

Fiber Optics
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Lecture – 27
Recap: Propagation Characteristics

Till now we have studied the propagation characteristics and salient features of optical fiber and optical wave guides. So, I think now it is time to stop for a while, and have a recap of whatever we have done till now.

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EM WAVE PROPAGATION IN INFINITELY EXTENDED DIELECTRIC

n^2 is independent of x , y , and z

An *em-wave* propagating in z -direction

$$\vec{E}(z, t) = \vec{E}_0 e^{i(\omega t - kz)}$$

$$\vec{H}(z, t) = \vec{H}_0 e^{i(\omega t - kz)} \quad E_{0z} = 0 \text{ and } H_{0z} = 0 \quad \rightarrow \text{Transverse waves}$$

For a linearly polarized wave which is polarized in x -direction and propagating in z -direction

$$\vec{E} = \hat{x} E_0 e^{i(\omega t - kz)}$$

$$\vec{H} = \hat{y} H_0 e^{i(\omega t - kz)}, \quad H_0 = \frac{k}{\omega \mu} E_0$$

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So, in order to understand propagation characteristics of optical fibers, we have first understood how an electromagnetic wave propagates in an infinite extended dielectric medium. So, infinitely extended dielectric medium n square is independent of x y and z . So, it is homogeneous medium then an em wave propagating in z direction, has an electric field and magnetic field given by $E(z, t)$ is equal to $E_0 e^{i(\omega t - kz)}$ and H is equal to $H_0 e^{i(\omega t - kz)}$. So, this has x y and z all the components, and we had shown that for such a case E_{0z} and H_{0z} are 0 so, this is a transverse wave

For a linearly polarized wave which is polarized in x direction and propagating in z direction, I can write the other components as E is equal to $\hat{x} E_0 e^{i(\omega t - kz)}$ and H is equal to $\hat{y} H_0 e^{i(\omega t - kz)}$, where

H_0 and E_0 are related by this. So, so I see two things one you should notice that this E_0 and H_0 they are constants they do not depend upon a spatial coordinates, and also that this is purely real. So, h and e are in phase. I can also write it down in terms of B , since B is equal to μ times h . So, B would be y cap $B_0 e$ to the power $i\omega t - kz$. So, the magnitude amplitude of B is 1 over v times E . And so the amplitude of B is really much smaller than the amplitude of e , then we had considered propagation in wave guides.

(Refer Slide Time: 03:02)

EM WAVE PROPAGATION IN WAVEGUIDES

$n^2(x, y)$ CHANNEL WAVEGUIDE

$$\vec{\mathcal{E}}(x, y, z, t) = \vec{E}_0(x, y)e^{i(\omega t - \beta z)}$$

$$\vec{\mathcal{H}}(x, y, z, t) = \vec{H}_0(x, y)e^{i(\omega t - \beta z)}$$

Modes of the system

$n^2(x)$ PLANAR WAVEGUIDE

$$\vec{\mathcal{E}}(x, y, z, t) = \vec{E}_0(x)e^{i(\omega t - \gamma y - \beta z)}$$

$$\vec{\mathcal{H}}(x, y, z, t) = \vec{H}_0(x)e^{i(\omega t - \gamma y - \beta z)}$$

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So, now if n square depends upon x and y , and the wave is propagating in z direction, then the electric and magnetic fields associated with the wave, light wave would be given by these and we notice here now these E_0 and H_0 are not constants, but they depend upon x and y . So, there is some profile, there is some electric field profile and magnetic field profile which propagates in z direction with certain propagation constant β . So, these are the modes of the system; if instead of having n square x y I have variation only in x direction.

So, n is a function of x only then it is a planar wave guide, and in such a case I can write down the electric and magnetic fields as E_0 of x e to the power $i\omega t - \gamma y - \beta z$, and h like this. So, here what I see that this E and H they depend only on x . So, they are the functions of x and they form the mode of the planar wave guide ok.

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PLANAR OPTICAL WAVEGUIDE
 $n^2(x)$ z -propagation

TE-MODES (Non-vanishing E_y, H_x, H_z)

$$i\beta E_y = -i\omega\mu_0 H_x$$
$$\frac{\partial E_y}{\partial x} = -i\omega\mu_0 H_z$$
$$-i\beta H_x - \frac{\partial H_z}{\partial x} = i\omega\epsilon_0 n^2(x) E_y$$

TM-MODES (Non-vanishing H_y, E_x, E_z)

$$i\beta H_y = i\omega\epsilon_0 n^2(x) E_x$$
$$\frac{\partial H_y}{\partial x} = i\omega\epsilon_0 n^2(x) E_z$$
$$-i\beta E_x - \frac{\partial E_z}{\partial x} = -i\omega\mu_0 H_y$$

TE-MODES

$$\frac{d^2 E_y}{dx^2} + [k_0^2 n^2(x) - \beta^2] E_y = 0$$

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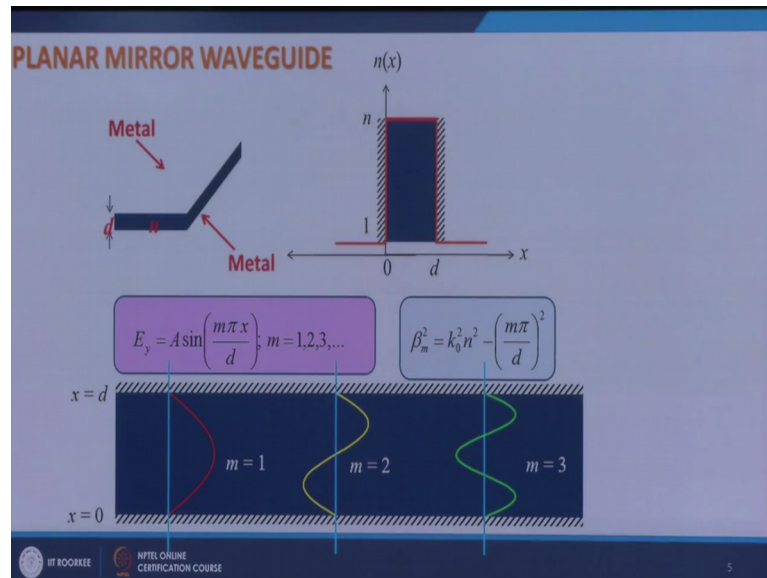
And I can choose either beta here or gamma depending upon in what direction the mode is propagating.

So, if I choose the direction of propagation is z then I can put gamma is equal to 0. Then we considered planar optical wave guide with n square of x variation and propagation in z direction then we had seen that we can categorize the modes of such a wave guide into two categories two polarizations, one is TE modes, where the non vanishing components of electric and magnetic fields are E_y, H_x and H_z , and these three components are related to each other by these three equations.

I can have another category that is TM mode TM polarization, where the non vanishing components are H_y, E_x and E_z and these components are related to each other by these three equations. So, for TE modes for example, I can form a wave equation I can eliminate H_x and H_z and get an equation in E_y , which tells me how E_y changes with x and this is what I want to find out because I want to find out E_y of x which is the mode.

So, that satisfies this equation. So, for any given n square of x can solve this equation find out the values of beta and corresponding E_y . So, these are the modes.

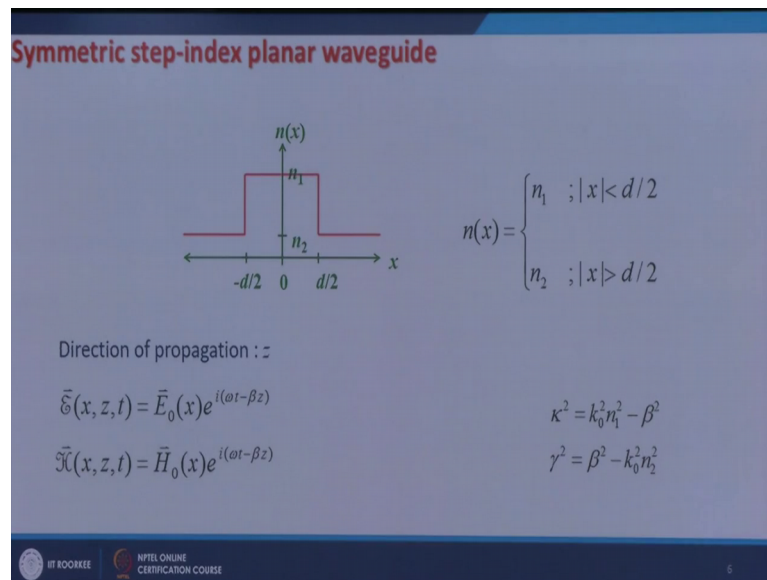
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I work out first an example of planar mirror wave guide where I have very thin layer of dielectric medium of refractive index n and width d , which is coated with metal on top and bottom. So, the refractive index profile looks like this and when we worked out the modes of this then the electric field associated with this is like this, and the propagation constants are like this.

So, I have modes discrete modes of such a wave guide corresponding to different values of m . So, they look like this they look like the modes of a string modes of a (Refer Time: 06:32) string; then we consider symmetric step index planar wave guide, dielectric wave guide where the refract index profile is like this.

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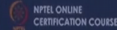


So, I have a high index layer sandwiched between two lower refractive index layers. The width of the high index layer is d which extends from x is equal to $-d/2$ to x is equal to $d/2$.

If the direction of propagation again I considered z , then the electric and magnetic fields associated with the light would be given by this and now my task is to find out this E_0 of x and H_0 of x . Refractive index profile of this wave guide is given like this where n is equal to n_1 for $|x| < d/2$, it is n_2 for $|x| > d/2$, and n_1 is greater than n_2 . We define some κ and γ as κ^2 is equal to $k_0^2 n_1^2 - \beta^2$, and γ^2 is equal to $\beta^2 - k_0^2 n_2^2$. So, when I find out the guided modes of this TE modes and TM modes.

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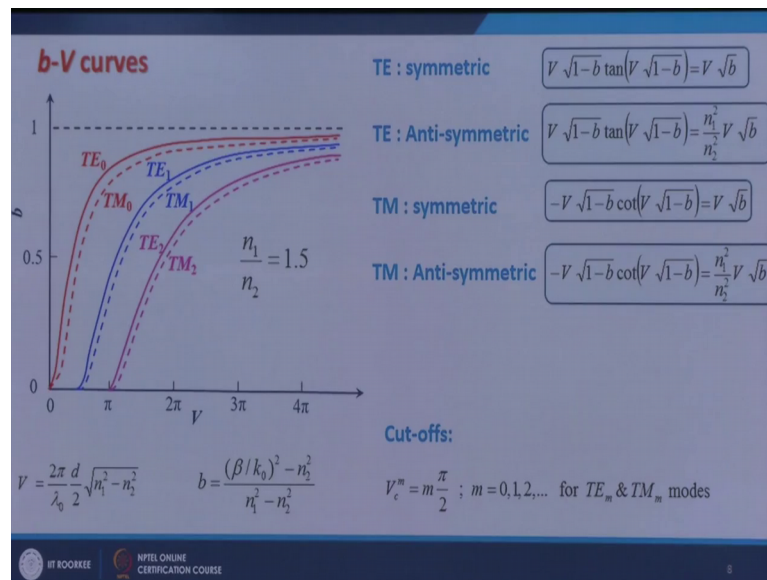
TE-Modes E_y, H_x, H_z	TM-Modes H_y, E_x, E_z
Symmetric modes	Symmetric modes
$E_y(x) = \begin{cases} A \cos \kappa x; & x < d/2 \\ C e^{-\gamma x}; & x > d/2 \end{cases}$	$H_y(x) = \begin{cases} A \cos \kappa x; & x < d/2 \\ C e^{-\gamma x}; & x > d/2 \end{cases}$
$\frac{\kappa d}{2} \tan \frac{\kappa d}{2} = \frac{\gamma d}{2}$	$\frac{\kappa d}{2} \tan \frac{\kappa d}{2} = \frac{n_1^2 \gamma d}{n_2^2}$
Anti-symmetric modes	Anti-symmetric modes
$E_y(x) = \begin{cases} B \sin \kappa x; & x < d/2 \\ D \frac{x}{ x } e^{-\gamma x}; & x > d/2 \end{cases}$	$H_y(x) = \begin{cases} A \sin \kappa x; & x < d/2 \\ D \frac{x}{ x } e^{-\gamma x}; & x > d/2 \end{cases}$
$-\frac{\kappa d}{2} \cot \frac{\kappa d}{2} = \frac{\gamma d}{2}$	$-\frac{\kappa d}{2} \cot \frac{\kappa d}{2} = \frac{n_1^2 \gamma d}{n_2^2}$

So, for TE modes again I have non vanishing components E_y , H_x and H_z , and if I solve the wave equation that the differential equation for E_y , then I find that I can further categorize these modes into symmetric and anti symmetric because their structure is symmetric about x is equal to 0. For symmetric modes the field profiles are given by E_y of x is equal to a cosine κx in the in the guiding film in the high index region, and it is exponentially decaying in the lower index surrounding. The values of beta they satisfy this equation this is the transcendental equations.

So, we solve this transcendental equation get the values of beta which are the solutions of this equation, and for those values of beta I find out E_y . So, in this way I get the complete solution the mode field profiles as well as the propagation constants. Similarly for anti symmetric modes this is the field solution and this is the transcendental equation. The same thing I can do for TM modes for which the non vanishing components are H_y , E_x and E_z , I have symmetric modes and anti symmetric modes and these are the transcendental equations. Now only difference that you would see in between this and this is a factor of n_1^2 over n_2^2 , this comes from the boundary conditions.

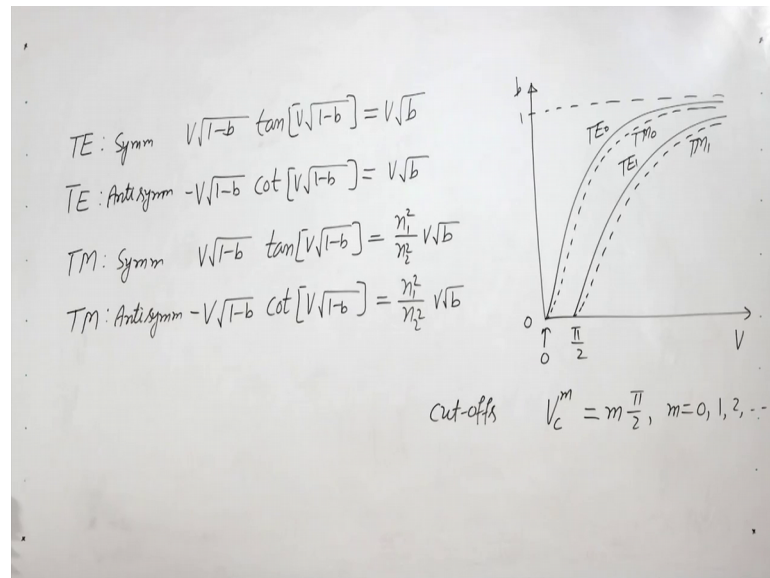
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So, in a wave guide a very important diagram is bV diagram, where b is the normalized propagation constant defined like this, and V is the normalized frequency defined like this which contains all the information which you require about the wave guide and the wave length.

So, I can now translate the transcendental equations corresponding to TE and TM symmetric and anti symmetric modes into these normalized parameters like this, and then I can solve these equations to generate these curves. The idea of and advantage of generating these curves is that these curves are universal. So, if I talk about TE modes, then this is the equation for symmetric and this is the equation for anti symmetric. There should not be this is the equation for TE is symmetric mode and this is the equation for TE anti symmetric mode.

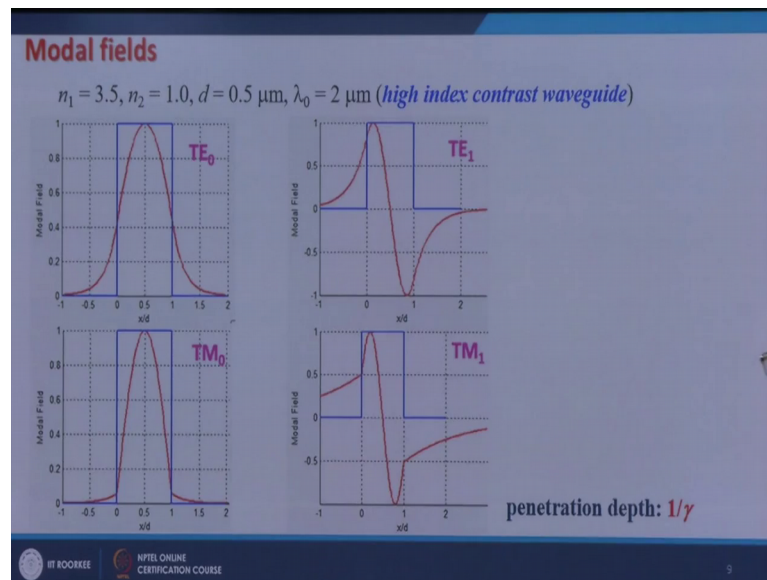
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So, let me make this correction here, TE symmetric is $V \sqrt{1-b} \tan V \sqrt{1-b} = V \sqrt{b}$ and TE anti symmetric is $-V \sqrt{1-b} \cot V \sqrt{1-b} = V \sqrt{b}$ TM symmetric is $V \sqrt{1-b} \tan V \sqrt{1-b} = \frac{n_1^2}{n_2^2} V \sqrt{b}$ and TM anti symmetric is $-V \sqrt{1-b} \cot V \sqrt{1-b} = \frac{n_1^2}{n_2^2} V \sqrt{b}$. So, I can have the transcendental equations in normalized parameters, and by solving these equations for different values of V , I can generate $v b$ curves. So, those $v b$ would look like this is V and this is b ; for guided modes b would vary from 0 to 1. So, this would be TE_0 mode TE_0 this would be TM_0 mode.

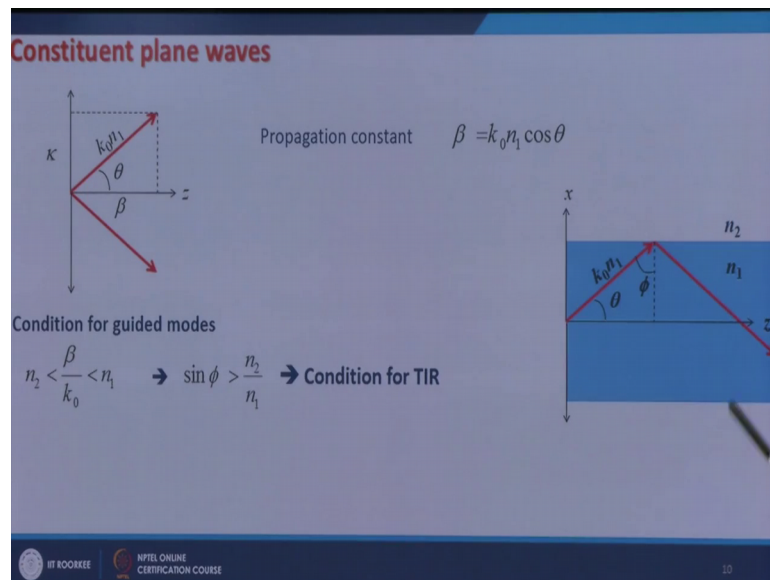
This would be TE_1 mode; this would be TM_1 mode and so on. So, I can generate these universal curves $b v$ diagrams, and if I now know the wave guide and the wave length then I can just calculate the value of V and go to these curves find out the value of V and find out the value of propagation constants. The cut offs of these modes are here. So, this is 0 this is $\pi/2$ and the cut offs are given by cut offs V_c^m for TE TM both is equal to $m \pi/2$, where m is equal to 0, 1, 2 and so on. So, how the modal fields look like.

(Refer Slide Time: 14:44)



So, if I take high index contrast wave guide, I have taken high index contrast wave guide here in order to show the difference between TE TM modes for n_1 is equal to this n_2 is equal to this, d half a micron λ_0 naught 2 micro meter. So, this is how the TE_0 mode would look like and this is how the TM_0 mode would look like. You can see the discontinuity in at in the slope at the boundaries. This is how TE_1 mode look like this is how TM_1 mode would look like I also define penetration depth, the extent up to which the field would penetrate into n_2 region. So, it is given by $1/\gamma$. We have seen that these modes are nothing, but the superposition of two plane waves one going in exact plane like this and one going in exact plane like this.

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So, upward and downward, making angles plus minus theta from z axis. So, what we see that there is one plane wave going like this, another plane wave going like this. So, if I take the components in x direction and in z direction, then in x direction I get two counter propagating waves which give you standing wave pattern. And these modes these modes are nothing, but the standing wave patterns. So, the field stands in x direction does not propagate in x direction, but this is standing wave pattern propagates in z direction with propagation constant beta, which can be given by $k_0 n_1 \cos \theta$.

And you had also seen that the condition for guided mode can be translated to the condition for total internal reflection. So, this is also we had seen then what is the power associated with the mode? To calculate the power associated with the mode we found out the pointing vector which gives us the intensity, and since the electric and magnetic field associated with light or fluctuating rapidly, k at a frequency of about 10^{14} to 10^{15} hertz. So, we need to take the average time average of these.

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Power Associated with Modes

Average Intensity $\langle \vec{S} \rangle = \langle \vec{E} \times \vec{H} \rangle$ $S_z = \frac{\beta}{2\omega\mu_0} E_y^2(x)$

Power (per unit length in y-direction) $P = \frac{\beta}{2\omega\mu_0} \int_{-\infty}^{\infty} E_y^2(x) dx$

TE-modes $P = \frac{\beta}{2\omega\mu_0} \frac{1}{2} A^2 \left[d + \frac{2}{\gamma} \right]$

TM-modes $P = \frac{\beta}{2\omega\epsilon_0 n^2} \frac{1}{2} A^2 \left[d + \frac{2(n_1 n_2)^2 k_0^2 (n_1^2 - n_2^2)}{\gamma (n_2^4 k^2 + n_1^4 \gamma^2)} \right]$

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So, when we do this, then we find that only z component of S the pointing vector survives, and it is equal to beta over 2 omega mu naught E y square of x, this is for TE modes. So, from here I can find out the power per unit length in y direction, because y direction is infinitely extended for planar wave guide. So, the power per unit length in y direction would be given by beta over 2 omega mu naught integral minus infinity to plus infinity E y square of x dx. So, when I calculate this for symmetric planar wave guide, then for TE modes the power is given by this and for TM modes the power would be given by this.

These powers are in watts per meter then we had extended the analysis to asymmetric planar wave guide, where I have asymmetry here this is the substrate this is the cover. So, the substrate and cover refractive index are different. So, n_s here is greater than n_c, and what we did we wrote down the fields in all the three regions. So, these are the fields electric fields corresponding to modes in all the three regions, and the beta the propagation constant will satisfy this transcendental equation. For TM modes the fields would be given like this and the transcendental equation becomes this.

(Refer Slide Time: 18:39)

TM-modes



$x < -d$ (substrate) $H_y(x) = D \exp(\gamma_s x)$
 $-d < x < 0$ (film) $H_y(x) = B \exp(ik_f x) + C \exp(-ik_f x)$
 $x > 0$ (cover) $H_y(x) = A \exp(-\gamma_c x)$

$$\tan(\kappa_f d) = \frac{\frac{n_f^2 \gamma_s}{n_s^2 \kappa_f} + \frac{n_f^2 \gamma_c}{n_c^2 \kappa_f}}{1 - \frac{n_f^2 \gamma_s}{n_s^2 \kappa_f} \frac{n_f^2 \gamma_c}{n_c^2 \kappa_f}}$$

Normalized Parameters

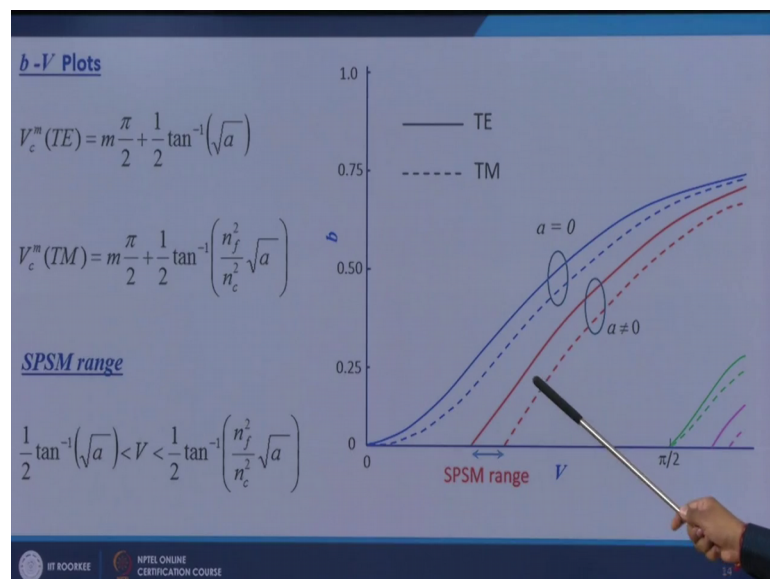
$$V = \frac{2\pi d}{\lambda_0} \sqrt{n_f^2 - n_s^2} \quad b = \frac{(\beta/k_0)^2 - n_s^2}{n_f^2 - n_s^2} \quad a = \frac{n_s^2 - n_c^2}{n_f^2 - n_s^2}$$

TE: $\tan(2V\sqrt{1-b}) = \frac{\sqrt{1-b} + \frac{b+a}{\sqrt{1-b}}}{1 - \frac{b}{\sqrt{1-b}} \frac{b+a}{\sqrt{1-b}}}$
TM: $\tan(2V\sqrt{1-b}) = \frac{\frac{n_f^2}{n_s^2} \frac{b}{\sqrt{1-b}} + \frac{n_f^2}{n_c^2} \frac{b+a}{\sqrt{1-b}}}{1 - \frac{n_f^2}{n_s^2} \frac{b}{\sqrt{1-b}} \frac{n_f^2}{n_c^2} \frac{b+a}{\sqrt{1-b}}}$



13

We had also defined normalized parameters for asymmetric planar wave guide. So, V is defined with respect to n_s . So, here you see n_f square minus n_s square not with n_c , because it is n_s that will govern the guiding condition. B is also defined with respect to n_s . So, beta square over k_0 square minus n_s square divided by n_f square minus n_s square, here we have another parameter which is asymmetry parameter which tells you how different n_s is with n_c . In normalized parameters I can have transcendental equation for TE modes and for TM modes like this, if I draw the $b-v$ curves here now then I see that the $b-v$ curves for asymmetric wave guides they go like this ok.

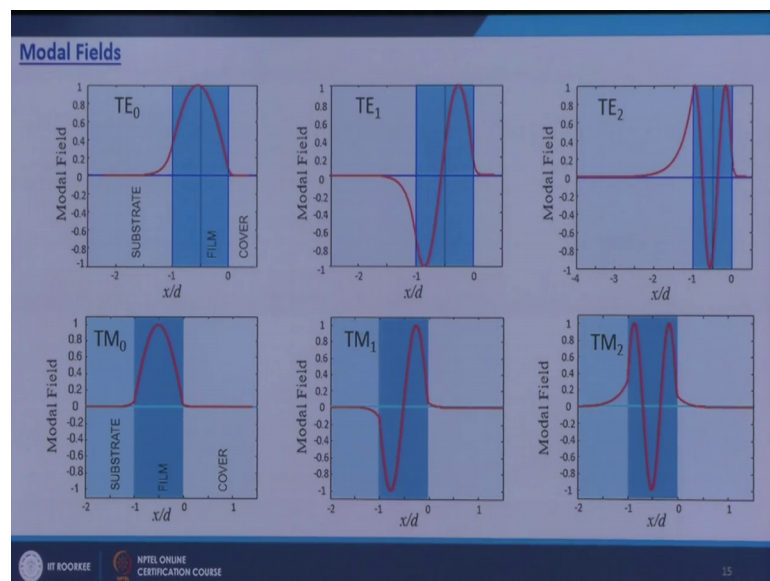
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What I notice that they do not have 0 cut off. TE 0 mode also does not have 0 cut off. If you look at the cut offs if you look at the cut offs then you can see that for symmetric planar wave guide the cut offs were $m\pi/2$, but now these cut offs would be shifted by this amount $\frac{1}{2} \tan^{-1} \sqrt{a}$ because of this asymmetry parameter. If a is equal to 0 then it becomes the symmetric planar wave guide. So, these I also notice that the cut offs of TE and TM modes are now different. In case of symmetric planar wave guide the cut offs of TE and TM both the modes were the same $m\pi/2$.

But now the TE and TM modes have different cut offs. So, this is TE 0 mode this is TM 0 mode TE 0 mode have cut off here, TM 0 mode has cut off here. Now if I have v which is somewhere here then I will guide only TE 0 mode and TM 0 mode is cut off. So, in the range of V from which goes from $\frac{1}{2} \tan^{-1} \sqrt{a}$ to $\frac{1}{2} \tan^{-1} \frac{nc^2}{a}$, I have single polarizations single mode wave guide, and this range is known as sp sm range. This is how the modal fields would look like, this is for TE 0 mode this is TM 0 mode this is TE 1, TM 1, TE 2, TM 2.

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We can notice that in the sub straight the field extends more because index contrast with the film is less while in the cover, the field very quickly decreases goes down to 0, because the index contrast is high. Then we worked out the modes of a step index optical fiber. So, in a step index optical fiber I have a core and cladding I have assumed to be extended to infinity.

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STEP-INDEX FIBER

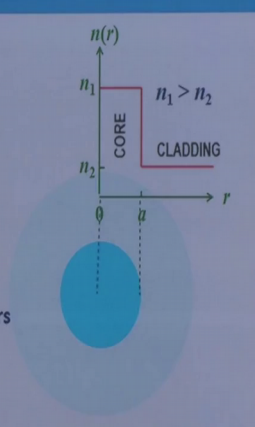
$$\psi(r) = \begin{cases} A J_l \left(\frac{Ur}{a} \right) \begin{cases} \cos l\phi \\ \sin l\phi \end{cases}; & r < a \\ B K_l \left(\frac{Wr}{a} \right) \begin{cases} \cos l\phi \\ \sin l\phi \end{cases}; & r > a \end{cases}$$

$$W = V\sqrt{b} \quad U = V\sqrt{1-b} \quad V = \frac{2\pi}{\lambda_0} a \sqrt{n_1^2 - n_2^2} \quad b = \frac{(\beta/k_0)^2 - n_2^2}{n_1^2 - n_2^2}$$

Transcendental equations in terms of normalized parameters

For $l = 0$ $V\sqrt{1-b} \frac{J_1(V\sqrt{1-b})}{J_0(V\sqrt{1-b})} = V\sqrt{b} \frac{K_1(V\sqrt{b})}{K_0(V\sqrt{b})}$

For $l \geq 1$ $V\sqrt{1-b} \frac{J_{l+1}(V\sqrt{1-b})}{J_l(V\sqrt{1-b})} = -V\sqrt{b} \frac{K_{l+1}(V\sqrt{b})}{K_l(V\sqrt{b})}$

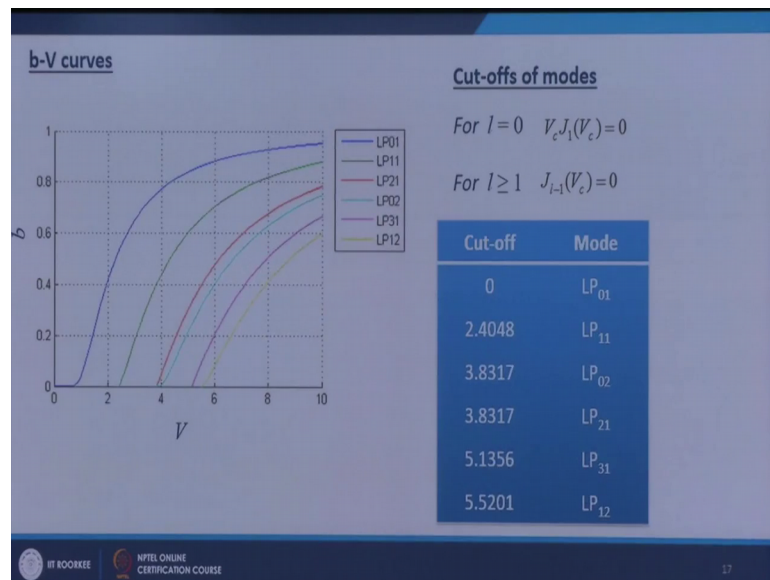


$$n(r) = \begin{cases} n_1; & r < a \\ n_2; & r > a \end{cases}$$

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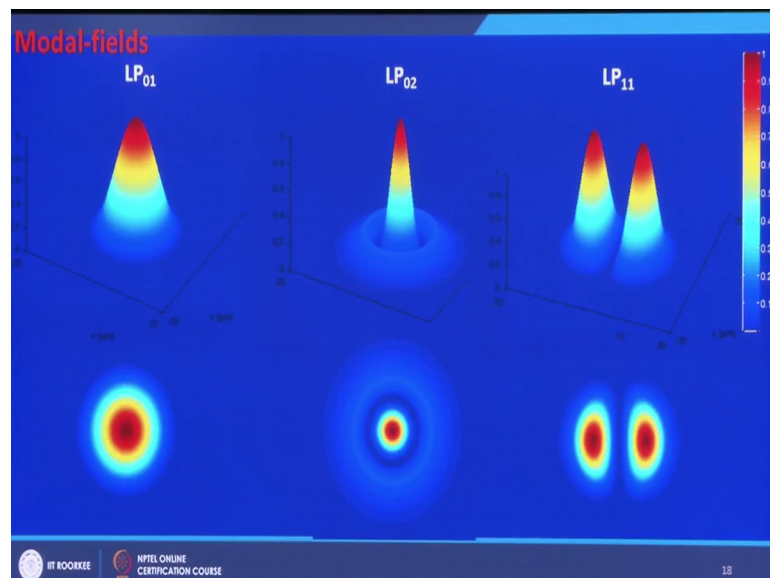
So, here I have the modes which are function of r and phi. So, this is the r solution $A J_l U r$ over a, and phi solution can be cosine l phi and sin l phi this is in the core, in the cladding the r solution is $k_l w r$ over a, and phi solution is again cosine l phi and sin l phi. Then we again we had defined TE normalized frequency and normalized propagation constant, and worked out the transcendental equations for different modes corresponding to l is equal to 0 and l greater than or equal to 1, because now I have phi directions. So, there would be discrete modes in phi direction also. So, which are governed by the values of l and the l assumes only integral values 0 1 2 3. So, for those I can then find out what are the r solutions, and the modes propagation constant would then be given by these two transcendental equations.

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So, if I do that. So, for different values of b different values of v I can find out the propagation constants and I can plot them. So, this is LP 0 1 mode, this is LP 11 mode this is LP 21, 0 2, 3 1, 1 2 and so on. The cut offs of the modes are not straight forward they are the solutions of these equations they are given by this. So, which a where j are the Bessel functions. So, by solving these by finding out the zeros of these Bessel functions I can find out the cut offs.

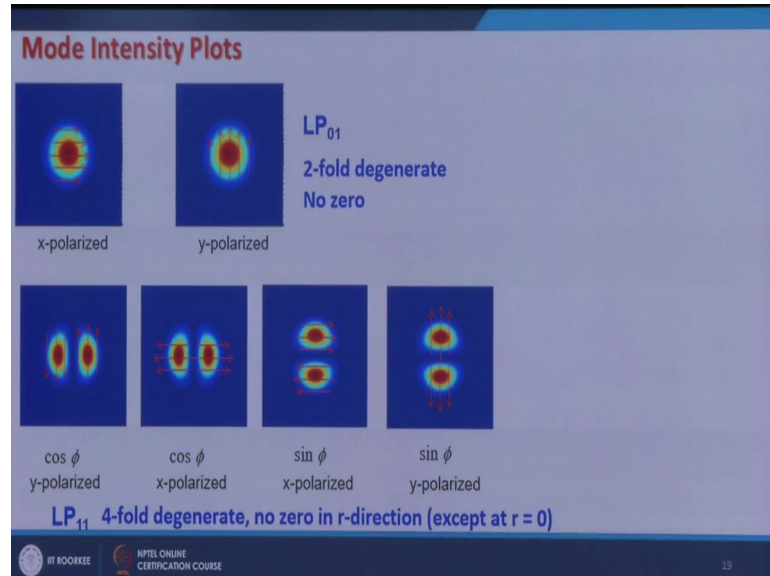
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So, these are the cut offs of various modes first few modes. This is how the modal fields

would look like this is LP₀₁ mode LP₀₂ mode LP₁₁ mode. So, these are 3 d plots and these are the plots.

(Refer Slide Time: 24:09)



When you see them from the top these are the contouring intensity plots. What I see that l is equal to 0 modes are 2-fold degenerate, because the ϕ solution is $\cos l\phi$ and $\sin l\phi$. So, if l is equal to 0 then there is no ϕ dependence there the degeneracy comes only from polarization, while l not equal to 0 modes are 4-fold degenerate. 2-fold degeneracy comes from ϕ and two-fold degeneracy comes from the polarization. Then we had seen the single mode fiber and a single mode fiber we had defined three very important parameters, one is cut off wavelength which is given by 2π over 2.4048 times a square root of $n_1^2 - n_2^2$.

(Refer Slide Time: 24:51)

CUT-OFF WAVELENGTH

$$\lambda_c = \frac{2\pi}{2.4048} a \sqrt{n_1^2 - n_2^2}$$

$\lambda > \lambda_c$: Single - mode
 $\lambda < \lambda_c$: Multi - mode

PROPAGATION CONSTANT

$$b(V) = \left(A - \frac{B}{V} \right)^2; \quad 1.5 < V < 2.5 \quad A = 1.1428 \quad B = 0.996$$

SPOT SIZE

$$\psi(r) = A \exp\left(-\frac{r^2}{w^2}\right) \quad w : \text{Gaussian spot - size} \quad 2w : \text{mode field diameter (MFD)}$$

$$\frac{w}{a} = 0.65 + \frac{1.619}{V^{3/2}} + \frac{2.879}{V^6}; \quad 0.8 < V < 2.5$$

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And what is the cut off wavelength of the fiber? It is the wavelength corresponding to the cutoff of LP 11 modes.

So, if lambda is greater than lambda c then the fiber is single modded, and if lambda is smaller than it is multi modded. The mode propagation constants of the fiber can be fitted with an empirical relation, if V lies between 1.5 and 2.5, and this empirical relation is given by this with these values of a and b. The spot size the mo the modal field profile of a single mode fiber if you look back to this, it resembles a Gaussian and it can be very well fitted with Gaussian. So, we can fit a Gaussian to this profile then we define this spot size as Gaussian spot size where the field intensity drops down to 1 over e square of it is value at the center.

So, w is Gaussian spot size and 2 w is mode field diameter, we can also fit an empirical relationship between excuse me w over a and v which is given by this. So, if I have a fiber I just calculate the value of v for a given wavelength, and I can get the value of w over a from this relationship. Then another important parameter for a single mode fiber is what is it is bend loss.

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BEND LOSS

$$\alpha(\text{dB/m}) = 4.343 \left(\frac{\pi}{4aR_c} \right)^{1/2} \left[\frac{U}{VK_1(W)} \right]^2 \frac{1}{W^{3/2}} \exp \left(-\frac{3W^3}{3k_0^2 a^3 n_1^2} R_c \right)$$

MODE MISMATCH LOSS $\alpha(\text{dB}) = -20 \log \left(\frac{2w_1 w_2}{w_1^2 + w_2^2} \right)$

TRANSVERSE OFF-SET LOSS $\alpha_t(\text{dB}) = 4.34 \left(\frac{u^2}{w^2} \right)$

ANGULAR OFF-SET LOSS $\alpha_{\text{angular}}(\text{dB}) = 4.34 \left(\frac{k_0 m w \theta}{2} \right)^2$

LONGITUDINAL OFF-SET LOSS $\alpha_L(\text{dB}) = 10 \log \left[1 + \left(\frac{D}{k_0 m w^2} \right)^2 \right]$

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So, bend loss is given by this, then their instances there are instances where we need to join two fibers together, and when we join two fibers together then there would be losses.

If two fibers are not identical, then even if there are no misalignment while joining the fibers there would be a loss which is known as mode mismatch loss, which is purely due to different spot sizes of the modes of two fibers. If you join two identical fibers and then there can be three kind of misalignments transverse, angular and longitudinal, and these are the misalignment losses or splice losses due to different kind of misalignments. So, this we had seen. Then in the end we had seen the wave guide dispersion.

So, ultimately we need to use this fiber in telecom system, and when we send pulses through this optical fiber then dispersion happens we had seen that in a fiber there is inter modal dispersion and material dispersion, when you use single mode fiber inter modal dispersion is diminished.

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WAVEGUIDE DISPERSION

$$D_w = -\Delta \frac{n_2}{c\lambda_0} V \frac{d^2(bV)}{dV^2}$$

Empirical Formula

$$b = \left(A - \frac{B}{V} \right)^2; \quad A = 1.1428, \quad B = 0.996 \quad \text{and} \quad V \frac{d^2(bV)}{dV^2} = \frac{2B^2}{V^2}$$

More Accurate Empirical Formula

$$V \frac{d^2(bV)}{dV^2} = 0.080 + 0.549(2.834 - V)^2$$

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So, we get rid off inter modal dispersion, but then there is another dispersion which is wave guide dispersion which comes out because of the dependence of mode propagation constant of fiber mode on wavelength ok.

So, the wave guide dispersion coefficient is given by this, and we can use empirical formula between b and V to calculate this vector, or we can use a more accurate empirical formula given by marquees for V times d 2 b V over dV square. So, in this way we had understood the salient features of optical fiber, how light is guided into optical fiber, what are the important parameters of a single modded fiber.

So, with all this background about optical fibers and the propagation characteristics of optical fiber, now we are ready to use this fiber in a system or to use this fiber to make components and devices, which we will do in the subsequent lectures.

Thank you.