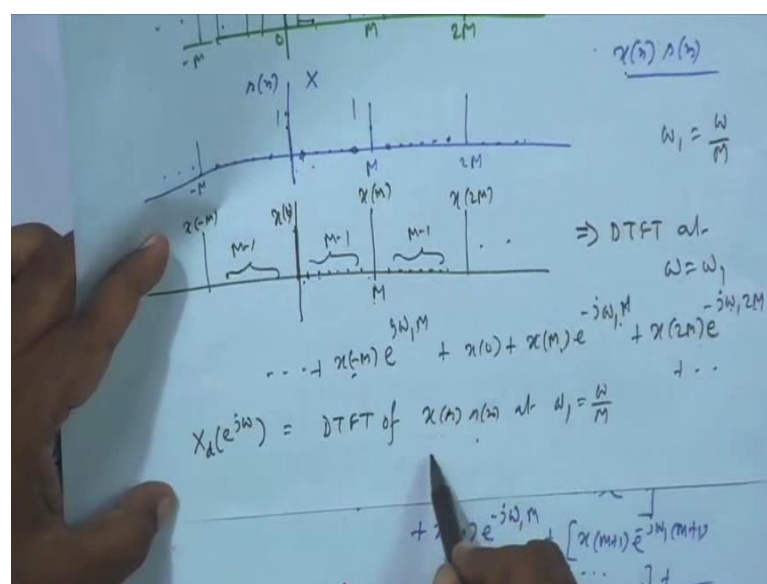
$$X(e^{j\omega_1}) = \dots + x(0) + [x(1) e^{-j\omega_1} + \dots + x(m-1) e^{-j\omega_1(m-1)}] + x(m) e^{-j\omega_1 m} + [x(m+1) e^{-j\omega_1(m+1)} + \dots]$$

This will work out today these not an easy thing; is a tricky thing. So, let us work out; x_d e to the power $j\omega$ means by this formula $x_d(n)$, d t f t formula and I am doing for d t f t same thing you can extend to the more generalize case of z transform same same approach. Now you see $x_d(0)$, $x_d(0)$ e to the power 0; then $x_d(1)$, e to the power minus $j\omega$, $x_d(2)$, e to the power minus $j2\omega$ dot dot dot dot and on the left-hand side $x_d(-1)$, e to the power $j\omega$; it is plus $j\omega$ and dot dot dot dot. But earlier we have seen $x_d(0)$ is same as $x(0)$, but $x_d(1)$ is a m th original sample; first new sample, decimated sample possible of the decimated sequence of the m th of the original. So, I just write x_n , these I keep as it is, in $x_d(2)$; second sample of the decimated sequence if the two m th sample of the original sequence, again minus oneth means originally this x alright.

Now, this ω I can write as ω by m into m , this I can write again ω by m into $2m$, this you can write into m like that. So, ω is radian; ω by m also radian you can call it ω_1 . So, this is x minus m ; e to the power j ; $\omega_1 m$, then $x(0)$, e to the power 0; $x(m)$, e to the power minus j ; $\omega_1 m$, $x(2m)$, e to the power minus j twice $\omega_1 m$ and dot dot dot dot. See, I have to manipulate so that this term becomes closer to the original x , e to the power $j\omega$. Now if you see x , e to the power $j\omega$, now if you see x e to the power $j\omega$; it will have $x(0)$, but after that

there is this terms you know $x[1]$, $e^{-j\omega_1}$, $x[2]$, $e^{-j2\omega_1}$, this is the original d t f t and these then $x[n]$, $e^{-jn\omega_1}$, $x[2M]$, sorry $x[M+1]$, $e^{-j(M+1)\omega_1}$; again dot dot dot dot some term and then again this term will be present. These are not present here, only this is present, this is present, this will be present like that, so these summation, these d t f is part of original d t f t at a frequency ω_1 , where ω_1 is these, ω_1 is these. So, original d t f t at ω_1 has got, you know say generalized expression, this has got the terms presenting these d t f t, there plus additional terms. Now, if I can make this additional terms 0 in the original d t f t, then I will get this that will become equal to the desired d t f t, this $x[n]$ $e^{-jn\omega_1}$. For this what I do is this, suppose this is your original $x[n]$, you may sample, zeroth sample, intermediate samples; I am not so bothered about again these, these intermediate samples; then $x[2M]$ on this side also you have got then $x[-M]$ dot dot dot dot these are original sequence $x[n]$.

(Refer Slide Time: 13:34)



Suppose I multiply it by that sampling sequence I told you; 1, 0, 0, 0 then again at m th point 1, 0, 0, 0; then again 2 m th point 1, 0, 0, 0, then 0, 0, 0, 0, 0 at minus $m-1$ dot dot dot. If I multiply by the 2, let me call it s_n sampling sequence. So, x_n into s_n suppose I do then what will I get, I will get x_0 into 1 and then these fellows will be raised to 0s, this into 0 is 0, this into 0 is 0, this into 0 is 0 then x_m into 1, so like that. So, this will be the decimated sequence x_0 , but in the decimated sequence this fellow x_m was called first sample; it was index one, but now these 0s are brought back, so it will be m th again; x_m into 1, so m th sample will be x_m , this into 0; 0, this into 0; 0. So, $m-1$ 0s and x_m again 0, 0, 0; $m-1$; 0s, x_{2m} because dot dot again $m-1$, zeros x_{-m} right.

If I take d t f t of this sequence at $\omega = \omega_1$, what will I get? x_0 as it is plus 0 into something, 0 into something they do not contribute anything. So, next is x_m , so x_0 then these 0s you forget, then next is x_{mM} , at capital M th point, so it will be minus j , frequency to $\omega_1 m$, then again 0s you forget, next is x_{2m} ; e to the power minus $j \omega_1$ into $2m$ and dot dot dot on these side these zeros you forget; x_{-m} ; e to the power; minus, minus plus will $j \omega_1 m$ and dot dot dot.

And this is what you have here you see, x_{-m} , e to the power $j \omega_1 m$, x_0 , x_m , x_{mM} ; e to the power minus $j \omega_1$ given m , x_{2m} ; e to the power minus $j 2 \omega_1 m$, $2m$ these 2; ω_1 or $\omega_1, 2$ same thing. So, this d t f t, the decimated sequence is same as d t f t of the product of these two sequences, original sequence multiplied by the sampling sequence, at a frequency ω_1 . Remember this d t f t was at a frequency ω , so decimated sequence it is d t f t at a frequency ω is same as d t f t of this product sequence x_n, s_n ; x_n was the original sequence, s_n is the sampling sequence. So, $x_n; s_n$ at ω_1 , where ω_1 is ω by m that is decimated sequence is d t f t at any these is d t f t of $x_n; s_n$ at ω_1 equal to ω by m alright because two expressions are same alright.

So, this we work out we tried to carry out d t f t of these for these two as a ω_1 , alright; that means, x_d , e to the power $j \omega$.

(Refer Slide Time: 18:06)

The image shows handwritten mathematical derivations on a whiteboard. The top part shows a double summation:
$$= \frac{1}{N} \sum_{k=0}^{N-1} \sum_{n=-\infty}^{\infty} x(n) e^{-j(\omega_0 - 2\pi k) \cdot n}$$
 A bracket under the inner summation is labeled with ω_1 and ω . Below this, it is shown that $\sum_{n=-\infty}^{\infty} x(n) e^{-j\omega_1 n} = X(e^{j\omega_1}) = \frac{\omega}{M}$. The bottom part shows the substitution into the main equation:
$$X_d(e^{j\omega}) = \sum_{n=-\infty}^{\infty} x(n) n(n) e^{-j\omega_1 n} = \frac{1}{M} \sum_{n=-\infty}^{\infty} x(n) \left(\sum_{k=0}^{N-1} e^{j2\pi k n / M} \right) e^{-j\omega_1 n}$$

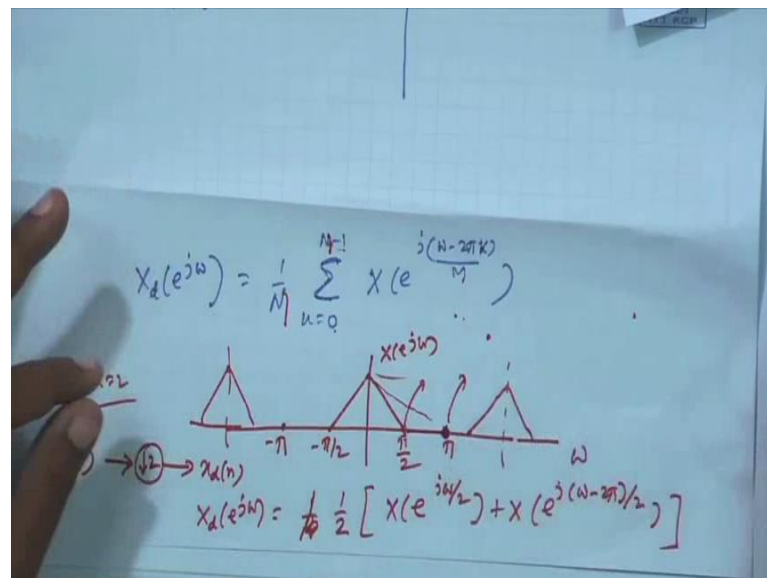
By this expression, it means a product $x(n)$; $s(n)$, e to the power minus $j\omega_1 n$ because at ω_1 , ω is a frequency where you have to find out this and $\omega_d t f t$. Now $s(n)$, I told you the sampling sequence can be represented, if these when are just a minute that time it was called x , now this is called s because call x tilde that time the same thing I am calling $s(n)$ now; the $s(n)$ can be represented like this. So, that I substitute here this $s(n)$ are replaced by that; 1 by capital n will go out, outer summation as before $x(n)$; $s(n)$.

So, summation a equal to 0 to m minus 1; e to the power $j2\pi$ that came from, here n from outside and then multiplied by e to the power minus $j\omega_1 n$; alright and now these we can simplify whenever I know that whenever there is a double summation, next step is to interchange the two summations. So, this summation over k will go out, n from sorry minus infinity to infinity; $x(n)$, e to the power minus j . Now ω_1 , I replace by ω by m , now this is not n , this will be m sorry I forgot to mention because it was; here the period was taken to be capital N ; 0 to n minus 1, but for our business, for our business it was if this is m right.

Just give me a minute, mth yeah here this is the sampling sequence. Earlier I took 0 to 0 and then capital N ; 2 n , three (Refer Time: 20:44) n m capital M ; 2 m . So, that is why this formula; I should not write n it should be m , it should be m ; everywhere should be m

and ω_1 I know is ω_1 is ω by m . So, now here I write $x[n]$, e to the power minus j , I take this common; not ω_1 ω by m , so ω minus; n I will put outside, $2\pi k$ by m into n ; ω by m , ω by m is ω_1 , ω_1 into n with minus, e to the power minus j ω_1 in there is present and another term is minus, minus plus $j 2\pi k$, $j 2\pi k$ small n by m small n by m alright. Now what is this summation, ω is radian; digital frequency minus 2π into k another radian, so ω minus $2\pi k$ is radian another digital frequency, divided by m still radian; another digital frequency, if you call ω' .

(Refer Slide Time: 22:39)



So, this is nothing, but this inner summation is what; summation $x[n]$, e to the power minus j ω' n alright which is nothing but dft at ω' . So, x e to the power j ω' , so this entire thing then turns out to be this; x_d , e to the power j ω is 1 by capital N , this much frequency ω' , so I am writing ω' ; I am writing this full, so this is the dft . Now what is the interpretation of this, for that we will take an example; I do not know whether I have that much time today, but nevertheless we will take an example of just m equal to 2 and suppose I take a first page, suppose this formula I need; suppose what is given is x e to the power j ω , again I tell you cannot plot a complex function by just one plot, I should have mod of these

versus ω and angle of these versus ω , but just to explain my point I am here doing that I am plotting capital X, e to the power $j\omega$ versus ω .

So suppose this is given to you like this okay, now this $x[n]$; that has done these $d[n]$ and these $x[n]$, I am decimating by a factor 2, to get $x_d[n]$, that is M equal to 2; question is what will be these, $x_d[n]$; e to the power $j\omega$ that by this formula will be 1 by n , so it is not n it is m ; everywhere I am making the mistake, there is no capital N here, it is M you be correct, this would have been M . So, now these summation means 1 by M , M is 2 , so 1 by 2 and there were only 2 cases; k equal to 0 and this is M . Look at here, this would be M , wherever you had N ; that becomes M , so M minus 1 , sorry for this mistake there is no capital N for our notation here, these all capital M . So, instead of capital N everywhere, it will become capital M , there is where is capital M ; capital M , half k equal to 0 and $M/2$ means there is 1 , so 0 case and 1 case; if you put k equal to 0 here, you have got $x[n]$, e to the power $j\omega$ by 2 because n is 2 and then $x[n]$, e to the power $j\omega$ minus 2π into 1 because k is 1 now, M is 2 alright if you plot this then you will get that.

This will be in the next class, you can see one thing if you look at this $d[n]$; π corresponds as I told you half some free frequency. Now this is the analog, this is the digital version of the analog band limiting frequency, so here half sampling frequency itself is twice with band limiting frequency; whereas to avoid earlier see it was enough if band limiting frequency equal to (Refer Time: 27:27) sampling frequency so that means, I can still bring down the sampling; rate half sampling frequency is π is giving you analog half sampling frequency going to π , analog band limiting frequency going to π by 2 . So, I am finding half sampling frequency is already twice the band limiting frequency; which is not required, half sampling frequency could be equal to band limiting frequency, and band limiting frequency is fixed. So, I can bring down the half sampling frequency and therefore, sampling produced by factor 2 , so in that case half sampling frequency is always, analog half sampling frequency will always map to π .

So, band limiting frequency will go like this then they will coincide with that and it will be like that, this will see in the next class. Thank you very much, you try this as an exercise by yourself; we will meet in the next class.

Thank you very much.