

Advanced Digital Signal Processing –Wavelets and Multirate
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Lecture No. # 34
Constructing the lattice And its Variants.

A warm welcome to the 34th lecture on the subjects of wavelets and multirate digital signal processing this lecture intends to build upon the structure that we started discussing in the previous lecture let us recall in a few words what we are trying to do in the previous lecture we are trying to bring a modular structure to realize the orthogonal analysis filter bank.

We noted that in building a modular structure we would have an inductive approach and an approach where we would construct the smallest order orthogonal filter of length 2 and we demonstrated how the Haar analysis filter bank can be. So, construct it and then to expand the length by 2 every time we would need to introduce an expansion module an inductive module we introduce that inductive module and we prove by mathematical induction that you get an orthogonal filter bank of length 2 more every time we introduce 1 instance of that module.

Now, this was constructive in the sense that we assumed that we were drawing a structure of this kind we were putting together modules of this kind and we noted that we had the conjugate quadrature filter relationship or relationship in which the orthogonal filter bank conditions on the analysis filter side were obeyed, but what we wanted to do was to go the other way.

Given an orthogonal filter bank we needed to construct such a lattice structure which could realize the orthogonal filter bank. So, for example, if you had a Daubechies filter of length four you would want to construct a lattice structure to realize it and it was that direction that we wrote down the recursions that governs the modular structure and you are trying to use those recursions to go 1 step back to peel off 1 layer as it were let us build on that peeling of process further today.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

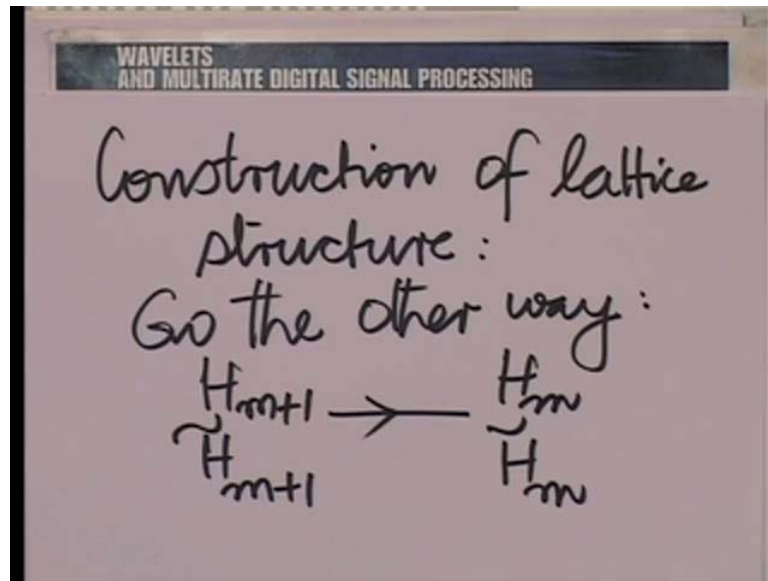
$$\tilde{H}_{m+1}(z) = \frac{-(z^{2(m+1)} - 1)}{z} H_{m+1}(-z^{-1})$$

Conjugate Quadrature relation is carried

Now, what we said was something like this we said you have 1 modular stage like this here you had the system functions $H_m(z)$ come up to here $\tilde{H}_{m+1}(z)$ come up to there and assuming that we had taken the down sampler well beyond the down sampler is at the end after all the modules what appears next here is as z to the power minus 2 following that we have the m plus one lattice coefficient as it were which we called k_{m+1} here and the negative of the same coefficients here minus k_{m+1} and this brought as to the next system function which we called $H_{m+1}(z)$ and $\tilde{H}_{m+1}(z)$ where we noted that the conjugate quadrature relation between H_m and \tilde{H}_m is preserved in H_{m+1} and \tilde{H}_{m+1} what we mean by that is given that $\tilde{H}_m(z)$ is $z^{2m-1} H_m^{-1}(z)$ we have $\tilde{H}_{m+1}(z)$ is indeed the $z^{2m+1-1} H_{m+1}^{-1}(z)$. So, the conjugate quadrature relation is carried it is carried pass the module.

That is what we are saying now what we want to do is to go the other way.

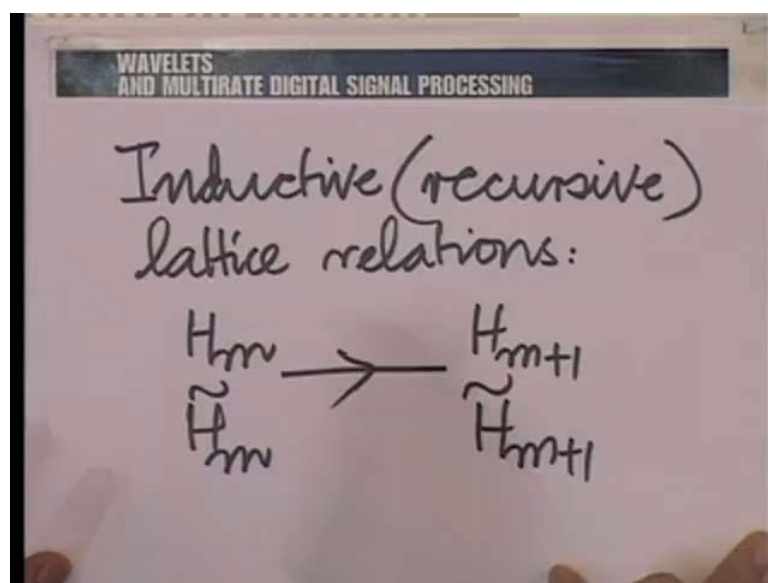
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So, the construction of a lattice structure essentially means go the other way from H_{m+1} and \tilde{H}_{m+1} to H_m and \tilde{H}_m now towards the objective we need to write down the recursive expression the inductive expression again that would give us a clue how to go back words.

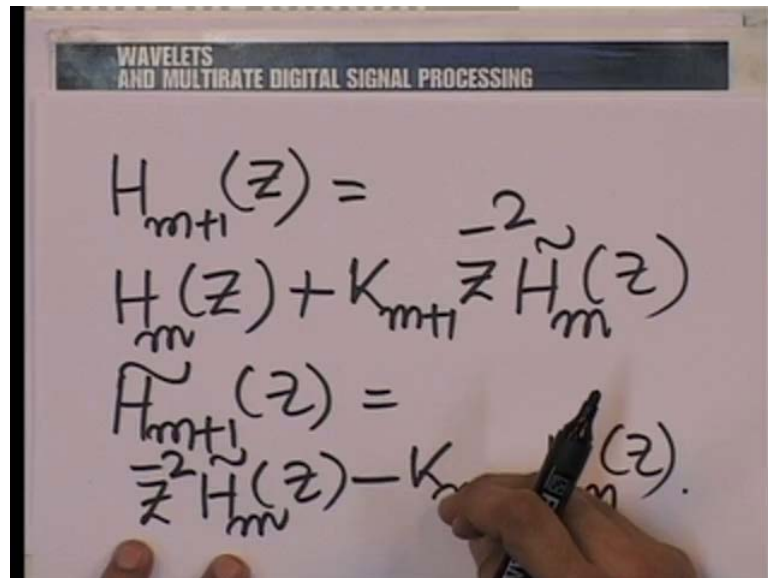
So, let us write down the inductive relations once again inductive or recursive lattice relations.

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Take you from H_m and \tilde{H}_m to H_{m+1} and \tilde{H}_{m+1} day and the relations are $H_{m+1} z$ is $H_m Z$ plus $K_{m+1} Z$ to the power minus 2 $\tilde{H}_m Z$ and $\tilde{H}_{m+1} Z$ a Z raise the power minus 2 $\tilde{H}_m Z$ minus $k_{m+1} H_m Z$.

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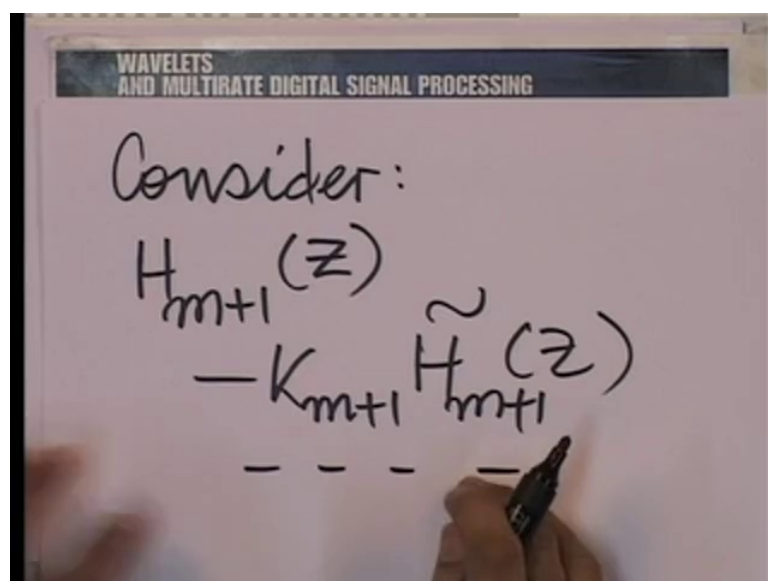
WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

$$H_{m+1}(z) = H_m(z) + K_{m+1} z^{-2} \tilde{H}_m(z)$$

$$\tilde{H}_{m+1}(z) = z^{-2} \tilde{H}_m(z) - K_{m+1} H_m(z)$$

Now, if you observe these 2 relations carefully and we wish to extract H_m in terms of H_{m+1} and \tilde{H}_{m+1} all that we need to do as we see is to cancel out this term and that is easily done by multiplying this expression by minus K_{m+1} .

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

Consider:

$$H_{m+1}(z) - K_{m+1} \tilde{H}_{m+1}(z)$$

If we multiply this equation by minus K_{m+1} and added to this **this** term would vanish what you are saying in a effect is consider $H_{m+1} Z^{-K_{m+1}} H_{m+1}$.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

Equal to:

$$H_m(z) + K_{m+1}^2 H_m(z) + K_{m+1} Z^{-2} H_m(z) - K_{m+1} Z^{-2} H_m(z)$$

Tilde Z very clearly this would be equal to $H_m Z$ plus K_{m+1} squared $H_m Z$ plus $K_{m+1} Z$ raise to the minus 2 H_m tilde a Z minus $K_{m+1} Z$ raise the power minus 2 H_m tilde Z and which of course, very easily seen to be $1 + K_{m+1}$ squared $H_m Z$. So, we are eliminated H_m tilde.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

$$= (1 + K_{m+1}^2) H_m(z)$$

(eliminated)

So, we have an easy way now of obtaining H_m from H_{m+1} and \tilde{H}_{m+1} now you know what we are doing here in terms of real computations or actual realization starting from a higher order filter and going 1 step lower in the lattice and what we just shown is if you know the higher order filter you know of course, it is conjugate quadrature filter once you know H_{m+1} it is easy to construct \tilde{H}_{m+1} all that you do is to replace Z by Z^{-1} in the argument and then multiply by Z raise the power are sufficient negative power. So, as to make it (()) we shall illustrate this specifically for length four in short while, but let us get the algebra complete first.

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The image shows a slide with the title "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING". The handwritten equation on the slide is:

$$H_m(z) = \frac{H_{m+1}(z) - K_{m+1} \tilde{H}_{m+1}(z)}{1 + K_{m+1}^2}$$

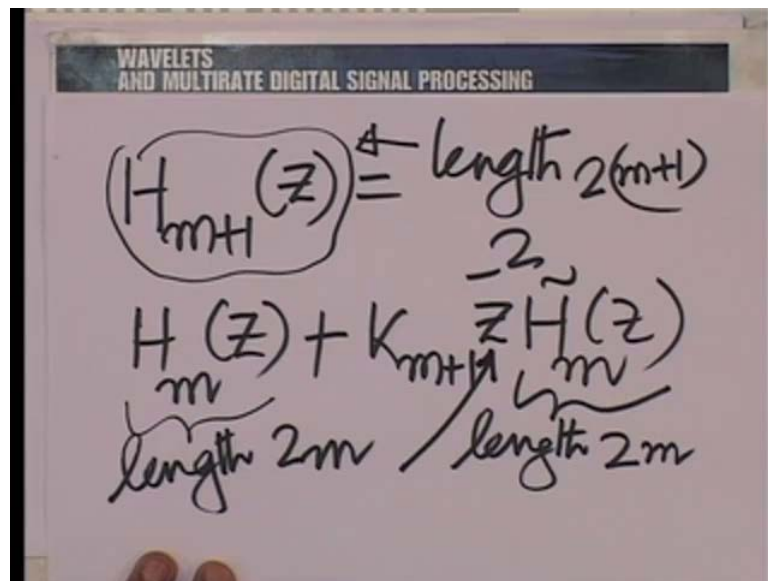
So, what we are saying in a effect is $H_m Z$ is $H_{m+1} Z$ minus $K_{m+1} \tilde{H}_{m+1} Z$ divided by $1 + K_{m+1}^2$ now let us look at the validity of an expression like this validity means is this expression computable does it make sense now of course, this definitely make sense there is no problem, but we do not know what K_{m+1} we need to reason that out and that will also require little bit of reasoning. So, if I new K_{m+1} this is. In fact, very easy to compute and.

So, is the denominator then. So, all that we have to do is to reason out have to find K_{m+1} and. In fact, that we shall do by looking back at the forward recursion once again, but you know as for this denominator is concerned here the validity of division by $1 + K_{m+1}^2$ cannot be questioned as long as K_{m+1} is real if K_{m+1} is

real we have no problem of validity at all because this is definitely going to be a non negative strictly non 0 quantity.

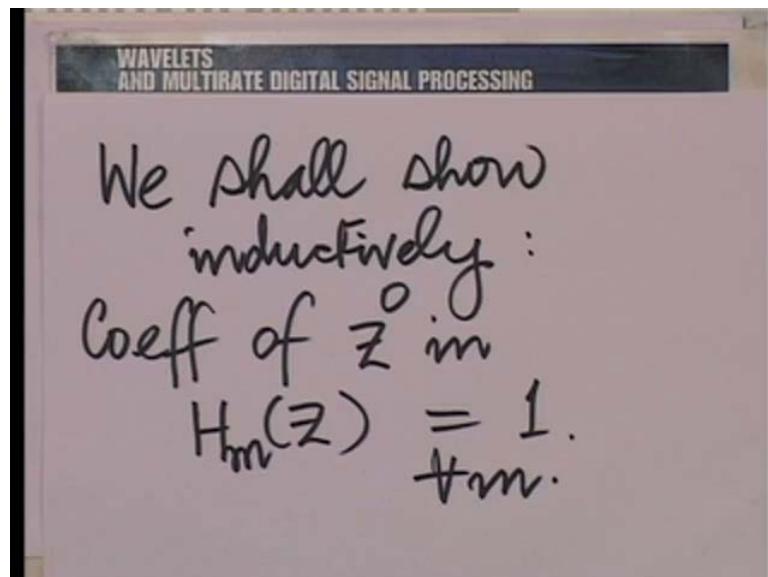
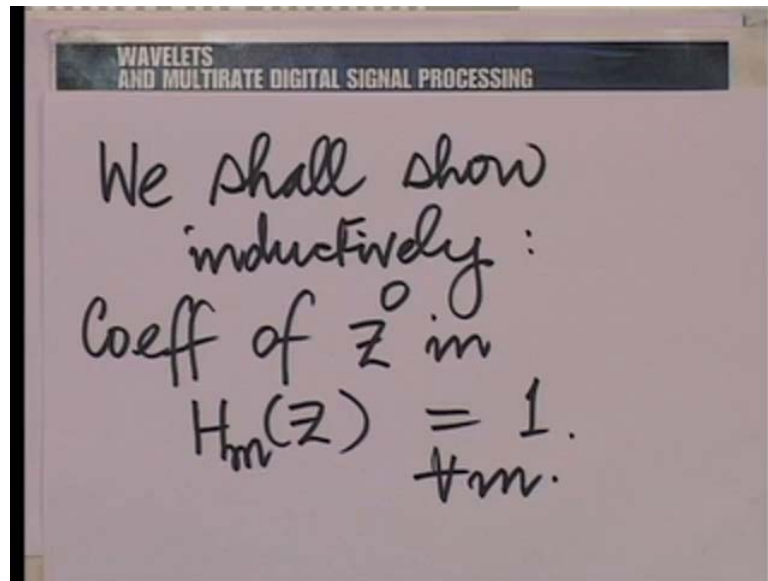
So, all that we need to do obtain K_{m+1} how and that is done very easily by looking back at the forward recursion. In fact, here again will need to do a bit of inductive reasoning.

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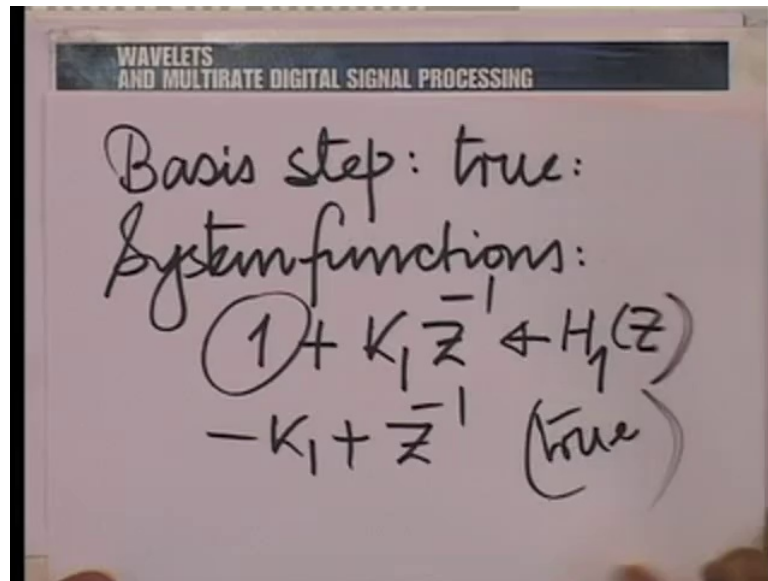
So, let us write down the main step of forward recursion once again $H_{m+1} z$ is $H_m Z + K_{m+1} Z^{-2} H_m$ now notice that H_{m+1} is going to be a length $2(m+1)$ and these are going to be of length $2m$. So, this factor Z^{-2} is going to push this length $2m$ filter 2 steps forward and that is how you are going to get a length of $2(m+1)$ you know this pushing forward by 2 is what increases the length by 2.

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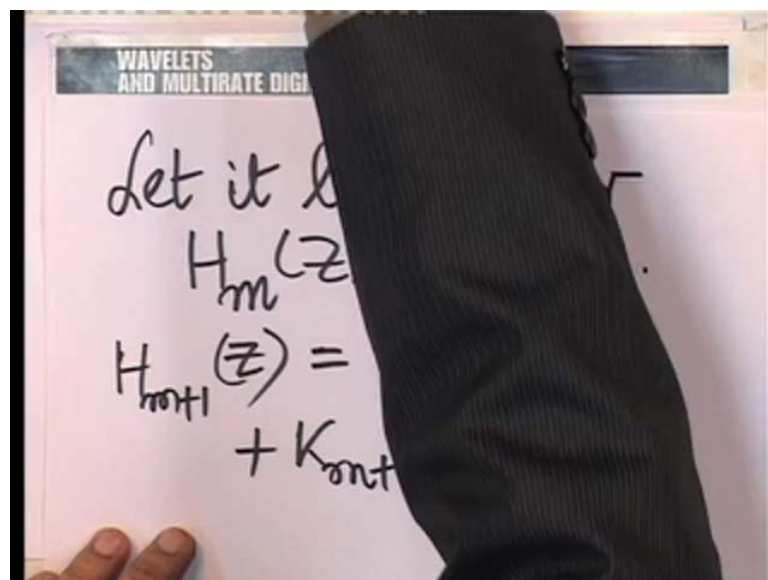
So, this is critical as you can see increasing the length now let us make a simple observation about the coefficient of Z raise the power of 0 we shall show now inductively coefficient of Z raise the power 0 in $H_m Z$ is always 1 for all m .

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In fact, this is true in the basis step you recall that in the basis step the system functions very simple they were essentially $1 + K_1 z^{-1}$ and $-K_1 + z^{-1}$ and here of course, this was the coefficient of z raise the power of 0 in $H_1(z)$.

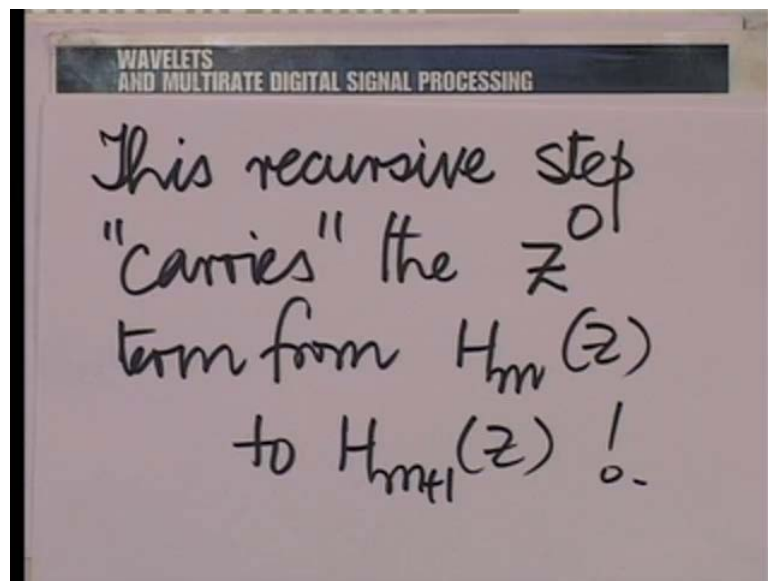
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If you please and therefore, it is true now we shall use the inductive step to continue this proof let it be true for $H_m(z)$ $m \geq 1$ now for us $H_{m+1}(z)$ is concerned $H_{m+1}(z) = H_m(z) - K_{m+1} + K_{m+1} z^{-1}$ and let us go back a couple of steps in a reasoning when we load this equation

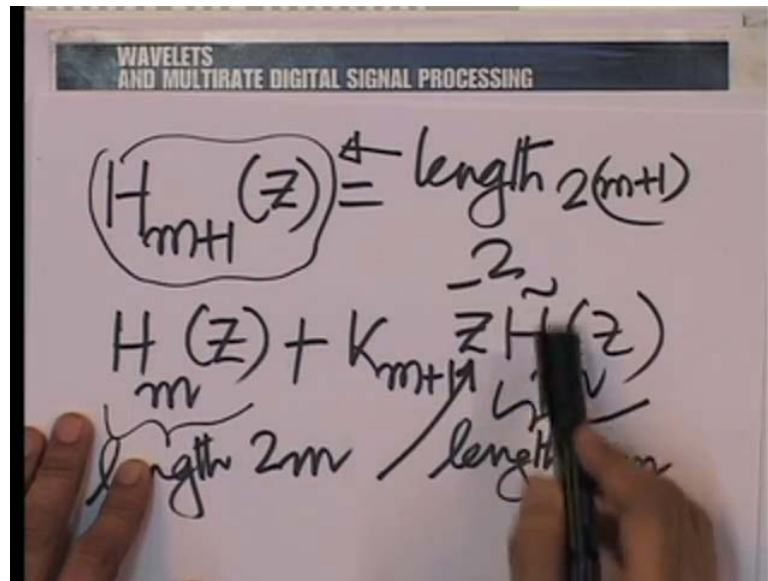
you know you can visualize the equation where would the Z raise the power of 0 coefficient here and if any in this term, but please note that in this term you have the Z raise the power of minus 2 common to all the term in this expansion. So, this was the length 2 m and by multiplying by the Z raise the power minus 2 the lowest power of Z inverse in this entire term is 2 now. So, you have no Z raise the power of 0 term here and that has been assumed by induction to be 1 and that is carried over here.

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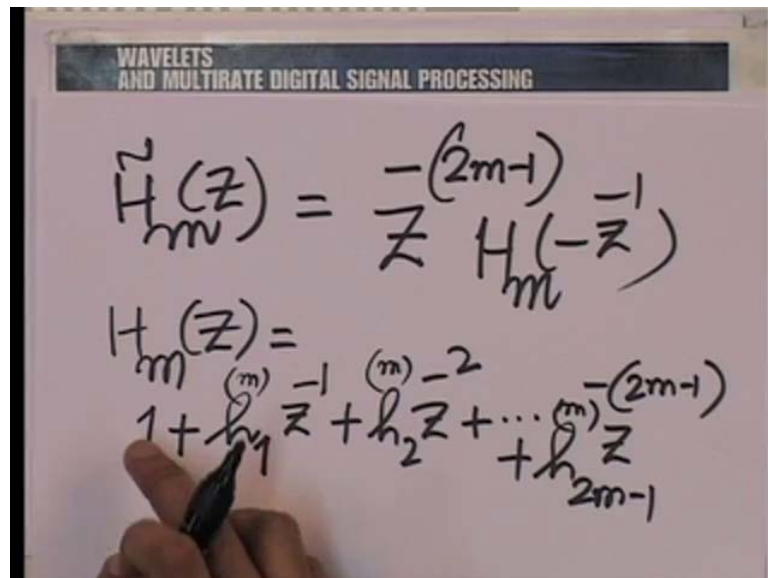
So, it is this equation which carries over the Z raise the power of 0 terms is carried from here to here and of course, by inductive assumption the Z raise the power of 0 term is 1 here it is also 1 there this completes the inductive proof now we know the coefficient of Z raise the power of 0 is always going to be 1 in all the H_m s now is very easy to see what k_m plus 1 is so.

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In fact, let us go back to the same equations that we use the minute ago let us look at these equations once again **ya** H_{m+1} is $H_m Z$ plus $K_{m+1} Z$ raise the power minus 2 H_m tilde Z now let us understand what H_m tilde Z is all about. So, H_m tilde Z we assumed.

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In fact, if shown inductively that it must be the form Z raise the power minus $2m$ minus 1 H_m minus Z inverse and H_m said we know what that looks like. So, m said looks like this it looks like 1 plus let us say $h_1 Z$ to the power minus 1 plus $H_2 Z$ raise the

power minus 2 plus $H^{2m-1} Z$ raise the power minus 2 $m-1$ now you know if you want to be very **very** careful you should subscribe this with m here to emphasize a this corresponds to the m th system function. So, i do that to be careful. So, all of saying is these are the impulse response coefficients are the impulse response points in H^m and you know that the coefficient of Z raise the power of 0 is 1.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

$$\tilde{H}_m(Z) = Z^{-(2m-1)} \left\{ 1 - h_{1,m} Z + h_{2,m} Z^2 - h_{2m-1,m} Z^{2m-1} \right\}$$

So, this is the form of the system function $H^m Z$ as account sequence H^m tilde Z is now going to be the following form you know H^m tilde Z is first going to have Z replaced by minus Z inverse here. So, to do that we have $1 - h_{1,m} Z + h_{2,m} Z^2 - h_{2m-1,m} Z^{2m-1}$ and this whole thing is then multiplied by Z raise the power minus 2 $m-1$. So, if you look at what is happening these positive powers are going to be shadowed over by this negative power. So, for example, the H^{2m-1} term is going to have Z raise the power of 0 with it now and then you are going to have increasing negative powers of Z as you come down and finally, this is going to correspond to Z raise the power minus 2 $m-1$ right there. So, let us write the expansion again.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

$$\tilde{H}_m(z) = -h_{2m-1}^{(m)} + \dots + h_2^{(m)} z^{-(2m-1)/2} + \dots + \frac{h_0^{(m)}}{z^{2m-1}}$$

So, $\tilde{H}_m(z)$ is going to have the form $h_{2m-1}^{(m)} + \dots + h_2^{(m)} z^{-(2m-1)/2} + \dots + \frac{h_0^{(m)}}{z^{2m-1}}$. So, on plus H_2 $m Z$ raise the power $2m-1$ plus 2 and as you keep going downwards then you have until you reach Z raise the power $2m-1$ here this is important and now making this observation the coefficient of the highest and of course, odd negative power of Z in \tilde{H}_m is 1 and we carry that here. So, the coefficient of the highest power of Z inverse in h_{m+1} is only going to come from here it cannot come from here the highest negative power here is Z raise the power $2m-1$, but this is of length 2 into $m+1$ the highest negative power of Z inverse here is going to be too larger than the highest negative power here and that can only come from here and. In fact, as you can see that must be k_{m+1} .

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WAVELETS
AND MULTIRATE DIGITAL SIGNAL PROCESSING

As a consequence of
 $H_{m+1}(z) = H_m(z) + K_{m+1} z^{-2} H_m(z)$
contributes $z^{-(2m+1)}$ term

S, what we are saying in a effect is as a consequence of this let us write it down clear as a consequence of H_{m+1} said is equal to $H_m Z$ plus $K_{m+1} Z$ to the power minus 2 H_m tilde Z and you know what the form of H_m tilde Z this is what H_m tilde Z looks like the Z to the power minus 2 m plus 1 comes only from here the highest negative power of Z and. In fact, the coefficient of Z raise the power minus 2 m plus 1 that is the highest negative power of Z is equal to K_{m+1} .

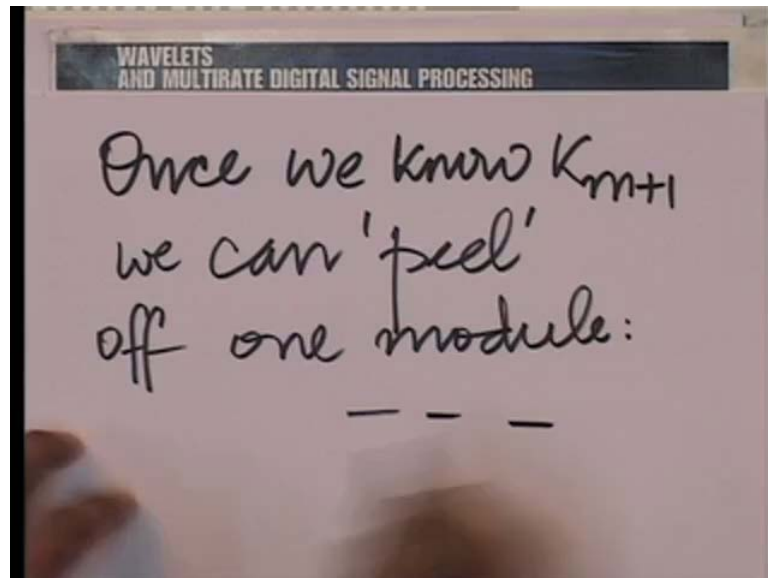
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WAVELETS
AND MULTIRATE DIGITAL SIGNAL PROCESSING

Coeff of $z^{-(2m+1)}$
(highest neg power of z)
 $= K_{m+1}$

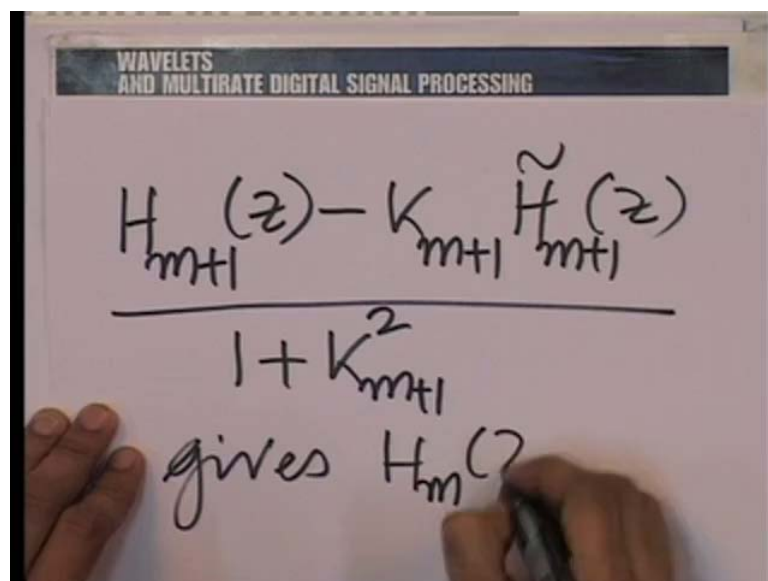
So, now, we have highest negative power of Z. So, once you have the final system function which you are trying to peel of stage by stage look at the coefficient of the highest negative power of Z and that itself is the first lattice coefficient simple.

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So, once you know k_{m+1} you know how to go down descend 1 step once we know K_{m+1} we can peel off of 1 module in other words.

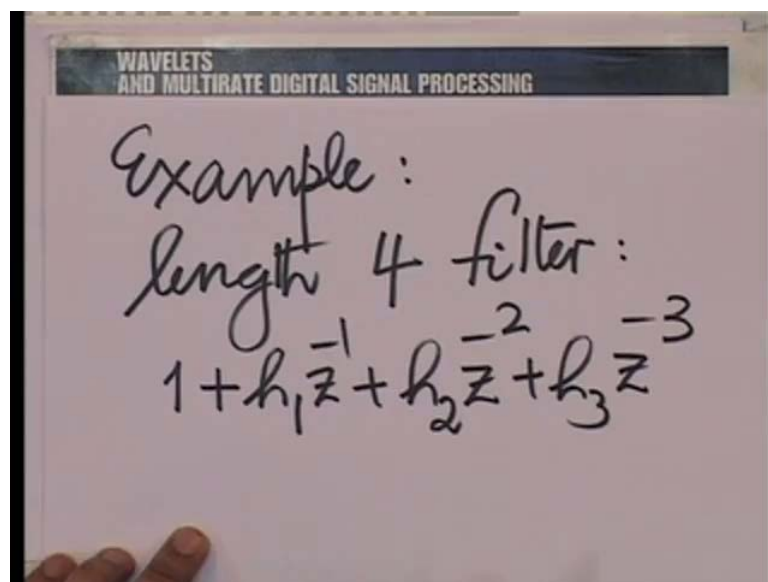
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We know H_{m+1} we can construct H_{m+1} tilde and then $H_{m+1} Z^{-K_{m+1}}$ divided by $1 + K_{m+1}^2$ gives $H_m z$. So, notionally we have

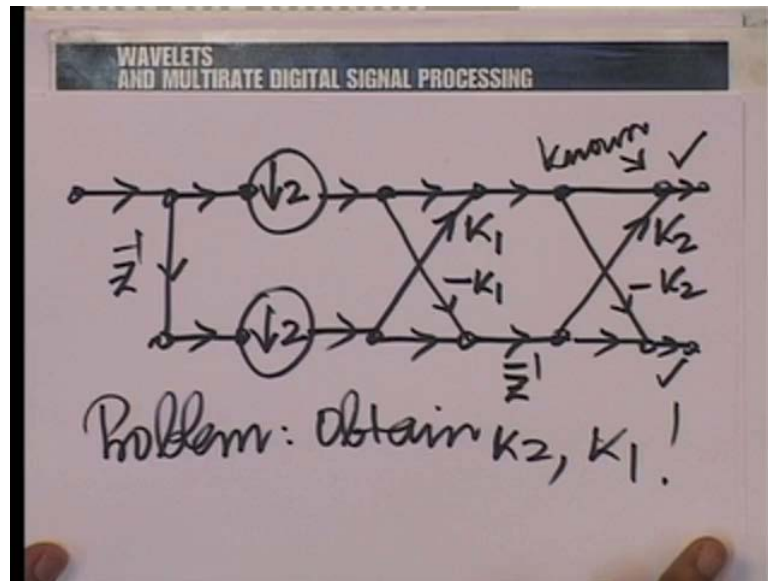
got $H_m(z)$ was the system function after the previous module, but this is notion what we need to worry about is does $H_m(z)$ really have a length of 2 less now instead of again trying to do the algebra generally here as we be in doing for phi. So, this was the general algebra it will be easier for us to take a specific example in see how this $(())$. So, let us take of lengthh four orthoonah filter that lengthh four filter could for example, be the dhobash length four filter, but anyway we shall not put specifically the numerical values of the impulse response coefficients, but we do not need to understand how this recursion words you would find it much easier to just carry out the recursion and demonstrate that it feasible.

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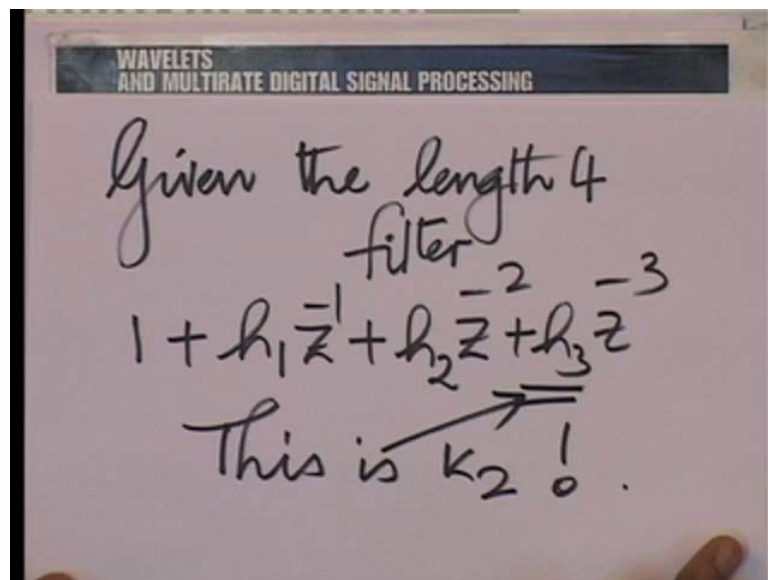
So, let us take an example lengthh four filter. So, if you had a form $1 + h_1 z^{-1} + h_2 z^{-2} + h_3 z^{-3}$ plus $H_2 Z$ raise the minus 2 plus H there z raise the power minus there. So, we wash to peel of 1 stage in other words what we are saying is we have this situation we have 2 stages of the lattice here. So, we know the system function here and therefore, also here known at this point and we wish to obtained K_2 and K_1 that was we are saying.

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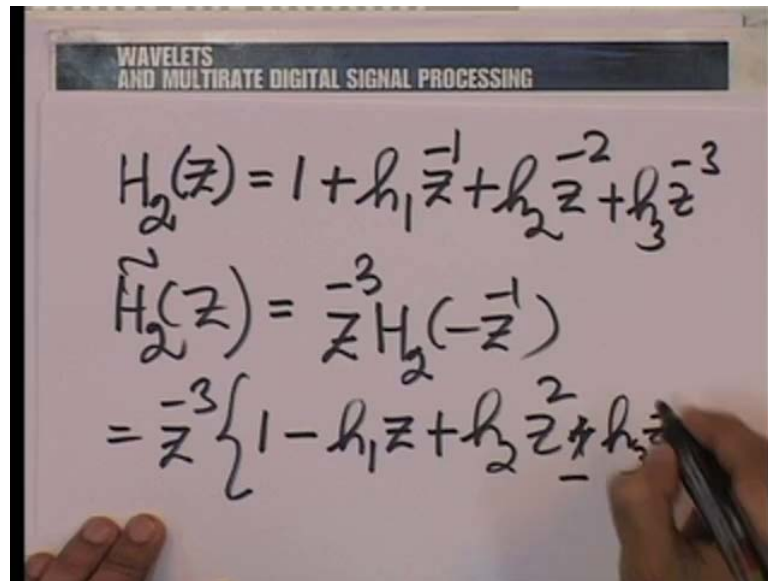
So, the problem is obtained K_2 and K_1 now what we just said makes it very easy for us to obtain K_2 given the length four filter.

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$1 + h_1 z^{-1} + h_2 z^{-2} + h_3 z^{-3}$ what we discussed a few minutes before this namely that the most negative power of Z reveals K_2 as. So, this essentially is K_2 and having motive that we can obtained 1 step lower.

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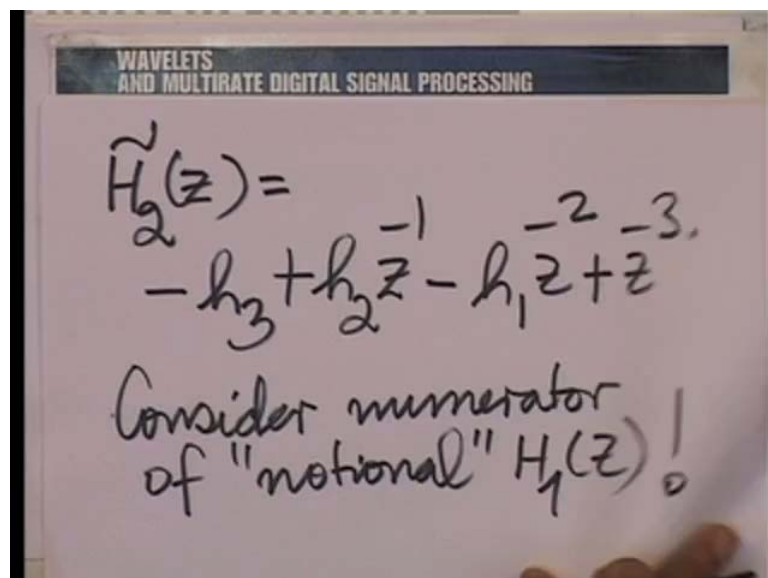


WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

$$H_2(z) = 1 + h_1 z^{-1} + h_2 z^{-2} + h_3 z^{-3}$$
$$\tilde{H}_2(z) = z^{-3} H_2(-z^{-1})$$
$$= z^{-3} \{ 1 - h_1 z + h_2 z^2 - h_3 z^3 \}$$

So, we can construct if we call this $H_2(z)$ or let us say $H_2(z)$ after the second stage then we can also construct $\tilde{H}_2(z)$ and $\tilde{H}_2(z)$ is going to be very simple it is going to be $z^{-3} H_2(-z^{-1})$ which is $z^{-3} (1 - h_1 z + h_2 z^2 - h_3 z^3)$. So, $\tilde{H}_2(z)$ can be simplified it is $z^{-3} (1 - h_1 z + h_2 z^2 - h_3 z^3)$.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

$$\tilde{H}_2(z) = -h_3 + h_2 z^{-1} - h_1 z^{-2} + z^{-3}$$

Consider numerator of "notional" $H_1(z)$!

So, all that we now need to do is to consider the numerator of notional $H_1(z)$ that numerator $H_1(z)$ notionally as you remember was $H_2(z) - K_2 \tilde{H}_2(z)$ divided by $1 + K_2^2$ and if you only K_2 are to expand this.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

$$H_1(z) = \frac{H_2(z) - K_2 \tilde{H}_2(z)}{1 + K_2^2}$$

So, you know to expand the numerator it will be easier for me to write down the expansion in 2 lines 1 line for $H_2(z)$ and 1 line for $-K_2 \tilde{H}_2(z)$.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

$$H_2(z) - K_2 \tilde{H}_2(z) \quad (K_2 = h_3)$$

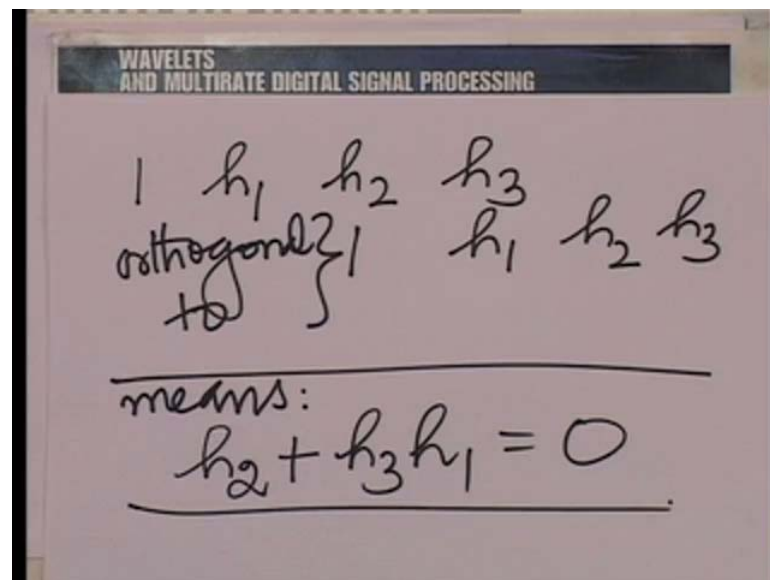
$$1 + h_1 z^{-1} + h_2 z^{-2} + h_3 z^{-3} - (h_3 z^{-2} - h_2 z^{-1} + h_1 z^0 - h_0 z^1)$$

$$0$$

So, $H_2(z) - K_2 \tilde{H}_2(z)$ and noting that K_2 is essentially H_3 we have $1 + h_1 z^{-1} + h_2 z^{-2} + h_3 z^{-3} - (h_3 z^{-2} - h_2 z^{-1} + h_1 z^0 - h_0 z^1)$

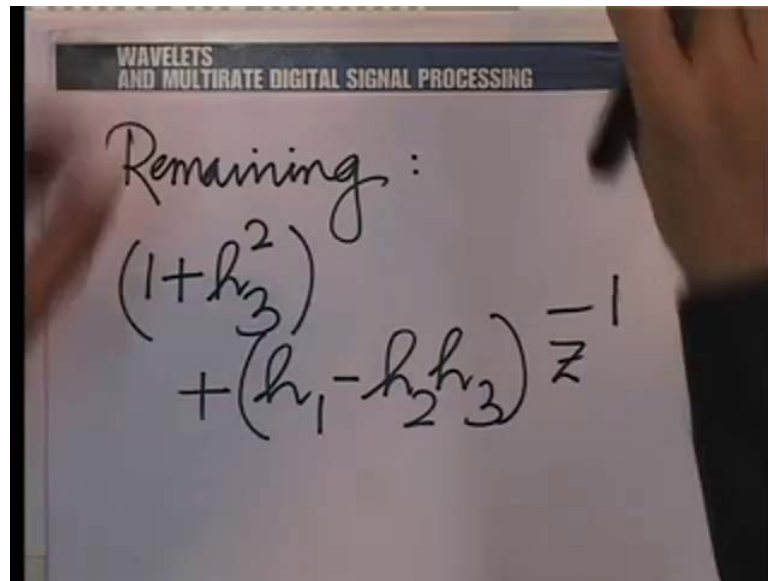
three minus $K^2 H^2 \tilde{z}$ which is minus $H^3 Z$ raise the power minus three plus $h_1 H^3 z$ raise the power minus 2 and something else here. So, of course, we just completed its minus $H^3 H^2 z$ raise the power minus 1 and of course, plus H^3 squared here now this is very interesting. So, let us look at this expression carefully. So, let us look at the coefficient of z raise the power minus three this anyway become zero identically what is interesting is this the coefficient of z raise the power minus 2 it is H^2 plus $h_1 H^3$, but let us go back to the basic requirement that we had on an orthogonal filter remember the whole idea in constructing the dhobash filter for example, was that the impulse response was orthogonal to its even translates let us put that condition down again in see what we gives us.

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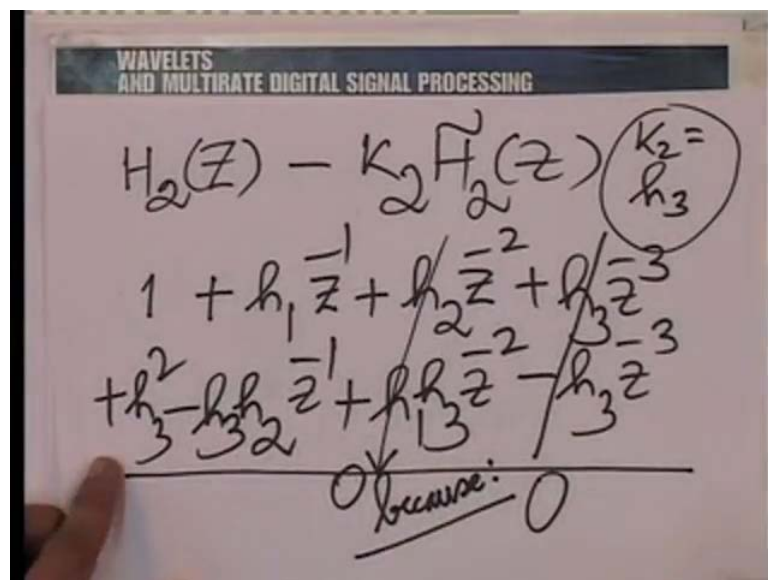
So, $h_1 H^2 H$ there is orthoona to $h_1 H^2 H$ there and; that means, H^2 into 1 plus H there into h_1 is 0 now go back to the coefficient of z raise the power of minus 2 here that is H^2 plus $h_1 H$ there low and be whole that the same as this and it is identically zero and account of the orthogonality to even translate. So, let us make an note of that.

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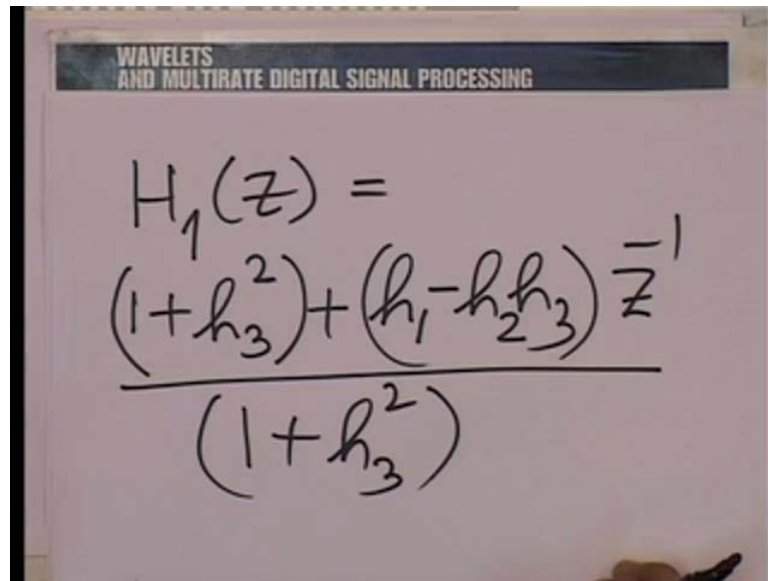
So, this is zero because of this orthogonality to even translates. So, what is left now remaining is 1 plus H_3 squared plus h_1 minus $H_2 H_3$ times z raise the power minus 1.

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So, we are arrange very good situation as far as $H_2 Z$ minus k_2 tilde z is concerned the length has gone down by 2 as we indeed wanted it 2.

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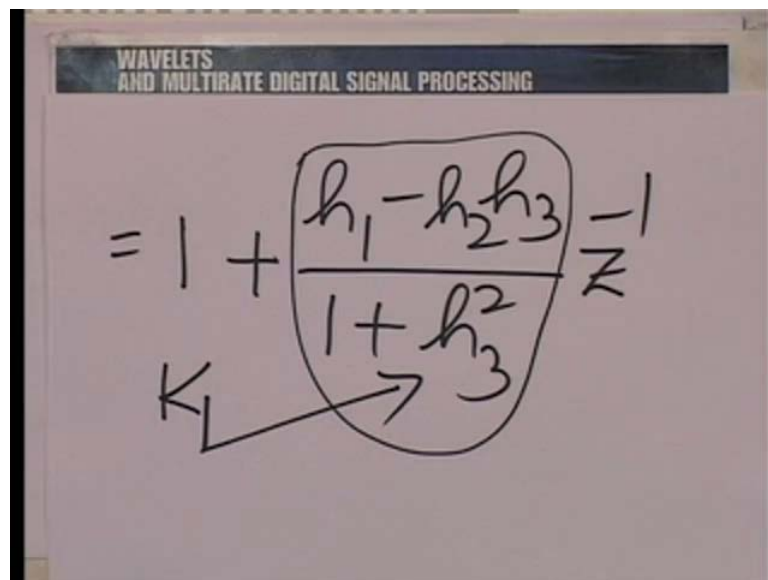


WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

$$H_1(z) = \frac{(1+h_3^2) + (h_1-h_2h_3)z^{-1}}{(1+h_3^2)}$$

Accordingly $h_1 z$ is now going to be $1 + H_3^2 + h_1 - H_2 H_3$ into z inverse divided by $1 + H_3^2$. Remember.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

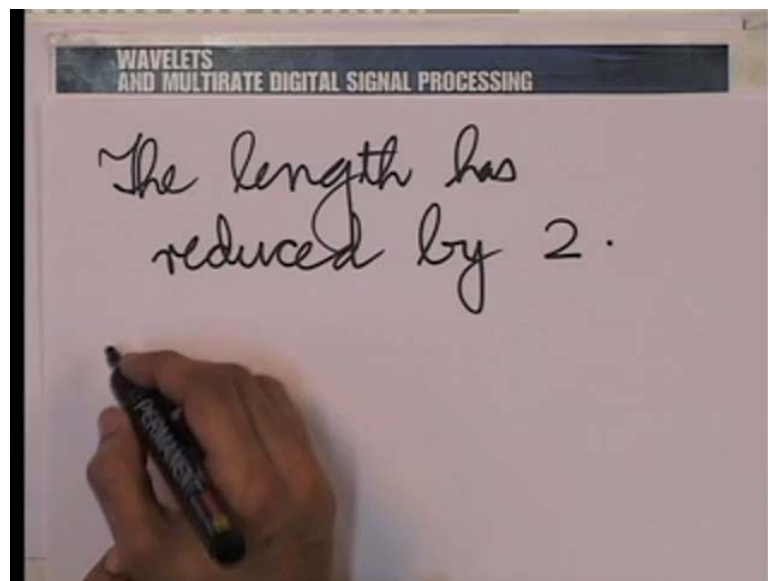
$$= 1 + \frac{h_1 - h_2 h_3}{1 + h_3^2} z^{-1}$$

K_1 →

here and that is very simple that is $1 + h_1 - H_2 H_3$ divided by $1 + H_3^2$ times Z inverse and low in behold also reveals to us k_1 direct. So, we completed the lattice construction as you can see for the dthobosh case is going to work definitely all that the required is orthogonality to even translates nothing else was needed because of the orthogonality to even translates when we carried out the numerator part of the

peeling process the length decreased by 2 now you know you must understand where there is something unusual here in going to H_m of that matter H_m tilde from H_{m+1} of that matter H_{m+1} tilde there was just a little bit of solution of 2 equations involved that could be an solve for any H_{m+1} and H_{m+2} tilde there is nothing very special about the form of H_m or H_m tilde of that matter in writing down that downwards. However.

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What was important that when you went from H_{m+1} to H_m you needed the length to go down not by 1, but by 2 and this decrease of length by 2 is what broke Then the requirement of orthogonality of even translates. So, it is only because of the orthogonality to even translates that the length decrease by 2 in the numerator part of the expansion and of course, once you assure at the coefficient k which you had peeled off of the coefficient k by looking at the coefficient of the highest negative power of Z in H_{m+1} gave us yield it clearly revealed the corresponding k of the outer most lattice stage. So, now, we can see how to peel of stage by stage we done it specifically for a 2 stage lattice and its not difficult to extend this to a there stage or a multistage lattice beyond there stages now a few remark are in order here you know what we done now is actually to demonstrate with an example that when you carry out this backward recursion again .

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WAVELETS
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The backward recursion

$$H_m(z) = \frac{H_{m+1}(z) - K_{m+1} \tilde{H}_{m+1}(z)}{1 + K_{m+1} z^{-1}}$$

Let us make this observation the backward recursion $H_m(z)$ is $H_{m+1}(z) - K_{m+1} \tilde{H}_{m+1}(z)$ divided by $1 + K_{m+1} z^{-1}$ is effected by the following step.

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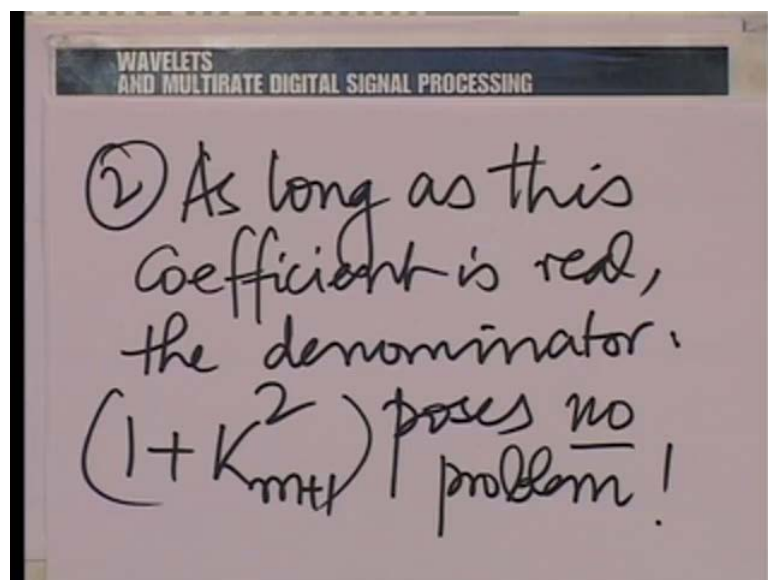
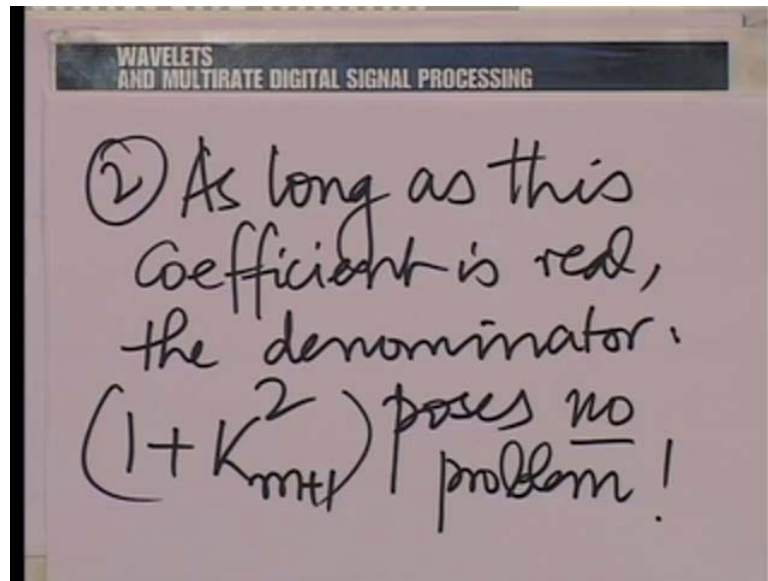
WAVELETS
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-- is effected by:

① K_{m+1} is the coefficient of highest negative power of z^{-1} in $H_{m+1}(z)$

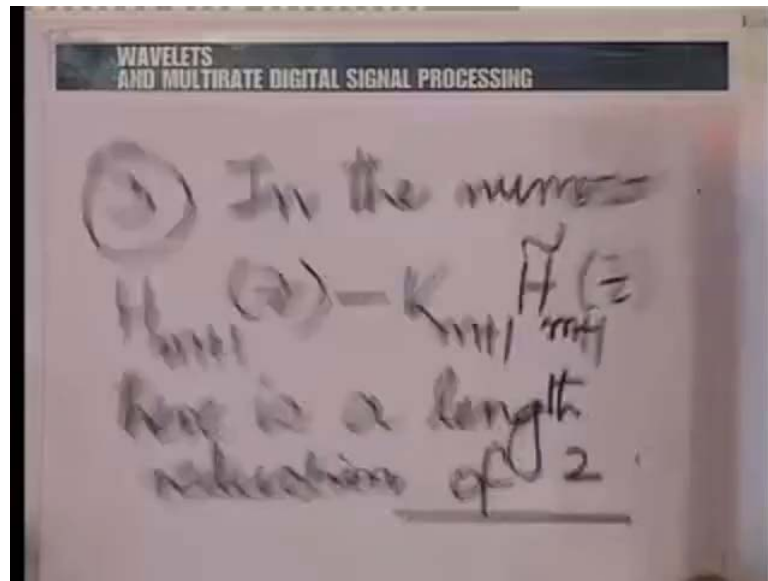
Set no1 K_{m+1} is the coefficient of the highest negative power of Z inverse in H_{m+1} .

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As long as this coefficient is real the denominator which is $1 + K_{m+1}^2$ squared poses no problem what I mean by that is a denominator can pose a problem if it is 0 you cannot divide by 0. So, you are assured that this is not going to be 0 if K_{m+1} is real or in other words the coefficient of the highest power of Z^{-1} is real you are guaranteed this $(($ $))$ does not go to division by 0.

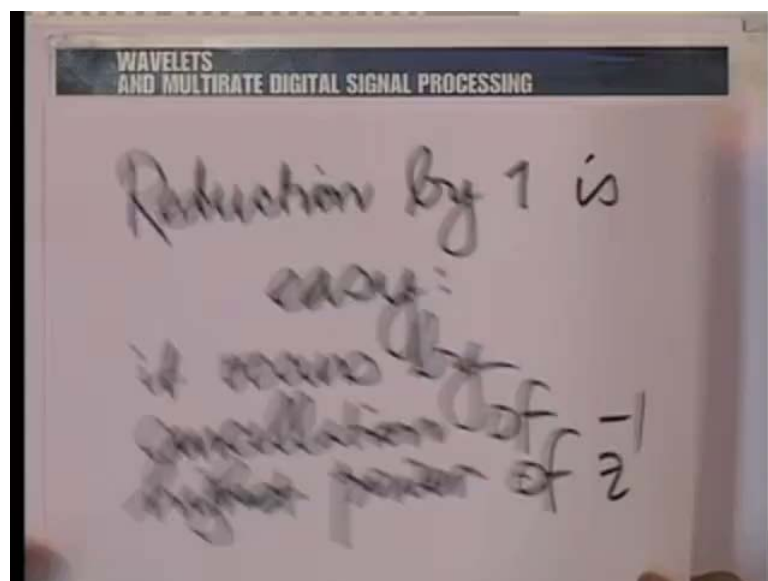
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having made that observation now the third observation is in the numerator namely $H_{m+1}(z) = K_{m+1} \tilde{H}_{m+1}(z)$ there is a length reduction of 2 and the justification you see reduction by 1 is easy essential.

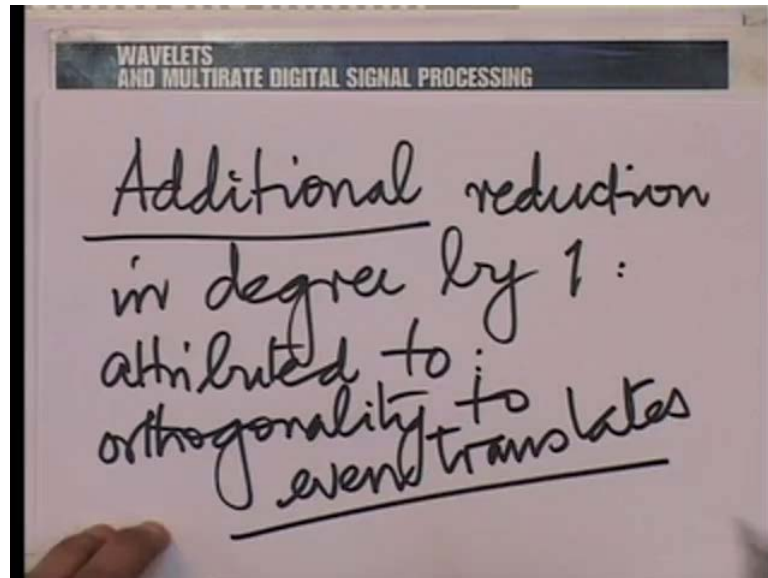
If you look at this numerator the coefficient of the highest power of Z^{-1} here is K_{m+1} and the coefficient of the highest power of Z^{-1} here without this would be a_1 , but with this its K_{m+1} .

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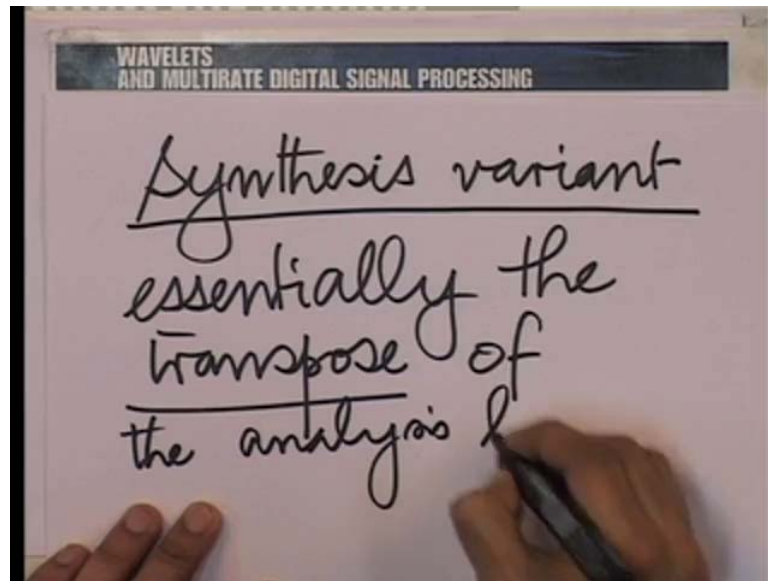
So, that would cancel out. So, essentially reduction by 1 is easy it occurs by cancellation of highest power of Z inverse, but the additional 1 step reduction.

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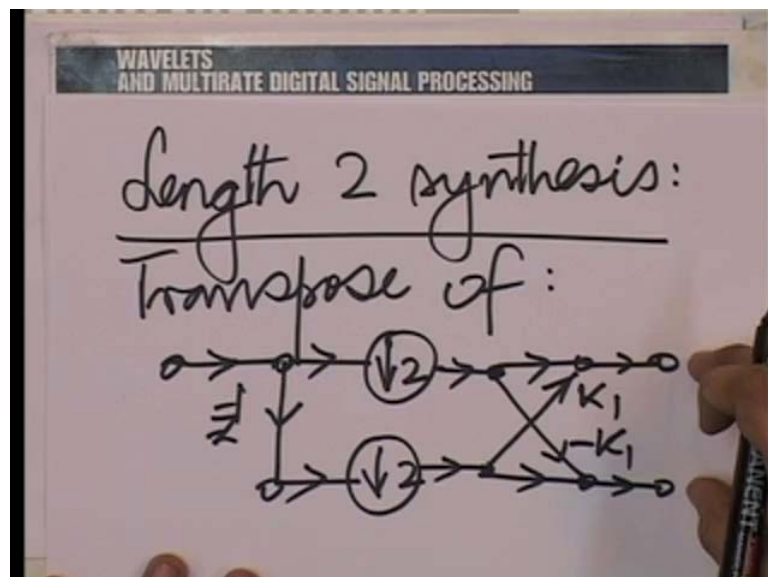
Additional reduction in degree by 1 is attributed to orthogonality to even translates this is a very important observation this is where the orthogonality of the filter bank where it not an orthogonal filter bank that we were talking about we would of course, have trouble getting in this additional reduction of technique now finally, we need to complete this lecture by putting in a variant and that variant is essentially to transpose the structure that we construct it in other words how could we use the lattice structure that we have construct it to build a lattice structure for the synthess side let us complete this lecture by discussing this variant.

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The synthesis variant its very easy to construct the synthesis variant essentially it is the transport of the analysis lattice and what i shall do is to construct the synthesis variant first for a length 2 and then for length four and that would illustrate and how do we do it in general.

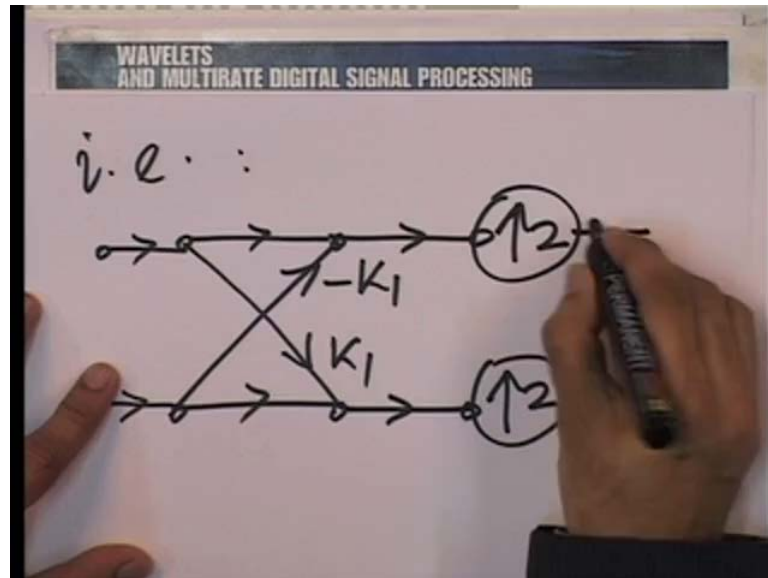
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So, length 2 synthesis lattice is essentially the transpose of this and you can visualize what the transport will look like let us just over it orally first and then draw it we would have the transport with all these arrows reversed here you would have up sampler at this

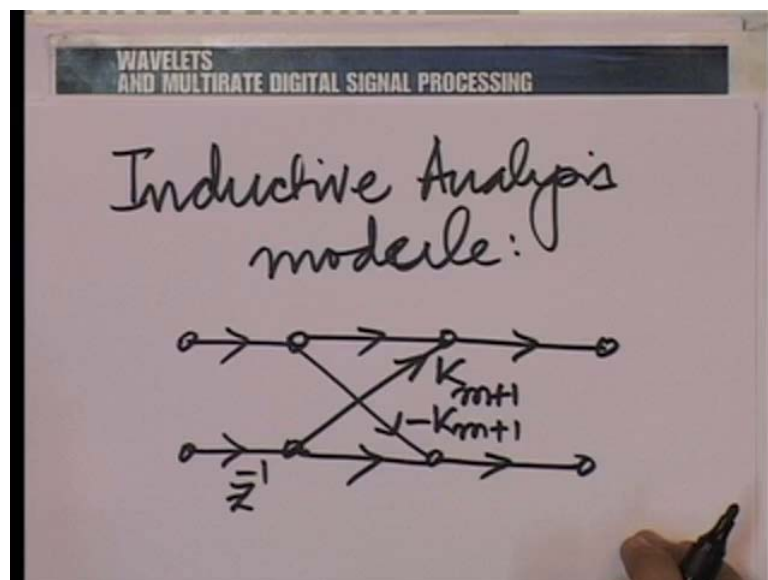
point these arrows reversed the Z inverse as it is this now becoming a summing point inside of a branching point and then resulting in the output here let us draw the transpose.

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They we are this is the transpose of length 2 stage now all that we need to do to complete this discussion is to show the transpose of the inductive module.

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So, the inductive analysis module was essentially this of course, this is after the down sampler let us say K_{m+1} here and minus K_{m+1} there and therefore, the corresponding synthesis inductive module would look like this once we have the

inductive module and the basis module it is easy to construct the complete lattice on the synthesis side and now I leave it as an exercise for the class at the end of this lecture to work out the same kind of recursions as we did for the analysis side to construct the synthesis filter bank leaving that exercise to you we conclude the lecture here to observe that we can have a beautifully computationally efficient structure called the lattice to realize an orthogonal analysis and now synthesis filter bank thank you.