

Advanced Digital Signal Processing - Wavelets and Multirate
Prof. V. M. Gadre
Department of Electrical Engineering
Indian Institute of Technology, Bombay

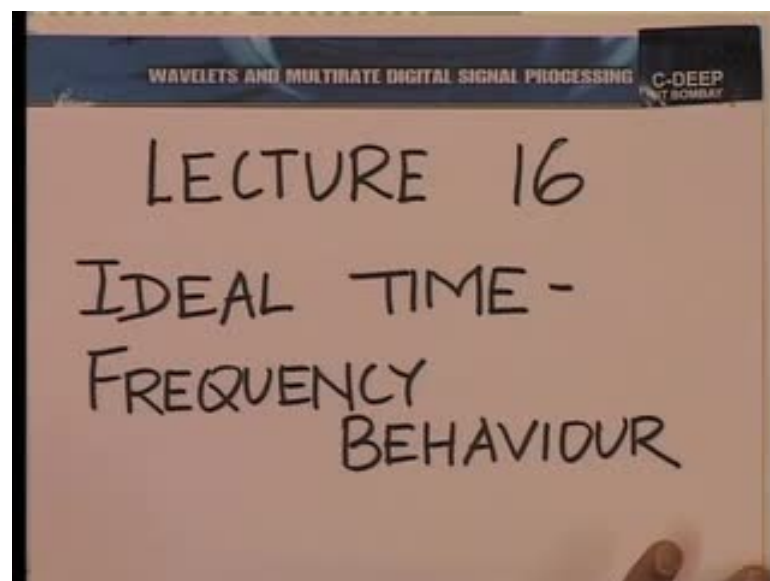
Lecture No. # 16
Ideal Time Frequency Behaviour

A very warm welcome to the sixteenth lecture on the subject of Wavelets and Multirate Digital Signal Processing. Let us put in perspective, what we are going to do in today's lecture, building up from what we did in the previous one.

In the previous lecture, we had looked at the Fourier transform of the scaling function or the so-called father wavelet $\phi(t)$, and the wavelet function or the so-called mother wavelet $\psi(t)$, in the Haar Multiresolution Analysis. What I shall do, is to begin with a description of where we wish to go from here.

You see, we had made some observations about the nature of the magnitude of the Fourier transform of $\phi(t)$ and $\psi(t)$; we also noted, that when we multiply a given function $x(t)$ by a translate of $\phi(t)$, the magnitudes of the Fourier transforms of x and ϕ are getting multiplied, and we saw that the nature of the Fourier transform of ϕ or for that matter even that of ψ was such that, it emphasized some band of the Fourier transform of the underline function x which was being studied.

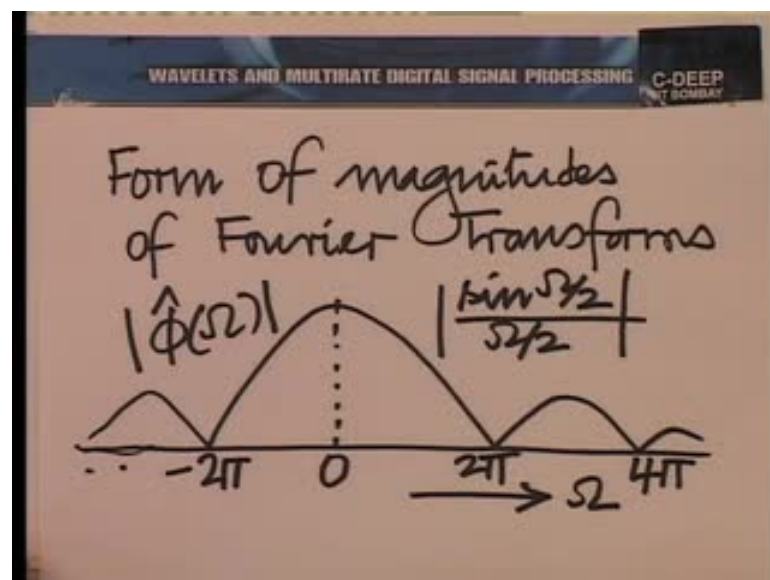
(Refer Slide Time: 02:16)



Now, what we intend to do today is to idealize from there, what is the ideal situation to which we strive. And therefore, I have put down the central theme in the lecture today, to be ideal time frequency behavior, what is the ideal, towards which we are trying to move.

Now, let me put before you, once again, the nature of the Fourier transforms of ϕ and ψ ; by nature, I am essentially going to refer to the magnitude; the phase, though important, in general, is not of prime importance at the moment, because it is the magnitude which makes a selection of band; so, let us put down the nature of the magnitude.

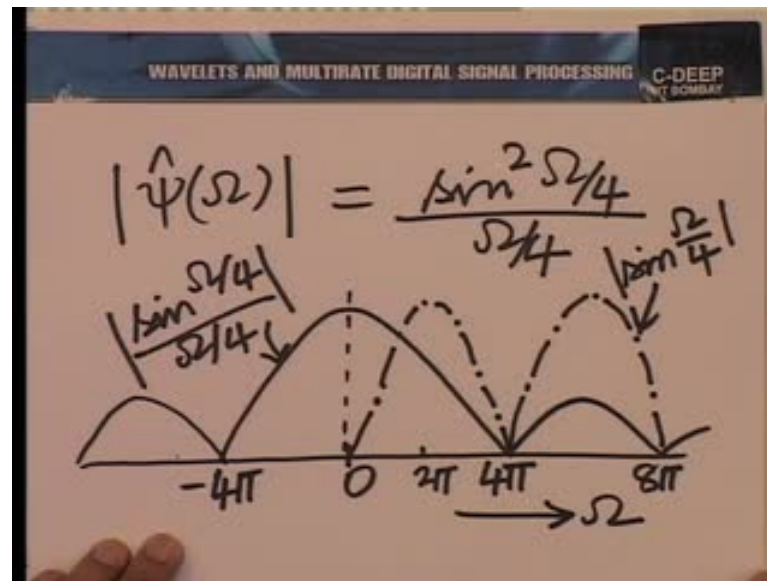
(Refer Slide Time: 02:57)



So, form of magnitude of Fourier transform, for ϕ , it had an appearance like this; so, this was 0 frequency and this was 2π here, and all multiples of 2π subsequently, and so on.

This is $\text{mod } \phi \text{ cap } \omega$, you see when I say **form**, what I imply is that I am not going to consider any constant of phase; constants will only scale this up or down, and the phase would not affect the magnitude, of course; so, let us also look at that of ψ .

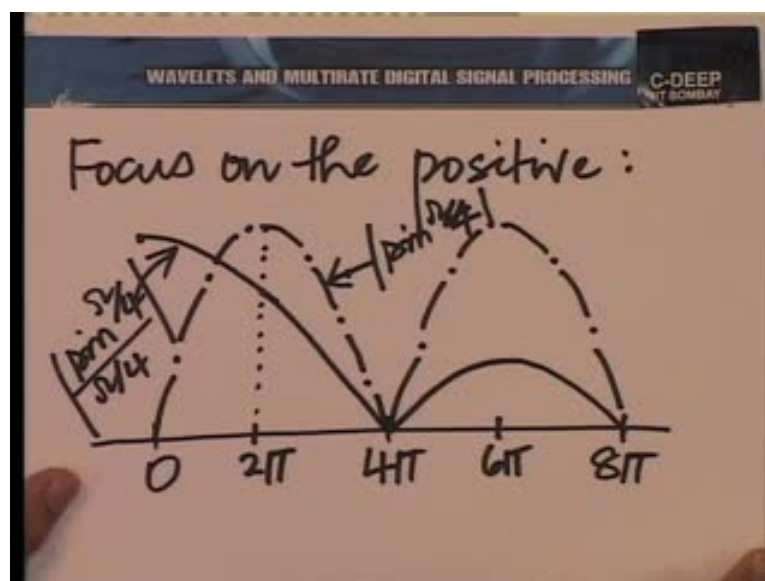
(Refer Slide Time: 04:38)



The form of the magnitude of the Fourier transform of psi looks something like this, sin squared omega by 4 divide by omega by 4. And we are had been attempt to sketch this last time, we first sketched sin omega by 4 by omega by 4; so, we said this was the form of mod sin omega by 4 divide by omega by 4, the solid line here, and I also drew a dotted line to indicate this timelets or wavelets, use a dot dash line to make a distinction from this margin or this axis.

So, let us use the dot dash line to denote the magnitude of the other term sin omega by 4; so that would have a peak at 2 pi, this is the form. Now, this format line is multiplied by this dot dash line here; so, of course, you must visualize this dot dash line being replicated on the negative side and as you know for a real function the Fourier transform is magnitude symmetric; so, it is enough for me to study the positive side of omega and the negative side would be a mirror image.

(Refer Slide Time: 07:06)



So, let me expand this part here, focus on the positive, a good mortaring general; you see this $\sin \omega$ by 4 by, ω by 4, this one has a monotonically decreasing character from 0 to 4π ; this one has a monotonically increasing character between 0 and 2π and then monotonically decrease.

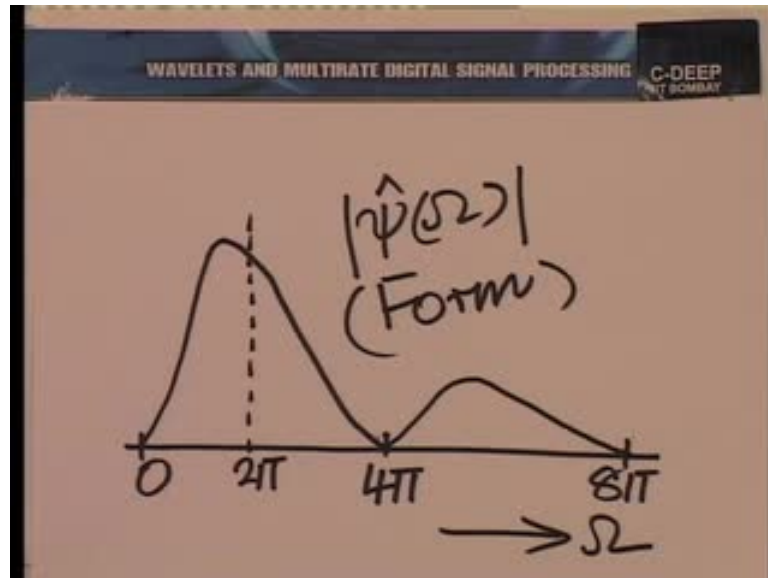
Now, it is very clear that from this point onwards, the product of these two is only going to decrease; so, you cannot possibly have a value of this, product this, of course, being $\mod \sin \omega$ by 4. You cannot have a magnitude of the product of this dot dash line with the solid line, greater in the segment between 2π and 4π , then it is at 2π .

So, in other words after 2π between 2π and 4π , this product is only going to decrease, and therefore, I can get a feel; you see it is clear that the product is 0 at ω equal to 0 whatever is, after 2π is going to be less than what is at 2π , somewhere in between it is going to achieve a maximum and then continue to drop; so, we get a feel of this, we must get a fine of feel of this, and we had the last time. In fact, let me also make one more remark.

You see if you look at the region between 4π and 8π , thus this situation is a little simple. There is a kind of tendency to a maximum, somewhere in between in both of these functions and then a drop; so that, similar pattern would be replicated in the product a maximum, somewhere in between will not quite at 6π , please remember this

is not quite symmetric; you must remember that although, **this is**, this is not quite symmetric.

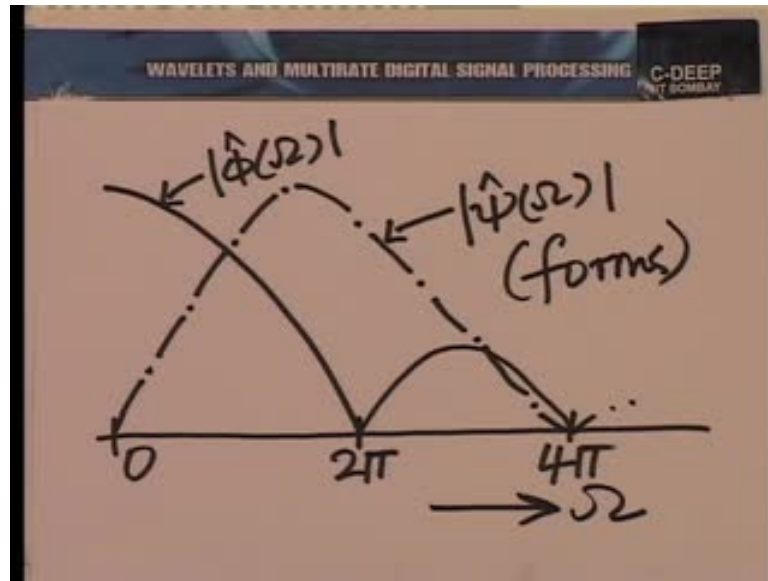
(Refer Slide Time: 10:39)



So, the maximum will be somewhere other than 6π , but that is not of, so much of concern then be a maximum, somewhere near 6π , and it would drop of on both sides; so, in total, this is, what the product would look like.

I would not mark this maximum, it is a little difficult to calculate, but this is the nature, this is the form. And now, let us take the trouble to draw them together, again only on the positive side of the frequency axis, the form of the Fourier transform of ϕ and of ψ .

(Refer Slide Time: 11:36)



Let us focus only between 0 and 4 pi; so, phi looks something like this, and psi looks something like this. And we had made a remark, and what phi does? And what psi does? Phi, in a fact, emphasizes those frequencies lying around 0 frequency, and psi emphasizes those frequency line around its maximum in the band between 0 and 4 pi, and deemphasizes frequency on either side. So, in fact, if you look at psi it deemphasizes frequencies around 0, and then after that band; so, it emphasizes a band of frequencies.

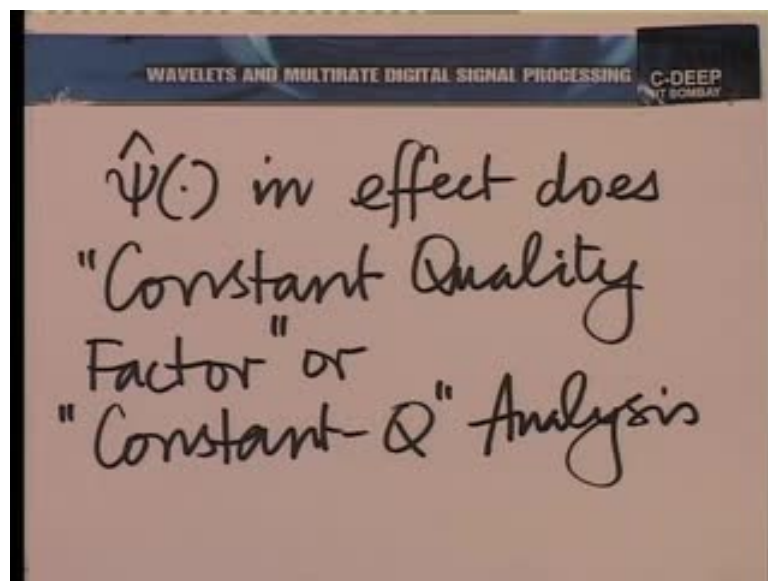
It is clear that psi has a band pass character, it is a band pass function; a band pass function is one which emphasizes frequencies around some called center frequency where the response is a maximum, and deemphasizes frequencies both around 0 and around infinity so to speak; so that, the finite band of frequencies one band, which that function emphasizes.

Loosely speaking, this psi omega here emphasizes, those frequencies line around its maximum here, and of course, psi emphasizes frequencies around 0. We also made one more remark on the distinction between phi and psi; you see we noted that when we contract or expand, so when we go up or down the ladder, what we are, we doing in the Fourier domain? When we go up the ladder, we are expanding in frequency, because we are contracting in time; when we go down the ladder, we are expanding in time, and therefore, we are contracting in frequency.

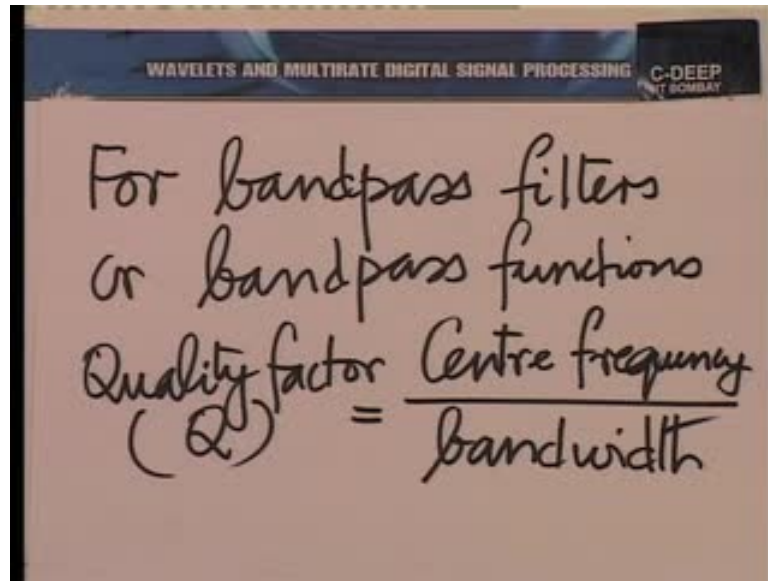
So, let us look at this figure once again, as we go down the ladder, we are contracting in frequency; so, we are emphasizing smaller in smaller bands around 0. And again, since we are contracting this as well we are emphasizing frequencies around a smaller and smaller center frequency. In fact, it is very easy to see that this center frequency, the point where there is a maximum, in the magnitude of ψ decreases geometrically or log arithmetically, as we go down the ladder in the haar multiresolution analysis, and the width of this band also decreases geometrically or logarithmically; this is something very interesting.

The band decreases geometrically, the center frequency also decreases geometrically. So, we have a situation where the ratio of the band to the center frequency is a constant; so, we have a name for that kind of analysis in the literature on wavelets or time frequency methods, we call it constant quality factor analysis or constant q analysis; let us write that down.

(Refer Slide Time: 16:01)



(Refer Slide Time: 16:40)



The image shows a handwritten formula on a slide. The slide has a header that reads "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP UNIT BOMBAY". The handwritten text says: "For bandpass filters or bandpass functions Quality factor (Q) = Centre frequency / Bandwidth".

$$\text{For bandpass filters or bandpass functions} \\ \text{Quality factor (Q)} = \frac{\text{Centre frequency}}{\text{Bandwidth}}$$

Psi in effect does constant quality factor analysis or constant q analysis, and this word quality factor comes from a term used in the context of band pass filter. For band pass filters or band pass functions, the quality factor or q, as it is often denoted in brief is the ratio of the center frequency to the band or bandwidth.

You know what bandwidth, of course, has to be taken with the pinch of salt, what does bandwidth mean? There are different definitions, particularly when you do not have a clear brick wall situation; you have a smooth variation of magnitude with frequency as you do here; so, there is a maximum and the frequency falls off on either side. Typically, we use the word bandwidth to denote that range of frequencies, within which the magnitude remains, within a certain percentage of the maximum magnitude.

So, for example, where the magnitude remains between 70 and 100 percent say of the maximum magnitude or where it remains even more specifically, for most situations. We talk about what is called the half power bandwidth? Where the amplitude or the magnitude response, falls to the square root of half from the maximum, and the square root of half has the significance, that at that point; where it falls to the square root of half, the power of a sin wave is half of, what it would be in proportion to the original has compared to the maximum point.

So, if at the center frequency, the point where the magnitude responses are maximum, the power ratio of input to output is say 100 units, then at the point where the power, the,

you know, the magnitude falls to one by square root of 2, the power would be only 50 units, the ratio, power ratio, output divide by input; so, it is called the half power point.

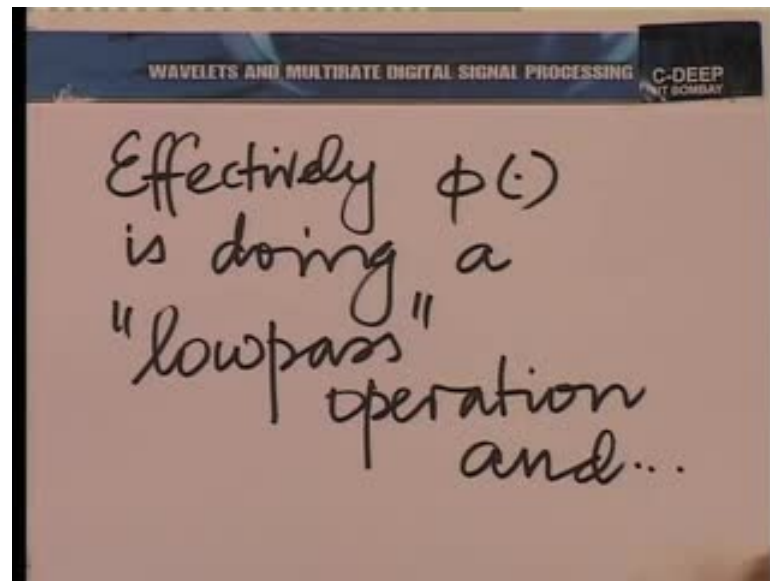
Very often we talk about half power bandwidth, in any case, it does not matter what percentage we use? 70 percent, so be 60 percent, so be whatever it be. With this notion of bandwidth, the ratio of the center frequency to the bandwidth in the sense, is a construct as we stretch or compress the Fourier transform of ψ , and there; so, of course, you know whatever you do in terms of stretching or compressing in the time domain, you are doing exactly the opposite in the frequency domain. So, as you go up the ladder you are going towards higher frequencies, and you are also spanning a larger bandwidth; as you come to lower steps, as you go descend, descend in the ladder, you go to lower rungs of the ladder, you are essentially going to smaller center frequencies and using a smaller bandwidth.

Now, let us bring the idea of time resolution and frequency resolution here; if you use bandwidth as a measure, please note as a measure of the range of frequencies that are emphasized by the function ψ ; now, why am I saying? Once again that these frequencies are emphasized, let me just recapitulate I am saying this again and again, because one must firmly understand this. I am saying that those frequencies are emphasized, because in finding the dot product of a function $x(t)$ with any translate of this function $\psi(t)$ or one of the stretch or compressed versions of $\psi(t)$.

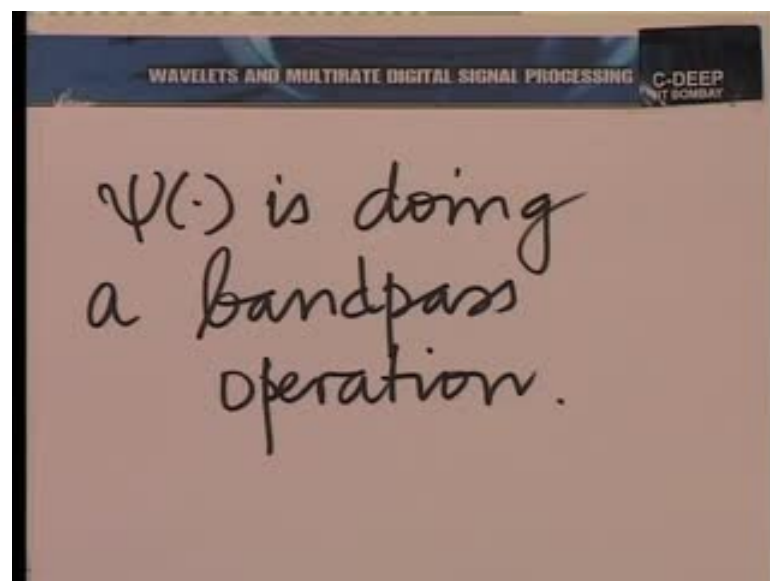
Parseval's theorem tells us that you are also multiplying the Fourier transform of x with the Fourier transform of that particular translate, and dilate of ψ of that matter of ϕ whatever it be. Now, we also understood that translation has no effect on the magnitude, dilation does, and when we multiply the Fourier transform of x by the Fourier transform of ϕ or ψ as the case may be appropriately dilated. One is automatically emphasizing multiplying that part of the band, which lies in the region of large magnitude of Fourier transform of ϕ or ψ , by a larger number and the other parts are been multiplied by a tapering number.

So, in fact, there is a filtering operation also being done by ϕ and ψ ; effectively, ϕ is doing a low pass filtering operation, and ψ is doing a band pass filtering operation; let us, make a note of this, this is very important.

(Refer Slide Time: 22:13)



(Refer Slide Time: 22:39)



So, effectively ϕ is doing a low pass operation, and ψ is doing a band pass operation. Now, then it almost seems trivial, what is so great? We could have built a band pass or low pass filter, otherwise why did we have to do all this haar business?

Well you see the beauty is in the two domains together, and this is where the whole catch lies, and this is where the whole struggle lies. You are able to do some kind of a crude low pass operation, I say crude, because nobody will agree, if you look at the frequency

response the Fourier transform of ϕ , that it is really very close to a good low pass filter crude in that sense.

You are doing a crude low pass filter operation, but with the provision that you are also confining yourself in time. So, you are saying, you are able to say with some confidents and that confidents depends on how well localize that Fourier transform is around 0 frequency? So, you are able to say with some confidents that when I multiply $x(t)$ by a certain dilate and translate of ϕ , I am emphasizing that band of frequencies around 0, which is covered by the appropriate dilate of ϕ .

So, if you take ϕ itself, and if you focus your attention on the main lobe of the Fourier transform, you may say, in a crude sense, that you are emphasizing the frequencies around 0 up to the extent of 2π , a main lobe goes up to 2π . And you are doing this, in a time region in which ϕ lies, in fact, that can be sending non crudely; so, ϕ is indeed very, very localized in time; I think nobody will disagree with that. **So is ψ** ; so when you multiply by a certain dilate in a translate of ψ , you are in effect doing a kind of localization in frequency around that point of maximum, as you saw it lay somewhere near 2π , before 2π actually.

And as you take different dilates of ψ , you are taking different bands and this is being done in the time zone covered by that particular translate. This is a serious statement, we are making; we are making a statement about localization in two domains simultaneously, in time and in frequency. And if you recall in the very first lecture, when I introduced the subjects of wavelets and time frequencies methods, this is one other things I mentioned, as a fundamental challenge and signal processing. In fact I went to the extent of saying the same challenge appears in different manifestations in different subjects.

In signal processing, we see it is a conflict between time and frequency, where is the conflict? The conflict is partly seen, now partly I say, you see as you notice in time we are very correct in saying that we have localized, after all $\phi(t)$ and $\psi(t)$ and they are translates and dilates are non 0 only over a finite region of time.

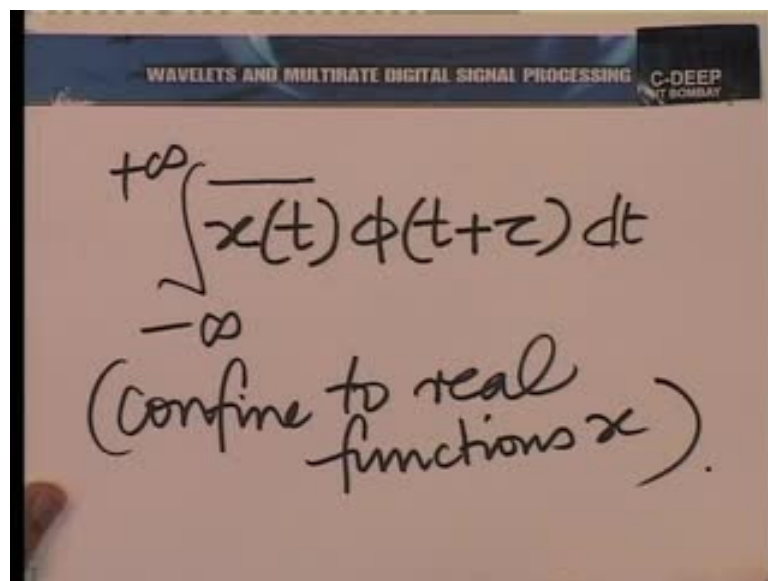
So, localization in time, in this case is not under question at all; it is localization in frequency which is somewhat suspect, we can crudely say that because if you focus your

attention in the main lobe, then in some sense it is localized, but there are the side lobes in the Fourier transform, both of ϕ and ψ .

So, now, we want to ask the question what ideal would I like to strive towards, if I were to have my way, how should I make the Fourier transform of ϕ and ψ look? We know how they should be in time, they should be packed into a finite region of time; we are able to do that. I would also like to pack them into a finite region of frequency simultaneously.

Now, what would that region of frequency be, let us use our understanding of signals and sampling a little bit here; you see, let us write down the dot product of x with a particular integer translate of ϕ has a sampling problem now.

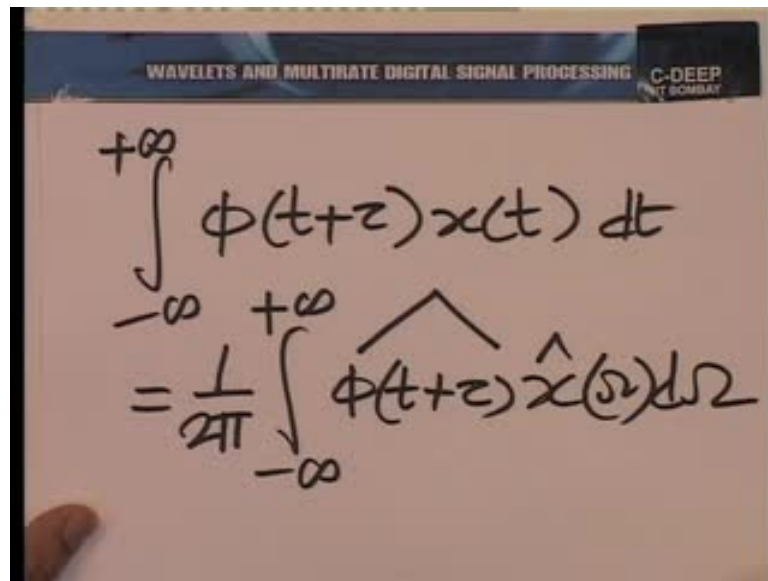
(Refer Slide Time: 27:43)


$$\int_{-\infty}^{+\infty} x(t) \phi(t+\tau) dt$$

(Confine to real functions x).

So, if you take this product, if you wish I can put complex conjugates maybe I should put a complex conjugate there, it would be as dot product in the strict sense, but even if I do not put a complex conjugate and confine myself to real functions x and doing rather well; in fact, we will do that for the moment, because we do not want to mix too many issues.

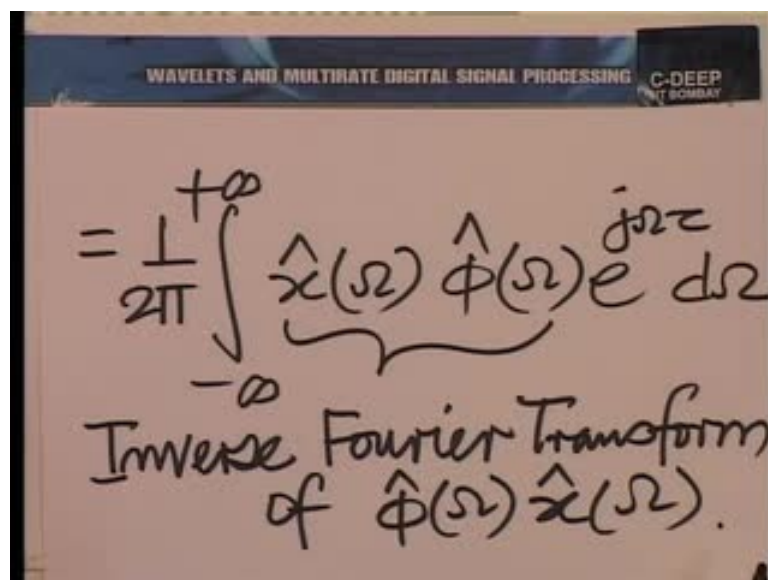
(Refer Slide Time: 28:32)



$$\int_{-\infty}^{+\infty} \phi(t+\tau) x(t) dt = \frac{1}{2\pi} \int_{-\infty}^{+\infty} \phi(t+\tau) \hat{x}(\omega) d\omega$$

Let us confine to real functions, and then I have, this is of course, equal from parseval's theorem to the Fourier transform of phi t plus tau times the Fourier transform of x integrate it to overall omega and this is easy to evaluate.

(Refer Slide Time: 29:09)



$$= \frac{1}{2\pi} \int_{-\infty}^{+\infty} \hat{x}(\omega) \hat{\phi}(\omega) e^{j\omega\tau} d\omega$$

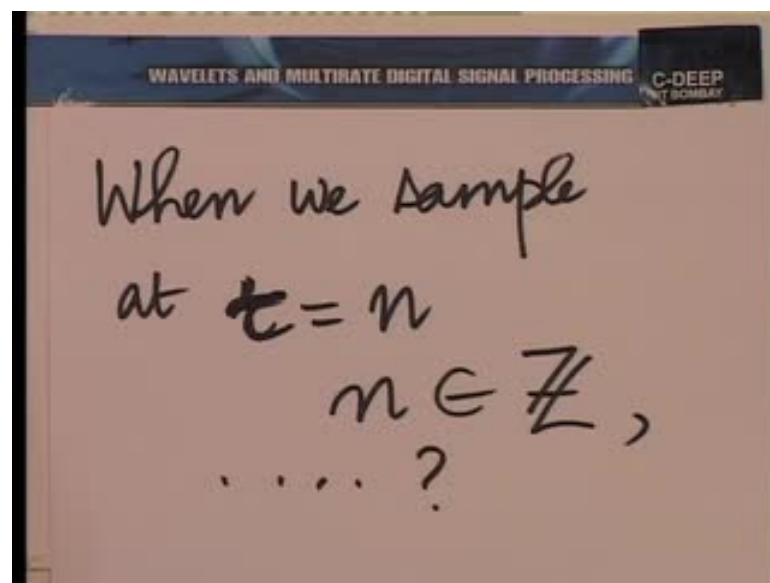
Inverse Fourier Transform of $\hat{\phi}(\omega) \hat{x}(\omega)$.

So, essentially we have a product of Fourier transforms x and phi, x cap and phi cap multiply it together, and then an inverse Fourier transform is being computed at the point tau. So, this is like, you know even if you were to use the complex function, the only change would be here, they would need you need to put a complex conjugate there, that

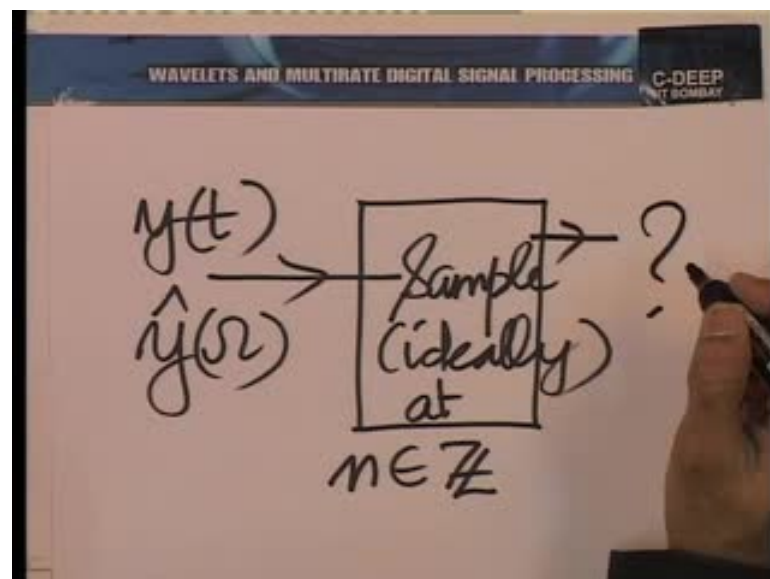
is why I said that? That is not such a serious issue at the moment; we will just focus on real functions and interpret.

So, here when you multiply by $\phi(\omega)$, you are in effect doing some kind of a low pass filtering. And when you take the inverse Fourier transform your calculating, what comes out of that crude low pass filter, whose impulse response is essentially ϕ , essentially ϕ ; I mean, do not worry about inversions or you know time inverse, it relates to ϕ , very closely ϕ .

(Refer Slide Time: 31:08)

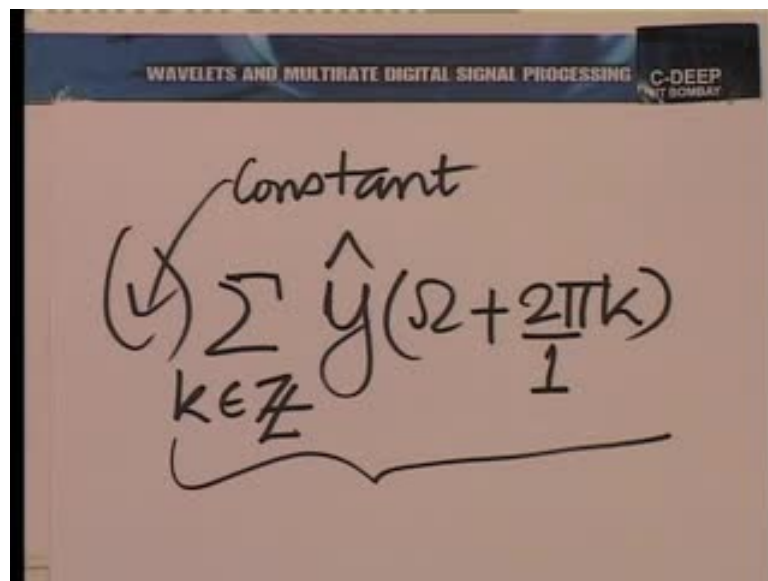


(Refer Slide Time: 31:33)



Now, what we are saying is when you sample this, when you put tau equal to all the integers; so, if you take this and substitute tau by different integer values. So, when we sample at tau equal to n, n all integers, what is going to happen? We are going to take the original Fourier transform, you see when we sample, if you take a function, let us say y t with Fourier transform y cap omega, and you sample, this sample ideally if you like at all integers that essentially, means, your sampling it sampling rate of one.

(Refer Slide Time: 32:28)

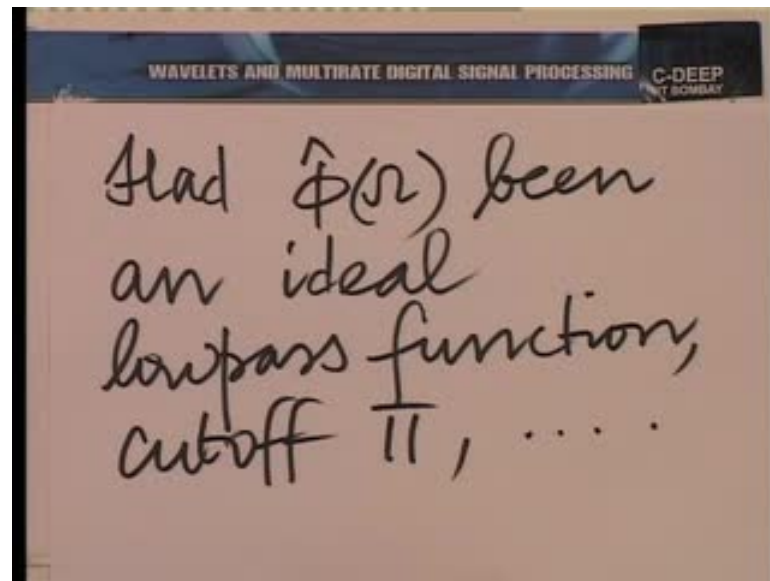


$$(\checkmark) \sum_{k \in \mathbb{Z}} \hat{y}(\omega + \frac{2\pi k}{1})$$

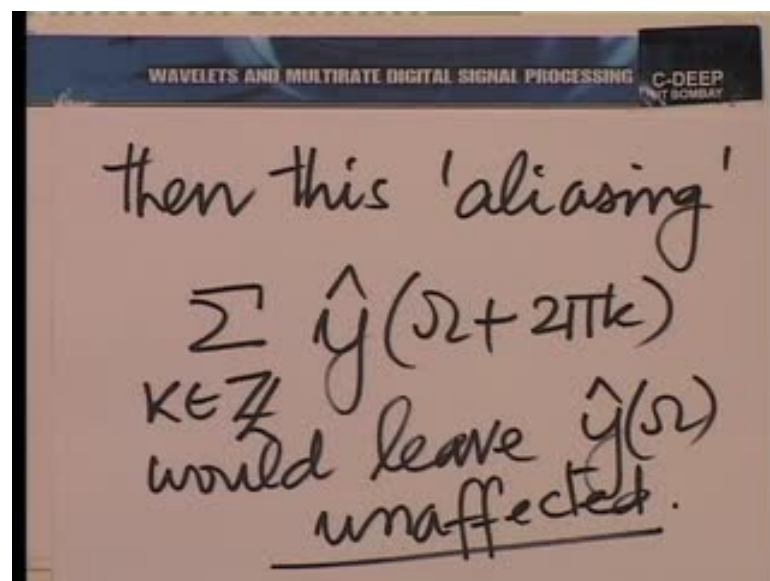
So, that amounts to taking the original Fourier transform, translating it by every multiple of 2 pi divided by 1, which is 2 pi on the angle of frequency axis, and adding up this translates. So, let me write that down in terms of an algebraic expression; what we are doing essentially is, we are taking the original Fourier transform translating it by every multiple of 2 pi divided by 1, if you please every multiple of that, and summing up these translates. Some constant possibly that constant relates to the sampling process; let us ignore that constant for the moment or attention is here.

So, in order to reconstruct y from its samples what should we have to desire? We should have desire that this translates do not interfere with the original. So, it would have really been nice, if we had been able to ensure that this carbon copy is created by y cap omega plus 2 pi k are non-overlapping with the original, and that is ensured by ensuring that the low pass filter cuts off at capital omega equal to pi.

(Refer Slide Time: 33:33)

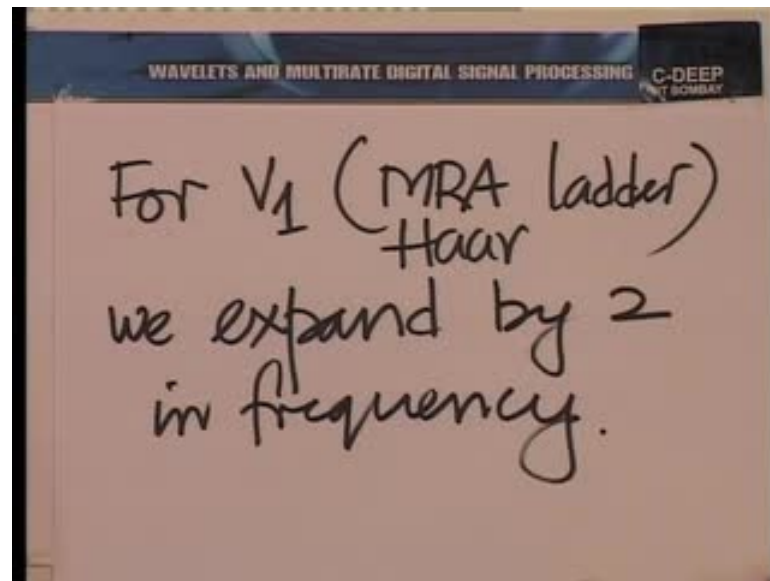


(Refer Slide Time: 34:00)

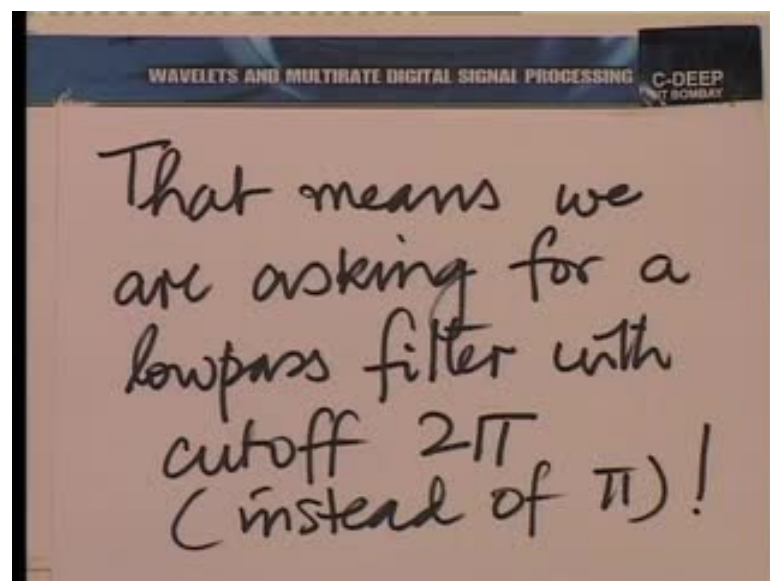


Let me sketch that for you; had ϕ cap ω being an ideal low pass function with a cutoff of π , then, then, this aliasing process would leave y cap ω unaffected; so, that is the ideal towards which we are striving as for as ϕ goes. Now, what is the ideal towards which we are striving as for as ψ goes, let us see.

(Refer Slide Time: 35:06)



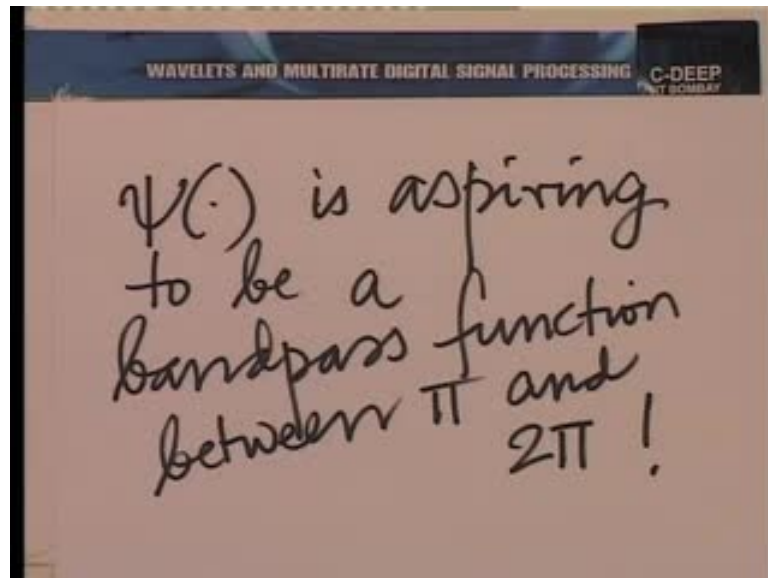
(Refer Slide Time: 35:40)



You see when you go from v_0 which is what brought us to ϕ to v_1 , what is v_1 ? Just essentially v_0 , but compressed our factor of 2 in time, and therefore, expanded by a factor of 2 in frequency. So, for v_1 , I am talking about the ladder - MRA ladder haar ladder, we expand by 2 in frequency; we are talking about frequency domain behavior. So, we expand by 2 in frequency; that means, we are asking for a low pass filter with cut off 2π instead of π .

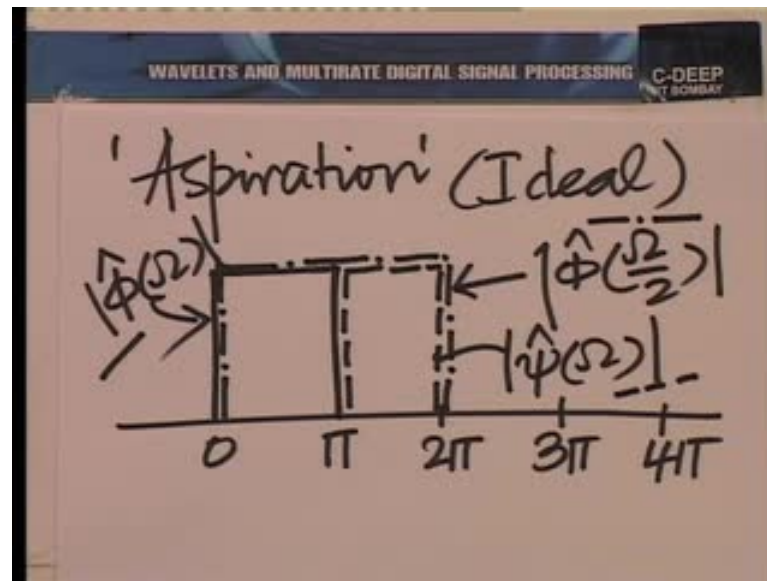
Now, we also have an interpretation for the incremental subspace; obviously, if v_0 is going to contain information between 0 and π , and v_1 is going to contain information between 0 and 2π , then the difference subspace w_0 should contain the information between π and 2π simple.

(Refer Slide Time: 36:40)



So, what we are saying in effect is, ψ is aspiring to be a band pass function between π and 2π . And of course, this is for going from v_1 , from v_0 to v_1 , when you go from v_{-1} to v_0 , you use the corresponding dilate of ψ which is aspiring to be a band pass function between $\pi/2$ and π . When you go from v_1 to v_2 , then you bring in a dilate of ψ which aspires to be a band pass function between 2π and 4π and so on.

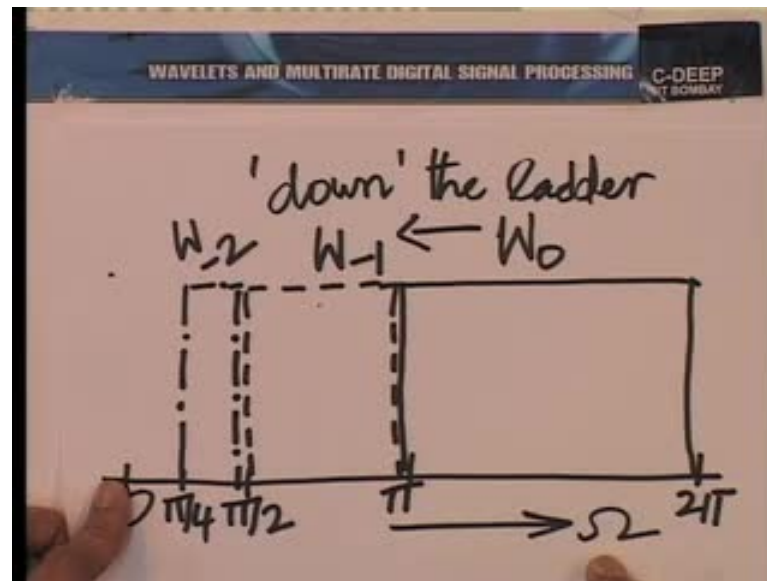
(Refer Slide Time: 37:50)



Let us draw the ideal situation. So, we aspiration the ideal is the following, this is the aspiration for ϕ at least in terms of magnitude; I will show the aspiration for the corresponding ϕ of ω by 2, and then I used a kind of dash line here, to show the aspiration for ψ .

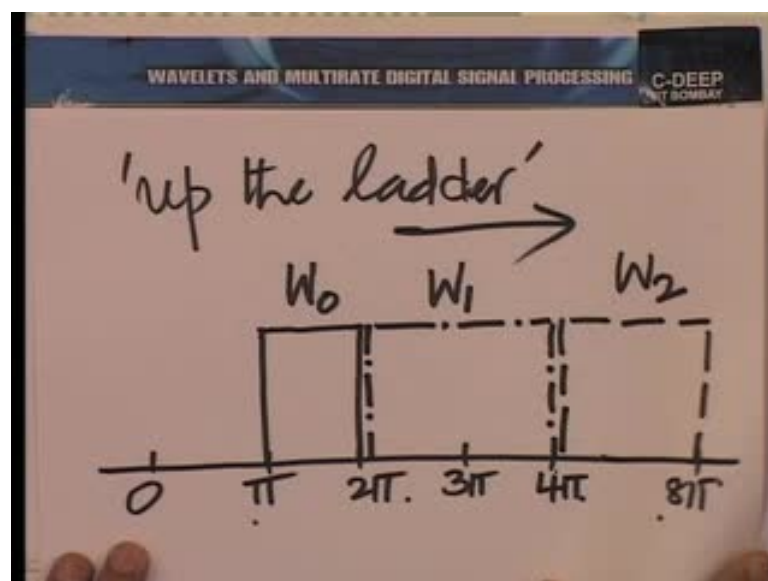
This is the ideal towards which we are going this one dot dash, this one solid, this one only dashes, now things are began to fall into place. In fact, now, we can also see what we need, when we forget about ϕ entirely? And use only ψ whatever we doing in the frequency domain or rather what are we aspiring to do? So, you know you can what I am saying is instead of thinking of all the shells up to a point, and removing one shell think of the whole **onion**, as only shells only sites.

(Refer Slide Time: 40:18)



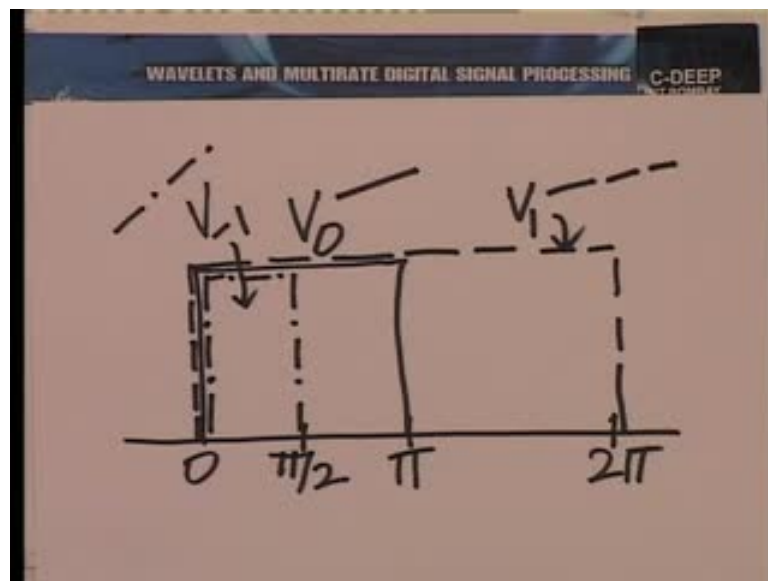
So, what is happening? Then the following is happening in the frequency domain. In fact, now I need to work carefully around 0; so, I will draw a big pi here, and a big 2 pi there; so we will start with w_0 so this is w_0 ; w_{-1} will essentially do this ideally; w_{-2} would be here between $\pi/4$ and $\pi/2$ and so on. Each time you go towards 0, you are contracting this band by a factor of 2, and therefore, both the center frequency and the bandwidth have been reduced by a factor of half, and of course, you can visualize going in this direction 2.

(Refer Slide Time: 41:55)



So, just for completeness, I think I should draw w_1 and w_2 , though not on the same graph, it is difficult to do; so, we will draw it separately. So, to be specific we should say down the ladder here, and up the ladder here. I will show two steps not quite proportional, but that is ok, this is w_0 ; w_1 will essentially take this from 2π to 4π ; w_2 will cover 4π to 8π here, this is 8π , please note. Again as I said forgive my drawing it is not quite proportional, but it is indicative π here, 2π here, 4π there 8π there.

(Refer Slide Time: 43:38)



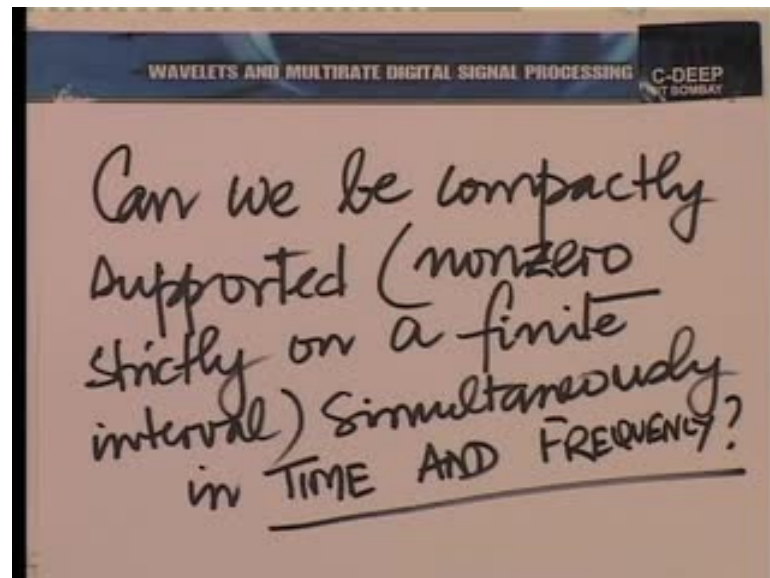
Now, we know what we are doing? As we go up the ladder, we are going to double the center frequency each time, and double the bandwidth ideally. And once again, let us show the behavior, as far as the spaces we go; so, here we show what happens with w , let show what happens with v . So, I could it show just on one, so I will just show for completeness three of them. This is what v_0 does, this is what v_1 does; so, v_0 is the solid line, v_1 is just the dash line and v_{-1} is the dot dash line, that I am drawing now.

So, these are what are called the complete subspaces, these are well, I should not use the word complete in the rigorous sense, when these are the entire set of shells up to that shell, and the others which we grow a minute ago; the w 's were just one peel or one shell at a time.

Now, we understand perfectly what we are doing in frequency, we are trying to do. And now, we also understand perfectly where the challenge lies, we are aspiring to do this

and we also wanted to do something similar in time. We want to confine ourselves to a certain region of time, and we also want to focus on a particular region of frequency, ideally focusing means being only in that region and 0 outside.

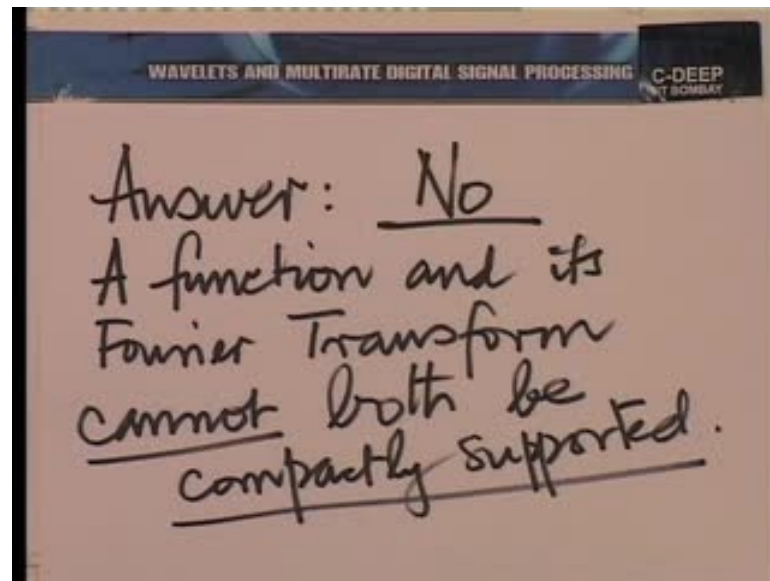
(Refer Slide Time: 45:56)



So, the first question that we need to answer is, is it exactly possible? Can we be compactly supported in time and frequency simultaneously? Let us put the question; question is also important here. Can we be compactly supported, this is the technical term compactly supported I shall spend too much of time on explaining the detail, but non-zero strictly on a finite interval is a simple way of saying it at the moment; simultaneously, in time and frequency

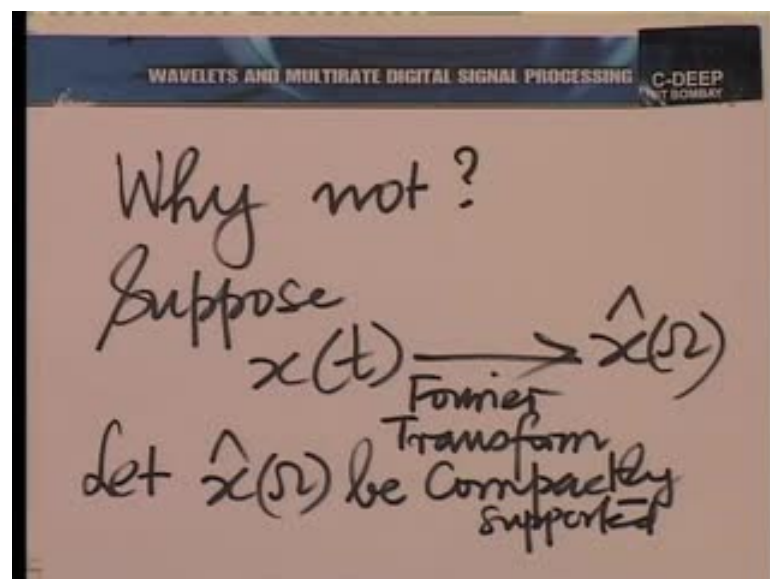
And unfortunately or may be fortunately, because it brings up and opens up a whole new subject, the answer is no. If you talk about exact behavior, it is impossible to be compactly supported in both domains; that is not a very deep result in the theorems of Fourier analysis, though it is an important result. It is the relatively weaker; weaker in the sense not of requirement, but in terms of the depth of proof or depth of implication.

(Refer Slide Time: 48:03)

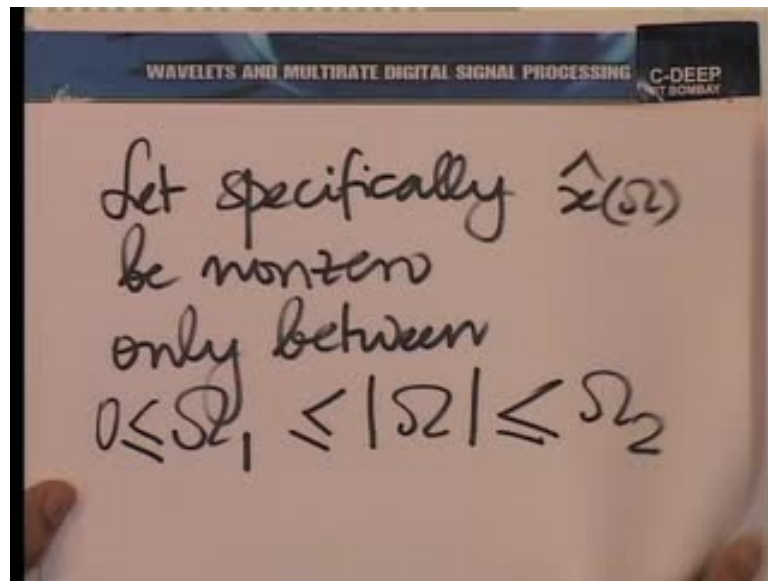


It is more easily proved, easier to indicate or to justify; you cannot be compactly supported in time and frequency simultaneously. Let us make that statement very clear; answer: no. A function and its Fourier transform cannot both be compactly supported; in fact, I shall give an indication of the idea behind the proof, and I shall leave it to the class, I shall leave it to the students who are listening here to develop deeper.

(Refer Slide Time: 49:04)



(Refer Slide Time: 49:51)



The idea behind the proof, why not? Well **suppose $x \in \mathcal{F}$** , suppose $x \in \mathcal{F}$ has the Fourier transform $\hat{x}(\omega)$ or $\hat{x}(\omega)$, let us take ω if you like. And let $\hat{x}(\omega)$ be compactly supported; in other words, let us specifically $\hat{x}(\omega)$, the non-zero only between ω_1 and ω_2 in magnitude; of course, needless to say ω_1 is greater than equal to 0, and therefore also ω_2 .

(Refer Slide Time: 50:39)

$$x(t) = \frac{1}{2\pi} \int_{-\omega_2}^{\omega_2} \hat{x}(\omega) e^{j\omega t} d\omega$$
$$+ \int_{-\omega_2}^{-\omega_1} \dots (\text{same integrand}).$$

Then it is very clear that the Fourier transform, the inverse Fourier transform which gives us back $x(t)$ is a finite integral. So, $x(t)$ is then going to have a finite integration involved plus the same thing on the negative, and the same integrand.

Now, the central idea in the proof is the following: I can take derivatives on both sides, and I remember I had a finite integral on both sides; when I take a derivative with respect to time of $x(t)$, then if I look at the integral here that derivative essentially acts only on the $e^{j\omega t}$ part, and that operation of taking the derivative into the integral is valid, because this is a finite integral; the same thing holds good for the second integral here. So, effect you are talking about a function which has an infinite number of derivatives, because after all each of the integrals involve could be a finite integral here.

So, I have, just I am not really giving you a rigorous proof, I am just indicating the central idea in the proof; it relates to the fact that the function which is compactly supported in the frequency domain must have a certain kind of smoothness as seen in time. No matter how many derivatives you take here, you do have an expression for the derivative there, that is derivative exist and in fact can also be shown to be continuous.

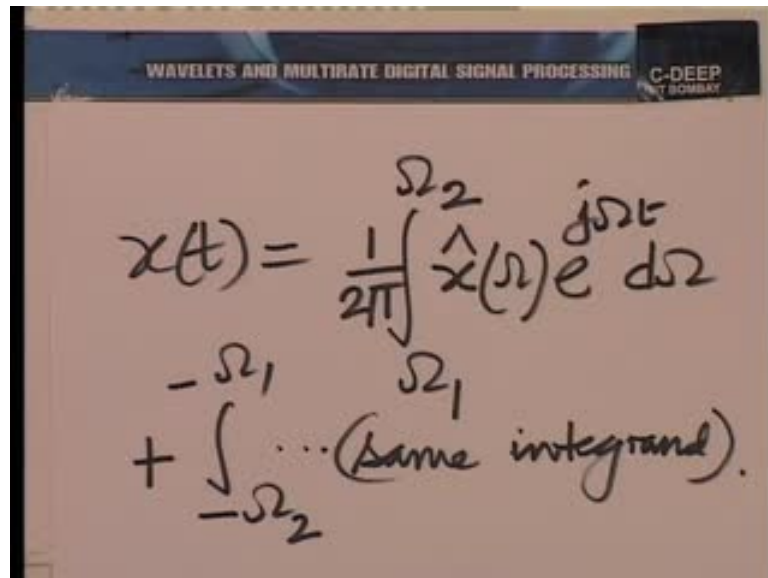
So, **there is a**, there is the quality of infinite smoothness in that function $x(t)$ in some sense as I said, all this is only indicative of the proof. Now, I encourage those of you, who are more mathematically minded to take this proof to completion; so that, because of this finite integral here, and the fact that the function must be smooth as much as you desire in terms of derivatives, it cannot be compactly supported in time as well. In effect what you saying is, you are asking for an analytic function, a function which has an infinite number of smooth derivative to be compactly supported in time, there is a problem there.

Well that was indicative of the proof that was indicative of the central idea as to why you cannot have compactly supported functions, both in time and frequency together. And this is, where the whole challenge starts, but now we need to ask a slightly more relaxed question, and that will be the issue that we shall discuss in much greater depth in due course now.

The question is suppose we do not ask for strict compact support, that means, suppose we are not saying that function must be **non-zero outside or sorry non-zero only inside**, a certain compact interval, only inside the certain finite interval, and 0 everywhere else we do not mind a certain amount of energy of that function or most of the function in a

certain sense, being concentrated in a certain region in time and also in frequency. Then can we get a function which is both compactly or not compactly, but in that sense restricted in time and frequency, and of course, as we expect the answer is - yes, if you are willing to give up a little bit we can get something.

(Refer Slide Time: 50:39)



$$x(t) = \frac{1}{2\pi} \int_{-\Omega_1}^{\Omega_2} \hat{x}(\Omega) e^{j\Omega t} d\Omega + \int_{-\Omega_2}^{-\Omega_1} \dots (\text{same integrand}).$$

If you are willing to give up exact compact support, so if you are willing to allow some leakage outside that region of time, and therefore, also outside the certain region of frequency, but be content with the fact, that in a certain weaker sense the function is concentrated in a certain region of time, and in a certain region of frequency; then can one first have this kind of broad concentration in time and frequency together, well the answer is yes, because it depends on what you mean by that weaker sense of concentration; in fact, ϕ and ψ are in that sense concentrated both in time and frequency in a weaker sense; if you focus only on the main lobe, and of course, the main lobe has a certain amount of the energy, and yes indeed, of course, ϕ is simultaneously localized in time and frequency.

So, what is that general sense that we are going to allow? Well at sense will come from essentially either what we might call the statistical property of variants or if you want to use a mechanical analogy, the idea of centre of mass and radius of gyration or the volumetric occupancy of a body.

So, we will think both of the function and its Fourier transform as one-dimensional bodies, and we can think of their center of masses. And then we could think of how much the body spreads around the center of mass, by using what is called the idea of radius of gyration. And other perspective is, if you think of probability density functions based on the functions, and it is Fourier transform, you could ask what is the mean of that density, either in the time domain or in the frequency domain, and then you could ask what are the variants of the density, again either in the time domain or in the frequency domain?

And now, there is a clear way to formulate, can we have finite variants both in time and frequency? And there as we expect the answers going to be yes, that is not a problem. Now, the more difficult question how small can the function be simultaneously in time and frequency, in this broader sense? So, how small can you make the variants in time and frequency simultaneously? That is the deeper question, and that is the whole idea behind the uncertainty principle.

In fact, now we are beginning to understand why we needed to go to better and better multiresolution analysis? Why could we not be happy with the haar? The haar is somewhat concentrated in frequency, but well concentrated in time. I had one point asked you to find out the Fourier transform of the dobash functions as well.

So, you know if you look at the dobash functions as you go from length 4 to length 6 to length 8, and if you look at the Fourier transform, you would find that they are slightly better approximations to that ideal low pass filter with cutoff π and ideal band pass filter with band between π and 2π as we desired.

So, what we are going to do subsequently now, **is essentially to**, essentially bring out this concept of uncertainty more deeply, and then to investigate whatever we have been doing in the language of uncertainty starting from this point onwards. Thank you.