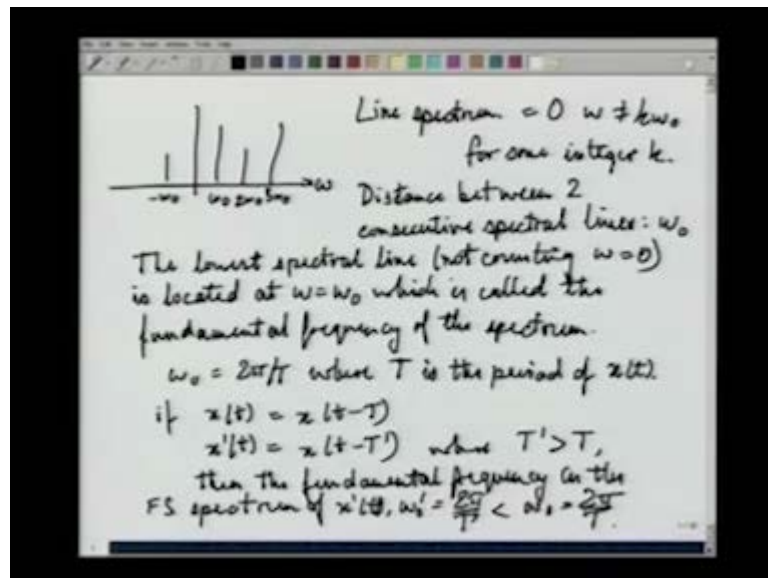


Signals and System
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Lecture - 27
Fourier Spectrum

(Refer Slide Time: 00:38)



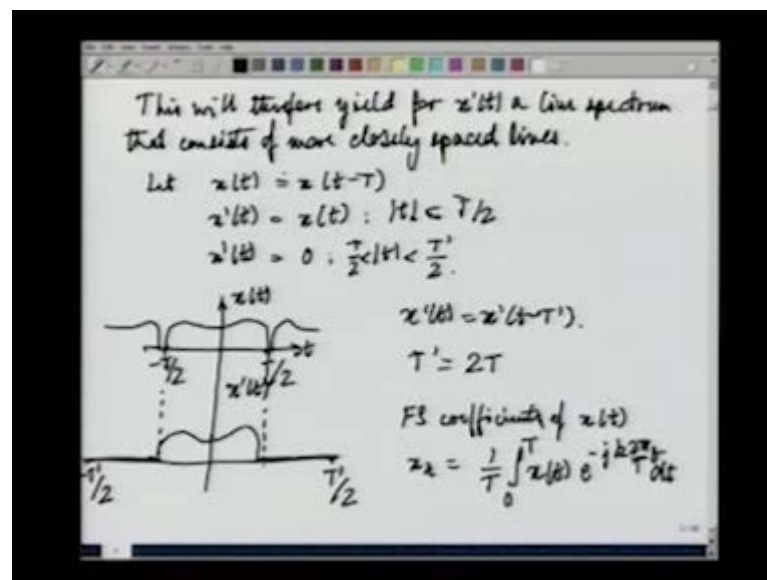
The Fourier series spectrum that we have learnt about is often called a line spectrum, and the reason is obvious this is because as we had made a few sketches. The last lecture, if ω is the variable against which we make the plot rather than k , then you see that the lines occur at ω_0 , $2\omega_0$, $3\omega_0$ and so on. In short there are values at specific discrete points which are multiples of a certain quantity ω_0 which is equal to $2\pi/T$, where T is the period of the signal whose Fourier series expansion has been found. At all other values of ω there is no value assigned to the Fourier spectrum, that is why this is called a line spectrum.

Now, as we can see this line spectrum is equal to 0, at all ω not equal to $k\omega_0$ for some integer k , right. Now, what is the distance between 2 consecutive spectral lines? This is ω_0 between 2 consecutive spectral lines. This is ω_0 , and the lowest spectral line is at frequency ω_0 not counting $\omega=0$. $\omega=0$ is located at ω_0 , which is called the

fundamental frequency of the spectrum. The physical meaning that is attached, that is often attached to the Fourier series spectrum of a periodic signal is the following, we say that if non zero Fourier series components x_k exists for larger values of k , then we will say that the signal the periodic signal $x(t)$ has high frequency components. If it has significant values only for lower values of k or lower values of modulus k , then we will say that it is a low frequency signal or a low pass signal.

So, that is the kind of meaning that one attaches to the existence or non existence or largeness or smallness of the various spectral lines that constitute the Fourier series spectrum. Now, one thing that is clear at the end of this discussion is that the distance between the lines is dictated by ω_0 , and ω_0 is nothing but $2\pi/T$, where T is the period of $x(t)$. So, if we take a different $x'(t)$ with a greater period, then it will give rise to a fundamental frequency ω_0' which has a smaller period, which has smaller fundamental frequency. Thus for example, if $x'(t)$ equals $x(t)$ minus T and $x'(t)$ equals $x(t)$ minus T' , where T' is greater than T , then the fundamental frequency in the FS spectrum of $x'(t)$ namely ω_0' equal to $2\pi/T'$ is less than ω_0 which is equal to $2\pi/T$.

(Refer Slide Time: 07:03)



This will therefore yield for $x'(t)$ a line spectrum, that is or that consists of more closely spaced lines why simply, because for $x'(t)$ not only is the fundamental frequency ω_0' less than ω_0 , which was the fundamental

understand, how their Fourier spectra relate the Fourier series lines spectral related to each other in order to do this lets for convenience make the assumption that T dash equals 2π .

Now, let us first extract the Fourier series components of $x(t)$, which are the Fourier series components of $x(t)$, FS components another co-efficients. So, this is what we have? Now note that, since $x(t)$ is a periodic signal, the integration to evaluate the Fourier series co-efficients, which we have been doing from 0 to T , could well be done over any interval any consecutive interval of T , it may not be from 0 to T , it could be from any point t_0 to $t_0 + T$ and we would still evaluate the same coefficients x_k , let me write this down.

(Refer Slide Time: 16:05)

$$x_k = \frac{1}{T} \int_{t_0}^{t_0+T} x(t) e^{-jk\omega_0 t} dt \quad -\infty < k < \infty.$$

$$= \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk\omega_0 t} dt$$

FS coeffs x'_k of $x'(t)$: $x'_k = \frac{1}{T} \int_{-T/2}^{T/2} x'(t) e^{-jk\omega_0 t} dt$

$$= \frac{1}{T} \left[x(t) e^{-jk\omega_0 t} \right]_{-T/2}^{T/2} + \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk\omega_0 t} dt$$

$\omega_0 = \frac{2\pi}{T}$

$$= \frac{1}{T} \left[x(t) e^{-jk\omega_0 t} \right]_{-T/2}^{T/2} + \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk\omega_0 t} dt$$

$= 0$ because $x'(t) = 0$ at $t = -T/2, T/2$

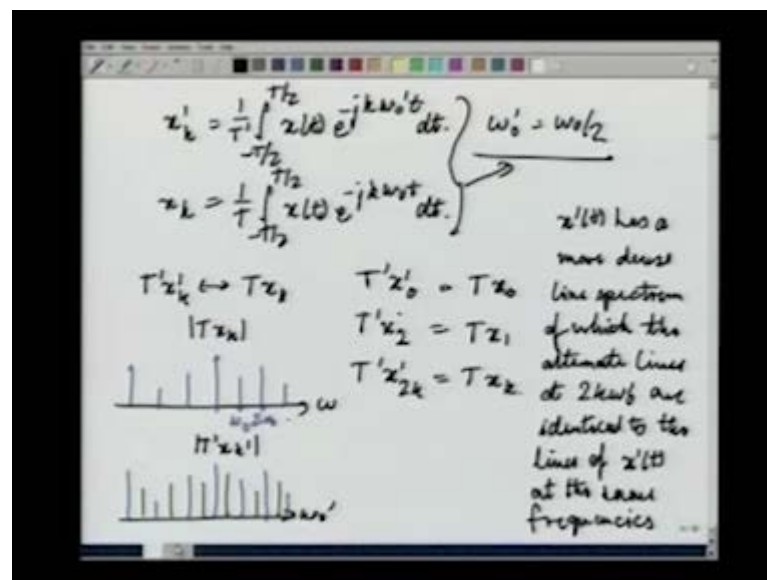
x_k equals $\frac{1}{T} \int_{t_0}^{t_0+T} x(t) e^{-jk\omega_0 t} dt$, let us say $t_0 = -T/2$ there is a t_0 to $t_0 + T$, that is any contiguous interval of length T to the minus $j k \omega_0 t$ dt, and this expression would hold for k . So, I could choose for example, this minus $T/2$ to $T/2$ which means I am choosing the interval of minus $T/2$ to $T/2$ x t e to the minus $j k \omega_0 t$ dt, suppose I choose this.

Now, I have all the Fourier series co-efficients x_k corresponding to the signal $x(t)$, right. Now, let us look at the Fourier series co-efficients x'_k of $x'(t)$, they would be given by using the same argument, and using the limits minus $T/2$ to $T/2$, we would have x'_k equals $\frac{1}{T} \int_{-T/2}^{T/2} x'(t) e^{-jk\omega_0 t} dt$

$x(t)$ to the minus $j 2 \pi k$ by T dash t $d t$, that is what you would get? But note that $x(t)$ as defined over the T dash length contiguous in 12 minus T dash by 2 to T dash by 2 has the same value of x as $x(t)$ over the interval minus T by 2 to plus T by 2 , and is anyway 0 elsewhere. Hence this integral can be equally written as 1 by T dash integral minus T dash by 2 to minus T by 2 $x(t)$ e to the minus $j k \omega_0$ naught dash t $d t$, where ω_0 naught dash equals 2π by T dash plus integral 1 by T dash again of course, integral minus T by 2 to T by 2 $x(t)$ e to the minus $j k \omega_0$ naught t $d t$ plus the last term 1 by T dash integral T by 2 to T dash by 2 $x(t)$ e to the minus $j k \omega_0$ naught t $d t$ here also ω_0 naught dash t here also ω_0 naught dash t $d t$.

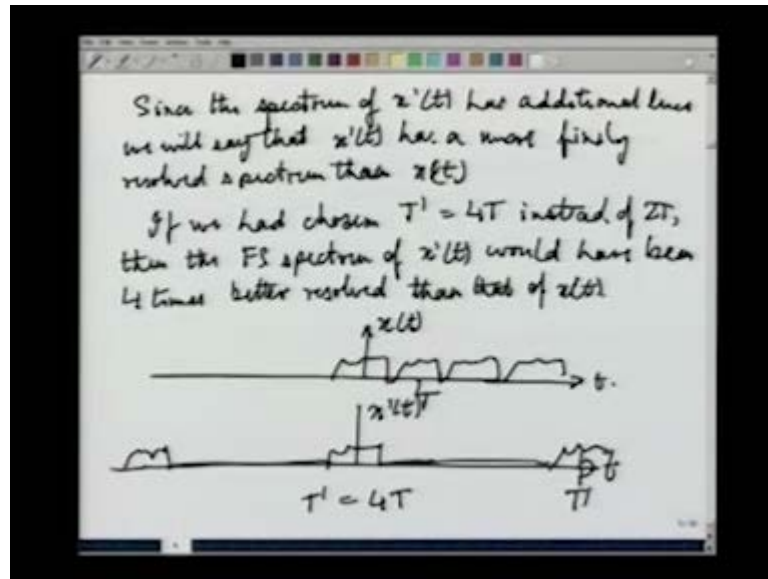
Now, this is 0, this is also equal to 0, because $x(t)$ equal to 0 in this interval for t minus T dash by 2 to minus T by 2 , this is 0 because x dash, fine. So, that gets rid of this term, that gets rid of this term. Now finally, just the middle term, and this middle term equals 1 by T dash integral minus T by 2 to T by 2 of $x(t)$, because in this interval minus T by 2 to T by 2 $x(t)$, if you go back to the earlier equations over here. $x(t)$ has been defined to be the same as that $x(t)$ has been defined to be same as $x(t)$ it is minus $j k \omega_0$ naught dash t $d t$, fine. So, this is what we have let us make sense out of this.

(Refer Slide Time: 22:34)



If you look at the equations we would finally get, it is the following we have x dash k equals integral one by T dash, sorry integral minus T by 2 to T by 2 $x(t)$ not x dash t $x(t)$ well x dash t also, but at anyway we can simplify to $x(t)$ e to the minus $j k \omega_0$ naught

(Refer Slide Time: 30:15)



So, let us put this down since the spectrum of x dash t has additional lines, we will say that x dash t has a more finely resolved spectrum, then x of t and it became more finely resolved simply by changing the periodicity of the signal without changing the shape of the signal adding it with extra 0 space, and doubling the period from t from capital to capital T dash. If you wish you can go back to the definition and take a look this was the definition of x dash t with respect to x t .

Now, this same argument can be extended and it is easy to see that if we had chosen T dash equal to $4 T$ instead of $2 t$, then the FS spectrum of x dash t would have been four times better resolved would have been four times better resolved than that of x t . Then we would have had a picture that goes like this x t would be a function that say. So, x t would have a shape like this followed by shape like this repeating itself well, I cannot write perfectly, but it is quite clear that I want to repeat the same thing again, and again whereas x dash t would have a far more sparse repetition.

Then a gap for three of these, and then again making a copy here and again gap on this side for three periods and then having a copy here, this would be x dash of t , this is what would happen? If you set T dash equal to $4 T$, because this would be T dash T dash whereas, this is t over here. So, T dash is four times t , and if you had this you would have a spectrum for T dash that has four lines in place of every single line of x of this spectrum of x t in short x k 's.

(Refer Slide Time: 34:33)

Handwritten mathematical derivation on a whiteboard:

$$\begin{aligned}
 &T x_k \quad X(k\omega_0) \quad T' = 4T \quad \omega_0 = 4\omega_0' \\
 &T' x_{k'} \quad X'(k\omega_0') \\
 &X(0) = X'(0) \\
 &X(\omega_0) = X'(4\omega_0') = X'(4\omega_0') \\
 &X(2\omega_0) = X'(8\omega_0') = X'(2\omega_0) \\
 &X(k\omega_0) = X'(4k\omega_0') = X'(k\omega_0) \\
 \\
 &\text{Let } T' \rightarrow \infty \\
 &\omega_0' \rightarrow d\omega \\
 &x(t) = \sum_{k=-\infty}^{\infty} x_k e^{jk\omega_0 t} \xrightarrow{T' \rightarrow \infty} \sum_{k=-\infty}^{\infty} \frac{X'(k\omega_0')}{T'} e^{jk\omega_0' t} \\
 &\text{In the limit, } x(t) = \int_{-\infty}^{\infty} X'(\omega) e^{j\omega t} \frac{d\omega}{2\pi}
 \end{aligned}$$

I am from, now on I will call the spectral lines $T x_k$ by the name x of k omega naught, because they are occurring at k omega naught and the spectral lines $T' x_{k'}$, I will call this as x k dash k omega naught dash this is just a notation. I am going to use now using this notation it is evident that with T' equal to four with T equal to 4 T , you would have x_0 equal to x dash 0, then x omega naught equal to x dash four omega naught dash equal to x dash omega naught. Because now remember that omega naught equals four omega naught dash, because of this. And next x_2 omega naught equals x dash 8 omega naught dash equals x dash 2 omega naught and so on.

So, in general in general you you would be able to write x of k omega naught equals x dash 4 k omega naught dash equals x dash k omega naught. So, this is the kind of spectrum you would get more, and more highly resolved spectrum results when you increase the period of x dash t with respect to x t . Now, we we just want to follow up on this lead, and find out what happens if we try to make T' go to infinity. So, let T' dash go to infinity, if T' dash goes to infinity you will see that omega naught dash goes to an infinitesimally small component which we will call d omega, fine.

And then the summation of a countable number of discrete frequency components that existed to construct to synthesis x t becomes for the case of x dash t and integration. So, what you have for x t is x t equals summation k equals minus infinity to infinity $x_k e^{jk\omega_0 t}$ this transforms to as T' dash tends to infinity an integral the

x_k 's. Now will be replaced by x dash k omega naught dash by T dash, we will make it go towards within the limit due to the $j k$ omega naught dash t , fine. Now, in the limit this becomes x dash t equals integral omega equals minus infinity to infinity, because essentially we were having discrete line frequencies k omega naught dash and omega naught dash is getting smaller, and smaller. And becoming d omega and x dash k omega naught essentially simply becomes x dash omega the continuous variable omega e to the $j k$ omega naught t simply becomes e to j omega t , and this 1 by T dash where T dash is tending to infinity simply becomes d omega by 2π . So, this is the expression for x dash t the synthesis expression for x dash t when x dash t becomes non periodic.

(Refer Slide Time: 40:45)

Handwritten notes on a whiteboard showing the derivation of the Fourier transform synthesis equation for non-periodic signals.

$$x'(t) \text{ with infinite period (non-periodic)}$$

$$x'(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X'(w) e^{jw t} dw$$

Fourier Transform Synthesis Equation
(Inverse Fourier Transform).

$$Tx_k = X(kw_0) = \int_{-T/2}^{T/2} x(t) e^{-jk w_0 t} dt$$

$$T'x'_k = X'(w) = \int_{-\infty}^{\infty} x'(t) e^{-jw t} dt$$

$$Tx_k = X'(kw_0)$$

So, to summarize we have x dash t with infinite period that is to say non periodic alright, it has infinite period and is given by this is nothing but the synthesis equation or the inverse Fourier transform equation. What we have just discovered is the Fourier transform though we have actually landed up directly with the inverse Fourier transform.

So, this is called the Fourier transform synthesis equation, because it synthesizes the function Fourier transform synthesis equation or the inverse Fourier transform, what about the analysis as long as we were dealing with the Fourier series for a periodic signal. We had discrete Fourier series coefficients x_k which when multiplied by the respective complex exponential e to the $j k$ omega naught t gave us, what we call the

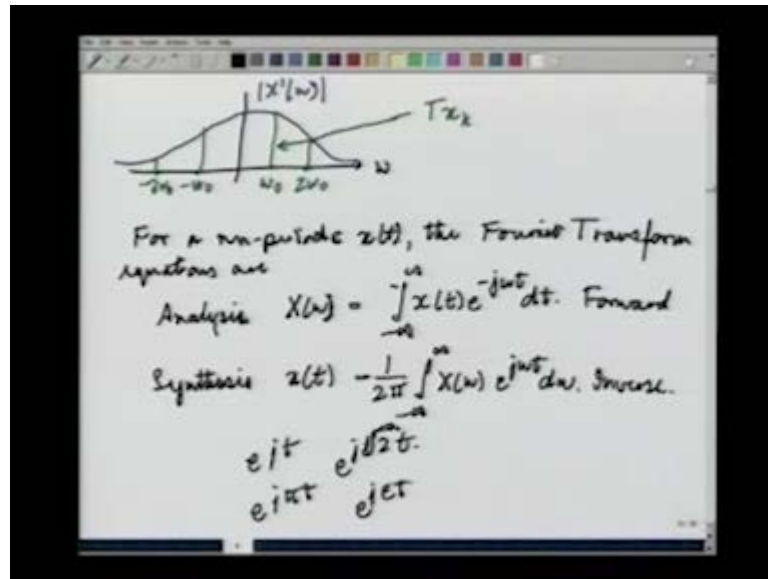
Fourier series components. So, the co-efficients over the x_k s the components for x_k multiplied by the respective complex exponential, this was the case for the Fourier series.

Now, from the equation of the synthesis we find that components are no longer at discrete positions alone, but they are everywhere for every value of ω , you have some components. And hence we have to ask the question what would be the form of the analysis equation here the analysis is nothing but evaluating x dash of ω for every value of ω , we want an expression for x dash of ω remember that t was taken over to the other side. So, we just had let us write the old equation first. So, that we remember what we had we had $t x_k$ which is nothing but x of $k \omega$ naught, and this was given by integral minus $T/2$ to $T/2$ $x(t) e^{-j k \omega t} dt$ this is what we had...

Now, T dash is standing to infinity and hence T dash x dash k , which is nothing but x dash of $k \omega$ naught will be equal to as T dash tends to infinity the limits of the integral also go to infinity. We are actually carrying out a co-relation as I said, but now a co-relation over infinite time of t going from minus infinity to t going to plus infinity x dash $t e^{-j k \omega t}$ now merges into the continuum ω t integrated with respect to t .

So, this is what we have this x dash $k \omega$ naught, if evaluated at frequencies $k \omega$ naught, you will get the same thing as this. And if evaluated for all values of ω , you will just get x dash of ω all we have done is to take the original $x(t)$ take just one period of the original $x(t)$. And instead of making identical copies of it to construct a periodic signal on both sides of the time axis, we have replaced all the periodic copies by 0 to realize a non periodic signal. And therefore, there is still reason to ask if there is any relation between the $T x_k$'s of the original $x(t)$ and the x dash ω of the non periodic modification of $x(t)$. So, we have $t x_k$ on the one hand side and what is this equal to this is equal to x dash of $k \omega$ naught, where ω naught is the fundamental frequency of the periodic signal. So, let us make a plot of this to make an idea.

(Refer Slide Time: 47:08)



We now have a continuous spectrum, and what am I actually trying to plot is against omega I am going to plot mod x dash omega. So, let us plot it in color maybe something like this, now how does this relate to the Fourier series terms, Fourier series components of the original x t, it is like this if the period was some t. So, that there was a fundamental frequency of omega naught, and let us say this was omega naught 2 omega naught minus omega naught, and minus 2 omega naught then the old spectrum which I am now plotting in green that is T x k's are in green then you will have I, then taking exactly these values at these points. So, this is the discrete lines spectrum in green, and how it relates to the continuous spectrum the infinitely resolved spectrum of x dash of t as shown in blue. So, let me let me just write that in blue, these are the lines spectral lines, and what we have I have here is mod x dash omega the periodic signal summarizing.

We now see how the periodic signal with its line spectrum has been transformed into the non periodic signal with the continuous spectrum, let us just put down the analysis, and synthesis equations for a non periodic x t the Fourier transform equations, equations are analysis x omega equals integral minus infinity to infinity x t. And infinite time co-relation minus integral minus j omega t d t, and the synthesis sorry x dash x of t given by one by 2 pi integral minus infinity to infinity, where omega is the variable of integration. Now, of x omega e to the j omega t d omega, this is what we have in the transform

domain, the inverse transform this is called the forward transform, and this is called the inverse transform this is called the inverse transform.

So, together we have completed the story of the construction of the Fourier transform in the form of evolving it by by a means of evolving it from the Fourier series, but then there are some interesting troubling questions to ask we had said, now before we ask the question ask those questions. Let us look at what we have. Now, we have a linear combination, if you look at the synthesis equation over here a linear combination though, now the linear combination is been in the form of a integral rather than a summation, we have a linear combination of what are still periodic complex exponential functions e to the $j \omega t$. And this linear combination of periodic complex exponentials is yielding at $x t$ which is non periodic, but how is that possible had we not already said at one time the sum of periodic functions will always be periodic.

Then how come we are adding e to the $j \omega t$ with weights $x \omega$, and generating a non periodic signal $x t$, how is it the addition of periodic signals is yielding a non periodic signal, this is something that we really should wonder about if that is we should wonder about, if we do not remember exactly what we said earlier we had not really said that the sum of 2 periodic functions is always periodic, we had only said that the sum of 2 periodic functions is periodic. If the periods of two functions being added are harmonically related here, that is clearly related not the case because e to the $j \omega t$ is a periodic function, but ω here can take all kinds of possible values; for example, you could be combining something like e to the $j t$ with e to the $j \sqrt{2} t$.

Now, in one case ω is equal to one and the other case ω equals $\sqrt{2}$, they are certainly not harmonically related. So, there is no problem such 2 signals, if they are added together will not yield a periodic signal or you could have e to the $j \pi t$, and you could have e to the $j e t$ where by e here, I mean the same base of the natural logarithm these two are again 2 irrational numbers, they are not harmonically related to each other their sum will not be a periodic signal. So, there is no inconsistency here that is all I wanted to point out.