

Advanced 3G & 4G Wireless Communication
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Lecture - 20
CDMA Near-Far Problem and Introduction to MIMO

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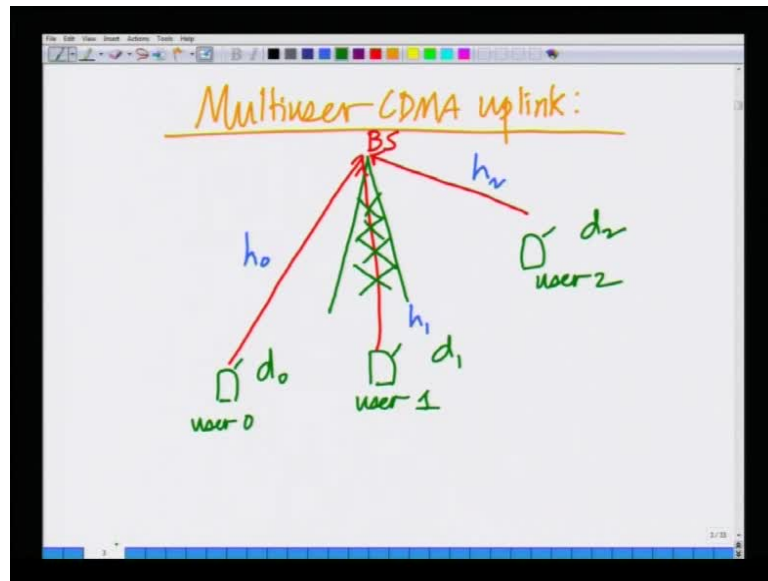
Downlink SNR
for multi-user CDMA
(desired user = user-0)

$$SNR = \frac{N ||h_0||^2 P_0}{\left(\sum_{k=0}^{K-1} P_k \right) ||h_0||^2 - P_0 \sum_{i=0}^{L-1} \frac{||h_0(i)||^4}{||h_0||^2} + \sigma_n^2}$$

Multipath diversity
 Spreading gain
 Power of desired user
 Multipath interference
 Thermal or AWGN

Welcome to the course on 3 G 4 G wireless mobile communication systems. In the last lecture, we have completed the analysis of the CDMA multi user downlink SNR, we looked at that expression which is given over here, which said the SNR is proportional to the spreading gain. It is proportional to the magnitude of the multi path channel between the user and the base station within the base station and the user that is this norm h_0 square represents the diversity its proportional to the power; And in the denominator, we have the multi user plus multi path interference plus thermal and additive white Gaussian noise. So, this is the comprehensive multi user SNR on the CDMA downlink.

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We also looked at the CDMA uplink situation which is shown here. However, uplink is slightly different from the downlink, because in the uplink the signal corresponding to different users traverse different channels and add up at the base station. Hence, this situation is slightly different from the downlink.

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The diagram shows a handwritten equation for Multisuser Uplink SNR. The equation is:

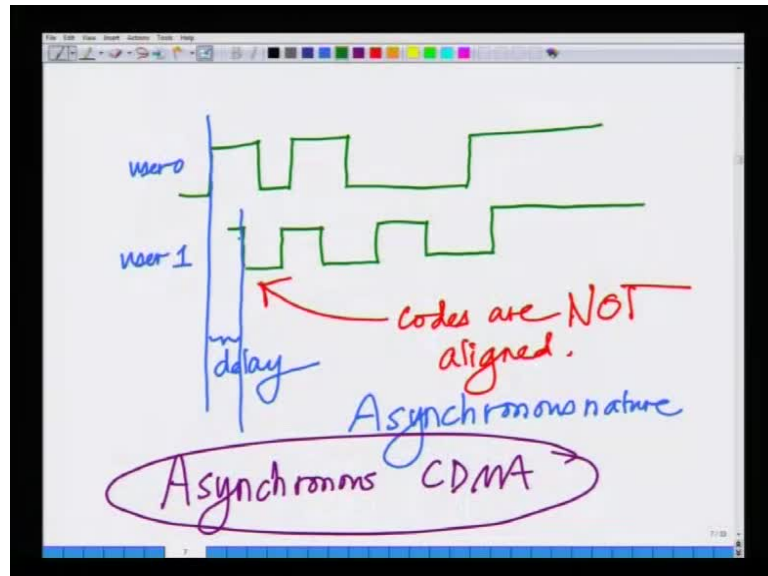
$$\text{SNR} = \frac{N \|h_0\|^2 P_0}{\sum_{k=0}^K P_k \|h_k\|^2 - \frac{P_0}{\|h_0\|^2} \sum_{d=0}^{L-1} |h_0(d)|^4 + \sigma_n^2}$$

The equation is circled in red. The title 'Multisuser Uplink SNR' is written in blue at the top.

However, the expression for the downlink can be suitably modified now instead of having some power into norm h_0 square, each power is affected by norm h_k square, where h_k is

the channel corresponding to user k that is the only change. So, x of that the expression is more or less the same.

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However, there is another challenge in uplink. Now, because different users are at different distances that is what we said because different users are different distances hence the propagation delays for the different signals are different. Hence, the signals at base station are asynchronous; that is if you look their edges they are not synchronized.

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$$E\left\{\left(\frac{1}{3N}\right)^2\right\} = \frac{1}{3N} + \frac{1}{3N}$$

$$= \frac{2}{3N}$$

asynchronous cross correlation Variance

And to analyze this situation, we have to derive the asynchronous cross correlation between the sequences or the asynchronous the (()) variance, the asynchronous interference. We said the variance of asynchronous interference is two-thirds over N, this also falls as 1 over N,0 in fact it is less than the synchronous interference because of more randomness it is two-thirds N.

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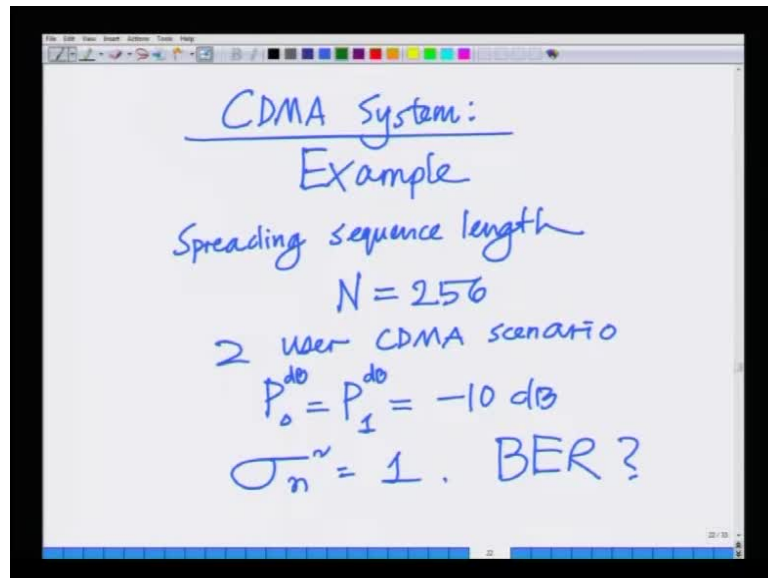
$$SINR_{u,a} = \frac{N \|h_a\|^2 P_a}{\sigma_n^2 + \frac{2}{3} \sum_{k=0}^K \|h_k\|^2 P_k - \frac{2}{3} \frac{P_a}{\|h_a\|^2} \sum_{d=0}^{L-1} |h_{ad}|^2}$$

async cross correlation variance.

uplink asynchronous multiuser CDMA SNR

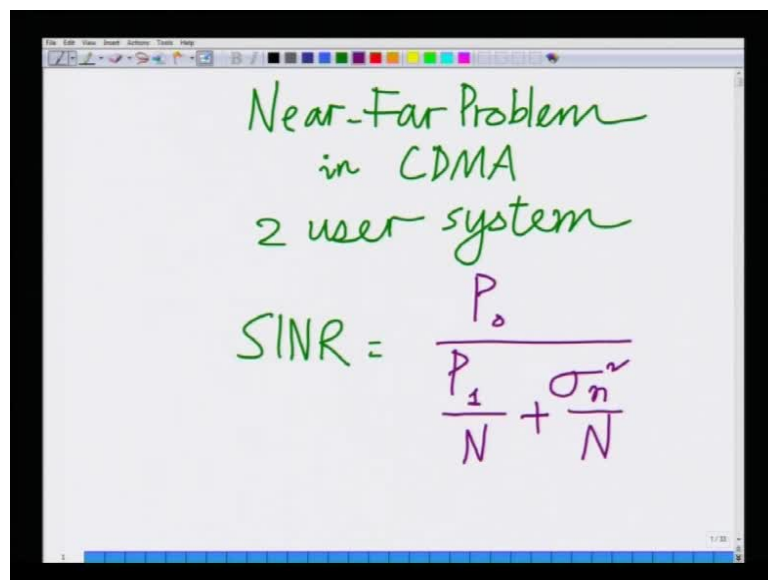
Then we simply derived the asynchronous uplink multi user SNR as $N \text{ norm } h \text{ square } P$ naught, accept that the multi user and multi path interference are replaced by a factor 2 over 3. That is the only difference between the synchronous and the asynchronous link that is the power the variance of the multi user interference is reduced by a factor two-thirds that is raising because of the asynchronous nature.

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Finally, we looked at an example of a CDMA system all right? So, that is where we stopped in the last lecture. Now, let us continue with today's lecture. Today, before I move on to a new topic, let me finish one last minor topic in CDMA systems.

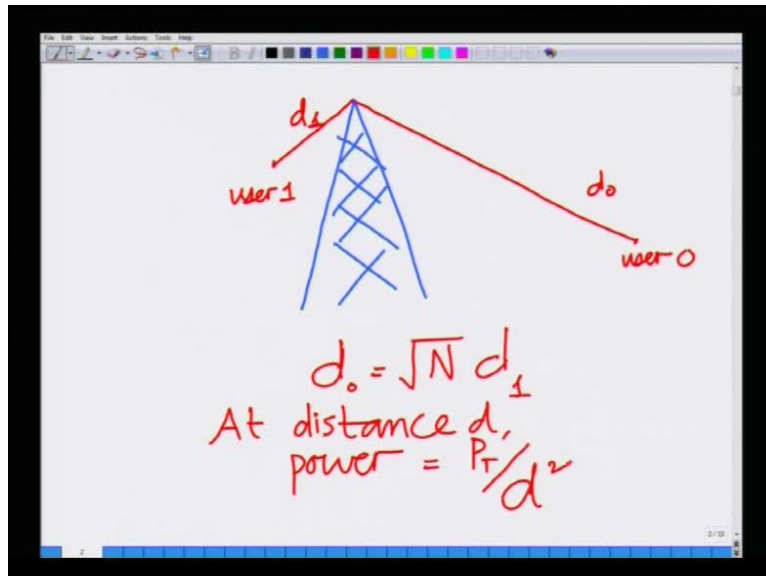
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That is I want to talk about one last topic that is the Near Far problem that is the Near Far problem in CDMA systems. Now, we know that in CDMA systems the SNR, let us consider a simple 2 user CDMA system consider a 2 user system. We know that the SNR is given as the

SINR at user 0 is given as $P_0 / (P_1 + \sum_{i=1}^N P_i)$, this is the multi user SNR in a CDMA system. Depic that scenario over here.

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So, I am considering a base station. Now, in this base station there is user user 0, who is at a distance d_0 and there is a user 1 who is closer to the base station at distance d_1 , all right? So, user 1 is closer to the base station compared to user 0, such that d_0 equals square root of N times d_1 . Let us consider a situation where d_0 square root of N times d_1 . We know square root of N is the length of the CDMA spreading sequence.

So, user 0 is at the edge of the cell he is much farther compare to user 1, who is closer to the base station that is the situation we are currently considering. Now, we also know that as the distance at a distance d , let us say the transmit power of the base station is P_T , then because of path loss the power the transmit power decays as a function of 1 over d square. That is if the transmit power is P_T at a distance d , the power received is P_T / d^2 at...

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$$P_0 = \frac{P_T}{d_0^2} = \frac{P_T}{N d_1^2}$$

$$P_1 = \frac{P_T}{d_1^2}$$

$$\text{SINR} = \frac{P_0}{\frac{P_1}{N} + \frac{\sigma_n^2}{N}}$$

So, let me write that at distance d the power received equals P_T over d square, which implies since P_0 is at a distance d_0 , the power received P_0 is P_T over d_0 square and P_1 equals P_T over d_1 . However, d_0 is square root of N times d_1 . Hence, P_0 is nothing but P_T divided by N times d_1 square since, d_0 is square root n times d_1 . Now, when we write the expression for the received SINR at user 0, the SINR equals P_0 divided by P_1 over N plus σ_n^2 over N . I will substitute for (σ_n^2) and P_1 in this expression. I will write SINR equals P_T over $N d_1$ square plus P_T over $N d_1$ square plus σ_n^2 over N . This can be simplified as P_T over P_T plus d_1 square σ_n^2 .

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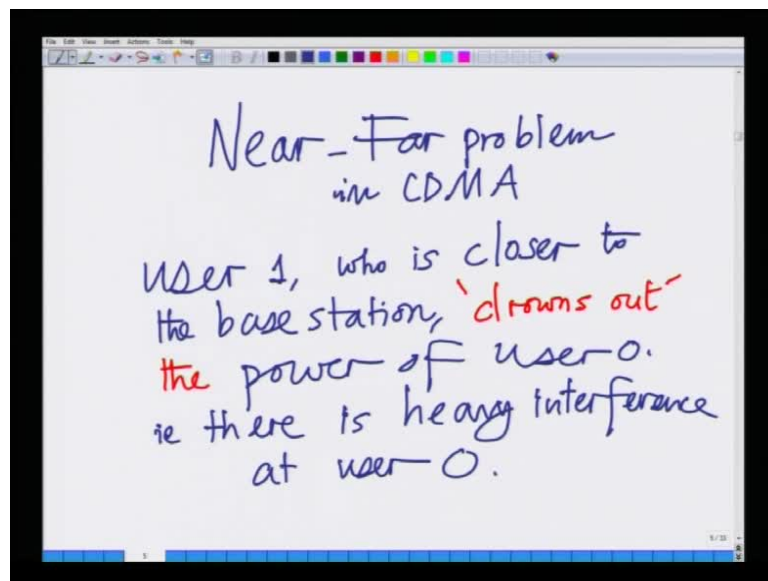
$$\text{SINR} = \frac{\frac{P_T}{N d_1^2}}{\frac{P_T}{N d_1^2} + \frac{\sigma_n^2}{N}}$$

signal Power \equiv P_T
and Multiuser interference are of same MAGNITUDE!

Now, you can see from this expression both the numerator signal power is P_T and more importantly the denominator is also P_T . So, what is happening is because user 0 is at a much larger distance compare to user 1 has received power is very small. In fact because of their choice of distances, I constructed this distances. Keeping this expression in mind in fact from our choice of distances it is so happening that because user 1 or user 0 is root N times the distance of user 1. His power is 1 over N times the power of user 1, which means he is essentially losing all his spreading gain.

Now, the signal power and interference are both the same magnitude. If you look at this expression the signal power and multi user interference are of same magnitude, which means the interference is very high compare to the multi user into compare to the signal power. Whereas, previously the interfering power was suppressed by a magnitude of N, now because one of the users is much farther compare to the other user his received power is very low. Hence, interference is very high, this is known as the Near Far problem in CDMA.

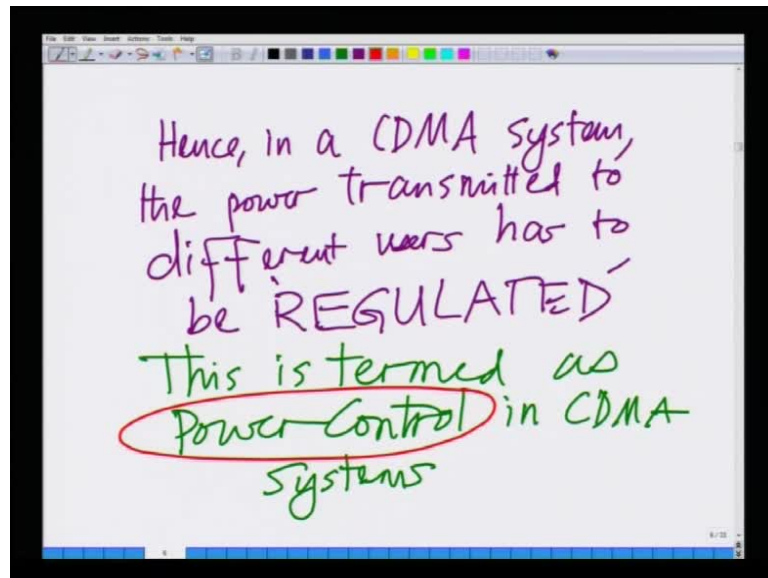
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So, this is precisely the Near Far problem in CDMA, where a user is closer to the base station because his transmitting his received power is much his his power is much higher drowns out the power of the farther user, all right? Or the user who is farther from the base station, so even though so user 1. So, in the previous situation user 1, who is closer to the base station, he sort of drowns out, he drowns out the power of user 0. That is he causes very heavy interference. That is, there is heavy, that is there is very heavy interference at user 0. This is

termed as Near Far problem. Hence, what is essential in every CDMA system is that, to the closer users you have to transmit lower power compare to the users who are farther to who you have transmit, more power. This is known as power regulation or power control.

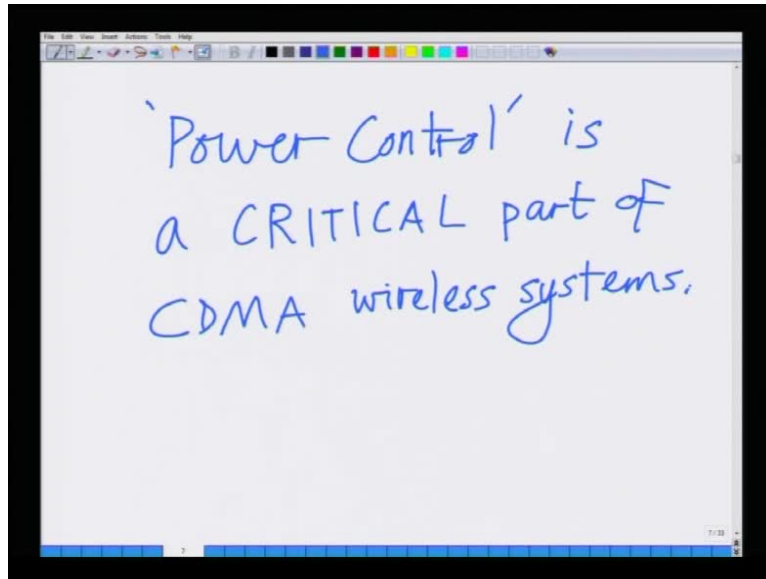
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Hence, in a CDMA system the power transmitted to different users has to be regulated. Hence, in a CDMA system the power transmitted to different users has to be regulated and this is termed at this process of regulation of the power, this process of transmitting more power to the farther farthest users and less power to the users who are closer to the base station is termed as power control in CDMA system. This is termed as power control in CDMA systems. Hence, power control is a key part in 2 G CDMA systems that is i s 95 based CDMA systems, 3 G CDMA systems the release 99 sort of systems power control is an extremely important part of any CDMA system.

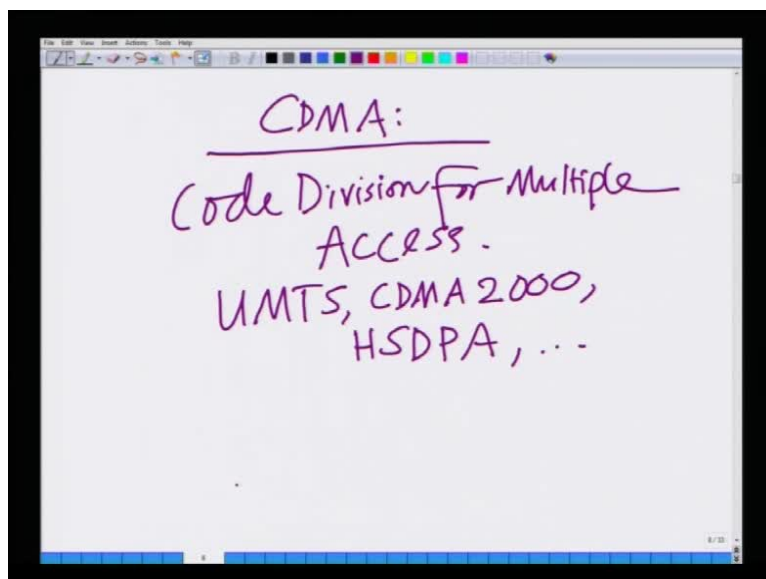
Because of the multi user interference nature of the CDMA system, there is no power control. Then the users who are closed by will drown out the signals of the users, who are farthest. Hence, it is important to transmit to regulate the power. So, that the user who are close to the base station have lower power compare to the users, who are farther from the base station to whom you have transmit have to transmit less power. This is known as power control and power control is a critical part of every CDMA system.

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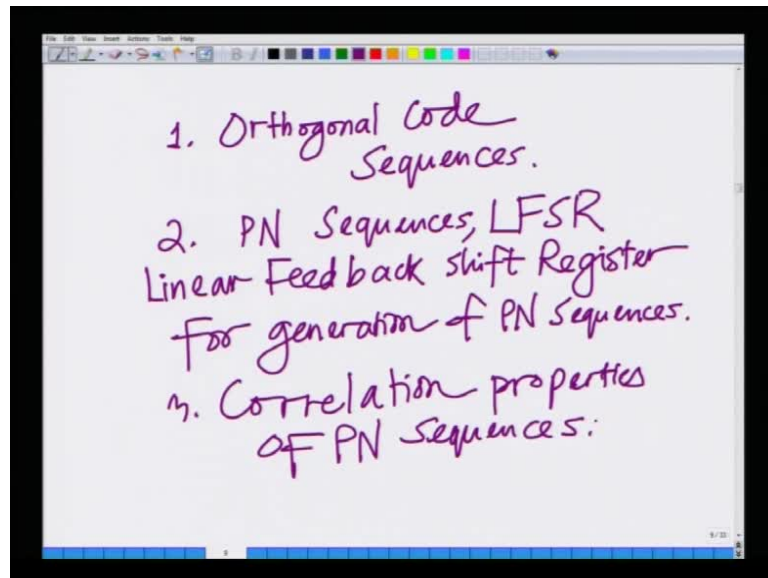
So, power control, let me make that final point. Power control is a critical part of CDMA wireless systems to avoid the near far problem power control is a critical part of every CDMA system, all right? So, with that we will essentially conclude our discussion of CDMA that is code division for multiple access, we have seen several topics in CDMA starting from thing such as code sequences.

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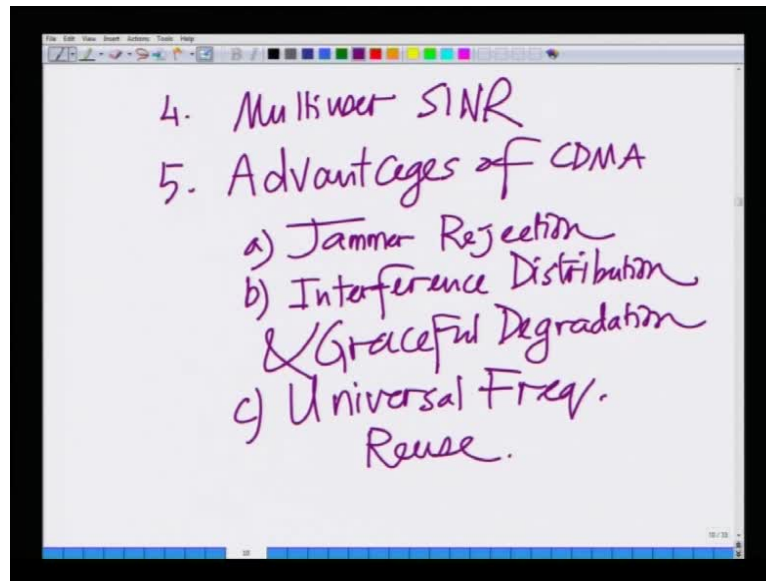
If I were to list the topics that we have seen in CDMA, CDMA we have covered the following topics; CDMA stands for code division for multiple access it forms the basis for all 3 G technology such as UMTS, CDMA 2000 HSDPA and so on.

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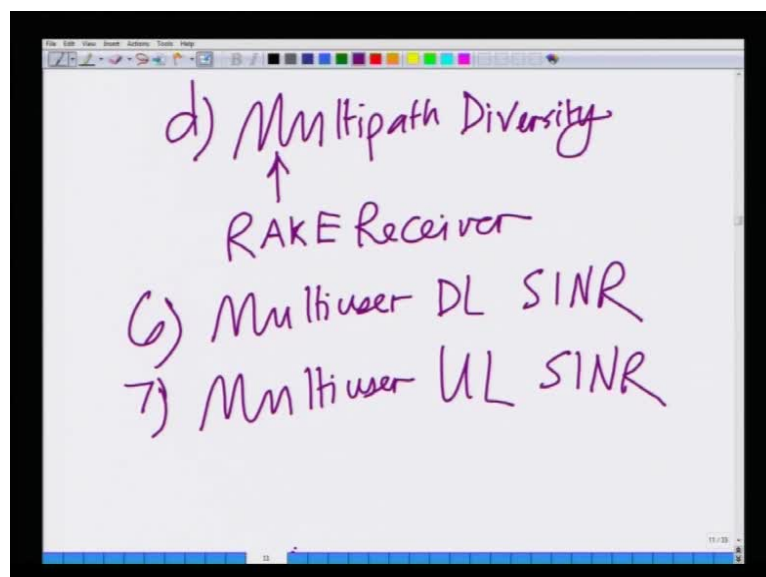
We have seen c CDMA several different properties of CDMA starting from the orthogonal spreading sequences or orthogonal code sequences. Then we looked at P N sequences and linear feedback shift register that is linear feedback shift register. That is linear LFSR linear for generation of P N sequences. Then we have seen the linear feedback shift register operation for generation of the P N sequences. Then we had seen the correlation properties of CPN sequences and that helped us characterize the SINR that is a simple multi user SINR of a CDMA system.

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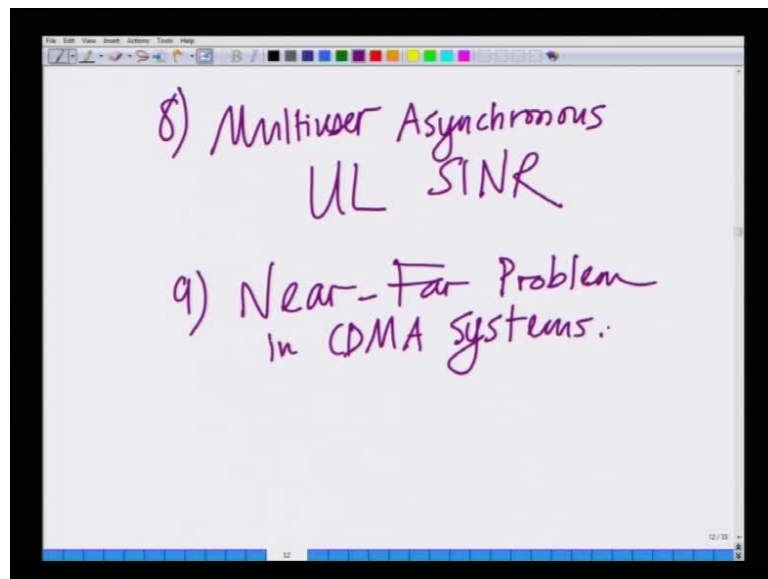
We also looked at the advantages of CDMA, which are a which is jammer, we said that CDMA helps in jammer rejection that is suppressing of malicious user, who are trying to disrupt communication by trying to transmit a signal of very high power and jam you signal. We have seen that CDMA helps in jammer rejection. CDMA also helps in interference distribution and graceful CDMA helps in interference distribution graceful degradation, which leads to universal frequency reuse. We have seen that because of universal frequency reuse every CDMA frequency band can be used in every cell, thereby greatly enhancing the capacity of a CDMA system.

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Hence, CDMA leads to universal frequency reuse and we have seen the fourth advantage is multi path diversity. CDMA can extract multi path diversity through the rake, this is a very important advantage of a CDMA system where previously the paths were combining beyond our control resulting in a fade in a CSMA system, because of the construction you can isolate each path combine path co paths coherently through the rake receiver to extract diversity. This is a very important aspect this is known as multi path diversity. Following this we derived the multi user downlink SINR of CDMA, we also derived the multi user uplink SINR.

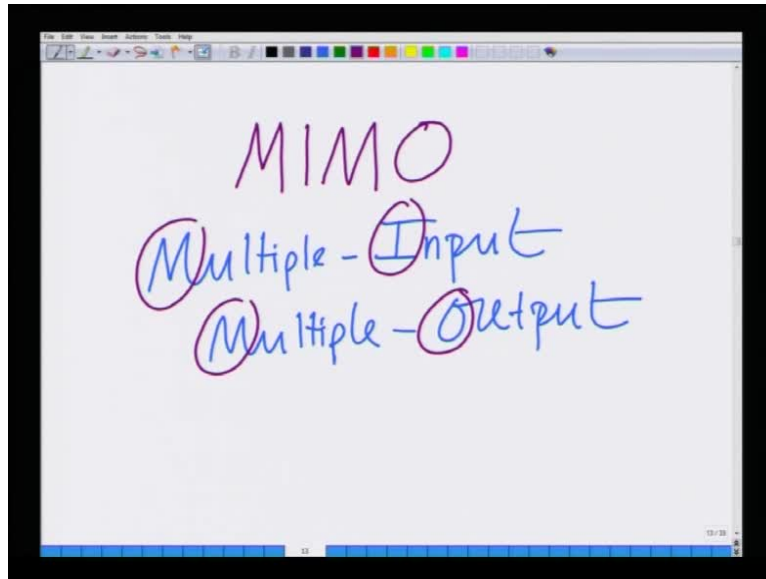
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We also derived the multi user asynchronous uplink SINR and finally, we looked at a problem of CDMA system, which is the Near Far, we looked at the Near Far problem in CDMA. We looked at the Near Far problem in CDMA systems, where a user who is closer to the base station can drown out a user who is farther from the base station if the power is not regulated, this leads to power controls. So, we have looked at several topics and comprehensively code division for multiple access system.

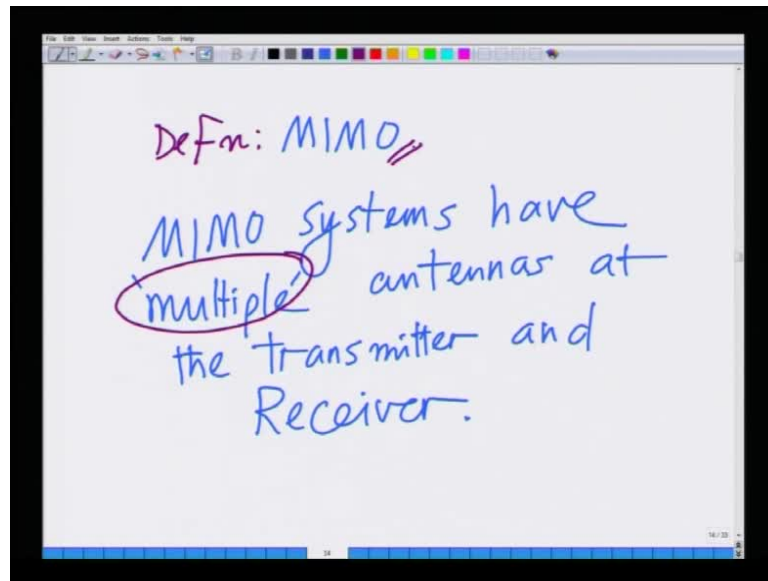
So, urge you to go back look into the notes and look into the lectures to understand this. A little better especially given a fact that CDMA expressions are fairly complicated the expressions for multi user SNR downlink uplink and especially the asynchronous, there are slightly complicated even though intuitive, they are slightly complicated. So, I urge you to again spend time trying to understand this better. So, with this we will move on to the next topic in this course of the next module.

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That is a very important topic, which is or a module which is MIMO, it stands for MIMO, so this is going to be abbreviated as MIMO, which stands for multiple input multiple output. So, MIMO stands for multiple input multiple output. That is MIMO are often called MIMO wireless system. So, we are going to talk discuss MIMO, the properties of MIMO wireless communication systems that is the next topic we are going to talk about.

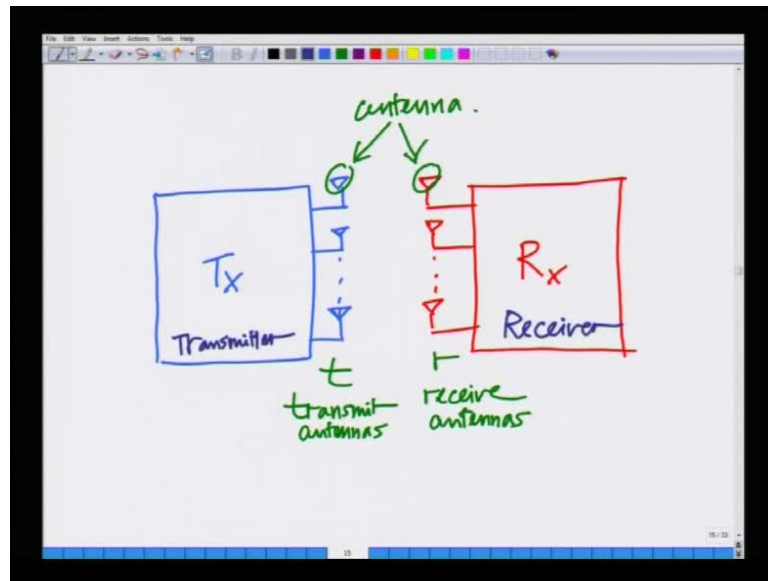
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So, what is MIMO, what is the definition of MIMO? Let me write the definition, MIMO systems are a special class of wireless systems, which have multiple antennas at both the transmitter and the receiver. So, MIMO systems have multiple antennas at the transmitter and receiver. So, MIMO, MIMO systems possess or have multiple that is a key word multiple antennas at the transmitter and receiver. As I said this is the key word multiple, which means greater than 1 that is 2, 3 and so on. Antennas at both the transmitter and receiver that is the definition of a MIMO system that is, we have multiple more than 1 antenna at the transmitter and the, and the receiver.

Now, conventionally as we know a transmitter has a single antenna a receiver also has a single antenna. This is known as a single input single output system. However, now you take the same transmitter and put multiple antennas on it and you take the same receiver and put multiple antennas in it, so the transmitter is still a single transmitter, the receiver is still a single receiver, however only the number of antennas is increasing, all right? This is not to be confused with situation, where you have multiple transmitter, it is the same transmitter on which you are placing multiple antennas same receiver on which you are placing multiple antennas, this is known as a MIMO system.

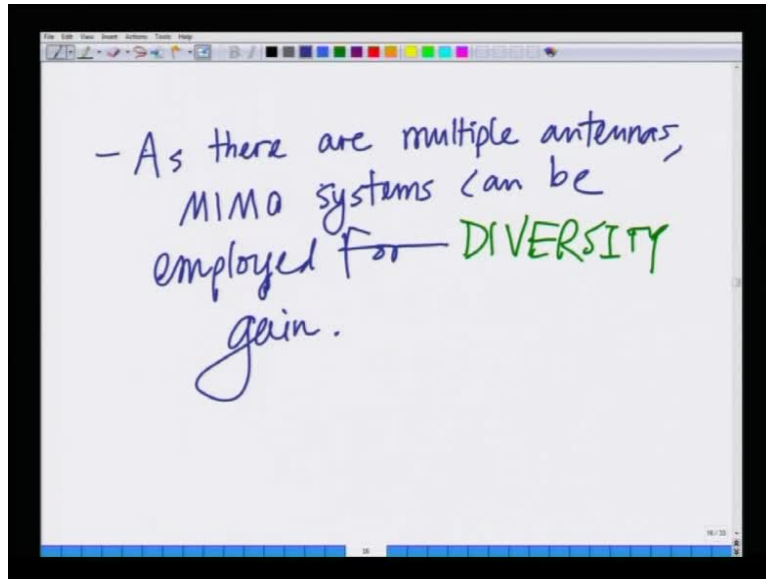
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So, let me describe it schematically, let me draw a simple schematic diagram of that. I have a transmitter block and previously I had 1 antenna. Now, I have multiple antennas. Similarly, I have a receiver block, this I will abbreviate as R_x and I have multiple antennas. Each of these triangular elements, each of this represents an antenna and the transmit block has multiple antennas. This is receiver block has multiple antennas, I will call the number of transmit antennas as t 's.

So, there are t transmit antennas and there are r and there are r receive antennas. There are t transmit antennas and there are r receive antennas. This is just for the sake of clarity, let me write down this as the transmitter and this as the receiver there is a transmitter there is a receiver. The transmitter has t transmit antennas and the receiver has r receive antennas. Now, as we have already seen before in a case of receive diversity that is we consider with system, which has multiple receive antennas having more antennas provides diversity in the system.

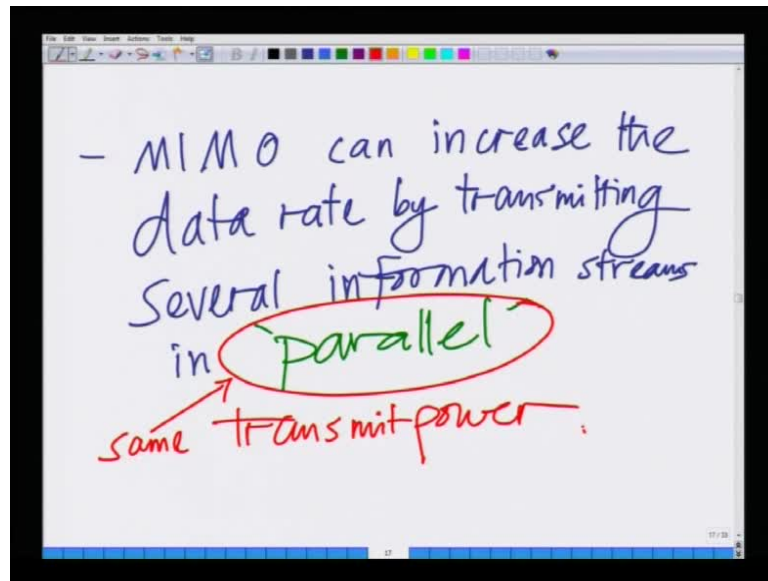
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So, I will write that down, as there are multiple antennas, as there are multiple... We simply said multiple antennas means multiple links. Hence, it means more robustness to fading. Hence, it leads to diversity as there are multiple antennas. This MIMO can be employed for diversity MIMO systems can be for diversity, for diversity gain as there are multiple antennas. We had seen this earlier, since there are multiple antennas there are multiple links. So, we can advantageously employ that to protect against fading that is by extracting diversity gain from this system. That is something we had already seen in the case of we have not seen it in the case of multiple transmit antennas.

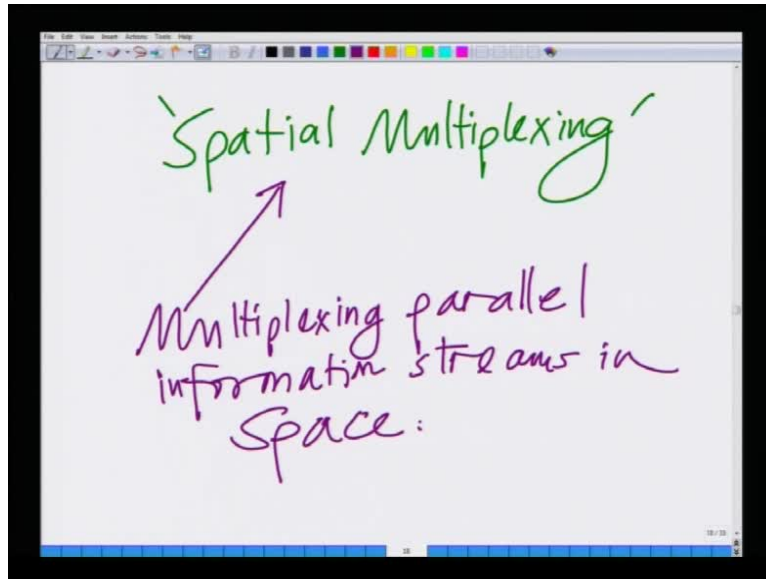
But we saw a case where there is one transmit antenna multiple received antennas. We said that is a received diversity system and can be used to extract diversity. Similarly, now we have multiple transmit antennas also. You can use that to again diversity, however something very different in when you have both multiple transmit and multiple receive antennas is the following. Now, you can not only gain diversity, but you can actually increase the data rate by transmitting several information streams in parallel. Now, you can also increase the data rate another advantage.

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MIMO can increase the data rate by transmitting several information's, so MIMO can increase the data rate by transmitting several information streams in parallel. That is I can transmit on the same frequency at the same time because I have multiple antennas. I can transmit several information streams between the transmitter and receiver in parallel and in fact at the same transmit power. So, I have quantified this. This is also at same transmit power it is as if I am taking this space and multiplexing different information streams across it that is between the in the space between the transmitter and receiver. I am multiplexing parallel information streams. Hence this is known in the context of MIMO as spatial multiplexing.

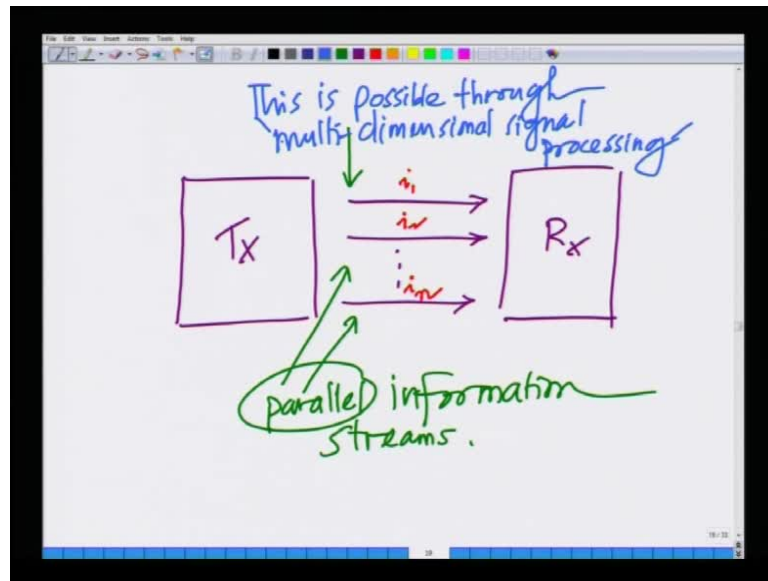
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So, this is known as spatial multiplexing, which literally means multiplexing parallel information streams, multiplexing parallel information streams in space, which is essentially multiplexing these information streams in space. So, this is known as spatial multiplex and that results in a tremendous increase in the data rate, which is a unique aspect of MIMO systems.

Hence, it is a key aspect of any 3 G and also 4 G wireless systems because as we said in the very beginning 3 G 4 G wireless systems are broadband systems characterized as having high data rate. Hence MIMO is a key technology, which in fact helps push up the data rates on this 3 G and 4 G wireless systems. Hence, it is a key component of 3 G 4 G systems. We are going spend a sufficient amount of time trying to understand the different properties and the different aspects of MIMO systems.

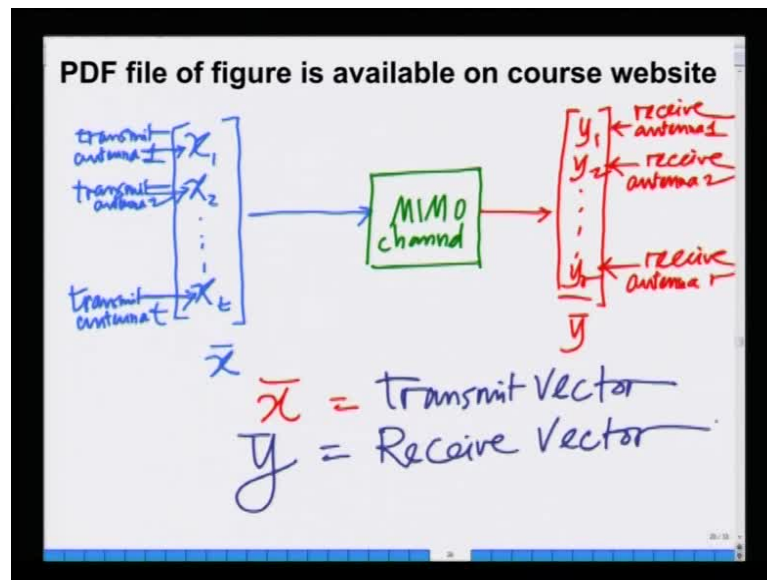
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So, to just schematically describe spatial multiplexing, I have a transmitter, I have the receiver and I am transmitting multiple information streams, i_1 , i_2 , up to i_n . That is these are the multiple parallel information. These are the parallel information, so these are the parallel information streams that we are transmit. This is a schematic, we are going to see the exact technique as we proceed through the contents of this topic we are going to see how it is possible to in fact transmit multiple independent information streams from the transmitter to receiver in a mimo system?

That is a very unique aspect and a very big advantage of MIMO or a multiple input multiple output wireless communication system. Let me just mention this is possible to complex multi dimensional signal processing, this is, this is possible through multi dimensional signal this is possible through multi dimensional signal processing. So, let us start with a basic description of MIMO.

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Let us start with getting a basic description of MIMO. In a MIMO system, I have t transmit antennas. So, I am transmitting t symbols from the t transmit antennas. Let me denote the symbol transmitted from the first transmit antenna as x_1 , the second transmit antenna as x_2 , so on and so forth up till the t th transmit antenna as x_t . This is a vector \bar{x} where x_1 denotes symbol from transmit antenna 1, the symbol transmitted from transmit antenna one x_2 is from transmit and x_t is from transmit, transmit antenna t . So, there are three sym, t symbols being transmitted from the t transmit antennas in parallel.

These pass through the MIMO channel, this is the MIMO channel. So, I am transmitting t transmit and a t symbols x_1, x_2, x_t from the t transmit antenna, so this is a vector \bar{x} of t dimension. There is a t dimension vector \bar{x} passing through the MIMO channel at the receiver. I have r receive antennas and I receive y_1, y_2 up to y_r at each of the receive antenna. So, at the receiver I will denote the received symbols as y_1, y_2 so on up to y_r where this is the vector \bar{y} and y_1 is received at receive antenna, y_2 is received at receive antenna 2 y_r is received at receive receive antenna r , all right?

So, I am transmitting a t symbol vector across the t transmit antennas goes through the MIMO channel at each of the r antennas, I am receiving y_1, y_2, y_r that is an r dimensional vector. So, \bar{x} is also known as the transmit vector and \bar{y} is also known as the receive vector. So, in this system \bar{x} is also known as the transmit symbol vector and \bar{y} is also known as the receive symbol vector. Now, I can represent this as follows there is n r dimensional

vector that is received at t dimensional vector that is transmitted. So, it must somehow go from a t dimensional transmit vector to an r dimensional receive vector, which means it is passing from a t dimensional space through to an r dimensional space. Hence, we know that this sort of a transformation can be represented by a matrix and that is what I am going to represent it as.

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$$\begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_r \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} & \dots & h_{1t} \\ h_{21} & h_{22} & \dots & h_{2t} \\ \vdots & \vdots & \ddots & \vdots \\ h_{r1} & h_{r2} & \dots & h_{rt} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_t \end{bmatrix}$$

r $r \times t$ t
 Flat fading channel coefficients. Channel matrix

So, I have the r dimensional receive vector y_1, y_2 up to y_r , which is some matrix times x_1, x_2 up to x_t . This is t dimensional this is r dimensional, hence we know that the matrix has to be r cross t dimensional and that can be represented as follows h_{11}, h_{12} so on up to h_{1t} h_{21}, h_{22} so on up to h_{2t} so on up to h_{r1}, h_{r2} so on up to h_{rt} . This is the r cross t dimensional matrix. This is also known as the channel matrix, each entry of this channel matrix is a flat fading coefficient, all right? h_{11}, h_{12}, h_{21} all these are flat fading channel coefficients. So, each of these are flat fading each of these is a flat fading channel coefficient.

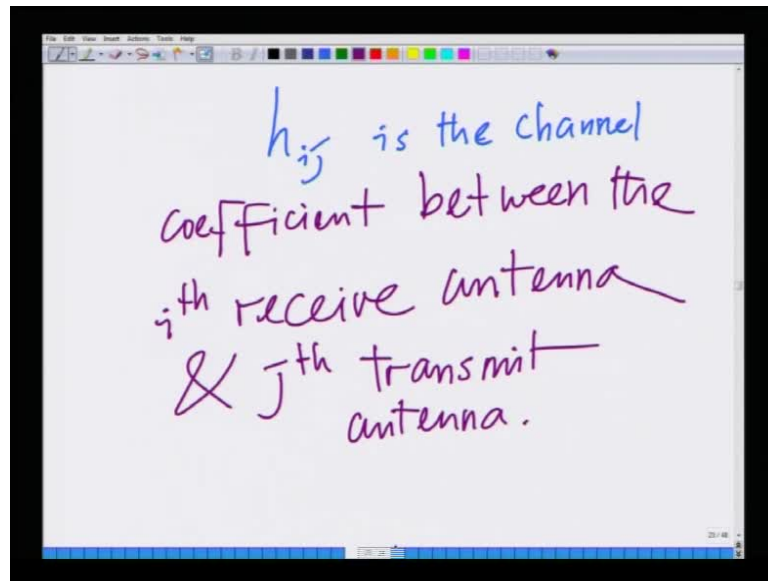
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The image shows a whiteboard with two equations and handwritten notes. The first equation is $y_1 = h_{11}x_1 + h_{12}x_2 + \dots + h_{1t}x_t$. Below it, a blue arrow points from x_1 to h_{11} , and a green arrow points from x_2 to h_{12} . Below the equation, it says "x₁, x₂, ... x_t interfere at y₁". The second equation is $y_2 = h_{21}x_1 + h_{22}x_2 + \dots + h_{2t}x_t$. Below it, it says "x₁, x₂, ... x_t interfere at receive antenna 2".

Now, let me just explicitly write out what y_1 is, y_1 equals $h_{11}x_1$ plus $h_{12}x_2$ plus so on $h_{1t}x_t$. Hence, you can see each of the transmit symbols x_1, x_2 so on up to x_t interferes at y_1 . So, all the t symbols, so x_1, x_2 so on up to x_t interfere at y_1 . Similarly, the signal y_2 at receive antenna 2 can be expressed as $h_{21}x_1$ plus $h_{22}x_2$ plus $h_{2t}x_t$ and you can see again x_1, x_2 up to x_t interfere at y_2 or it receive antenna 2. Hence, x_1, x_2 up to x_t interfere at receive antenna 2. So, x_1, x_2, x_t interfere at receive antenna 1, they interfere at receive antenna 2 and they interfere.

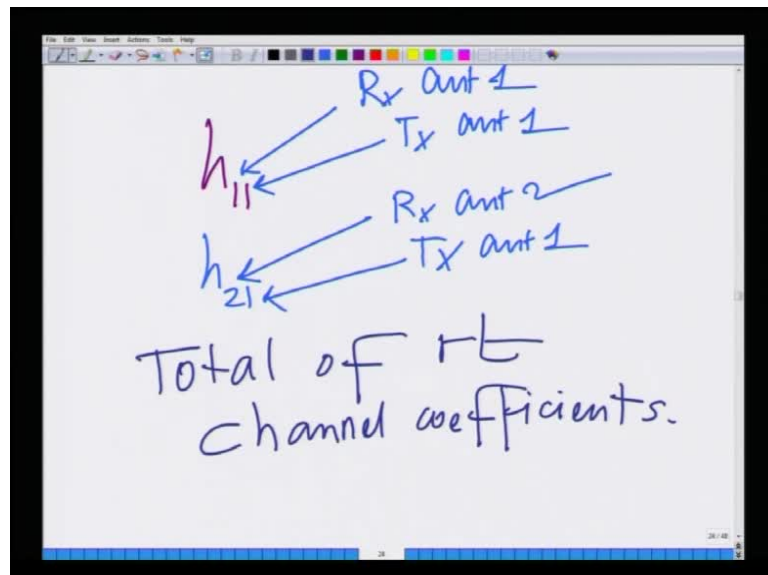
In fact at each of the r receive antenna, so the net signal that you get at each of the r receive antenna has contains a component from each of the transmit antenna. So, each of the transmit antenna goes through the channel the signal from each of the transmit antenna is received at each of the received antenna. So, its interfere it is it is a mingled signal from each of the transmit antenna, That is received at across the r receive antennas that is the system model in fact. Now, we can say that h_{ij} , if you look at this expression h_{11} is the channel between x_1 y_1 , which is transmit antenna 1 and receive antenna 1.

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h_{12} is the channel between x_2 and y_1 that is transmit antenna 2 receive antenna 1. h_{21} is the channel between x_1 and y_2 that is transmit 1 and receive antenna 2. Hence, we can say that h_{ij} is the channel between i^{th} receive antenna and j^{th} transmit antenna. h_{ij} is the channel or the channel coefficient, is the channel coefficient between the i^{th} receive antenna and j^{th} transmit

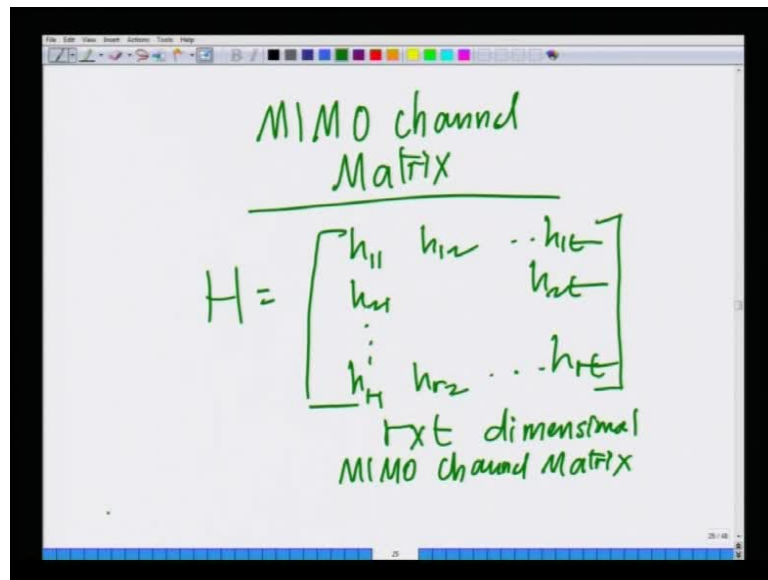
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In fact if you look at h_{11} that is transmit antenna 1, that is transmit antenna 1 and that is receive antenna 1. h_{21} is I am sorry this has to be the other way round, this is R_x antenna 1 and T_x antenna 1. h_{21} is between R_x antenna 2 and T_x antenna 1. Hence, between each

of the R receive antennas and T of the transmit antennas. There is a flat fading channel coefficient. Hence, there is a total of R times T channel coefficients. Hence, total of $R \times T$ channel, hence there is a total of $R \times T$ channel coefficients and the MIMO channel matrix as we already said is presented as the MIMO channel matrix.

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The image shows a whiteboard with the title "MIMO channel Matrix" written in green. Below the title, the MIMO channel matrix H is defined as an $R \times T$ dimensional matrix. The matrix is shown as a square with rows and columns. The first row contains elements $h_{11}, h_{12}, \dots, h_{1T}$. The first column contains elements $h_{11}, h_{21}, \dots, h_{R1}$. The last row contains elements $h_{R1}, h_{R2}, \dots, h_{RT}$. The last column contains elements $h_{1T}, h_{2T}, \dots, h_{RT}$. Below the matrix, it is noted that it is an "R x T dimensional MIMO channel Matrix".

$$H = \begin{bmatrix} h_{11} & h_{12} & \dots & h_{1T} \\ h_{21} & & & h_{2T} \\ \vdots & & & \vdots \\ h_{R1} & h_{R2} & \dots & h_{RT} \end{bmatrix}$$

R x T dimensional MIMO channel Matrix

The MIMO channel matrix is nothing but h equals $h_{11}, h_{12}, h_{21}, h_{R1}, h_{1T}, h_{2T}, h_{R2}$ so on up to h_{RT} . This is the R cross T dimensional MIMO, this is the R cross T dimensional MIMO channel matrix. Hence, the MIMO system model can be expressed as follows. So, we said that the channel matrix is R cross T dimensional, it contains $R \times T$ elements, which are the coefficients between each of that transmit each transmit and receive and each receive and transmit antenna pair. Since, there are R receive antennas T transmit antennas. There are total of R times T pairs and this channel matrix, hence $R \times T$ dimensional. Now, let us go to the MIMO system model.

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The image shows a whiteboard with the title "MIMO system Model:" written in green. Below the title, the vector equation is written in green ink:

$$\begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_r \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} & \dots & h_{1t} \\ h_{21} & & & \\ \vdots & & & \\ h_{r1} & & & h_{rt} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_t \end{bmatrix} + \begin{bmatrix} n_1 \\ n_2 \\ \vdots \\ n_r \end{bmatrix}$$

Below this, the compact matrix form is written:

$$\bar{y} = H \bar{x} + \bar{n}$$

The MIMO system model can be described as follows y_1, y_2 up to y_r equals h_{11}, h_{12}, h_{21} so on that is the channel matrix h_{r1} so on h_{1t} so on up to h_{rt} times $x_1 x_2$ up to x_t . That is the received vector is matrix h times transmit vector plus. Of course, there is going to be noise, which is n_1, n_2, n_r at each of their receive antennas there is going to be noise, so n_1 is the noise at receive antenna 1, n_2 is the noise at receive antenna 2, so on n_r is the noise at receive antenna r . So, I can represent this as $\bar{y} = H \bar{x} + \bar{n}$.

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The image shows the same whiteboard with the equation $\bar{y} = H \bar{x} + \bar{n}$ from the previous slide. This time, the variables are annotated with their dimensions in green ink:

- An arrow points from the text " r dimensional Receive vector" to \bar{y} .
- An arrow points from the text " t dimensional transmit vector" to \bar{x} .
- An arrow points from the text " $r \times t$ channel matrix" to H .
- An arrow points from the text " r dimensional noise" to \bar{n} .

So, let me write this down again succinctly over here \bar{y} equals $H \bar{x}$ plus \bar{n} , let me again label each of this components. \bar{y} is the r dimensional, this is the r dimensional receive, vector H is the r cross t dimensional channel matrix, \bar{x} is the t dimensional transmit vector and \bar{n} is the r dimensional. Hence, I have \bar{y} that is r dimensional receive vector equals H times \bar{x} plus \bar{n} \bar{y} is the r dimensional receive vector. H is the r cross t dimensional matrix \bar{x} is the t dimensional transmit vector plus \bar{n} , which is the r dimensional noise vector and a word about the noise now the noise is a vector and this vector noise can be described as follows. Let us look into before we proceed into noise. Let us look at some special cases.

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Special Cases:

$t = 1$

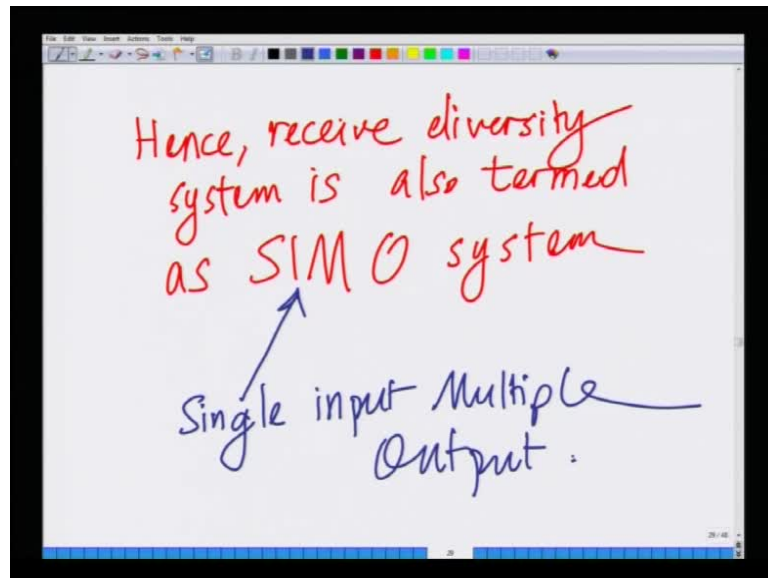
$$\begin{bmatrix} y_1 \\ y_2 \\ \vdots \\ y_r \end{bmatrix} = \begin{bmatrix} h_1 \\ h_2 \\ \vdots \\ h_r \end{bmatrix} x + \begin{bmatrix} n_1 \\ n_2 \\ \vdots \\ n_r \end{bmatrix}$$

$\bar{y} = \bar{h}x + \bar{n}$

Receive diversity system.

Now, if r equals or if t equals 1, we have a special case that is number transmit antennas equals 1. We have a special case that we have seen before we have y_1, y_2 up to y_r equals h_1, h_2 . Since, there is only one column of coefficients, I am going represent it by h_1, h_2 up to h_r . That is h_1, h_2 up to h_r, x plus n_1, n_2 up to n_r , the noise at each receive antenna. This we have seen before this is \bar{y} equals $\bar{h} \bar{x}$ plus \bar{n} . This is also the essentially the receive diversity system. We have seen this before in our discussion on diversity, this is a receive diversity this is a receive diversity system. In the context of MIMO this is known as look at this, I am transmitting a single symbol. I am receiving r symbols across the r receive antennas.

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Hence, this is single input multiple output hence this is also known as the SIMO, SISO or a single input multiple output system. Hence in diversity system is also termed as SIMO system where SIMO stands for single input multiple output SIMO also stands for single input multiple output this is termed as a SIMO system.

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$$r = 1$$
$$y = [h_1 \ h_2 \ \dots \ h_t] \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_t \end{bmatrix} + n$$
$$= \underline{\underline{h}}^T \underline{\underline{x}} + n$$

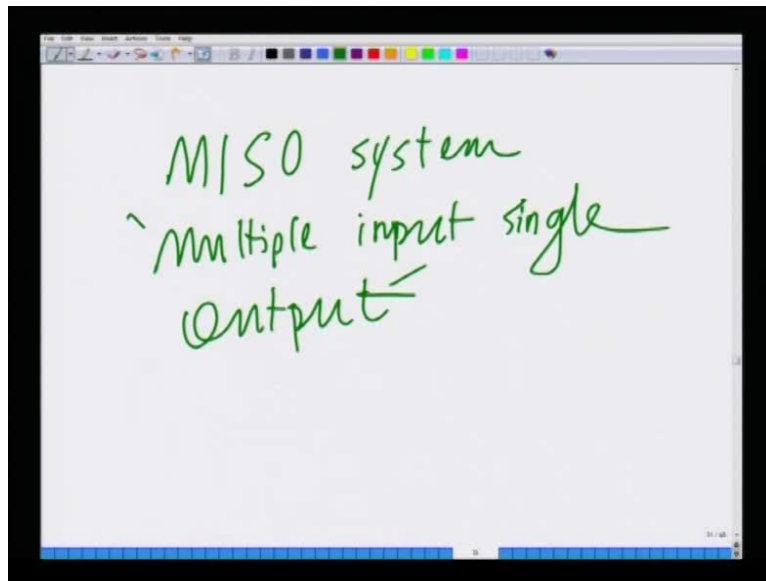
Transmit Diversity system

The image shows handwritten equations on a whiteboard. At the top, it says 'r = 1'. Below that is the equation $y = [h_1 \ h_2 \ \dots \ h_t] \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_t \end{bmatrix} + n$. This is followed by $= \underline{\underline{h}}^T \underline{\underline{x}} + n$, where $\underline{\underline{h}}$ and $\underline{\underline{x}}$ are underlined twice. Below the equation, it says 'Transmit Diversity system' in green ink.

Let us consider the other special case, which is t equals 1, which is r equals , and certain number of transmit antennas. In that case we have y equals h 1, so we are considering the case where number of receive antennas r equals 1, we have not seen this case before and y

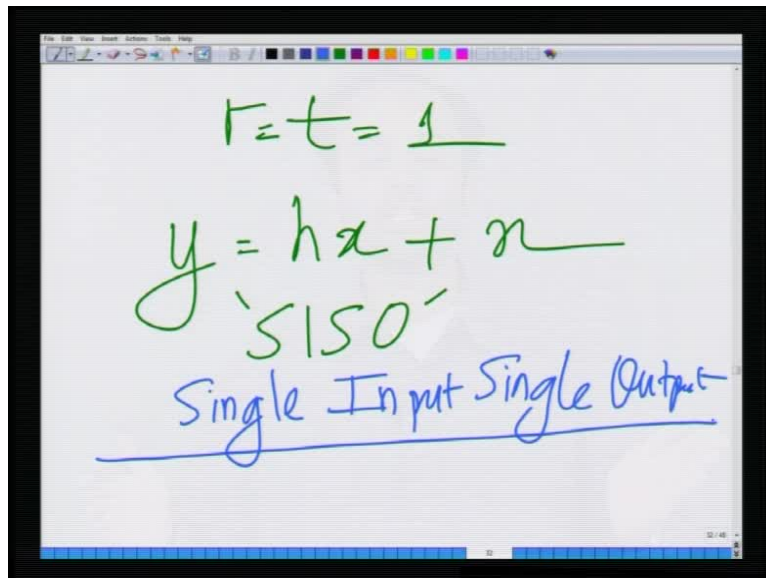
equals... Now, since I have many transmit antennas one receive antenna, I have row and this can be represented as $y = [h_1 \ h_2 \ \dots \ h_t] \times [x_1 \ x_2 \ \dots \ x_t]^T + n$. This is the same as $\mathbf{h}^T \mathbf{x} + n$. Now, the, I am saying there are t transmit antennas, but only one single antenna. Since, channel matrix is now a row vector, this is known as a transmit diversity scenario or a MISO system. That is multiple input single output system, this is also known as transmit diversity system.

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Also in the context of transmit diversity is known as MISO that is MISO system, which is multiple input single output. This is known as a MISO system, which is essentially have multiple input single output system. So, in the context of MIMO, we have two special cases when t equals 1, we know, know it as a, we term it as a SIMO system. That is single input multiple output when r equals one we term it as a MISO system, which is multiple input single output.

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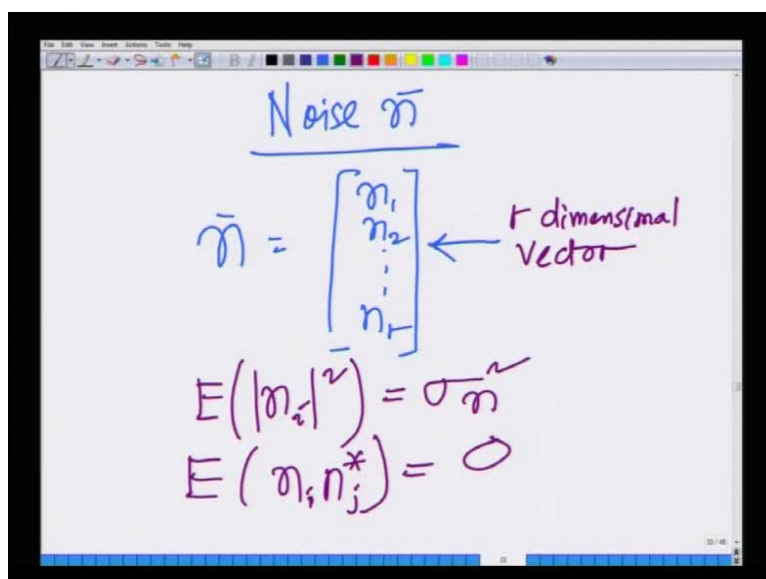

$$r=t=1$$
$$y = hx + n$$

'SISO'

Single Input Single Output

Now, obviously the next question is a special case what happens when r equals t equals 1? When r equals t equals 1, we have the traditional scalar fading coefficient channel, which is y equals $h x$ plus n . This is naturally termed as a SISO channel, which stands for single input this naturally stands for single input, single output. That is the traditional one transmit antenna of one receive antenna system is also termed as a SISO system, which is single input single output system.

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Noise \bar{n}

$$\bar{n} = \begin{bmatrix} n_1 \\ n_2 \\ \vdots \\ n_r \end{bmatrix} \leftarrow \begin{array}{l} r \text{ dimensional} \\ \text{Vector} \end{array}$$
$$E(|n_i|^2) = \sigma_n^2$$
$$E(n_i n_j^*) = 0$$

Now, let us come to a slight discussion on the noise. We have said the noise is vector \bar{n} , which is given as n_1, n_2 so on up to n_r . This is r dimensional vector, hence this is n_r dimensional. This is n_r dimensional vector, so this noise is an r dimensional vector. That is it as r noise components corresponding to the noise component at each receive antenna. We will assume the following, we will assume the per power of each noise component n_i square equals σ_n^2 . That is the power in each noise component at each receive antenna power in the noise. Also in the context of transmit diversity is known as MISO. That is MISO system, which is multiple input single output this is known as a MISO system, which is essentially have multiple input single output system.

So, in the context of MIMO, we have two special cases, when t equals 1, we know, know it as we term, it as a SIMO system, that is single input multiple output when r equals 1. We term it as a MISO system, which is multiple input single output. Component to each receive antenna is σ_n^2 further, we will assume that the noise processes at different receive antennas are uncorrelated. Since, it is Gaussian independent, which is also a natural things to assume because there are different receive antennas, so that noise processes are uncorrelated. Hence, we are going to also assume that expected $n_i n_j^*$ equals 0. That is a correlation between the noise and the i th antenna and the j th, i th and j th receive antenna is 0.

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The image shows a handwritten derivation of the covariance matrix of the noise vector \bar{n} . The equation is written as:

$$E(\bar{n} \bar{n}^H) = E \left(\begin{bmatrix} n_1 \\ n_2 \\ \vdots \\ n_r \end{bmatrix} [n_1^* \ n_2^* \ \dots \ n_r^*] \right)$$

An arrow points from the text "Covariance matrix" to the expression $E(\bar{n} \bar{n}^H)$. The equation is then expanded to show the elements of the matrix:

$$= E \left(\begin{bmatrix} n_1 n_1^* & n_1 n_2^* & \dots & n_1 n_r^* \\ n_2 n_1^* & n_2 n_2^* & \dots & n_2 n_r^* \\ \vdots & \vdots & \ddots & \vdots \\ n_r n_1^* & n_r n_2^* & \dots & n_r n_r^* \end{bmatrix} \right)$$

The diagonal elements $n_i n_i^*$ are circled in red, and the off-diagonal elements $n_i n_j^*$ are circled in blue.

Hence, we can write now the multi dimensional correlation or the co variance matrix of this noise as expected $\bar{n} \bar{n}^H$ her Hermecian. This is also the multi dimensional analogue of

the correlation between the noise. This is also known as a covariance matrix this is known as the covariance this covariance matrix is expected n_1, n_2 so on up to n_r times n Hermejian, which is you take transpose and conjugate. That is n_1 conjugate n_2 conjugate n_r conjugate, which is nothing but expected n_1, n_1 conjugate, which is norm in 1 square $n_2 n_1$ conjugate $n_1 n_2$ conjugate so on up to $n_1 n_r$ conjugate.

This is magnitude into square so on up to $n_r n_1$ conjugate so on up to magnitude n_r square. Now, what you can observe, when you take the expectation operator inside is expectation of low operator is linear. Hence, it goes inside that each of this components on the diagonal is σ_n^2 and each of this cross terms which are half diagonal are essentially 0.

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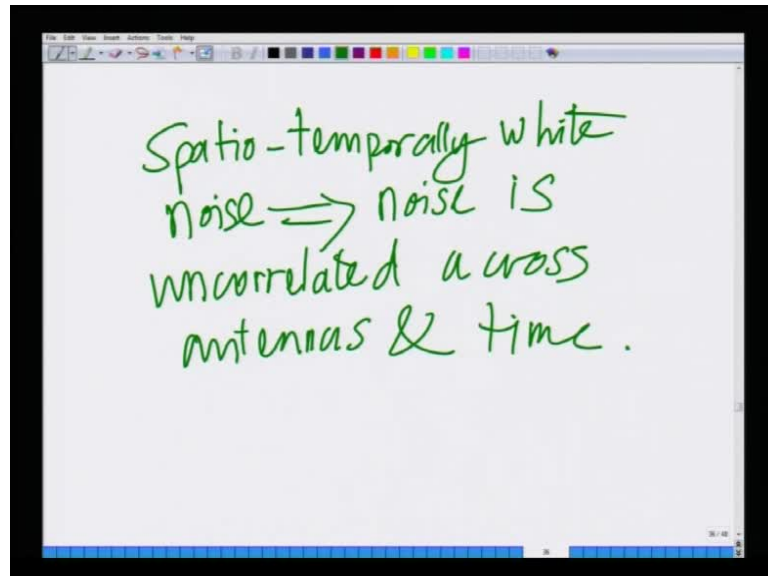
The image shows a handwritten derivation on a whiteboard. At the top, the equation $E(\bar{n}\bar{n}^H) =$ is followed by a matrix with diagonal elements σ_n^2 and off-diagonal elements 0. Below this, a green arrow points from the text "Spatially white noise" to the equation. To the right of the text, the equation $= \sigma_n^2 I$ is written, where I is the identity matrix.

Hence, this matrix which is the covariance matrix of the noise vector at the receiver is nothing but expected n bar n bar Hermejian equals only the diagonal terms survive. This is hence equals σ_n^2 identity matrix. Hence, the covariance matrix of this noise is σ_n^2 times identity, this is also termed as being spatially white. That is across space the antenna, the noise at each antenna each of pairs of receive antennas uncorrelated. That is it is white across space.

Similarly, you know that if the antennas if distinct samples of the noise are uncorrelated, then you know it is temporally uncorrelated. Now, we have seen the definition of spatially uncorrelated or spatially white noise. Now, If the noise is both spatially and temporally, that is

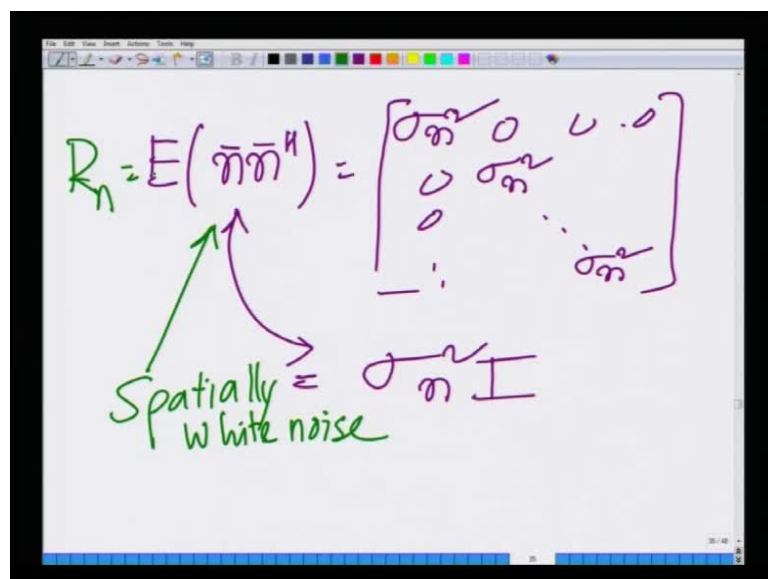
if each the noise samples are uncorrelated across antennas and if they are also uncorrelated across times that is known as spatio temporaly white noise.

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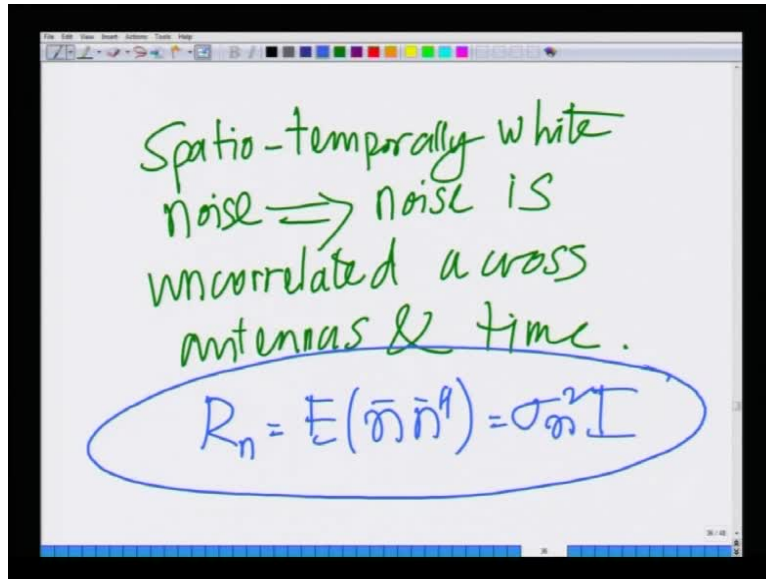
So, spatio white noise implies noise is uncorrelated across antennas and times that is spatio temporally uncorrelated noise, which has a covariance of σ_n^2 . We are going to divide denote this as r_n , which is the covariance.

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Hence r_n equals expected $\bar{n} \bar{n}^H$ Hermician equals $\sigma_n^2 I$. This is the covariane matrix of the noise at the receiver. So, this is the covariance matrix.

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So, that completes our discussion of the system model of the MIMO system model, which you have seen is consists of r dimensional receive vector, which is the symbols received at r antennas r cross t channel matrix consisting of r t flat fading coefficients. t dimensional transmit vectors consisting of symbols transmitted at each transmit antenna plus the r dimensional noise vector, which consists of the noise samples at each receive antenna. The covariance of that noise is given as r n is expected n bar n bar Hermitian, which is σ_n^2 identity. So, with this we have set down the system model of a MIMO system. With the next lecture we, in the next lecture we will study several other properties and start analyzing MIMO systems. So, because of so at this point I will conclude this lecture and I will see you again in the next lecture.

Thank you very much.