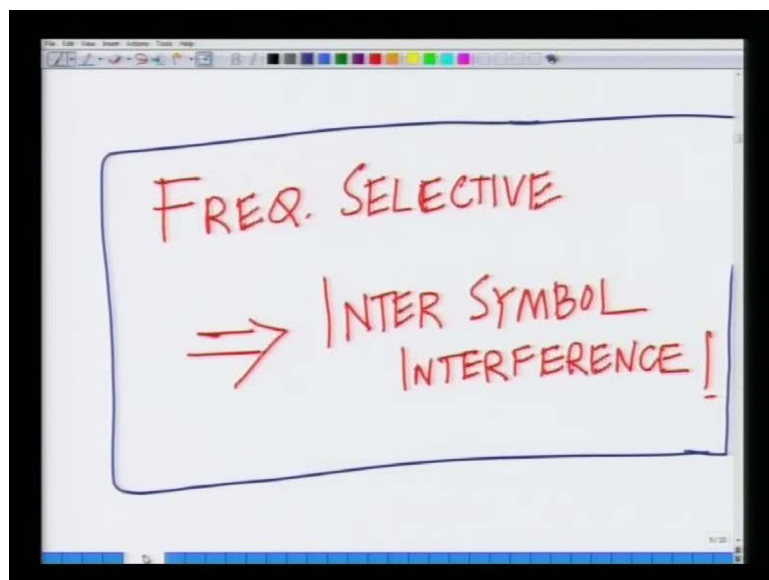


Advanced 3G and 4G Wireless Communication
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Indian Institute of Technology, Kanpur

Lecture - 12
Doppler Spectrum and Jakes Model

Welcome to another lecture in the course on 3G, 4G Wireless Communication systems. In the last lecture, we were discussing or completing our discussion on delay spread and frequency selective channels.

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Specifically, we said if the channel is frequency selective that is, if the signal bandwidth is greater than the coherence bandwidth, then in time domain it means that the delay spread is greater than the symbol time, hence the channel results in inter symbol interference, which causes distortion at the receiver. So, frequency selectivity and the frequency domain corresponds to inter symbol interference in the time domain.

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Handwritten equations on a whiteboard:

$$f_r = f_c + f_d$$
$$= f_c + \left(\frac{V \cos \theta}{c} \right) f_c$$

Re frequency due to Doppler

$0 \leq \theta \leq \pi/2$ MS \rightarrow BS

$\pi/2 \leq \theta \leq \pi$ MS \leftarrow BS

$\theta = \pi/2$

We also said, we also introduced the notion of Doppler resulting from a moving, moving mobile that is, has the mobile moves with the velocity V , the received carrier frequency changes. In particular that is given by this relation F_c plus $V \cos \theta$ over c times F_c , where $V \cos \theta$ over $c F_c$ is the Doppler shift, V is the velocity of mobile and θ is the angle of the velocity with line joining the base station and the mobile.

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Diagram showing a mobile station (MS) moving towards a base station (BS) with velocity V at an angle θ . The component of velocity towards the BS is $V \cos \theta$.

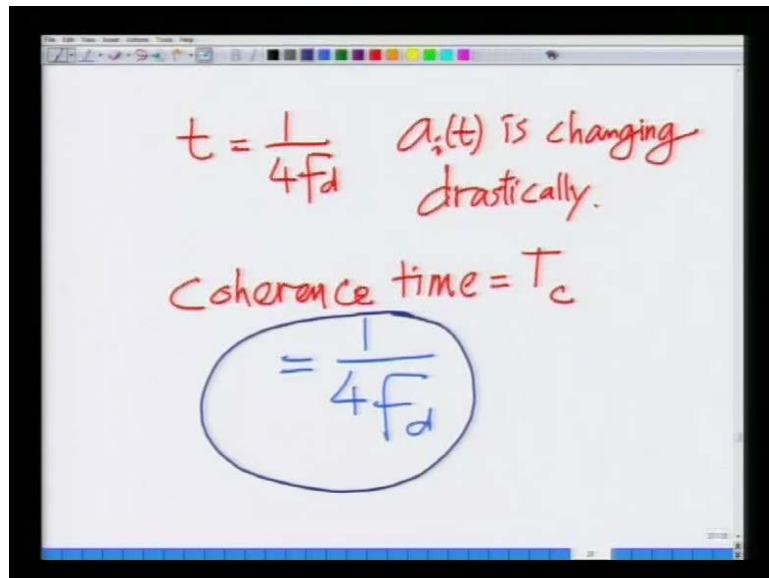
$$\tau_s(t) = \tau_s - \frac{V \cos \theta t}{c}$$

Initial delay

We said this Doppler or this velocity of the mobile results in a time varying delay, previously we had our delay fixed, which τ_i . But now because the mobile is moving that delay itself is

changing, which means now, the channel itself is changing, because the delay is changing. Hence, now Doppler results in a time varying or also a time selective channel, this is what we said in the last lecture.

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The image shows a digital whiteboard with handwritten text in red and blue ink. The text is as follows:

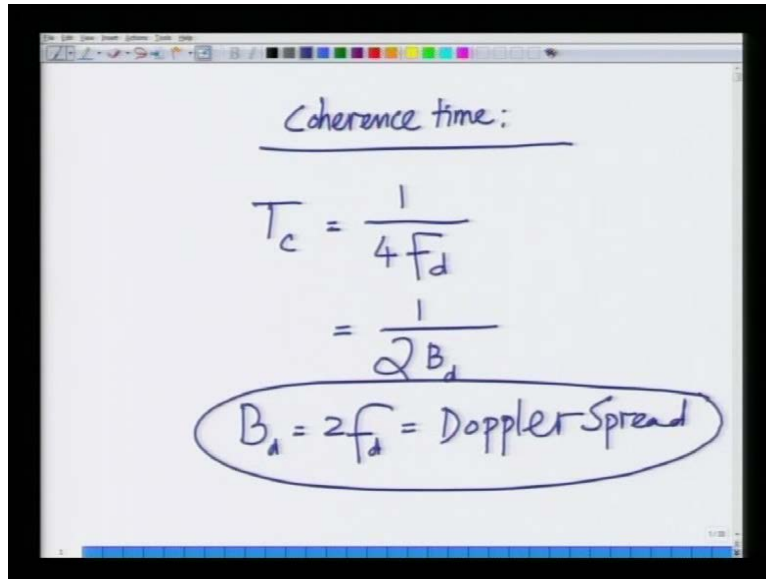
$$t = \frac{1}{4f_d} \quad a_i(t) \text{ is changing drastically.}$$
$$\text{Coherence time} = T_c$$
$$= \frac{1}{4f_d}$$

The expression $= \frac{1}{4f_d}$ is circled in blue ink.

And finally, we said the time of significant change of the channel is t equals 1 by $4 F d$, where $F d$ is the Doppler frequency. The t for change is 1 over $4 F d$, which is T_c the coherence time, which is that interval of time over which the channel is approximately constant. That is the channel is changing, it is changing from time to time but we can assume that it is approximately constant over one coherence time interval, which corresponds to 1 over $4 F d$.

Also observe that $F d$ is proportional to the velocity, which means if the velocity is high, $F d$ is high which means the coherence time T_c which is 1 over $4 F d$ is low, so as the velocity is increasing the coherence time is decreasing which is intuitive. Because, if the vehicle or the mobile terminal is moving faster and faster, then the channel is also changing faster and faster, which means the coherence time is lower and lower.

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Handwritten notes on a digital whiteboard:

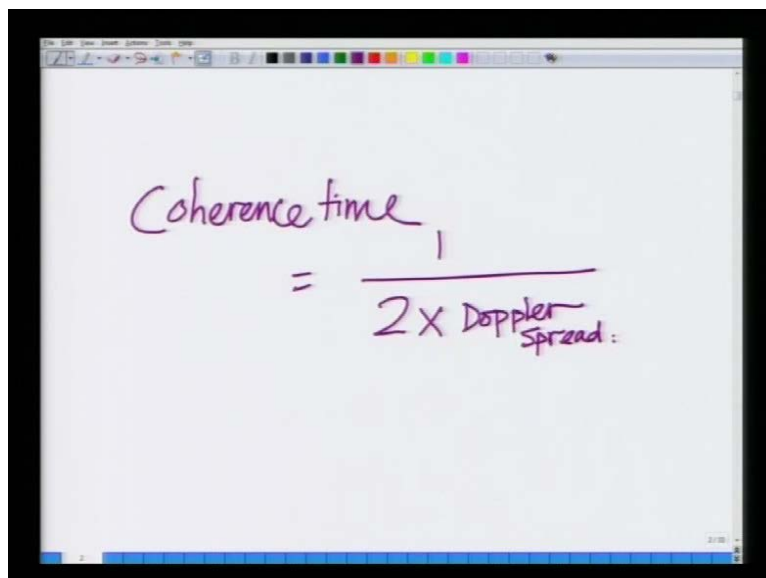
Coherence time:

$$T_c = \frac{1}{4f_d}$$
$$= \frac{1}{2B_d}$$

$B_d = 2f_d = \text{Doppler Spread}$

So, with that notion, let us start today's discussion let us start today's discussion that coherence time, let us start today's discussion let us refine this notion of coherence time. Coherence time T_c equals 1 over $4f_d$, I can also write this as 1 over $2B_d$, where B_d equals $2f_d$ equals the Doppler spread, I will call, I will define a new quantity to f_d define as the Doppler spread. That is twice f_d is Doppler spread, hence the coherence time T_c is 1 over $2B_d$.

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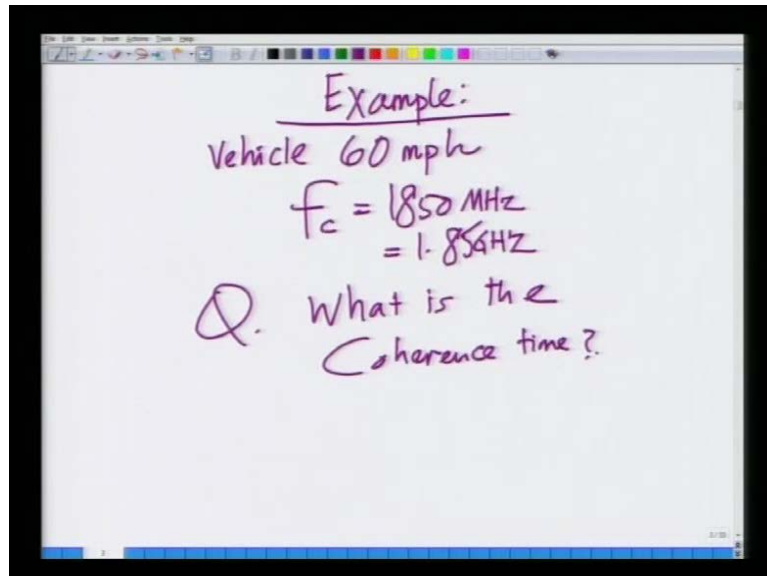
Handwritten notes on a digital whiteboard:

Coherence time

$$= \frac{1}{2 \times \text{Doppler Spread}}$$

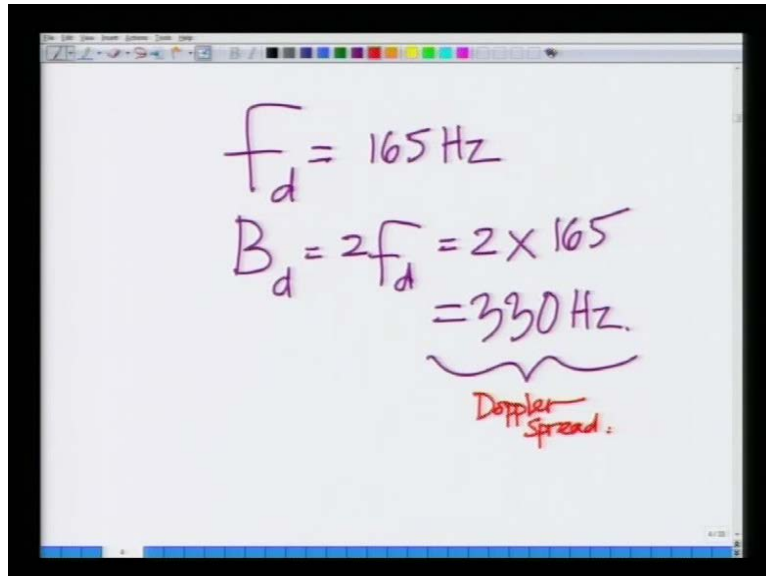
Let us write that clearly coherence time equals 1 over twice the Doppler spread of the channel, which is d ; let us do an example to understand this better.

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Let us do an example of the 3G 4G wireless communication system, let us go back to our earlier example, where there was a vehicle or there was a person in a vehicle which was moving at 60 miles per hour that is vehicle at 60 miles per hour moving directly towards the base station. At carrier frequency F_c equals 1850 mega hertz or in other words, 1850 mega hertz is the same as 1.8 Giga hertz or 1.85 Giga hertz. Now, we want to find out what is the coherence time, so want to find out what is the question, question is what is the coherence time of this channel, let us compute that.

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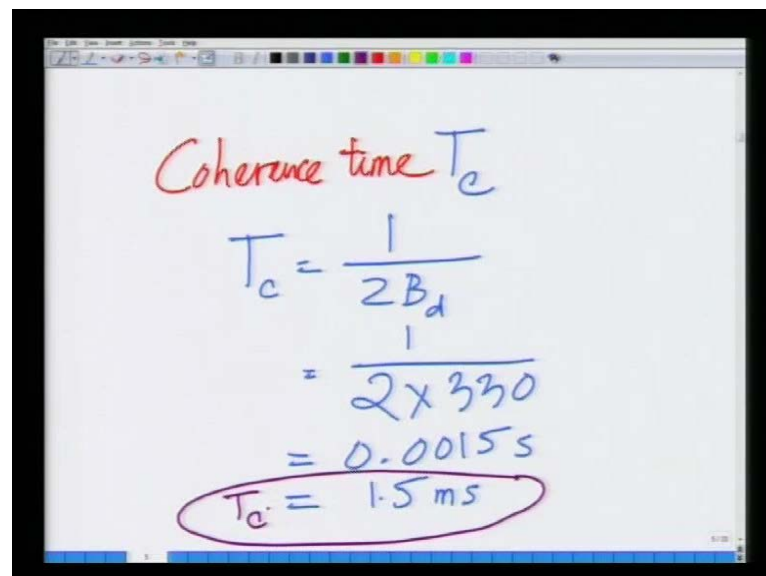
A whiteboard with a black border showing handwritten calculations in purple ink. The first line is $f_d = 165 \text{ Hz}$. The second line is $B_d = 2f_d = 2 \times 165$. The third line is $= 330 \text{ Hz}$. A wavy line underlines the result, and the text "Doppler Spread:" is written in red below it.

$$f_d = 165 \text{ Hz}$$
$$B_d = 2f_d = 2 \times 165$$
$$= 330 \text{ Hz}$$

Doppler Spread:

We computed earlier that for a vehicle moving at 60 miles per hour in the direction of the base station, at carrier frequency is 1.8 Giga hertz, the Doppler shift F_d is nothing but, 165 hertz, the Doppler shift is 165 hertz, which means the Doppler spread B_d equals twice F_d equals 2 into 165 equals 330 hertz that is the Doppler's spread, this is nothing but, the Doppler spread. That is the Doppler spread of a vehicle moving at 60 miles per hour with a person sitting in it, or a mobile moving at 60 miles per hour, directly towards the base station at and communicating at a carrier frequency of 1.85 Giga hertz is 330 hertz.

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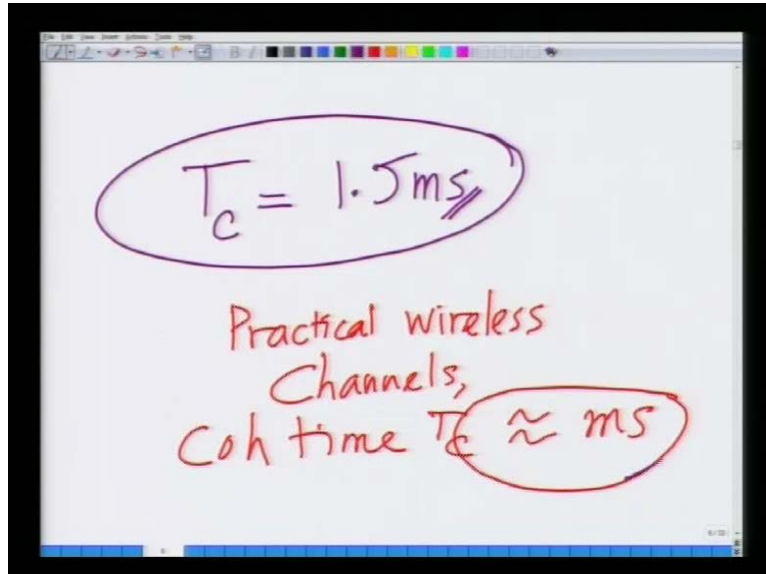
A whiteboard with a black border showing handwritten calculations in blue ink. The first line is "Coherence time T_c ". The second line is $T_c = \frac{1}{2B_d}$. The third line is $= \frac{1}{2 \times 330}$. The fourth line is $= 0.0015 \text{ s}$. The final line is $T_c = 1.5 \text{ ms}$, which is circled in purple.

Coherence time T_c

$$T_c = \frac{1}{2B_d}$$
$$= \frac{1}{2 \times 330}$$
$$= 0.0015 \text{ s}$$
$$T_c = 1.5 \text{ ms}$$

Hence, the coherence time T_c , T_c equals $1 / 2 B_d$ equals $1 / 2 \times 330$ hertz, this is also essentially 0.0015 seconds, or I can write the same thing as 1.5 mille seconds.

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$$T_c = 1.5 \text{ ms}$$

Practical wireless
Channels,
Coh time $T_c \approx \text{ms}$

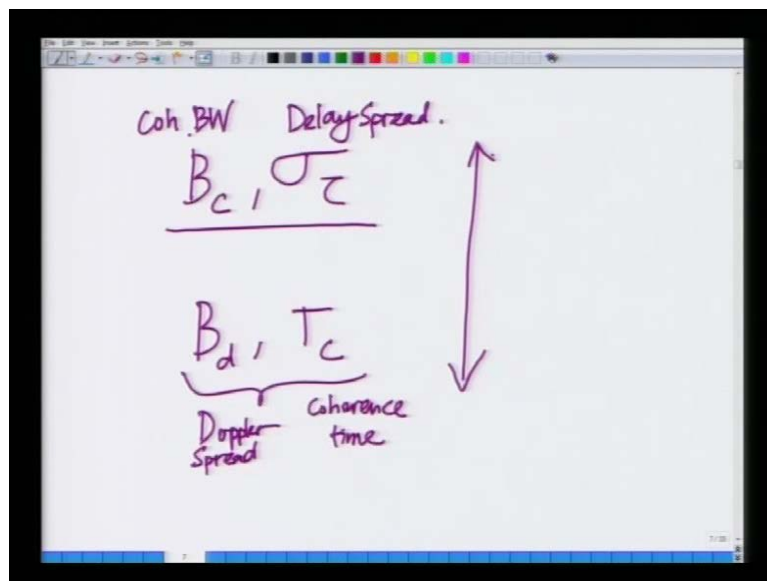
Hence, the coherence time is 1.5 mille seconds, since let me write this down here, T_c is coherence time equals 1.5 mille seconds, so what does this mean, this means that when a moving at 60 miles per hour towards the base station, roughly the period of time over which the channel is constant can be assumed to be 1.5 mille second, after 1.5 mille second, the channel changes to something different. That is, if I looking at the fading coefficient H in one period of 1.5 mille seconds is approximately constant, in the next period it has changed to some value, a different value which is $L H \Delta$ and so on, , so that is the importance of channel coherence step.

And you can also see that this coherent time is approximately of the order of mille second, so let me write down that property also here, in real wireless channel, so in practical wireless channels in practical wireless channels coherence time T_c approximately of the order of mille seconds. So, look at this what I am saying is, just like we said in practical outdoor wireless channels the delay spread approximately of micro second duration; coherence time which is another important parameter, which is related to how fast the channel is varying is roughly of the order of mille second.

These are two important quantities that remember, one is the delay spread which is micro second, one is the coherence time which is of the order of mile seconds aright; and both of

these are fundamentally different. One is in the delay spread is related that interval of time over which signal energy is arriving, while coherence time is related to the time duration over which the channel is constant; these are fundamentally different quantities one is not related to another, so I urge you again to revise this concept, so that you understand it clearly. Again to point out coherence time is related to the Doppler spread, while the coherence bandwidth is related to the delay spread.

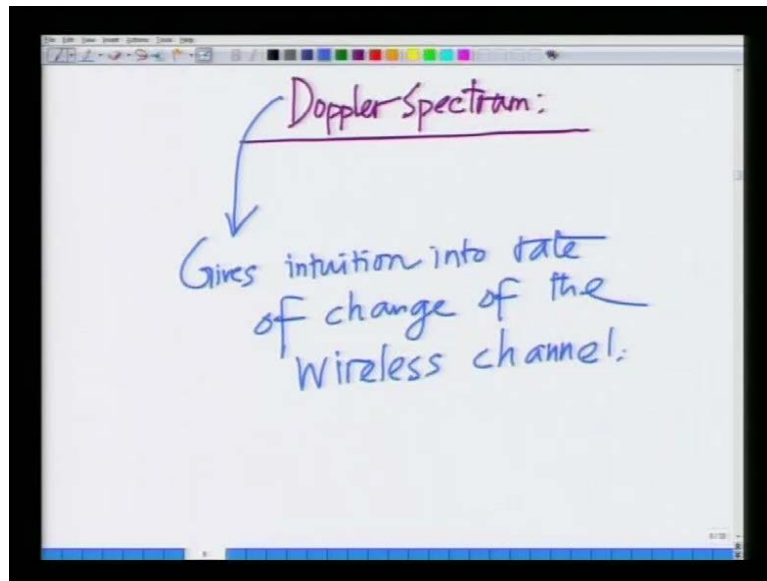
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So, these two, so let we write them down here B_c and coherence bandwidth, and the delay spread σ_τ are related to each other, this is coherence bandwidth, coherence bandwidth, and the delay spread are related to each other. And the Doppler spread, and coherence time these two are related to each other, Doppler spread and coherence time; I think it is essential to point out that sometime little confusing to think, that because the coherence word applies in both the bandwidth.

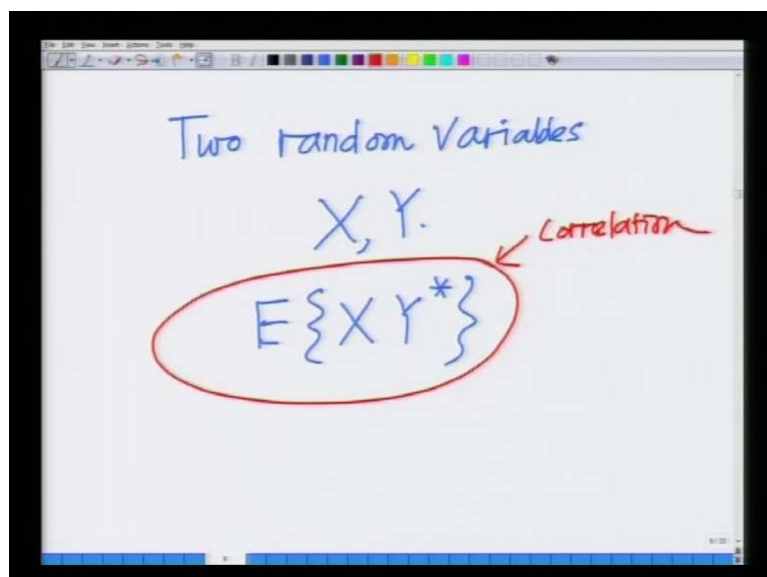
And the coherence time it is commonly, it is confusing some times and people relate coherence bandwidth to coherence time, however coherence bandwidth is related to delay spread. And Doppler's is related to coherence time, it is important to understand these two concepts very clearly, so I urge you again to revise this discussion in this lecture, in the pass lecture; so you have to understand these concepts clearly. Now, let us going to a more detailed analysis of the Doppler fading.

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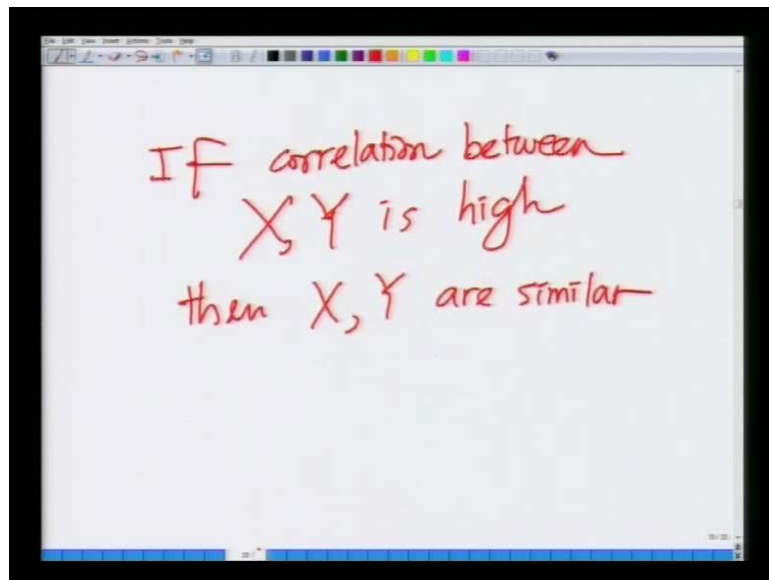
Let us compute what is known as the Doppler's spectrum, the Doppler's spectrum for this I will need what is known as the correlation function of the channel, so first I want to compute the Doppler spectrum. Because, it gives me a better idea, it gives me a more clearer idea of how faster, how slow is the channel varying with respect to time, so I want to compute the Doppler's spectrum. So, because it gives me, gives intuition into the rate of change of the wireless, of the 3G 4G wireless channel. So, if you want to analyze this rate of change in greater detail, I want to compute something known as the Doppler spectrum; now, for this I will embark on a related, I will go on a related discussion.

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Let us consider two random variables, two random variables X , Y we know that the correlation defined as expected X into Y conjugate, this is the correlation, this is the measure of how similar or dissimilar these random variables are, this is expected X Y conjugate is also known as the correlation.

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If the correlation is high if the correlation is high let me write them here, if correlation between X , Y is high, then X and Y are similar, then X , Y are similar. So, if the, if there are two random variables X and Y , I compute the correlation between X and Y given by expected X Y conjugate, if there if the correlation is high, then X and Y are similar that is they take values high values and low values, roughly related to each other.

However, in X and Y the correlation is low, then it means X and Y have no bearing on each other, what the value that X takes is not related to the value that Y takes. So, correlation is an important property of random variables, again please revise this concept from knowledge of random process probability. Now, so correlation indicates how similar or how dissimilar two the random variables are?

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$a_i(t)$ is the channel coefficient @ time t

$$a_i(t) = a_i e^{-j2\pi f \tau_i} e^{j2\pi f_d t}$$

time Varying channel.

time Varying phase factor

Now what I want to do is, I want to compute the following correlation, I want to compute we know that $a_i(t)$ is the channel coefficient at time t , $a_i(t)$ is the channel coefficient at time t of the i the part, and we also have an expression for $a_i(t)$, $a_i(t)$ is nothing but, a_i which is the attenuation times e power minus $j 2 \pi f \tau_i$ times e power $j 2 \pi f_d t$. We said $a_i(t)$ we earlier derive this expression, $a_i(t)$ is a_i the attenuation time e power minus $j 2 \pi f \tau_i$, where τ_i is the initial delay; times e power $j 2 \pi f_d t$ we said, this is the time varying, time varying face factor, this is the time varying face factor. This results in, this is what results in the time varying nature of the channel, this results in time varying wireless channel.

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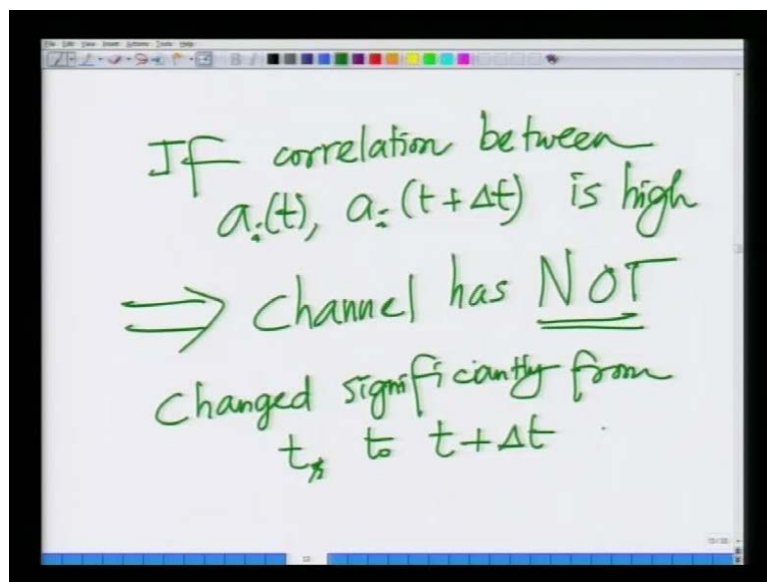
$$E\{a_i(t) a_i^*(t + \Delta t)\}$$

Coefficient Δt later

Now, what I going to compute is I want to compute the time correlation of this channel as follows, I will compute the quantity expected $a_i(t) a_i^*(t + \Delta t)$, that is I am taking $a_i(t)$ that is at a given time I am looking at the channel coefficient of i th path, I am computing its correlation with $a_i^*(t + \Delta t)$, which is the coefficient Δt later.

So, this is coefficient this is the coefficient Δt later, now if a if this correlation is high this means $a_i(t)$ and $a_i(t + \Delta t)$ have high correlation, which means the channel variation from t to $t + \Delta t$ is lower. However, if $a_i(t)$ correlated with $a_i^*(t + \Delta t)$ is low, it means the random variables are dissimilar, hence the channel has changed significantly from t to $t + \Delta t$.

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So, I will summarized that out here, if correlation between $a_i(t)$ and $a_i(t + \Delta t)$ is high, this implies channel has not remember has not changed significantly from t to $t + \Delta t$. So, if the correlation between $a_i(t)$ and $a_i(t + \Delta t)$ is high, the channel has not significantly changed from t to $t + \Delta t$, because $a_i(t)$ and $a_i(t + \Delta t)$ are similar, however if the correlation between them is low, then the channel has. Then $a_i(t)$ and $a_i(t + \Delta t)$ are different hence, the channel has changed significantly from $a_i(t)$ that from t to $t + \Delta t$. Now, let us compute this correlation, so that the correlation as the function of Δt gives me an idea how fast the channel is varying, as we just described before.

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$$a_i(t) = a_i e^{-j2\pi f_d \tau_i} e^{j2\pi f_d t}$$

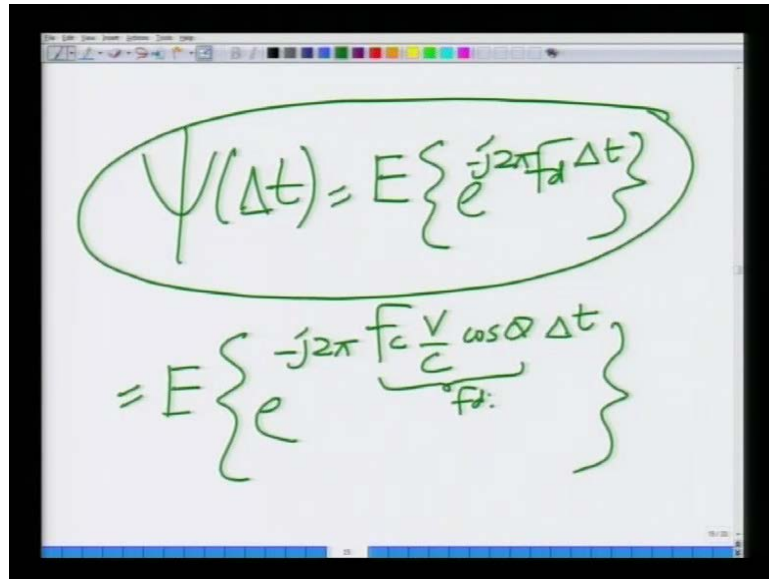
$$a_i(t + \Delta t) = a_i e^{-j2\pi f_d \tau_i} e^{j2\pi f_d (t + \Delta t)}$$

$$\psi(\Delta t) = E \left\{ \underbrace{|a_i|^2}_{|a_i|^2 = 1} \times 1 \times e^{j2\pi f_d \Delta t} \right\}$$

Now, $a_i(t)$ is nothing but, $a_i e^{-j2\pi f_d \tau_i} e^{j2\pi f_d t}$, and $a_i(t + \Delta t)$ equals $a_i e^{-j2\pi f_d \tau_i} e^{j2\pi f_d (t + \Delta t)}$, what I have done here, is I have written $a_i(t)$ and $a_i(t + \Delta t)$, now I need to compute $a_i(t)$ into $a_i^*(t + \Delta t)$, and take the expected value. So, I will write this as the correlation coefficient $\psi(\Delta t)$ as expected $a_i(t)$ into $a_i^*(t + \Delta t)$, which if I take this product a_i into a_i^* that gives me magnitude a_i^2 into $e^{-j2\pi f_d \tau_i} e^{j2\pi f_d \tau_i}$ into $e^{-j2\pi f_d t} e^{j2\pi f_d t}$ into $e^{-j2\pi f_d \Delta t}$.

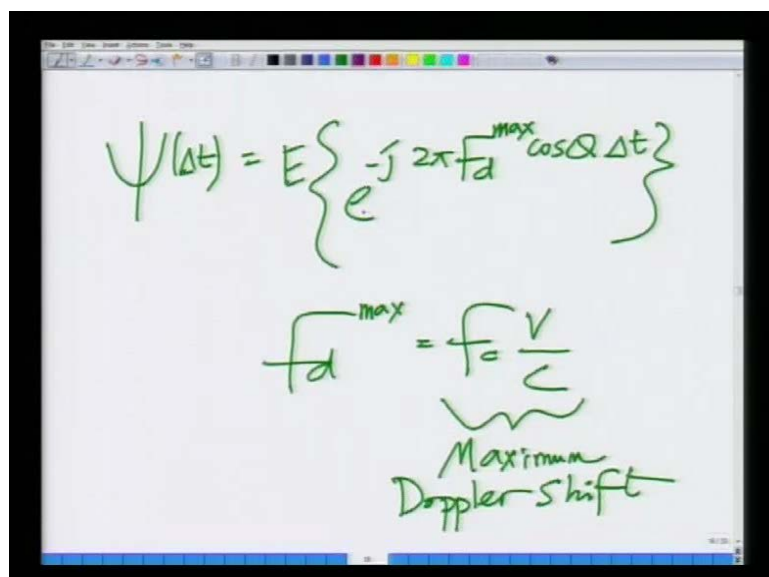
That gives me 1, into $e^{-j2\pi f_d t}$ into $e^{j2\pi f_d t}$ plus Δt that gives me $e^{-j2\pi f_d \Delta t}$. So, $\psi(\Delta t)$ which is the correlation between $a_i(t)$ and $a_i(t + \Delta t)$ is expected magnitude a_i^2 into $e^{-j2\pi f_d \Delta t}$, I will assume that this part is normalized to unit power that is, let us take this power out of the picture by normalizing. Because, this power depends on that transmitted signal power and so on, let us normalize this to unit power by taking power then take power out of the picture.

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$$\Psi(\Delta t) = E\{e^{j2\pi f_d \Delta t}\}$$
$$= E\left\{e^{-j2\pi f_c \underbrace{\frac{V}{c} \cos \theta}_{f_d} \Delta t}\right\}$$

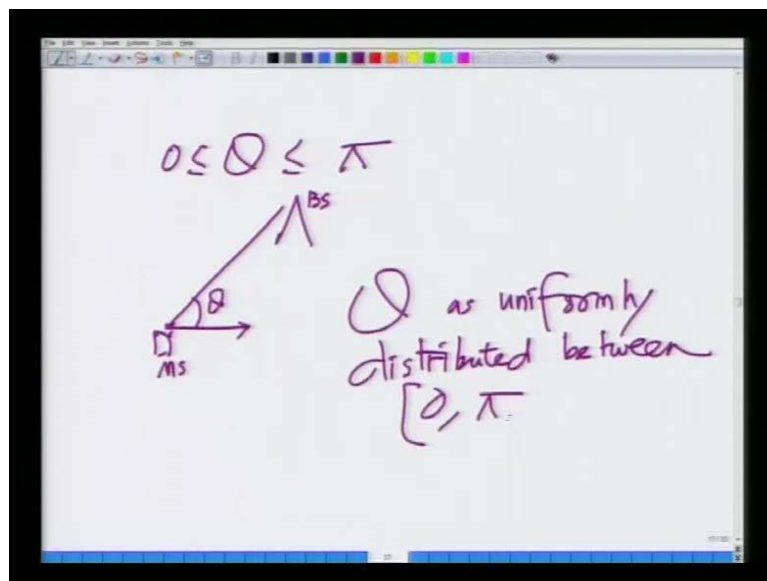
And now this becomes the correlation coefficient, now becomes psi delta t is expected e to power of minus J 2 pi f d delta t, this is the correlation coefficient, remember we still have to take the expected value we still have to take the expected over F d, over f over F d that is important . So, I will write this as, as follows I will write this as expected e to the power of minus J 2 pi and F d I will write as F c, V over C cosine theta into delta t, I am writing not doing anything it is complicated, I am simply writing F d. As F c V over C cosine theta, where V is the velocity, theta is the angle, this is nothing but F d I am substituting F d in the relation above.

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$$\Psi(\Delta t) = E\left\{e^{-j2\pi f_d^{\max} \cos \theta \Delta t}\right\}$$
$$f_d^{\max} = \underbrace{f_c \frac{V}{c}}_{\text{Maximum Doppler Shift}}$$

That gives me ψ of Δt , look at this $F_c V \cos \theta$ is nothing but, the maximum Doppler shift that corresponds to $F_c V \cos \theta$, when $\theta = 0$, this is the maximum Doppler shift that is $F_c V$, which is when the mobile is directly moving towards the base station or away from the base station. So, I will write this as expected value to the power of minus $\frac{1}{2}$ $F_c V \cos \theta$, where $F_c V \cos \theta$ equals $F_c V$, at this is the maximum Doppler, this is the maximum Doppler shift.

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Now, I have to take the expected value over θ , because remember θ is still unknown, because θ we said lies between 0 and π depending on the direction of the velocity of the line joining the mobile station and base station, θ can lie between 0 and π , so θ is a random quantity depending on the direction of the velocity of the line joining the mobile station and base station, θ can lie in between 0 and π .

Hence, we will assume θ as uniformly distributed between 0 and π , this is the reasonable assumption to make, because in the mobile cellular network the direction of velocity at any point is random. So, each user, each mobile station is moving randomly in one particular direction, that direction can be assumed to be uniformly distributed between 0 and π .

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$$\psi(\Delta t) = \int_0^\pi \frac{1}{\pi} e^{-j2\pi f_d^{\max} \Delta t \cos \theta} d\theta$$

$$\psi(\Delta t) = J_0(2\pi f_d^{\max} \Delta t)$$

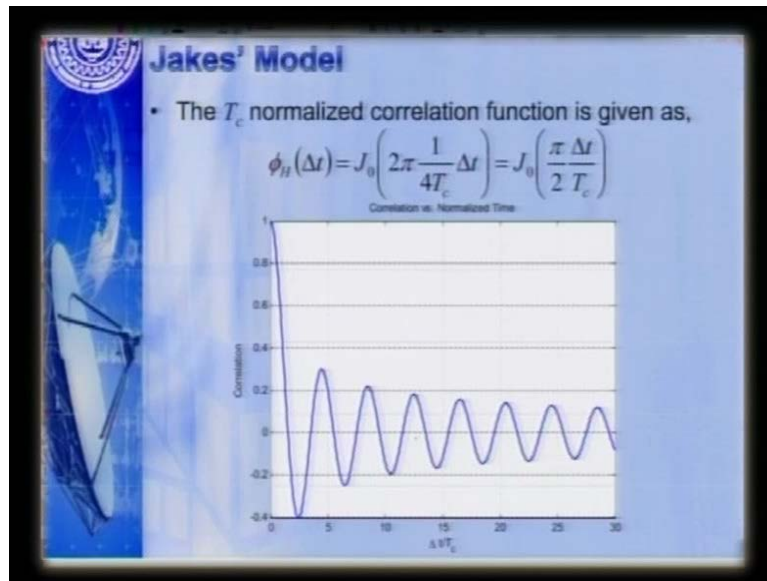
Bessel function of 0th order

Now, I can take the expected value of this correlation coefficient, ψ of Δt the average correlation is nothing but, $\int_0^\pi \frac{1}{\pi} e^{-j2\pi f_d^{\max} \Delta t \cos \theta} d\theta$. Look at this the correlation coefficient is the function of θ , that is $e^{-j2\pi f_d^{\max} \Delta t \cos \theta}$, I am saying θ is distributed uniformly in 0 to 2π . So, I am multiplying by $\frac{1}{\pi}$ that is the distribution, and averaging it from 0 to π by integrating this, hence this is nothing but, and hence this is the correlation.

Now, I am going to simplify this integral, this is $e^{-j2\pi f_d^{\max} \Delta t \cos \theta}$ times $d\theta$ this nothing but, the Bessel function, so this value is $J_0(2\pi f_d^{\max} \Delta t)$. So, I have derived the structure of the correlation, and that correlation is given by as a function of Δt is given as $J_0(2\pi f_d^{\max} \Delta t)$, where J_0 is the Bessel function, Bessel function of 0^{th} order.

So, we have now very interesting relation, we said to the correlation coefficient between the channel at t and $t + \Delta t$ can be derived as $J_0(2\pi f_d^{\max} \Delta t)$, the maximum Doppler shift corresponding to velocity V times Δt , where J_0 is the Bessel function of the 0^{th} order. Hence, this gives me the correlation as a function of time, let me show you a plot of this.

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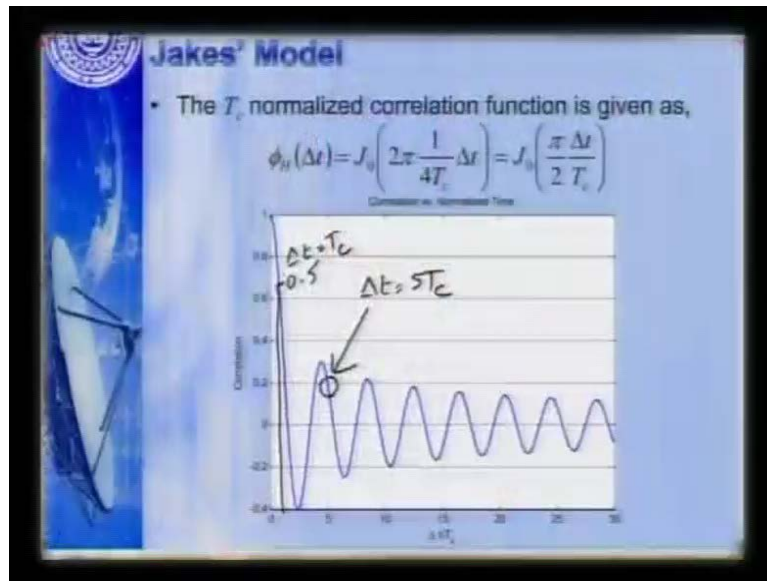
For instance, if you go here, here in this, I have plotted the correlation, before that let me do minor simplification to this.

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$$\begin{aligned} \psi(\Delta t) &= J_0(2\pi f_d^{\max} \Delta t) \\ &= J_0\left(2\pi \frac{1}{4T_c} \Delta t\right) \\ \boxed{\psi(\Delta t) &= J_0\left(\frac{\pi \Delta t}{2 T_c}\right)} \end{aligned}$$

So, this is J_0 , so $\psi(\Delta t)$ is $J_0(2\pi f_d^{\max} \Delta t)$, this can also be written as $J_0(2\pi \frac{1}{4T_c} \Delta t)$, remember $1/4T_c$ coherence that is $1/4$ times coherence time is f_d^{\max} , so f_d^{\max} I will write as $1/4T_c$ times Δt I can also write this as $J_0(\frac{\pi \Delta t}{2 T_c})$. So, this coherence time or this correlation function can also be expressed $J_0(\frac{\pi \Delta t}{2 T_c})$ that is correlation, at the channel Δt away.

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Now, let us go to the plot, I have plotted here $J_0 \pi \Delta t / 2 T_c$, you can see as Δt increases it is decreasing, but it is the Bessel function, it decreases in a cyclic kind of a. So, if you look at this this means at Δt equals $5 T_c$, let us look at this point Δt equals $5 T_c$, this is Δt equals $5 T_c$, that is at a time duration that is five coherence times away, the correlation is only 0.2 which means the correlation is low.

So, the channel has changed significantly in fact, if you look at Δt equals T_c somewhere around this point, Δt equals T_c you will find the correlation is 0.5, this is at Δt equals T_c . And hence, we are saying that at time lag that is T_c away the correlation between $a(t)$ and $a(t + \Delta t)$ is only 0.5. So, the channel has kind of changed significantly from $a(t)$, since the correlation is low; any correlation less than 0.5 can be considered as a low correlation.

If two random variables are significantly correlated with in the correlation has to be approximately goes to 1 so on so forth, which means at Δt equals T_c the correlation is 0.5, hence the channel has been changed significantly. And that also forms the basis of the rationale for earlier reason, argument that in one after one coherence period T_c , the channel changes significantly. Now, let us go back to the lecture, and again let us analyze this further, now I have the coherence function, I want to compute what is known as now, and I can from the correlation function.

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The image shows a handwritten derivation of the Doppler spectrum formula on a whiteboard. The title "Doppler Spectrum:" is written at the top. Below it, the formula is derived in two steps. The first step shows the Doppler spectrum $S_H(f)$ as the Fourier transform of the correlation function $\psi(\Delta t)$ with respect to Δt . The second step shows that $\psi(\Delta t)$ is a Bessel function $J_0(2\pi f_d \Delta t)$, and the integral is taken from $-\infty$ to ∞ .

$$S_H(f) = \int_{-\infty}^{\infty} \psi(\Delta t) e^{j2\pi f(\Delta t)} d(\Delta t)$$
$$= \int_{-\infty}^{\infty} J_0(2\pi f_d \Delta t) e^{j2\pi f(\Delta t)} d\Delta t$$

I can compute the Doppler spectrum, the Doppler spectrum S_H of F of a channel is given as minus infinite to infinite, it is given as the Fourier transform of the correlation function. So, I take the correlation function ψ of Δt , I take it is Fourier transform, remember Fourier transform with respect to Δt not with respect to t $d\Delta t$, this is the Fourier transform. Now, ψ of Δt we know it is Bessel function, so this is minus infinite to infinite, this is $J_0(2\pi f_d \Delta t)$ times $e^{j2\pi f \Delta t}$ $d\Delta t$, this can be given, this is given by an expression. This Fourier transform of the Bessel function is the standard known function, that is given by an inverted u shaped function.

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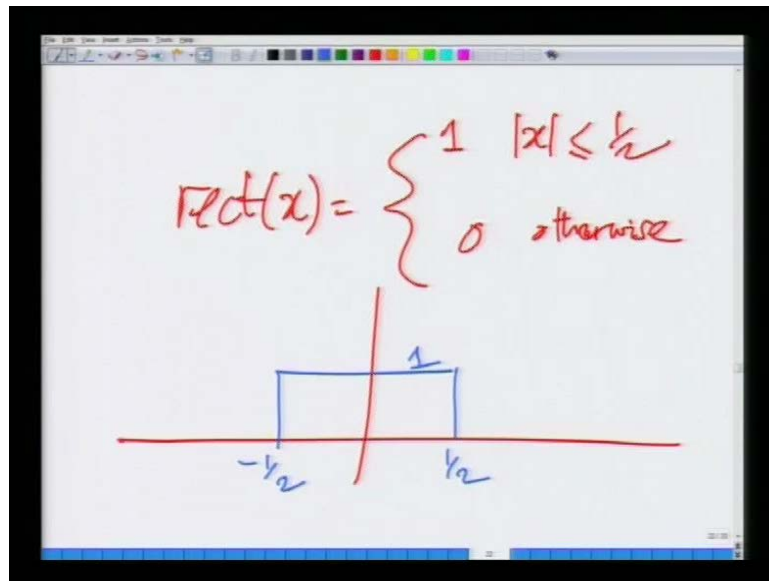
The image shows the final expression for the Doppler spectrum $S_H(f)$ written on a whiteboard. The formula is enclosed in a red box. Below the box, the words "Doppler spectrum" are written in red, with a red arrow pointing up to the $S_H(f)$ term in the formula.

$$S_H(f) = \frac{1}{\pi f_d} \frac{\text{rect}\left(\frac{f}{2f_d}\right)}{\sqrt{1 - \left(\frac{f}{f_d}\right)^2}}$$

Doppler spectrum

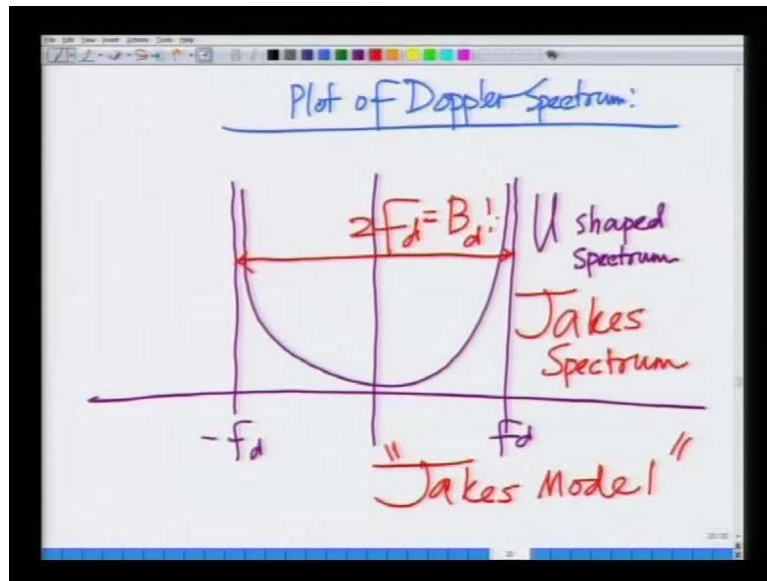
So, let us write that down that is a S H of F is nothing but, $\frac{1}{\pi F d} \max \text{rect}$, that is the rectangular function $\frac{F}{2 F d}$ divided by $\frac{1}{\sqrt{1 - \frac{F}{F d}^2}}$. Now, we have derived the Doppler's spectrum, this is what this is the Doppler spectrum, and we are saying that is the Doppler, that is Fourier transform of the correlation function, which is in our case is the Bessel function, which can be written as $\frac{1}{\pi F d} \text{rect} \frac{F}{2 F d} \sqrt{1 - \frac{F}{F d}^2}$.

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The rect function is given as follows, the rectangular x equals 1 is mod x less than equal to half 0, otherwise so the rectangular function looks as follow that is simply between minus half and half it is 1, and this is how the rectangular function looks.

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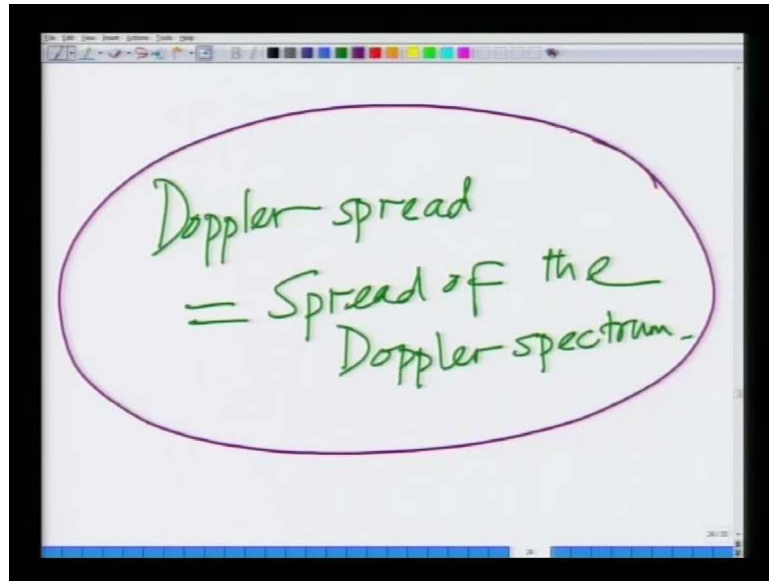


And now let me approximately plot this spectrum, Doppler's spectrum, so a plot of Doppler spectrum that looks as follows, now look at this Doppler's spectrum at f equals f_d , this is f by f_d is 1, so $1 - 1$ is 0 which means this shoots up to infinite. And for f greater than f_d , this is not defined, because then $1 - f$ over f_d is less than 0, hence the square root is not defined, hence it exists only for f between minus f_d and f_d , and towards f_d it shoots up to 0, this Doppler's spectrum is given as follows, it is symmetric, about f_d and it looks like this, at f_d and minus f_d , it shoots up to infinite.

It is an inverted U, it is increasing and limited between minus f_d and f_d , hence at f_d and minus f_d it shoots up to infinite, hence this is a U-shaped spectrum, this is a U-shaped spectrum. This also has a name, this is very popular in the context of 3G 4G wireless communication, even 2G wireless communications, this is known as the Jakes spectrum. This model is known as the Jakes model of a wireless communication channel, this spectrum is known as the Jakes spectrum and look at this, look at the spread of this, this spread between minus f_d and f_d is nothing but $2f_d$ which is nothing but B_d the Doppler spread that we defined earlier.

So, the Doppler spread is nothing but, the spread of the Doppler spectrum, so first thing is that this inverted U which shoots up to infinite at minus f_d and f_d is the standard Jakes spectrum, as part of the Jakes model. And this has the Doppler spectrum of $2f_d$, which is the B_d that is the Doppler spread that we defined earlier.

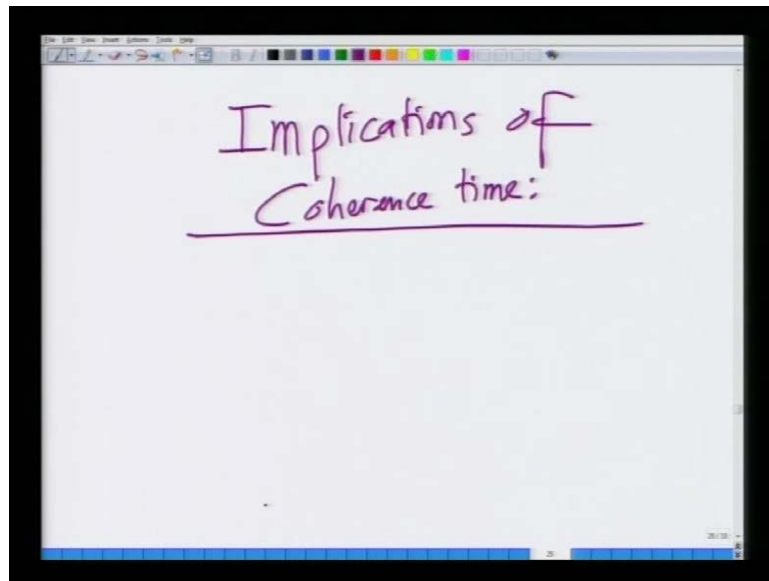
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Hence, we can now write Doppler's spread is nothing but, spread of the Doppler, the Doppler spread is nothing but, the spread of the Doppler's spectrum, as we have seen this is the Doppler spectrum, the spread is $2 F_d$ between minus F_d and F_d . So, the Doppler's spread is the spread of the Doppler spectrum, hence the Doppler spectrum gives us an idea how fast the channel is varying, if you want to get an intuitive to feel how fast the channel is varying. We compute the correlation function, from that we compute the Fourier transform which is the Doppler's spread.

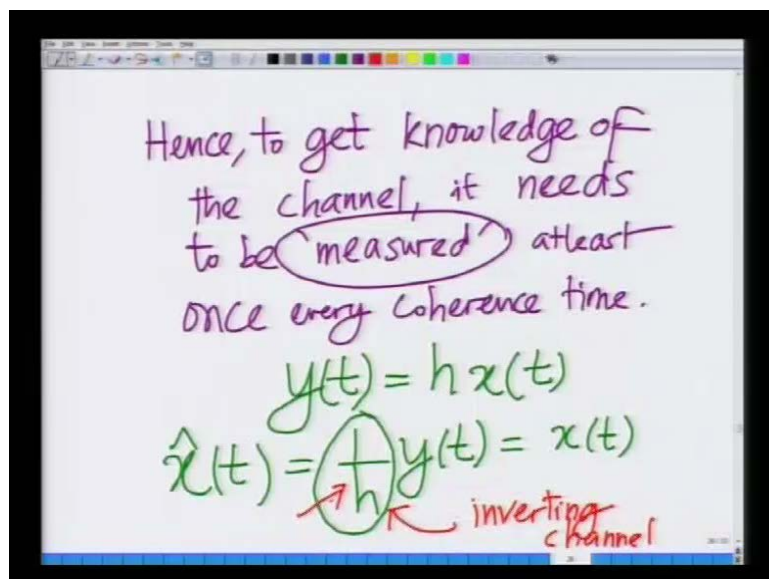
This Doppler spread you can measure the bandwidth of the Doppler of the Doppler spectrum that gives you the Doppler spread, and 1 over twice the Doppler spread, which is the coherence time which gives an idea of how the channel is varying. This is the complete procedure to compute the coherence time of any given wireless channel statistically; so this is very important concept the concept of Doppler spectrum.

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Now, what are the implications of coherence time, let us try to understand what the implications are, remember channel is changing constant for coherence time, and channel is changing from coherence time to coherence time. Hence the channel, if you want to know about the channel you need to measure it at least once every coherence time, hence to get knowledge of the channel it needs to be measure at least once every coherence time.

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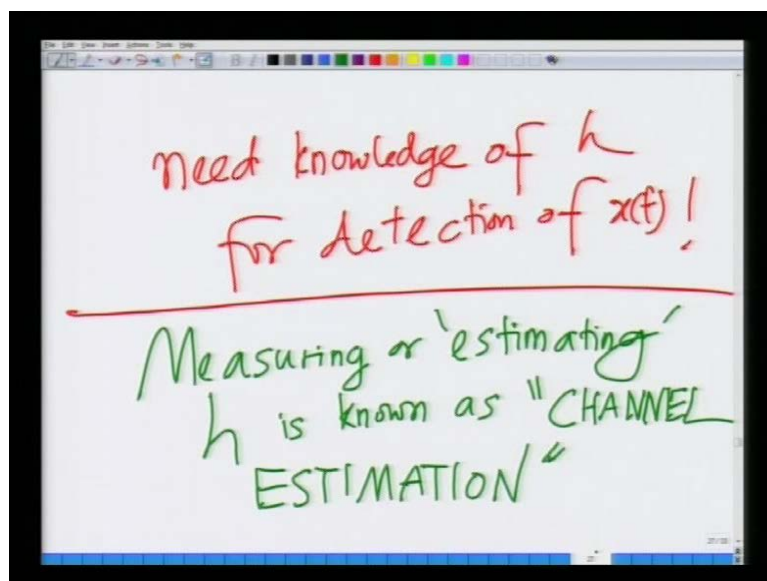


Let me write that over here, hence to get knowledge of the channel, it needs to be measure, this is the key word here measured at least once every, it needs to be measured at least once

every coherence time. Why do we need to measure it once every coherence time, look at this we said over channel received signal $y(t)$, is given as $y(t)$ equals h times $x(t)$ plus some noise, let $y(t)$ equal to h times $x(t)$ plus some noise. Now, let me assumed that the noise is insignificant that is noise is small, so I will erase this, so $y(t)$ is approximately h of $x(t)$.

Now, if I want to record that transmuted signal $x(t)$ at receiver that is if I want to recover $\hat{x}(t)$ of t , what do I need to do, I need to take $y(t)$ that is the received signal and divided by h ; which means $\hat{x}(t)$ is 1 over h $y(t)$. So, I need to take the received signal and divided by h to get back my $h(t)$, what am I doing here; here I am dividing by 1 over h . which is also known as channel inversion. So, I am inverting the channel, because h is multiplied with the channel, h is the fading coefficient which is sort of corrupting the signal, I need to invert this h by 1 over h multiplying it with the $y(t)$ that gives me $x(t)$.

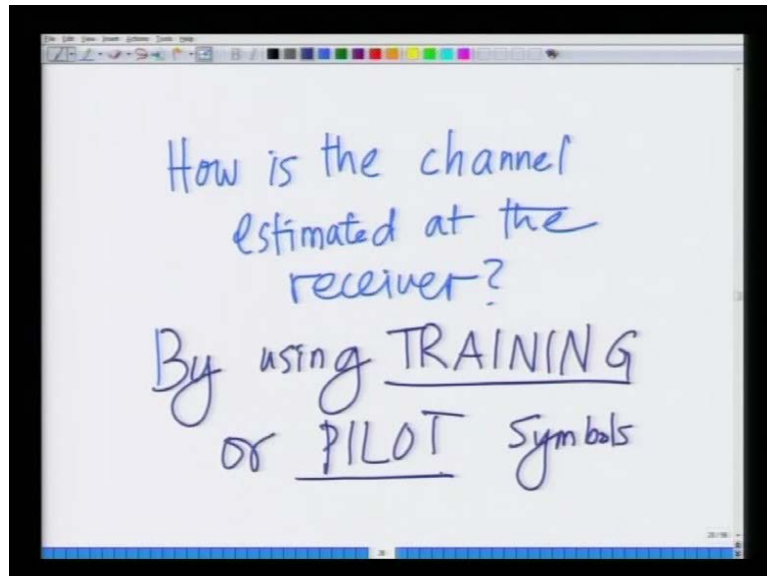
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Now, to do this I need knowledge of h , which is the channel coefficient, so I need knowledge of h , why I need the knowledge of h , I need knowledge of h for detection of $x(t)$. Hence, h needs to be measured at the receiver this process of measuring the h is a very standard procedure in 3G wireless, 4G wireless communications, it has a specific name that this procedure has a specific name, this known as channel estimation. Hence, measuring or in fact estimating, the technical word is estimating the channel coefficient h is known as channel estimation is a key procedure in every 3G 4G wireless communication system; it is very critical aspect of a 3G 4G wireless communication system. Because as we say we need h

for detecting that transmitted signal, hence the channel estimation is a key procedure for every 3G 4G wireless communication system.

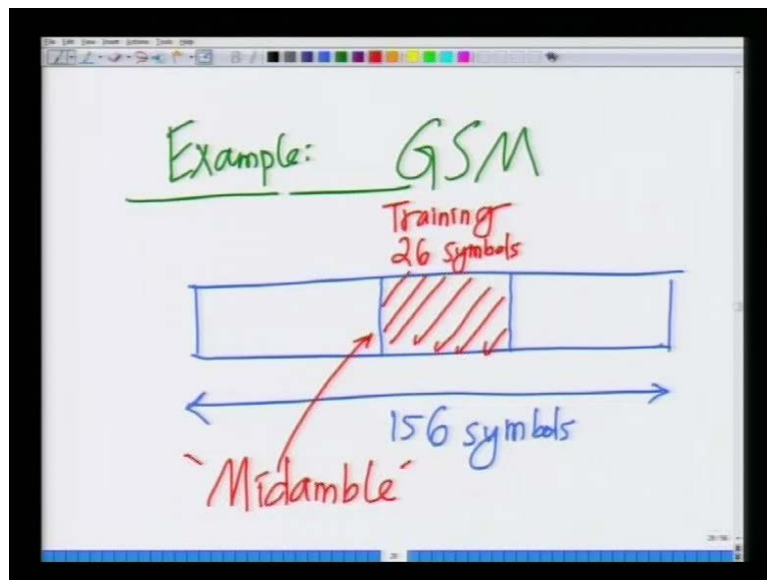
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So, now a pertinent question to ask is, how is this channel is estimated, how is the channel estimation done at the receiver, so let us ask the question is how is the channel estimated at the receiver. And the answer to that question is by, by using what are known as, training or pilot symbols. So, channel estimation has to be carried out the receiver, as we know this is a critical aspect, how is the channel estimation carried out the receiver we employed training, we train the receiver to give it the measure estimate of the channel; these are known as trainee or pilot.

Pilot are something that are transmitted in front of the actual things of the pilot symbols, or symbol that attracts transmitted in front of the information symbol, that is before transformation of the information symbols. These are employed essential to estimate the channel at the receiver, let me give you an example for instance.

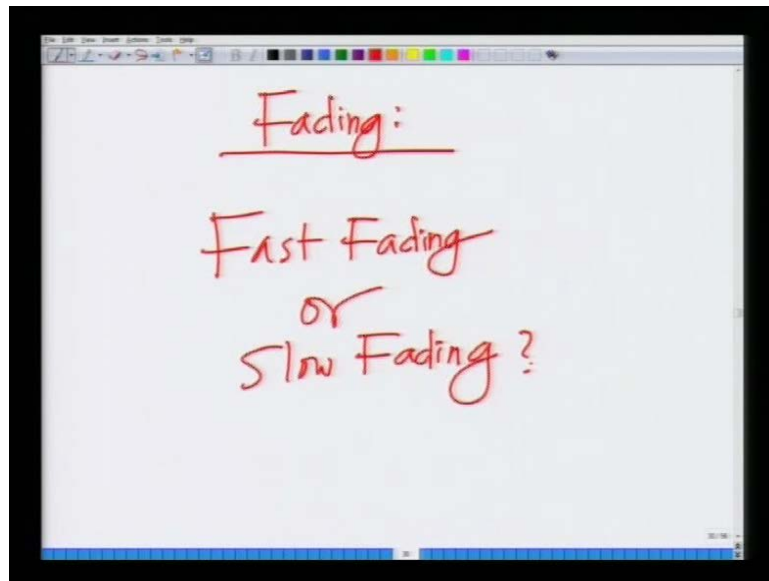
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Let us look at this example in a GSM system, let us consider the example of 2G GSM, it is very popular 2G GSM, every symbol, every frame there is a frame for a user, for every there is slot for every user. Every slot consists of 156 symbols, every slot consist of 156 symbol out of which 26 symbols that is 26, these are this is the set of 26 symbols, which constitute the training symbol, these are the training. In fact this is in the middle of, the middle of the slot, if anything is the in front of something of the something this is known as a preamble, if it is the middle it is known as midamble.

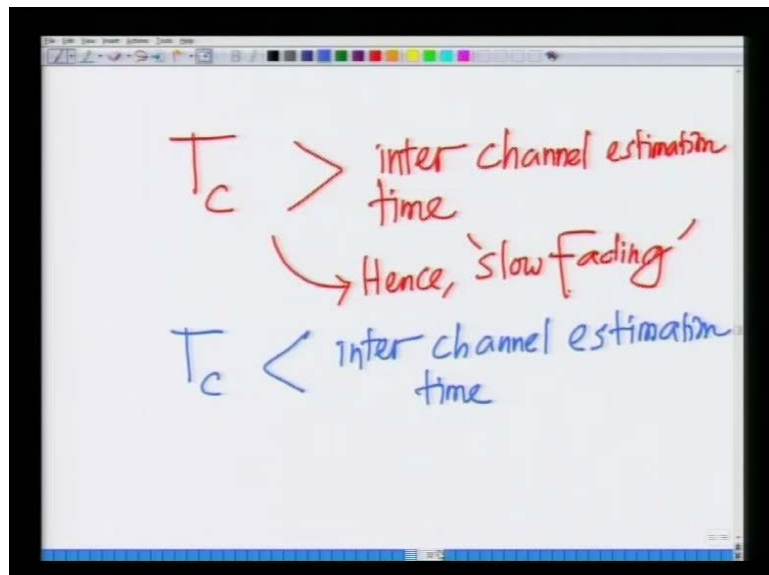
So, this also (()) or technically even referred to as midamble, so what I am saying is that GSM for every 156 symbol of the user, there are 26 symbol, there are transmitted in middle in the middle which is known as midambled. These are all, are this is also the training, and these are known as training or the pilot symbols, these are used to estimate the channel at the receiver. So, that you can use this channel estimate to detect yours signal that is essential idea.

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Now, fast versus slow fading, let us come to now introduce the concept of fading, now what is when do we say channel is fast fading or slow fading. So, is that fading fast or it is fading first or it is fading in time, so is it fast fading or slow fading, the answer to this question is it depends on the velocity and the system.

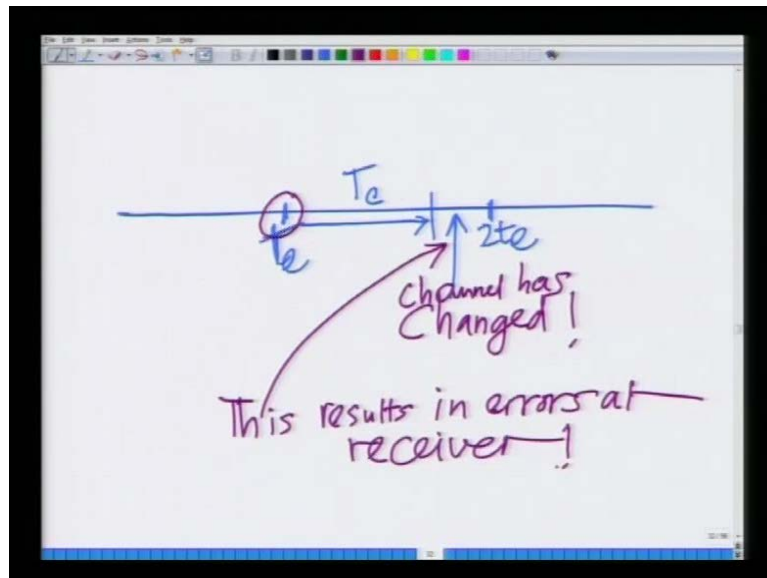
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Roughly speaking a channel if T_c , the coherence time is greater than the interval, successive intervals of carrying out channel estimation is, is greater than the inter channel estimation time, that is coherence time is greater than that time interval between two channel estimation

procedure. Then the channel is not changing during one channel estimation procedure, hence this channel is, hence this is a slow fading channel, however if T_c is less than the inter channel estimation time, that is I have estimated the channel.

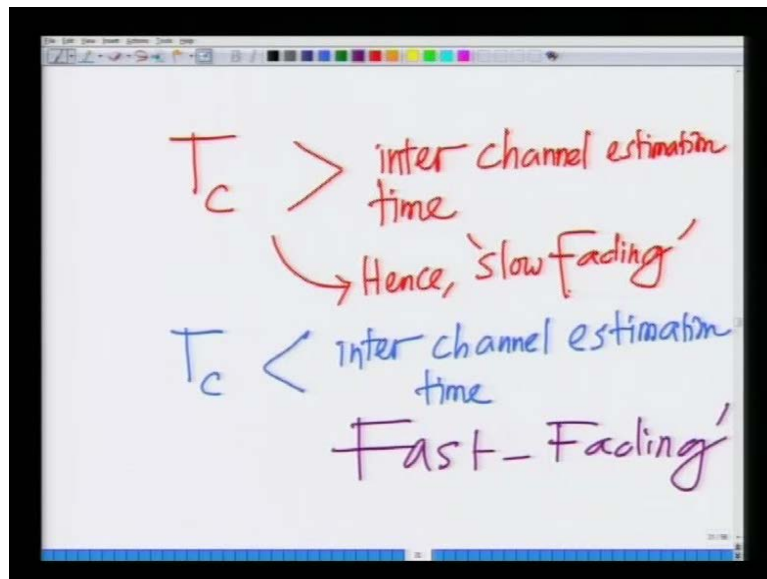
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Let me draw this pictorially I have estimated the channel that is, this is a this is the time at which the channel is estimated my next channel estimation time is t_e plus $2t_e$, but coherence time is only this much which is T_c , which is less than the inter channel estimation time, which means by the time I get here the channel has changed. So, if the channel if the coherence time that is the time for which the channel is constant is less than the inter channel estimation time.

Then my channel has changed which means I my receiver will be error, because if am using it to detect the receive signal but, my channel has change, so I am using wrong estimate. So, the channel has changed but, the receiver is still employing the wrong estimate corresponding to this channel estimation time, hence these results in errors at receiver.

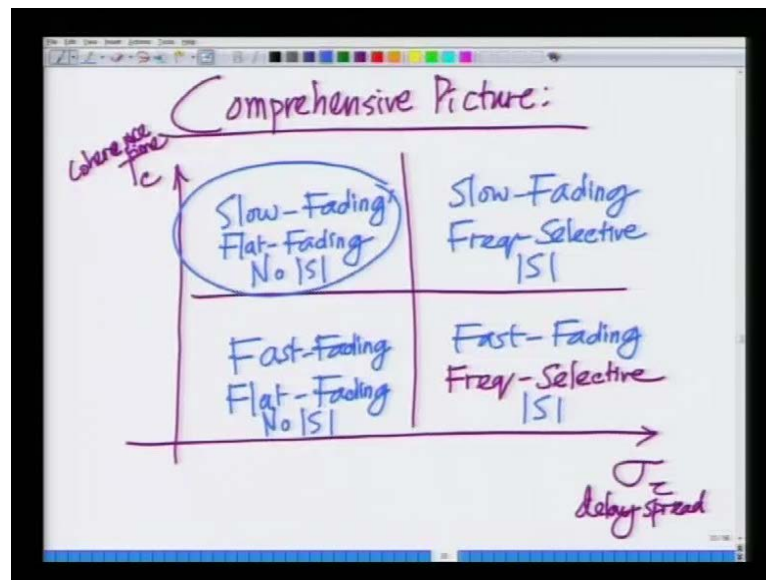
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Hence going back to the previous slide, if T_c is less than the inter channel estimation time, the coherence time less than the inter channel estimation time, then my channel is fast fading. Remember, we said something like this before in the case of coherence bandwidth, we said if the delay spread is greater than the symbol time, then the channel has inter symbol interference, which causes distortion at the receiver. Similarly, in concept to context of coherence time if the coherence of time if the coherence time is smaller than the inter, than the inter channel estimation time, that is the inter channel estimation time is larger than the coherence time.

Then the channel changes at very fast rate, at fast rate faster than the rate at which you measure the channel, hence it causes errors when use this erroneous estimate of the channel at the receiver for detection of the symbol. So, that is essential idea here of coherence time, and that is the importance of coherence time, now let us put all these ideas together, and formulated a comprehensive picture about the nature of the wireless channel.

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So, let me draw a comprehensive picture comprehensive picture, can be drawn as follows, let us say I am plotting the delays spread σ_{τ} here, this is the delay spread. And here on this axis I am plotting the coherence time, coherence time I will divide this into 4 quadrants, now very high delay spread, the delay spread becomes greater than the symbol time. So, this channel becomes frequency selective, similarly here it is also frequency selective, so very high delay spread the channel is frequency selective.

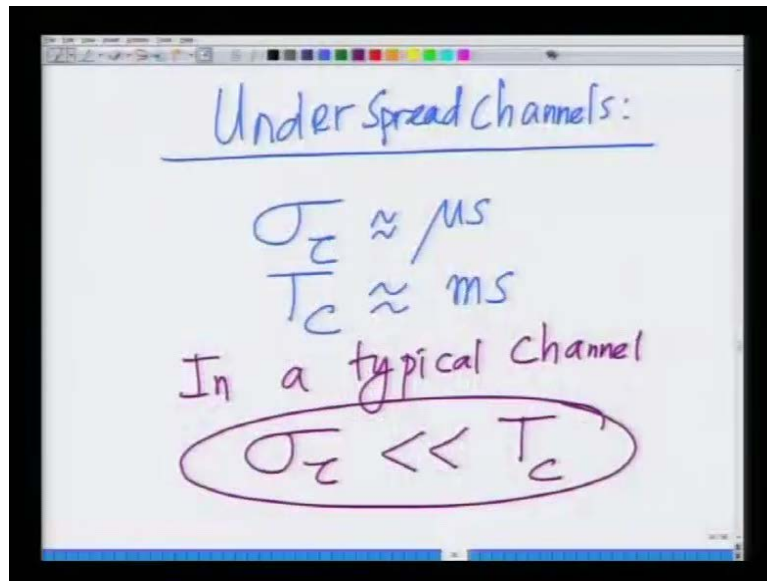
For low delay spread, the channel is flat fading, flat that is irrespective of the coherence time in low delay spread the channel flat fading, in high delay spread the channel is frequency selective. Now, let us look at what happens at high coherence time, the coherence time is high, then the channel is varying at the slow rate, hence the channel is slow fading, here also the channel is slow fading irrespective of the delay spread.

Similarly, if coherence time is low, then the channel is varying at very fast rate, so irrespective of the delay spread it is fast varying, and this is fast fading. Also if the delay spread is large it results in inter symbol interference inter symbol interference here there is no ISI, no ISI. Now, we have comprehensive picture for instance, let us look at this quadrant, in this quadrant the delay spread is high, but the coherence time is low, hence the channel is frequency selective, it has inter symbol results in the inter symbol interference.

And it is fast varying, because of the coherence time is low, look at this quadrant, in this quadrant the delay spread is low, but the coherence time is high which means no inter symbol

interference. The channel is fast fading, and it is slow fading, because the coherence time is high, so picture sort of gives you a comprehensive idea of different aspects of the wireless channel. And this understanding, this clarity of understanding about the nature of wireless channel is extremely important to understand different aspect of the 3G 4G wireless communication systems.

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Under Spread Channels:

$$\sigma_{\tau} \approx \mu s$$
$$T_c \approx ms$$

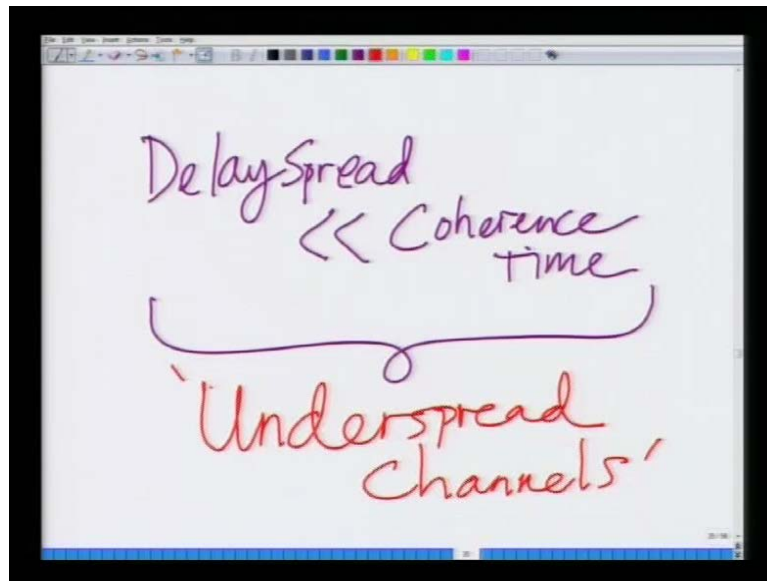
In a typical channel

$$\sigma_{\tau} \ll T_c$$

Let us now go to one final aspect in this channel characterization, which is that of under spread, this is under spread wireless channel, if you looked before, we said the delay spread of an outdoor wireless channel is approximately of the order of micro seconds. And the coherence time is approximately of the order of mille second, look at this, we looked at both the delay spread of the outdoor channel, and the coherence time.

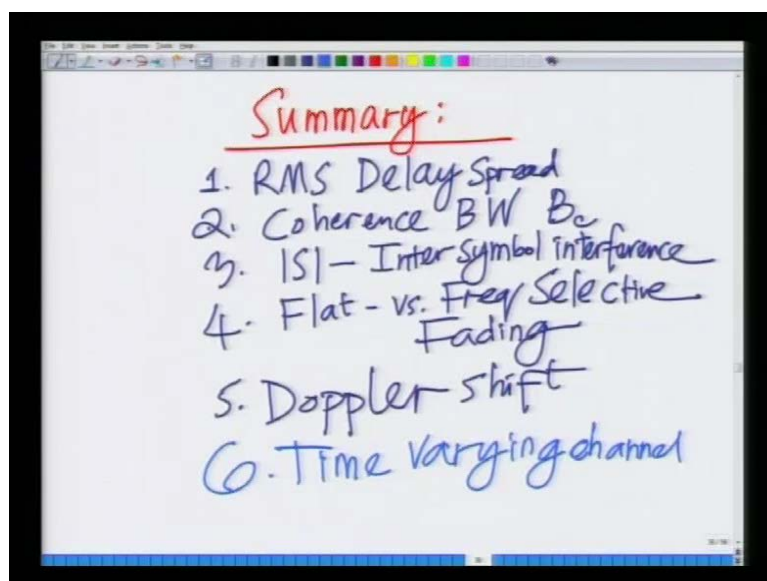
We said that delay spared is typically of the order of micro seconds, and the coherence time is typically of the order of mille seconds, which implies that in a typical channel in a typical channel, the delay spread sigma tau is much smaller than the coherence time. So, in a typical wireless channel the delay spread sigma tau, which is of the order of micro seconds is about a 1000 times smaller than the coherence time, which of the order of mille second.

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This property channels that satisfies this property channels that satisfy delay spread much smaller than coherence time, channels that satisfies this property have special name, they are known as under spread. Channels that satisfy the property that the delay spread is much less than the coherence time, are known as under spread channel, so this the last idea and all typical wireless channels outdoor, indoor under spread channel. So, this is the last idea of with which, we will conclude this section on channels, let me revise again, let me again list what topics we have covered in this context, in this section we have covered topics such as...

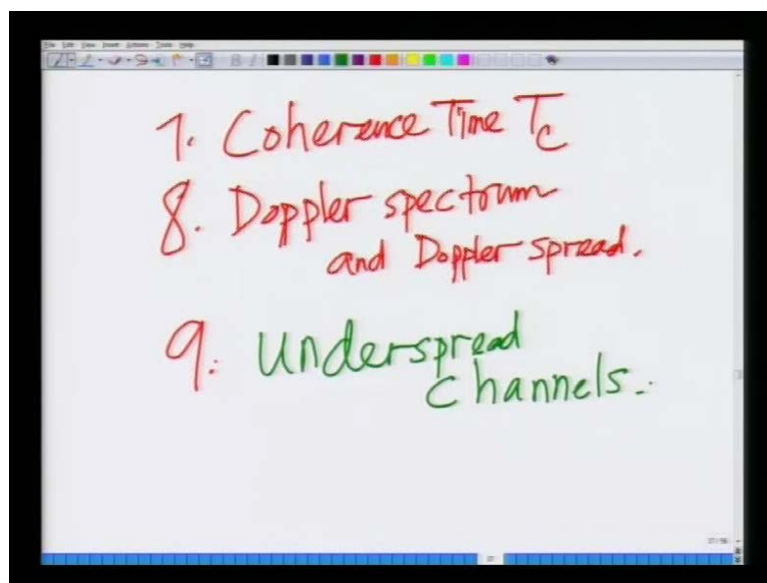
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So, list of topics that is covered, so let me write the summary, in summary the list of topics that we are covered is first, the RMS delay spread, we said this the interval over which the signal power is arriving. We covered concept of coherence bandwidth, which is B_c , we have covered ISI which is inter symbol interference, we have covered frequency we said that, if the delay spread is greater than the symbol time it results inter symbol interferences.

And frequency domain, it means that signal bandwidth is greater than the coherence bandwidth, so we consider the concept of flat versus frequency selective fading. Then we introduced the notion of the Doppler shift arising from the velocity, if a mobile is moving towards or away from the base station, we said that results in the Doppler frequency shift. We said this inter results in something important, this is a time varying channel, we said this results in a time varying channel.

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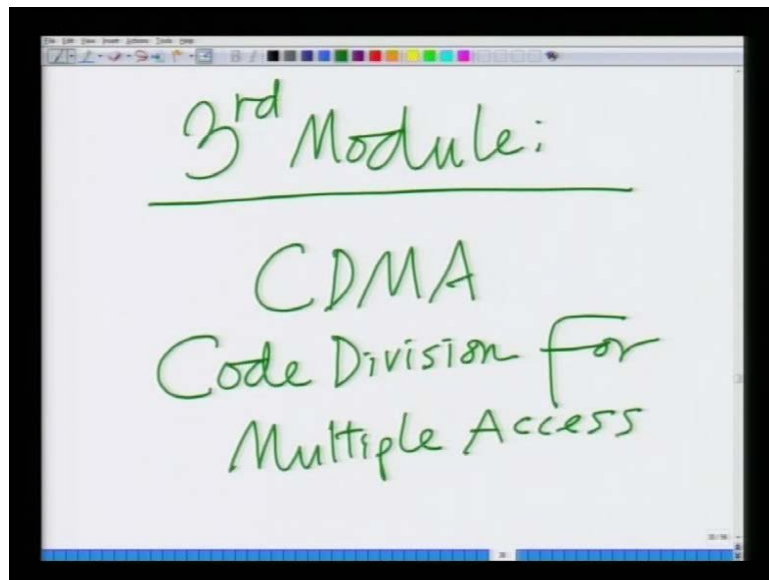


Hence what becomes important is what we considered next is the notion of coherence time T_c , this is the approximate time for which the channel can be assumed to be constant; it changes from coherence time to coherence time, which means we have to periodically every coherence time, we have to do channel estimation. Then we looked at the Doppler spectrum and the Doppler spread and then we also looked at finally, the last concept today, we looked at under spread.

That is channel whose delay spread is much smaller than the coherence time are known as under spread channel. These are the topics that we looked at today, this completes our 2nd

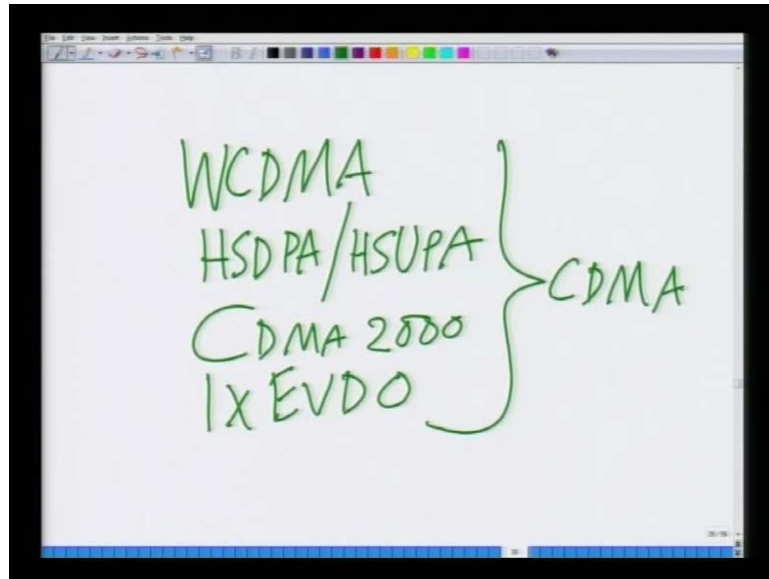
module on understanding the properties of the wireless channel, characterization of the properties of the wireless channel. Hence, now we will go on to the 3rd module of the course, on 3G over 4G wireless communications, which is the module on core division for multiple access.

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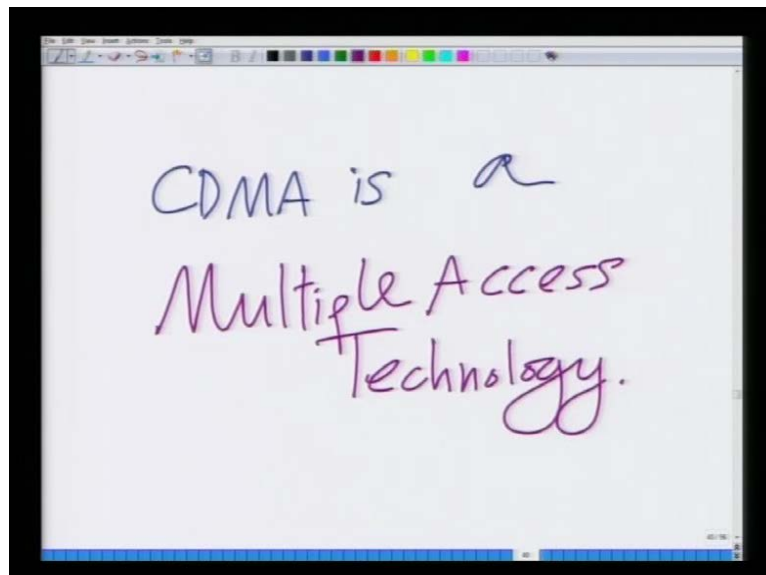
So, now let us start, the 3rd module this is the module on CDMA which also stands for Code Division for Multiple CDMA stands for Code Division for Multiple Access. CDMA is a very important physical layer technology, or a important technology for third generation 3G, third generation wireless communication system. In fact, I urge you to revise the earlier or recollect your 1st lecture, very 1st lecture, we did in the course on 3G 4G wireless communication system, we said 3G systems are all based on CDMA.

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Namely we said, WCDMA, and HSDPA slash HSUPA, and also CDMA 2000, 1 XEVDC and so on, and so forth. All these are based on CDMA, hence CDMA is a key technology as the name implies CDMA is multiple access technology, it is remember CDMA stands for code division for multiple access.

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Hence CDMA is a multiple access, this notion is an important notion, what is the multiple access technology, so with this, because short this point I will conclude this lecture. We will

start with a discussion introduction to CDMA, and analyzing CDMA, a core division for multiple access in details starting in the next lecture.

Thank you for your attention.