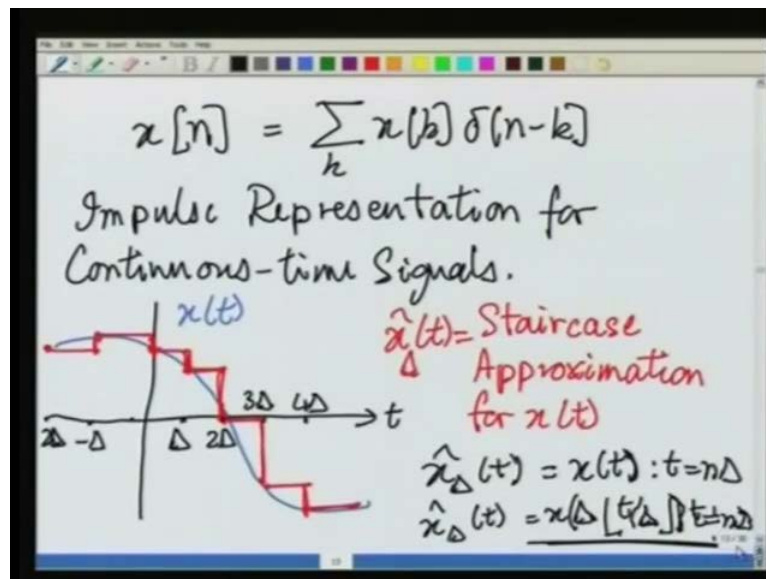


Signals and Systems
Prof. K. S. Venkatesh
Department of Electrical Engineering
Indian Institute of Technology, Kanpur

Lecture - 13
Representation of Continuous Time Convolution

(Refer Slide Time: 00:22)



Our discussion of the continuous time case. So, in order to find an impulse representation for continuous signals; continuous time, we first need to find out, how to go about it. Let us try a crude approach, the crude approach would be to get a very coarse representation, instead of an accurate representation very coarse approximate representation for a continuous time signal to begin with. So, in order to do this, let us first take some continuous time signal, a graph of a continuous time signal and see what we can do in it. So, here it goes this is the time axis and on this let sketch some continuous time signal $x(t)$, let this be $x(t)$.

Now, we want to representation for this, the first thing I will try to do is to form, what I call as staircase approximation for $x(t)$. Let us call this signal $\hat{x}_\Delta(t)$ the idea of a staircase representation is that $\hat{x}_\Delta(t)$ does not try to be equal to $x(t)$ every point t , instead at periodic intervals at periodic instance of time. Namely, t equal to 0, t equal to Δ , t equal to 2Δ and so on. $\hat{x}_\Delta(t)$ will be equal to $x(t)$ at other points between this point of equality $\hat{x}_\Delta(t)$ will be constant at the most recent value. So,

in order to do this, let me first mark out on the time axis. This is the time axis on the time axis let me just mark out first the points Δ , 2Δ and soon. We will also have of course, $-\Delta$, -2Δ and soon.

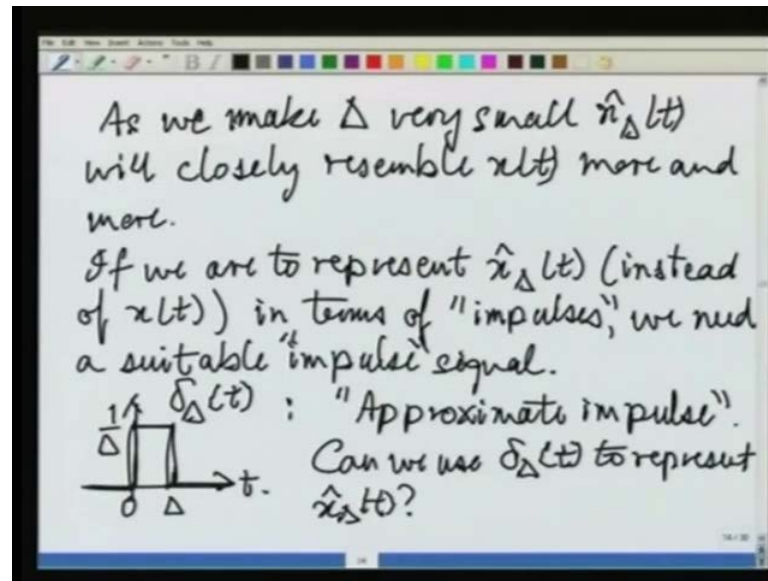
Now, let us start what does this $\hat{x}(t)$ do at $-\Delta$? $\hat{x}(t)$ will try match $x(t)$ exactly. So, at t equal to $-\Delta$, this will be a point on a $\hat{x}(t)$, what is this value of $-\Delta$? It is value at $-\Delta$ against to be equal to $x(t)$ in between these two points $\hat{x}(t)$ should remain constant at the value, it found at x at t equal to $-\Delta$. So, it stays constant in between this two points.

Now, at t equal to 0 $\hat{x}(t)$ will have to take the value of x of 0 , $\hat{x}(0)$ has to be equal to x of 0 , but between t equal to $-\Delta$ and t equal to 0 , it will step out at the value \hat{x} of $-\Delta$ or x of $-\Delta$. So, it will remain constant like this. Next, at t equal to Δ we have this value and from there up to this point it will have this value. Likewise if I go on doing this I would get the this is the staircase case, just for psychological convinces I will have rather for the staircase case as well I connect these parts of the staircase case by vertical line knowing fully well that that is found upon my the purest. So, this red function is the staircase case approximation $\hat{x}(t)$ of $x(t)$.

Now, because it has discretized the continuous time using a unit that we have called Δ I will actually, call this $\hat{x}_{\Delta}(t)$ and what can I say about the value of $\hat{x}_{\Delta}(t)$. I can say the following $\hat{x}_{\Delta}(t)$ equals $x(t)$ at t equal to $n\Delta$ at other points its value will be equal to the most recent value of $x(t)$, most recent sampled value of $x(t)$. So, thus \hat{x}_{Δ} at 2Δ will be equal to x of 0 , \hat{x}_{Δ} at 3.2Δ will be equal to x of $till$ and soon.

So, we have this staircase and at points not equal to $n\Delta$ at point of time not equal to $n\Delta$ we will get $\hat{x}_{\Delta}(t)$ equals x of Δ times the floor of t by Δ , as we have written over here. This completely describes the staircase case approximation in term of the original signal, it is clearly an approximation. Now, the important things to recognize the about this approximation is that.

(Refer Slide Time: 08:40)



As we make Δ very small $\hat{x}_\Delta(t)$ will closely resemble $x(t)$ more and more. So, though at present and in the kind of sketches we have drawn $\hat{x}_\Delta(t)$ looks very crude, $\hat{x}_\Delta(t)$ looks very crude compare to the original signal. This can be improve if you only make Δ smaller and smaller, but that as separate issue right now. Suppose, we are prepared to except $\hat{x}_\Delta(t)$ in mu of $x(t)$ $\hat{x}_\Delta(t)$ in mu, mu of $x(t)$ as the signal that we want to study.

Then can be find something like an impulse representation for $\hat{x}_\Delta(t)$ the nice thing about $\hat{x}_\Delta(t)$ is that it has only finitely, many different values between within a finite interval of the time axis. If you go back to the earlier figure, we find that in this interval of the time axis from minus 2Δ to approximately 4Δ , we have 1, 2, 3, 4, 5, 6 different values of $\hat{x}_\Delta(t)$ unlike the original $x(t)$, which had a infinite number of different values in this same interval.

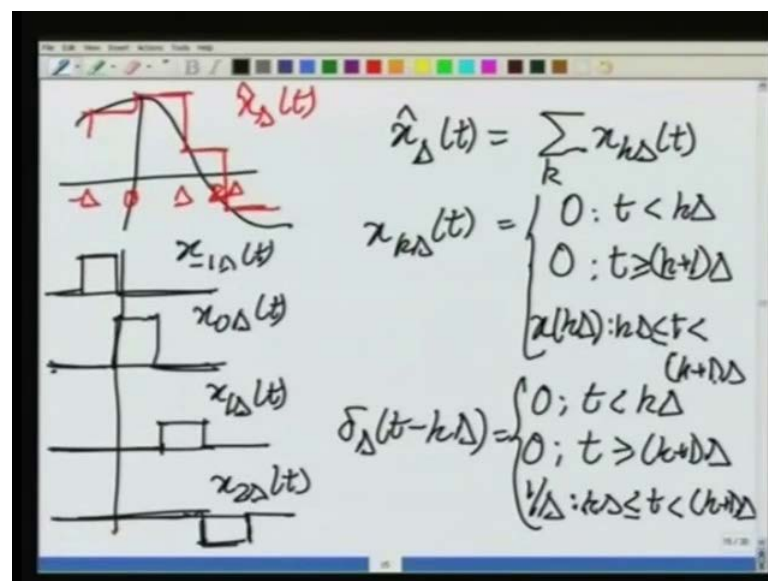
So, in some sense we have reduced the problem of continuous time signal representation to something that resembles the discrete time signal representation. So, let us say if we are ask to represent $\hat{x}_\Delta(t)$ using impulses, represent $\hat{x}_\Delta(t)$ instead of $x(t)$ in terms of impulses. We still do not know what we mean by impulse, but let us see. How would be go about it? We would try to defined a suitable impulses.

Impulse signal in this case it turn out to be convent to define that signal as a function sketched here, this is what I will call for lake of a better name I will call $\delta_\Delta(t)$

capital delta of t, this is the time axis. This signal we will have a non zero support that is a support from 0 to delta, it is 0 outside this interval and within this interval it is constant and take on a value equal to 1 by delta. This is what I call delta delta of t, I will call this now for again lack of better name, I will call it the approximate impulse. Delta delta of t is called the approximate impulse can I use delta delta of t to represent x delta of t, x hat delta of t.

The answer is yes and we could do it by decomposing x hat delta of t into x hat delta k of t, as we did last time. Each of this x hat delta k of t would be non zero only in the interval k delta to k plus 1 delta and in this interval it would that the value of x t at t equal to k delta that is it would be equal x of k delta. It would be constant in that interval at all other end points of time it would be 0.

(Refer Slide Time: 14:41)



So, suppose we had X_t as we are taken earlier then here is our $\hat{x}_\Delta(t)$. Now, I will represent this as some of shifted and scaled approximate impulses, but before I do that I will first explicit as some of simpler function, these simpler functions would be like this these is minus delta 0, delta 2 delta so on. I have this, so I would express this as some of the following signals. In this place I would just have 1 pulse like this, this I will call $x_{-1\Delta}(t)$. Then I would have the next, function $x_{0\Delta}(t)$. Then the next function $x_{2\Delta}(t)$ sorry $x_{1\Delta}(t)$, then the next member would be 0 up to here and then being negative this would be $x_{2\Delta}(t)$.

So, if you added $x_1 \Delta t$ and $x_2 \Delta t$ plus $x_3 \Delta t$ plus so on, you would get $\hat{x} \Delta t$. So, we can write $\hat{x} \Delta t$ equals summation $x_k \Delta t$ for different case so far so good. We seem to have completely side track the original problem of finding a representation for $x(t)$, we have now become completely preoccupied with finding a representation for $\hat{x} \Delta t$. We have set our height a little lower we have recognize that handling $x(t)$ directly, would be biting of more than we can choose. So, we have biting of only $\hat{x} \Delta t$ for time being and we want to represent this.

That representation can be made in the following manner. Now, each $x_k \Delta t$ can now be represented in term function Δt . How do we do it? Shift them and scale them and add them appropriately the game is very similar, what was done in the discrete time case. Let us try to characterized the $x_k \Delta t$, what can you say about $x_k \Delta t$ $x_k \Delta t$ equals 0 for t less than $k \Delta t$ it is also equal to 0, for t greater than equal to $k \Delta t$ plus Δt . If non zero only between these two points and what value is it between these two points, it is equal to x the original signal x at $k \Delta t$ in the interval $k \Delta t$ less than t less than equal to t less than $k \Delta t$ plus Δt . This is our characterization of $x_k \Delta t$ contrast, this with our characterization of Δt shifted to the point $k \Delta t$.

So, we want Δt minus $k \Delta t$ Δt minus $k \Delta t$ equals 0 t less than $k \Delta t$ 0 t greater than equal to $k \Delta t$ plus Δt . It is equal to 1 by Δt in the interval $k \Delta t$ less than equal to t less than $k \Delta t$ plus Δt . So, this is Δt minus $k \Delta t$ that is the Δt of x and the approximate impulse shifted to the point t equal to $k \Delta t$. This is the component of the stare case approximation the k th component of this stare case approximation $\hat{x} \Delta t$, this k th component is what we have called $x_k \Delta t$.

So, you will see that both $x_k \Delta t$ and Δt minus $k \Delta t$ are constant in the interval from $k \Delta t$ to $k \Delta t$ plus Δt . One of them is equal to x of $k \Delta t$ the other is equal to 1 by Δt . So, all we need to do is to find the appropriate scale factor and we would have to being with a means of representing the different component, simpler component signals $x_k \Delta t$ in terms of the approximate impulse. So, let us do this now [FL] we have.

(Refer Slide Time: 23:09)

$$\begin{aligned}
 & \left. \begin{aligned} x_{k\Delta}(t) &= x(k\Delta) \\ \delta_{\Delta}(t-k\Delta) &= 1/\Delta \end{aligned} \right\} k\Delta \leq t < (k+1)\Delta \\
 & x_{k\Delta}(t) = x(k\Delta) \delta_{\Delta}(t-k\Delta) \Delta \\
 & \underline{\hat{x}_{\Delta}(t) = \sum_{k=-\infty}^{\infty} x_{k\Delta}(t) = \sum_{k=-\infty}^{\infty} x(k\Delta) \delta_{\Delta}(t-k\Delta) \Delta}
 \end{aligned}$$

$x(k\Delta)$ of t in the interval of our interest, this is equal to x of $k\Delta$ and we have $\delta_{\Delta}(t - k\Delta)$. Assuming a value equal to $1/\Delta$ both this in the region $k\Delta \leq t < (k+1)\Delta$. So, clearly by comparing this two and both these functions are 0 outside this interval of $k\Delta$ to $(k+1)\Delta$. So, in order to make $\delta_{\Delta}(t - k\Delta)$, in order to express $x(k\Delta)$ of t in terms of $\delta_{\Delta}(t - k\Delta)$, all we have to do is scale $\delta_{\Delta}(t - k\Delta)$ by $x(k\Delta)$ which is x of $k\Delta$.

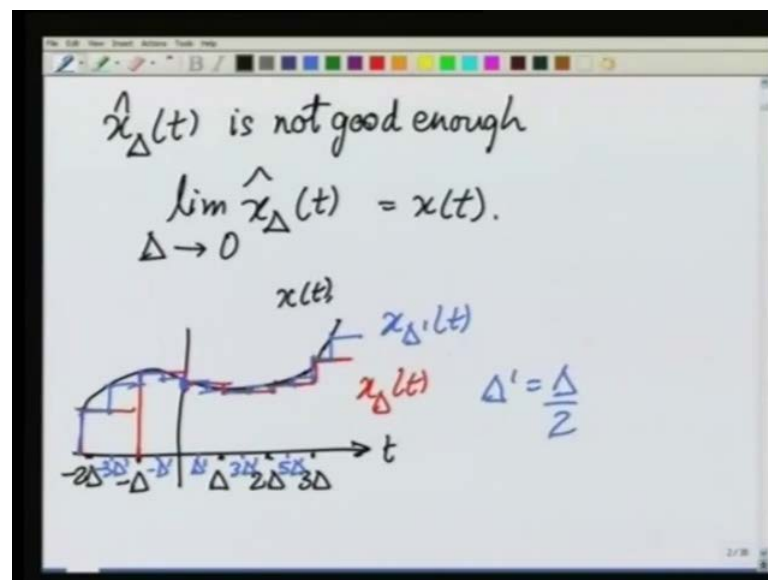
We can therefore, write that $x(k\Delta)$ of t equals $x(k\Delta) \delta_{\Delta}(t - k\Delta)$ this is almost okay, except for this one deficiency that $\delta_{\Delta}(t - k\Delta)$ has amplitude of $1/\Delta$ not of unity. Therefore, we have to further multiply this by Δ this gives us the expression for $x(k\Delta)$ of t , in terms of our approximate impulse $\delta_{\Delta}(t - k\Delta)$, the shifted scaled approximate impulse. This is the equivalent to our earlier expression where we had $x(k\Delta)$ in terms of $\delta(k\Delta - n)$ times x of k . So, with this in hand, it is very simple to what we do next.

Remember that our original exercise is to represent $x(t)$, but our intermediate exercise is to represent $\hat{x}_{\Delta}(t)$, which is the staircase approximation of $x(t)$. So, the staircase approximation is what we will now construct using shifted, and scaled approximate impulses. For all k that is running from minus infinite to infinite, this is what we know is

valid. Now, with our new information we can express this as k equals minus infinite to infinite, $x[k\Delta] \Delta$ minus $k\Delta$ times Δ .

So, this is the expression we were looking for. If our exercise was only to represent this staircase approximation then we have already achieved our objectives, we have now the means of representing $\hat{x}_\Delta(t)$ in terms of shifted and scaled impulses, but we have still got some distance should go from here because we actually, want to express $x(t)$ in terms of $\hat{x}_\Delta(t)$.

(Refer Slide Time: 27:19)



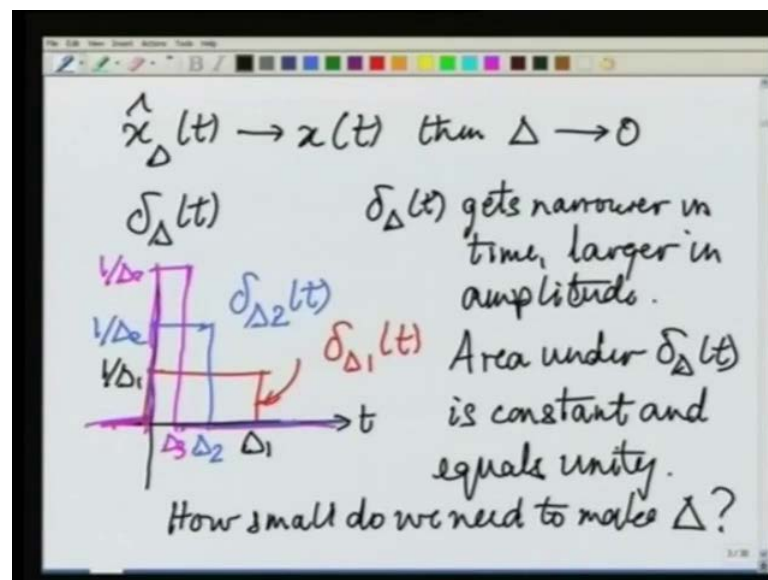
So, $\hat{x}_\Delta(t)$ is not good enough we have to see how to make $\hat{x}_\Delta(t)$ become close to or equal to $x(t)$. That is reasonable easy to see its evident that as these we discretion size Δ , discretion unit Δ tense to 0 $\hat{x}_\Delta(t)$ will become more and more close to $x(t)$. Thus we can write that the limit as Δ tense to 0 of $\hat{x}_\Delta(t)$ equals $x(t)$. Now, when we do this we will have a larger and larger number of the component signal $x[k\Delta]$ because in any given interval, we will have a larger number of units as Δ tense to 0 because Δ is the unit. Thus we will have more and more crowded representation plus try to make a diagrammatic exhibition of this fact.

Suppose, this is x and originally Δ is this big 2Δ , 3Δ minus Δ then the first staircase approximation would be like this, but we have sketched here the same signal $x(t)$ in black then there is a signal sketched in blue in red and that is $\hat{x}_\Delta(t)$. There is a third signal sketched in blue which is $\hat{x}_{\Delta'}(t)$, we have chosen Δ

prime to be equal to delta by 2 half delta. We can see from the horizontal time axis what relationship delta prime bears to delta.

We can see that delta prime is half of delta we have mark 3 delta, 5 delta, 3 delta prime 5 delta prime here in this axis. We have constructed this two stare case approximation x delta prime of t and x delta of t, x delta of t is seen to be more distanced from x of t then x delta prime of t. This is achieved simply because x delta prime of t checks the values of x t every time k delta prime, which is as smaller time interval then k delta. Thus it evident that as delta tense to 0 x delta of t will tend x delta hat of t will tend x t. So, let see what this n tells, if you wish to make x hat.

(Refer Slide Time: 31:47)



Delta of t tends to x of t then delta must tend to 0, let us see the consequence of the delta tending to 0. If we want every step of our derivation made until now to be valid then we should make appropriate definitions for the approximate impulse, the approximate impulse was earlier define as delta delta of t. Now, this approximate impulse is going to undergo modifications of a suitable kind, when capital delta the discretion step size tends to 0. Thus for example if you make a few versions of delta delta of t for varying values of delta you would get a diagram like this.

For large delta we would get something like this and we would say that this is let say delta 1. If this is delta 1 then this is 1 by delta 1 and we have this diagram. So, this is one choice of delta in which case, this function that we are just drawn will be what you can

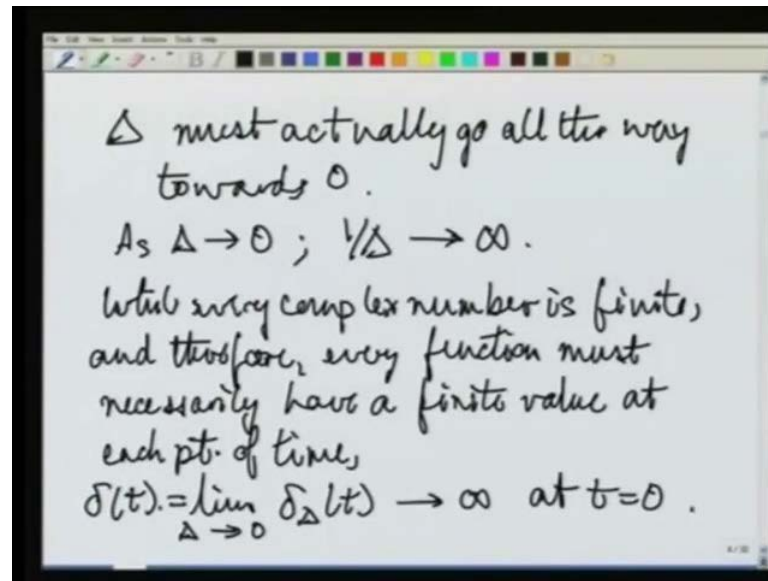
call δ_1 of t . If you make δ smaller let us make δ_2 that is half the size of δ_1 , then the height of this new function in its non zero region in this support must be equal to $1/\delta_2$ which is therefore, twice of $1/\delta_1$.

So, we get a new function that has this shape this is δ_2 and what we just have drawn is δ_1 of t we could make a third one if you like which had an even smaller support the δ_3 . Then we would have an even taller function, because $1/\delta_3$ would be even smaller than $1/\delta_1$ or $1/\delta_2$ would be even larger than $1/\delta_1$ or $1/\delta_2$ we would get this is not actually draw to scale, but this would be $1/\delta_3$ or to be $1/\delta_3$ and what we just drawn earlier is $1/\delta_2$.

So, if we remain consist with the way we have defined our approximate impulse. Then it turns out that as the discretion step get smaller and smaller, these approximate impulses get narrower in time and taller in amplitude. It gets undergoes this two changes in a very systematic manner and that systematic fact is that if its width is δ . Then this amplitude is $1/\delta$. So, that the area under δ of t is constant and equals unity. So, we have to maintain the area of δ of t to be 1 as the discretion step is made smaller and smaller.

So, finally the question is how small do you need to make δ , the answer unfortunately is that no finite size is small enough because we want to approximate a continuous time function $x(t)$ and a continuous time function uses time as a real variable. Now, a real variable the real line is infinitely divisible. So, theoretically $x(t)$ can vary from instant to instant, you can take as smaller interval of $x(t)$ as you like and still observe variation in it. So, if want $\hat{x}_\delta(t)$ to be faithful to $x(t)$ at whatever level of magnification, we wish to observe a $\hat{x}_\delta(t)$ in. Then it is imperative that δ must actually go all the way to 0.

(Refer Slide Time: 37:58)



Only then can we hope that the staircase approximation $\hat{x}_{\Delta}(t)$ will become actually, equal to $x(t)$. So, this is what we have to aim for. Now, let us see what happens to the approximate impulses as the unit of the discretization goes to 0. The area as we said must remain fixed at unity, the width of the support gets smaller and smaller and tends to 0. If both these constraints had to be made then necessarily, the amplitude $1/\Delta$ should tend to infinity.

So, as Δ tends to 0, $1/\Delta$ will tend to infinity there is no help for us. So, what do we have in the limit? What you have in the limit is an acquired mathematical entity, it has in many ways it lacks the basic properties of a function, because a function as we have understood till now is a map from the real line (time axis) into the range, which we expect to be the set of complex numbers. Every complex number in the set of complex numbers is finite. However, $1/\Delta$ as we observed here is infinite.

So, while every complex number is finite and therefore, every function must necessarily have a finite value at each point of time $\hat{x}_{\Delta}(t)$ will not satisfy these as Δ tends to 0. So, $\hat{x}_{\Delta}(t)$ as Δ tends to 0 of $\hat{x}_{\Delta}(t)$ will tend to infinity at $t=0$. This is a paradoxical situation and this is why $\hat{x}_{\Delta}(t)$ in the limit, which we call in fact we have a name for it we simply call it $\delta(t)$. We define the limit of this sequence of functions of $\hat{x}_{\Delta}(t)$ as capital

delta tends to 0 as an object that we call delta of t. So, let us say what delta of t, what you can say about delta of t.

(Refer Slide Time: 42:17)

Handwritten notes on a digital whiteboard defining the Dirac delta function:

$$\delta(t) = 0 : t \neq 0.$$

$$= ? : t = 0.$$

$$\int_{-\infty}^{\infty} \delta(t) dt = 1 = \int_{-a}^a \delta(t) dt.$$

$\delta(t)$: impulse Dirac delta. $a > 0$

$$x_{b\Delta}(t) = \delta_{\Delta}(t - h\Delta) \cdot x(h\Delta) \Delta.$$

Delta of t is the limit of this sequence of delta delta of t and hence, delta of t equals 0 t naught equal to 0 this much we can say, because this is a common property of all the delta delta of t outside there are support and the support of delta of t which is the limit of that sequence is in this the 0, it has the width of the support of delta of t is actually 0. So, there is no problem it will be 0 all t naught equal to 0. Now, what value will it have at t equal to 0 we can't answer this question because the value it will take is not within the set of complex numbers, infinite is not within the set.

This value that is tense to take will be very, very large. However, we can this seize to discuss the property, this particular property of what value it takes at t equal to 0 and instant focus on more secure things, things we are more comfortable with the things that we are more confident about. One thing we are very confident about of delta of t is the area under the curve, the area under the delta of t must be same as the area under every number of the sequence because we have set that area constant and equal to unity. Thus we can say that the area elevated of delta of t d t equals 1.

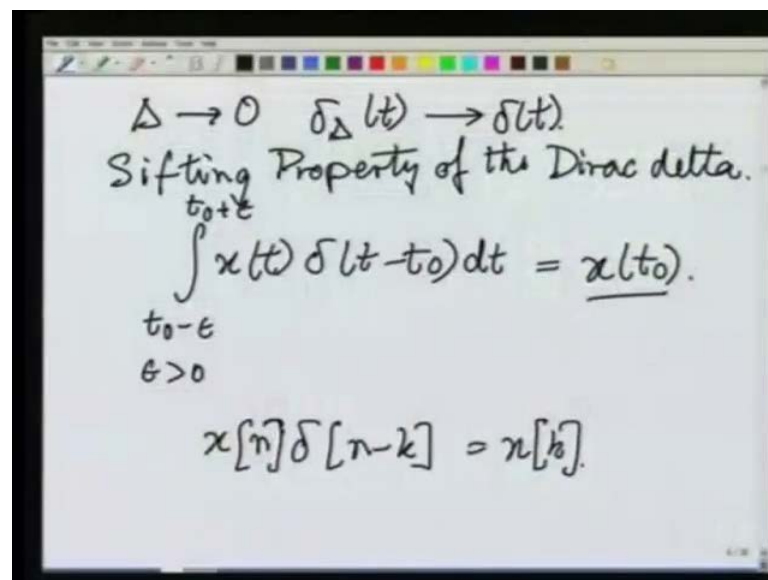
Since, its value is 0 for all t naught equal to we could take any interval containing 0, and integrate delta t over that interval we would still get 0. In short this is equal to integral from any minus a to a where a is greater than 0, delta of t d t. This is the same it will

have a area of unity. This delta this delta of t now called the impulse, simply called the impulse because it is no longer approximate it is the ultimate impulse, it is what we were hoping to construct.

Impulse for continuous time signal is just called the continuous time impulse it is also called the Dirac delta. So, all that we have said on the panel is the only if statement that we can make with certain t about the Dirac delta. Something we can say is that the Dirac delta is not a function because it is not possessed values within the set of complex number for all time, in particular at t equal to 0, where the value that the Dirac delta takes does not lay in the set of complex numbers.

So, it is not a function it is an entity that is all we will say it is right now, with this property that area under entity in the normal way we defined the area under any function evaluates to 1. It will evaluate one is that integral if carried out over any interval that contains t equal to 0. We can also say with certain t that delta of t is 0 for all points for all points of time not equal to 0. So, this much we have said there is one important property that delta of t will have.

(Refer Slide Time: 47:50)



Handwritten notes on a whiteboard showing the sifting property of the Dirac delta function:

$$\Delta \rightarrow 0 \quad \delta_{\Delta}(t) \rightarrow \delta(t)$$

Sifting Property of the Dirac delta.

$$\int_{t_0-\epsilon}^{t_0+\epsilon} x(t) \delta(t-t_0) dt = \underline{x(t_0)}$$

$\epsilon > 0$

$$x[n] \delta[n-k] = x[k]$$

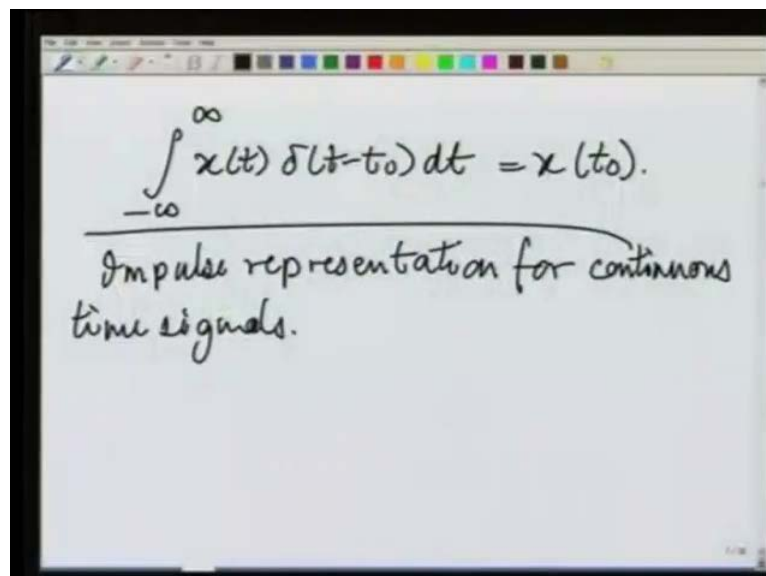
Remember that in our process in our exercise of coming to delta of t we have gone through the representation of x k delta of t, and we said that x k delta of t is just equal to delta delta of t minus k delta times x of k delta times delta. So, if we extend this property all the way until delta tense to 0, what we will get? We have essentially, here multiplied

the value of $x(t)$ at t equal to $k\Delta$, we have we have multiplied that with the approximate impulse and scaled it by Δ and this gives the value of $x(t)$ at t equal to $k\Delta$. So, as we make Δ tense to 0.

Δ of t as now become $\delta(t)$ the exact impulse or the ultimate impulse and $\delta(t)$ will still, we used to sample the value of $f(t)$ not in gross intervals of width Δ , but at a particular point. In short there is something called the shifting property of the Dirac delta, the shifting property of the Dirac delta is simply this. If I try to take any function $x(t)$ multiply it with a shifted Dirac delta located at some t_0 that is $\delta(t - t_0)$ and evaluate this product.

The integral of this product over any interval of time that includes the point of occurrence of the impulse, which is now t_0 say, $t_0 - \epsilon$ to $t_0 + \epsilon$, where ϵ is of course, greater than 0, then this will be equal to $x(t_0)$ it is this property that was what, that was that that we saw at when we wanted an impulse representation for continuous time signals. In the discrete case we wrote that $x[k]\delta[n - k]$ or rather $x[n]\delta[n - k]$ just one second, $x[n]\delta[n - k]$ was equal to $x[k]$ that is the equivalent property we want here.

(Refer Slide Time: 51:19)



$$\int_{-\infty}^{\infty} x(t) \delta(t - t_0) dt = x(t_0).$$

Impulse representation for continuous time signals.

There we were able to sample the value of the function discrete function $x[n]$ discrete time function $x[n]$ at the point k by multiplying it $x[n]\delta[n - k]$ which is the shifted discrete chronicle impulse at n equal to k . Here we have the Dirac impulse located at t

equal to t naught and we wish to evaluate $x(t)$ naught, we wish to express the number $x(t)$ naught using the shifted Dirac impulse to sample $x(t)$. So, we have this. So, in general we always write.

The shifting property as integral minus infinity to infinity, we can always use a larger interval because the Dirac impulse is anyway 0. For any point outside the point of its occurrence $x(t - \delta(t - t_{\text{naught}}))$ equals $x(t)$ naught, this completes our impulse representation for continuous time signals. So, we have an impulse representation. Now, what we do next we have to carry out the same set of steps, that we used in order to obtain the convolution of discrete time signals.

We apply homogeneity, we apply time invariance then we apply the property of linearity of the system in question. Then we find that $y(t)$ the output of a continuous time system, which as have been expressed, which has been impressed with an input $x(t)$ may be computed using something that is also called the impulse response here of the continuous time system.