

Quantum Electronics
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Module No. # 02
Nonlinear Optical Effects Nonlinear Polarization
Lecture No. # 05
Non - Linear Optics

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The image shows handwritten equations on a grid background. The top equation is a vector equation:
$$\vec{P} = \epsilon_0 \chi \vec{E} + \epsilon_0 \chi^{(2)} \vec{E} \vec{E} + \epsilon_0 \chi^{(3)} \vec{E} \vec{E} \vec{E} + \dots$$
 Below it is the component form:
$$P_i = \epsilon_0 \chi_{ij} E_j + \epsilon_0 \chi_{ijk}^{(2)} E_j E_k + \epsilon_0 \chi_{ijkl}^{(3)} E_j E_k E_l + \dots$$
 In the bottom left corner, there is a logo for NPTEL (National Programme on Technology Enhanced Learning) and the text 'NPTEL' below it. In the bottom right corner, there is a small text 'IIT DELHI'.

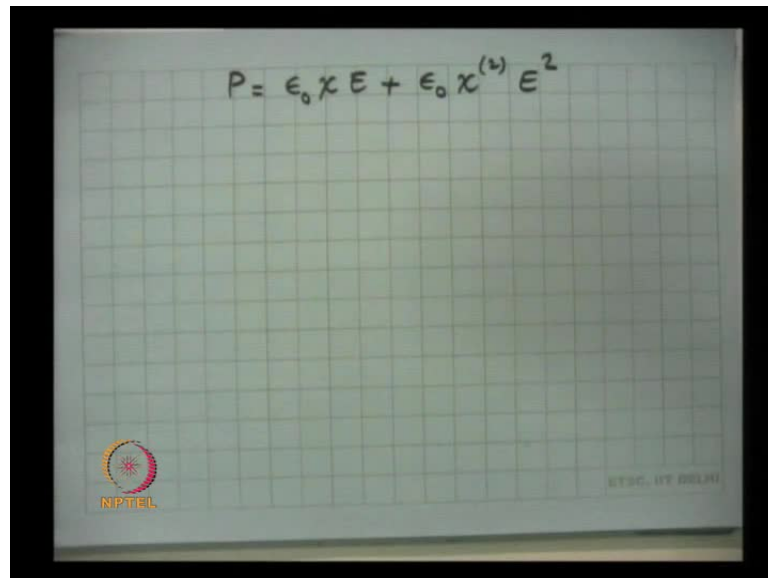
So, we continue with our discussion on non-linear optics. Let us recall that when the electric field of the light wave becomes strong, then the polarization is no more proportional to the electric field - it has also higher order terms - so you have something like $\epsilon_0 \chi^{(2)} E E$ plus $\epsilon_0 \chi^{(3)} E E E$ and so on.

Now, this equation is essentially in component form, an equation of the type $\epsilon_0 \chi_{ij} E_j$ plus $\epsilon_0 \chi_{ijk}^{(2)} E_j E_k$ plus $\epsilon_0 \chi_{ijkl}^{(3)} E_j E_k E_l$ and so on.

So, please note that, this $E E$ is not a dot product or a cross product, it is just E times E . And this is the short form of representing the summation over $\chi_{ijk} E_j E_k$ with

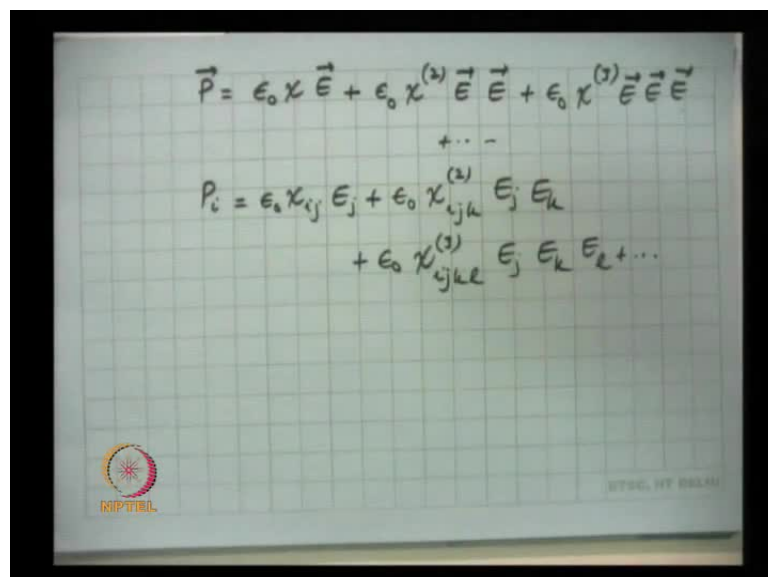
component of polarization is proportional to the j th component of electric field through a tensor χ_{ijk} - the linear susceptibility tensor. And these are the non-linear susceptibility tensors; the second 2 here means its E square term, 3 here is E cube term. So, this is responsible for second order nonlinearities, this is responsible for third order nonlinearities (Refer Slide Time: 02:15).

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$$P = \epsilon_0 \chi E + \epsilon_0 \chi^{(2)} E^2$$

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$$\vec{P} = \epsilon_0 \chi \vec{E} + \epsilon_0 \chi^{(2)} \vec{E} \vec{E} + \epsilon_0 \chi^{(3)} \vec{E} \vec{E} \vec{E} + \dots$$

$$P_i = \epsilon_0 \chi_{ij} E_j + \epsilon_0 \chi_{ijk}^{(2)} E_j E_k + \epsilon_0 \chi_{ijkl}^{(3)} E_j E_k E_l + \dots$$

Now, before we discuss the components of these tensors, we started looking at a simplified form of the polarization, to understand how new frequency can get generated.

So, recall that what we had done is, if I write in scalar form, if I write the second order term, I can write this as plus epsilon 0 chi 2 E square; I just write a scalar equation representing the vector equation. We will come back and discuss in more detail, how to write this equation in terms of an effective chi 2 susceptibility and so on.

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$$P = \epsilon_0 \chi E + \epsilon_0 \chi^{(2)} E^2$$

$$E = E_1 \cos(kz - \omega t)$$

$$E = E_1 \cos(k_1 z - \omega t) + E_2 \cos(k_2 z - 2\omega t)$$

$$k_1 = \frac{\omega}{c} n(\omega) = \frac{\omega}{c} n_1$$

$$k_2 = \frac{2\omega}{c} n(2\omega) = \frac{2\omega}{c} n_2$$

But right now, we want to understand the fundamentals of the new frequency generation and so, we will start with some scalar equation here. And remember, as we discussed last time, if I assume an electric field of the form $E_1 \cos k z - \omega t$, you found that if you substitute this into this equation, there is a linear component of polarization. And E^2 gives me a component of polarization at 2ω frequency; and polarization is nothing but dipole moment per unit volume. So, it implies that the dipoles are also oscillating at frequency 2ω .

So, when a dipole oscillates at a frequency 2ω , it generates electromagnetic waves at 2ω frequency; so, that is the origin of 2ω . But once remember, so if I take a medium a non-linear medium, I launch ω frequency into the medium; so, inside the medium, I will generate 2ω also.

Now, this equation actually, E is the total electric field. So, once 2ω starts to get generated inside the medium, now I have electric fields at ω and 2ω . So, **if I for example** write the total electric field as $E_1 \cos$, now let me differentiate the 2 frequencies like this; so, I have k_1 is the propagation constant of the medium at

frequency ω plus, I have an electric field now corresponding to frequency 2ω , and k_2 is the propagation constant of the medium at frequency 2ω .

So, k_1 will be ω by c into the refractive index at frequency ω , which I call as n_1 ; so, ω by c into n_1 . k_2 is the propagation constant of the medium at frequency 2ω , so 2ω by c into refractive index at 2ω , which I call as n_2 ; so 2ω by c into n_2 .

Please note, that the refractive index of the medium depends on frequency. So, electromagnetic wave at frequency ω and the electromagnetic wave at frequency 2ω will in general, not see the same refractive index. n_1 is the refractive index of the medium at frequency ω ; n_2 is the refractive index of the medium at frequency 2ω .

So, when I launch inside the medium - a frequency ω , through this E^2 term I find that electric fields at 2ω also get generated. So, the total electric field now becomes the sum of these two. So, what will happen if I substitute this into the total electric field here? So, let me look at for example the second term only.

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The image shows a handwritten derivation of the non-linear polarization term P_{NL} on a grid background. The equations are as follows:

$$\begin{aligned}
 P_{NL} &= \epsilon_0 \chi^{(2)} E^2 \\
 &= \epsilon_0 \chi^{(2)} E_1^2 \cos^2(k_1 z - \omega t) \\
 &\quad + \epsilon_0 \chi^{(2)} E_2^2 \cos^2(k_2 z - 2\omega t) \\
 &\quad + 2\epsilon_0 \chi^{(2)} E_1 E_2 \cos(k_1 z - \omega t) \times \cos(k_2 z - 2\omega t) \\
 &= \frac{\epsilon_0}{2} \chi^{(2)} \left[E_1^2 + \cos 2(k_1 z - \omega t) \right] E_1^2 \\
 &\quad + \frac{\epsilon_0}{2} \chi^{(2)} \left[1 + \cos 2(k_2 z - 2\omega t) \right] E_2^2 \\
 &\quad + \epsilon_0 \chi^{(2)} E_1 E_2 \left[\cos[(k_1 + k_2)z - 3\omega t] \right. \\
 &\quad \left. + \cos[(k_2 - k_1)z - \omega t] \right]
 \end{aligned}$$

A small logo for NPTEL is visible in the bottom left corner of the slide.

So, let me look at the non-linear polarization term; so, this is $\epsilon_0 \chi^{(2)}$ into E^2 ; so, I substitute for E from this equation into this E^2 term here. So, what will I get? I will have $\epsilon_0 \chi^{(2)} E_1^2 \cos^2(k_1 z - \omega t) +$

$\epsilon_0 \chi^2 E^2 \cos^2 k_2 z - \omega t + 2 \epsilon_0 \chi^2 E_1 E_2 \cos k_1 z - \omega t$ into $\cos k_2 z - \omega t$; a plus b whole square.

So, now, I write all the terms in terms of cosines. So, \cos^2 will be in terms of $\cos 2\theta$; and then, this is $\cos a \cos b$ in terms of $\cos a + b$ and $\cos a - b$. so, what will I get? I will get $\epsilon_0 \chi^2 E_1^2 \cos^2 k_2 z - \omega t$ for the first term; then, I will get $\epsilon_0 \chi^2 E_1 E_2 \cos k_1 z - \omega t$; **there is an E_1 , yes, E_1^2 square is outside**; this is $1 + \cos 2k_2 z - \omega t$ into $E_1^2 \cos^2 k_2 z - \omega t$ plus, so this is $2 \cos k_2 z - \omega t$; so, I can write this as $\epsilon_0 \chi^2 E_1^2 E_2^2 \cos k_1 z - \omega t$ into $\cos k_1 z - \omega t$ plus $\cos k_2 z - \omega t$ and then I will have one more term, which is plus $\epsilon_0 \chi^2 E_1 E_2 \cos k_2 z - \omega t$; $\cos a \cos b$ is $\cos a + b + \cos a - b$ divided by 2.

So, if you look at this terms now, this 1 here and this 1 here give you optical rectification; that means, some DC polarization term. This is that frequency 2ω , this term gives you frequency 2ω , this term gives you frequency 4ω , because the 2ω frequency once it is launched inside the medium, generated inside the medium, the nonlinearity at 2ω frequency will now generate 4ω frequency - the second harmonic of 2ω ; it also generates a 3ω term and $n\omega$ term.

So, the total electric field now will consist of ω , 2ω , 3ω , 4ω . So, you have to substitute back into the equation and recalculate everything. Now, it so happens as we will discuss, that to generate 2ω from ω , I must satisfy some conditions called phase matching conditions. If I do not satisfy those conditions, the efficiency of the generation of new frequency is very little. I will explain this phase matching condition mathematically and physically, what is the meaning of phase matching condition, but as we will see to generate 2ω from ω , I need to satisfy certain phase matching conditions.

Similarly, if you generate 2ω from 2ω a frequency 4ω , I need to satisfy another set of phase matching conditions. Now, usually what happens? Even to satisfy one phase matching condition is not easy. I have to design my polarization states, the propagation direction of the crystal, etcetera, to satisfy even one of the phase matching conditions.

So, normally what happens is, we will be able to satisfy only one of the phase matching conditions; for example, for omega to 2 omega conversion. So, when I satisfy the condition for omega to 2 omega, I may not be simultaneously satisfying the condition for 2 omega to 4 omega term. So, this 4 omega electric field will hardly be generated. This term will not be able to generate electromagnetic wave at frequency 4 omega; this one 3 omega term also requires another phase matching condition; so, usually that is also not satisfied.

But look at this term, this is a term which is generating omega. How is it generating omega? It is mixing the 2 omega and omega terms; it is the difference frequency. Omega frequency and omega frequency add to give you 2 omega frequency; 2 omega frequency and omega frequency mix to give you omega frequency, because omega plus omega is 2 omega and 2 omega minus omega is omega. So, this leads to, as you can see here, a non-linear polarization term at frequency omega. So, the same nonlinearity actually will generate from omega, it will generate 2 omega. And once 2 omega gets generated inside the medium, this 2 omega mixes with omega to generate back omega.

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$$\begin{aligned}
 P_{NL} &= \epsilon_0 \chi^{(2)} E^2 \\
 &= \epsilon_0 \chi^{(2)} E_1^2 \cos^2(k_1 z - \omega t) \\
 &\quad + \epsilon_0 \chi^{(2)} E_2^2 \cos^2(k_2 z - 2\omega t) \\
 &\quad + 2 \epsilon_0 \chi^{(2)} E_1 E_2 \cos(k_1 z - \omega t) \times \cos(k_2 z - 2\omega t) \\
 &= \frac{\epsilon_0}{2} \chi^{(2)} \left[E_1^2 + \cos 2(k_2 z - 2\omega t) \right] E_1^2 \\
 &\quad + \frac{\epsilon_0}{2} \chi^{(2)} \left[1 + \cos 2(k_2 z - 2\omega t) \right] E_2^2 \\
 &\quad + \epsilon_0 \chi^{(2)} E_1 E_2 \left[\cos[(k_1 + k_2)z - 3\omega t] \right. \\
 &\quad \left. + \cos[(k_2 - k_1)z - \omega t] \right]
 \end{aligned}$$

And as I will show you the phase matching condition, for omega to 2 omega conversion is the same as the phase matching condition for 2 omega to omega conversion back. That is, omega plus omega giving you 2 omega and 2 omega minus omega giving you omega, are the same phase matching conditions.

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$$P = \epsilon_0 \chi E + \epsilon_0 \chi^{(2)} E^2$$

$$E = E_1 \cos(k_1 z - \omega t)$$

$$E = E_1 \cos(k_1 z - \omega t) + E_2 \cos(k_2 z - 2\omega t)$$

$$k_1 = \frac{\omega}{c} \cdot n(\omega) = \frac{\omega}{c} \cdot n_1$$

$$k_2 = \frac{2\omega}{c} \cdot n(2\omega) = \frac{2\omega}{c} \cdot n_2$$

So, what will happen is, if I take this medium and launch 2 omega frequency with omega frequency into the medium, this omega will generate 2 omega; 2 omega and omega mix to generate omega. So, that is a coupling, that is taking place between omega and 2 omega frequency.

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$$P_{NL} = \epsilon_0 \chi^{(2)} E^2$$

$$= \epsilon_0 \chi^{(2)} E_1^2 \cos^2(k_1 z - \omega t) + \epsilon_0 \chi^{(2)} E_2^2 \cos^2(k_2 z - 2\omega t) + 2\epsilon_0 \chi^{(2)} E_1 E_2 \cos(k_1 z - \omega t) \times \cos(k_2 z - 2\omega t)$$

$$= \frac{\epsilon_0}{2} \chi^{(2)} [1 + \cos 2(k_1 z - \omega t)] E_1^2 + \frac{\epsilon_0}{2} \chi^{(2)} [1 + \cos 2(k_2 z - 2\omega t)] E_2^2 + \epsilon_0 \chi^{(2)} E_1 E_2 [\cos(k_1 + k_2)z - 3\omega t] + \epsilon_0 \chi^{(2)} E_1 E_2 \cos[(k_2 - k_1)z - \omega t]$$

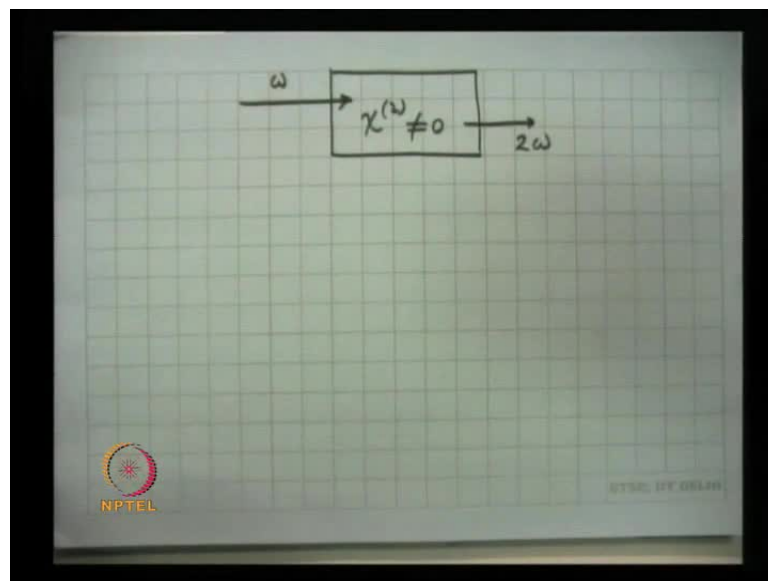
And this process I will show you is efficient only under certain conditions called phase matching conditions. So, although this term is generating new frequencies, it will so happen that the electric field generated at these new frequencies will be negligible.

So, usually, we will neglect all the other kinds of frequencies that get generated in the process and focus on the set of frequencies for which I am satisfying the phase matching condition; and hence, I will only worry about those electric fields.

So, please note that, when I launch an ω into a medium, I will generate non-linear polarization at 2ω ; that non-linear polarization at 2ω will generate electromagnetic wave at 2ω . The electromagnetic wave at 2ω will mix with the electromagnetic wave at ω , to generate waves at ω also.

So, there is a non-linear polarization term at 2ω , there is a non-linear polarization term also at ω . You cannot have coupling in only one direction; you have ω to 2ω transfer, 2ω to ω transfer is also taking place simultaneously.

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Now, what is the objective now? What is objective? So, problem is, I am given a medium in which this non-linear susceptibility is not 0. As I mentioned to you last time, if the medium has a center of inversion symmetry, the χ_2 tensor is 0; all elements of χ_{ijk2} are 0. You cannot have this process taking place in a medium, which has center of inversion symmetry.

So this, let me assume this is a medium which has no center of inversion symmetry and hence can generate second order processes. So, what is my problem? My objective is, I launch a frequency ω into the medium, I want to calculate how much of power am I

generating at 2ω . I launch electric field at ω inside the medium with a certain amplitude, the medium converts ω to 2ω , so what is the generated power at 2ω at the output? What does it depend on? How do I maximize this efficiency of conversion? And what are the conditions under which this efficiency will be maximum and how do I achieve those conditions in practice?

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$$\vec{P} = \epsilon_0 \chi \vec{E} + \epsilon_0 \chi^{(2)} \vec{E} \vec{E} + \epsilon_0 \chi^{(3)} \vec{E} \vec{E} \vec{E} + \dots$$

$$P_i = \epsilon_0 \chi_{ij} E_j + \epsilon_0 \chi_{ijk}^{(2)} E_j E_k + \epsilon_0 \chi_{ijkl}^{(3)} E_j E_k E_l + \dots$$

So, the problem is essentially boils down solving Maxwell's equations, with the condition that the polarization now is not just the first term here; it also contains higher order terms. And right now, we are looking at only this term; we will forget about all higher order terms; we will look at the contribution due to the second order nonlinearity in the Maxwell's equations.

Now, to keep the mathematics simple and to understand the physics, what we will do is, we will assume a scalar equation - scalar analysis - in the beginning, discuss the generation efficiencies and so on. And then, I will come back and look at this equation little more carefully, because finally if I give you a medium, I will tell this medium has this tensor - χ_{ijk} elements - I will give you and I will ask you, what polarization of the input should I choose, what is the polarization that comes out of the medium at 2ω frequency, what condition should I satisfy to maximize this efficiency.

So, this involves little more analysis using these tensor properties of this, but right now, to understand even the basic generation of second harmonic, we will assume a scalar equation and analyze the problem.

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$$\begin{aligned}
 \vec{D} &= \epsilon_0 \vec{E} + \vec{P} \\
 &= \epsilon_0 \vec{E} + \vec{P}_L + \vec{P}_{NL} \\
 &= \epsilon_0 \vec{E} + \epsilon_0 \chi \vec{E} + \vec{P}_{NL} \\
 &= \epsilon_0 (1 + \chi) \vec{E} + \vec{P}_{NL} \\
 &= \epsilon \vec{E} + \vec{P}_{NL}
 \end{aligned}$$

Now, before we do that, I must derive a wave equation in the presence of nonlinearity. So, let me go back and let me write this equation; so, D is equal to remember, epsilon 0 E plus P; this is the defining equation for the displacement vector. Now, I can write this as epsilon 0 E plus P linear plus P non-linear, P linear is epsilon 0 chi E plus P non-linear, this is equal to epsilon 0 into 1 plus chi E plus P non-linear and this is nothing but epsilon into E plus P non-linear; epsilon is the dielectric permittivity of the medium.

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$$\nabla \times \vec{E} = -\frac{\partial \vec{B}}{\partial t}; \quad \nabla \times \vec{H} = \frac{\partial \vec{D}}{\partial t}$$

$$\vec{B} = \mu_0 \vec{H}; \quad \vec{D} = \epsilon \vec{E} + \vec{P}_{NL}$$

$$\nabla \times (\nabla \times \vec{E}) = -\mu_0 \frac{\partial (\nabla \times \vec{H})}{\partial t}$$

$$\nabla (\nabla \cdot \vec{E}) - \nabla^2 \vec{E} = -\mu_0 \frac{\partial}{\partial t} \left(\epsilon \frac{\partial \vec{E}}{\partial t} + \frac{\partial \vec{P}_{NL}}{\partial t} \right)$$

$$-\nabla^2 \vec{E} = -\mu_0 \epsilon \frac{\partial^2 \vec{E}}{\partial t^2} - \mu_0 \frac{\partial^2 \vec{P}_{NL}}{\partial t^2}$$

$$\nabla^2 \vec{E} - \mu_0 \epsilon \frac{\partial^2 \vec{E}}{\partial t^2} = \mu_0 \frac{\partial^2 \vec{P}_{NL}}{\partial t^2}$$

So, the linear part of the polarization has been taken into account in epsilon and this second term is the non-linear term. So, normally, D would have been equal to epsilon E; so, that is the expression for D. So, let us go back and write this equation, so we have del cross E is equal to minus del B by del t and del cross H is equal to del D by del t from the no free currents, no free charges.

Now, we are going to assume, as I mentioned before, B is equal to mu naught H, and D is equal to epsilon E plus P non-linear. So, let me take a curl of the first equation, so del cross del cross E is equal to minus mu 0, now I interchange the time and space derivatives, so I get del by del t of del cross H.

The first the left hand side is gradient of divergence E minus del square E is equal to minus mu naught del by del t of del cross H, which is epsilon del E by del t plus del P non-linear by del t. Now, how about divergence E? Divergence E is 0 in isotropic media.

In anisotropic media, remember, D is perpendicular to k vector, E is not in general perpendicular to k vector divergence; E is equal to 0 implies transfer condition. That means, in anisotropic media divergence E is usually not 0, but again to simplify the analysis here, we are going to forget about divergence E term here, but in principle actually divergence E is not 0 in anisotropic media. But as I left a problem to you, you can check the angle between E and D vector is not very large; usually is very small - it is a couple of degrees.

So, I am going to assume, as a simplification divergence E is equal to 0 and so this is minus del square E is equal to minus mu 0 epsilon del square E by del t square minus mu 0 del square P non-linear by del t square. So, this is my wave equation, del square E minus mu 0 epsilon del square E by del t square is equal to mu 0 del square P non-linear by del t square.

Normally, when you derive the wave equation, the right hand side is 0, because normally you do not take into account any nonlinearity in the medium. So, the wave equation normally you would have obtained is, del square E minus mu 0 epsilon del square E by del t square is equal to 0. The right hand side actually acts as a source for this electromagnetic field; it is an inhomogeneous equation. This P non-linear is the source of the electromagnetic fields; and please note that this equation has to be satisfied for each and every frequencies.

So, the frequency omega will satisfy this equation - for omega frequency, where epsilon will be at omega frequency, E will be at omega frequency, P non-linear at omega frequency. For the 2 omega frequency, the electric field at 2 omega frequency, the epsilon at 2 omega frequency, here P non-linear at 2 omega frequency.

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$$\nabla^2 \vec{E}^{(\omega)} - \mu_0 \epsilon(\omega) \frac{\partial^2 \vec{E}^{(\omega)}}{\partial t^2} = \mu_0 \frac{\partial^2 \vec{P}_{NL}^{(\omega)}}{\partial t^2}$$

$$\nabla^2 \vec{E}^{(2\omega)} - \mu_0 \epsilon(2\omega) \frac{\partial^2 \vec{E}^{(2\omega)}}{\partial t^2} = \mu_0 \frac{\partial^2 \vec{P}_{NL}^{(2\omega)}}{\partial t^2}$$

$$\begin{aligned} E^{(\omega)} &= \tilde{E}_1 \cos(kz - \omega t + \phi) \\ &= \frac{1}{2} \left[\tilde{E}_1 e^{i\phi} e^{i(kz - \omega t)} + cc \right] \\ &= \frac{1}{2} \left[E_1 e^{i(kz - \omega t)} + cc \right] \end{aligned}$$

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So, for example, I would have to satisfy these equations, del square E at omega minus mu 0 epsilon of omega del square E of omega by del t square is equal to mu naught del square P non-linear at omega by del t square; and del square E of 2 omega minus mu 0

ϵ of 2ω $\Delta^2 E$ of 2ω by Δt^2 is equal to $\mu_0 \Delta^2 P$ non-linear at 2ω by Δt^2 .

I showed you that there is also a non-linear term at frequency ω , because of **mixing** between ω and 2ω frequencies. There is a non-linear polarization at 2ω ; there is a non-linear polarization at ω . And these two equations describe the propagation of frequency ω and 2ω through the medium and the interaction between these two frequencies is brought about by the nonlinearity in the medium.

Normally, if you have a linear medium, if you launch a frequency ω and 2ω simultaneously, they do not interact; ω frequency propagates independently, 2ω frequency propagates independently. But when there is nonlinearity, the presence of nonlinearity couples these two frequencies through the non-linear polarization term.

So, in principle, if you launch only one of the waves, it can lead to the generation of the new frequency. So, the condition I have is, I launch electromagnetic field at frequency ω , this electromagnetic field will generate P non-linear at 2ω , this P non-linear at 2ω will generate E at 2ω , E at 2ω and E at ω mix to generate P non-linear at ω , which will then influence the ω frequency.

So, these two equations will give me a pair of coupled equations which I need to solve. I launch ω frequency, this will generate P non-linear at 2ω , this P non-linear at 2ω generates E at 2ω , E at 2ω and E at ω mix to generate P non-linear at ω , which will then modify this equation. So, these two equations are to be solved simultaneously.

Now, I want to do a simplified analysis, where I will instead of writing this vector equation, we will try to solve a scalar equation, which means we will assume that, these are waves propagating along z direction, plane wave propagating along z direction and these are linearly polarized electromagnetic fields. So, E is along x direction; for example, it is propagating along some direction z and I will try to solve these equations in a scalar approximation, and later on I will come back and tell you, this approximation where from am I getting this approximation. If I am actually given a vector field which is different, how do I reduce the problem to this problem which I am solving now?

So, remember I also mentioned that, in the presence of nonlinearity, I have to use real fields always; that is why I wrote cosine there. I can actually overcome this, I can go back again to complex notation provided, the total is still real. So, the way I solve it is the following; so, for example, if I write E at ω is equal to $E_1 \cos(kz - \omega t + \phi_1)$, suppose this is my electromagnetic field at frequency ω .

So, I will write this as half of $E_1 \exp(i\phi_1) e^{i(kz - \omega t)}$, let me write k_1 here, $k_1 z - \omega t + \text{complex conjugate } c$; c stands for complex conjugate. So, I am writing this cosine as this exponential plus a complex conjugate divided by 2. And this one I write as, $E_1 e^{i\phi_1} e^{i(kz - \omega t)}$, where E_1 is the complex electric field at frequency ω ; it has also a phase term ϕ_1 . Because it is of the form this plus complex conjugate divided by 2, this is always real. I am still using real notation except that, instead of writing cosine, I am writing exponential plus its complex conjugate.

Yes.

Sir, E_1 is also complex amplitude.

No, in the real equation, E_1 is real. When I write $E_1 \cos(kz - \omega t + \phi_1)$ - everything is real, E_1 is real, ϕ_1 is real, everything is real. But what I am doing is, I am absorbing the phase ϕ_1 into the E_1 term and defining an E_1 - electric field, which is now in general complex. So, if the phase changes, E_1 will change in complex phase. And when you add the complex conjugate, you will get that phase inside the cosine term again. So, change of phase inside the cosine term will imply that E_1 is changing with z , because phase change is with z .

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$$\nabla^2 \vec{E}^{(\omega)} - \mu_0 \epsilon^{(\omega)} \frac{\partial^2 \vec{E}^{(\omega)}}{\partial t^2} = \mu_0 \frac{\partial^2 \vec{P}_{NL}^{(\omega)}}{\partial t^2}$$

$$\nabla^2 \vec{E}^{(2\omega)} - \mu_0 \epsilon^{(2\omega)} \frac{\partial^2 \vec{E}^{(2\omega)}}{\partial t^2} = \mu_0 \frac{\partial^2 \vec{P}_{NL}^{(2\omega)}}{\partial t^2}$$

$$\vec{E}^{(\omega)} = \vec{E}_1 \cos(k_1 z - \omega t + \phi_1)$$

$$= \frac{1}{2} \left[\vec{E}_1 e^{i\phi_1} e^{i(k_1 z - \omega t)} + cc \right]$$

$$= \frac{1}{2} \left[E_1 e^{i(k_1 z - \omega t)} + cc \right]$$

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So, whether phase change is with z or amplitudes change is with z, E 1 will change with z. So, E 1 is a complex electric field; and the electric field at omega is written as half of E 1 plus E 1 exponential i k 1 z minus omega t plus its complex conjugate.

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$$\vec{E}^{(2\omega)} = \frac{1}{2} \left[E_2 e^{i(k_2 z - 2\omega t)} + cc \right]$$

$$P_{NL} = \epsilon_0 \chi^{(2)} E^2$$

$$= 2 \epsilon_0 d E^2 \quad d \approx 10^{-12} \text{ m/V}$$

$$E = E^{(\omega)} + E^{(2\omega)}$$

$$P_{NL} = 2 \epsilon_0 d \frac{1}{4} \left[E_1 e^{i(k_1 z - \omega t)} + cc + E_2 e^{i(k_2 z - 2\omega t)} + cc \right]^2$$

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Similarly, I will write for E2, E 2 omega half of E 2 exponential i k 2 z minus 2 omega t plus complex conjugate, where E 2 is the complex electric field at frequency 2 omega; k 2 is the propagation constant of the medium at frequency 2 omega; k 1 is the propagation constant of the medium at frequency omega.

So, if it is an ordinary wave or extraordinary wave, k_1 will have to be taken the correct value. In general, the medium is anisotropic; so, k_1 and k_2 are the two propagation constants of these plane waves propagating along z direction.

Now, let me tell you that this z is the z direction - arbitrary some z direction; this is not the z direction of the crystal. This is an $x y z$ coordinate system in which I am solving the problem and this $x y z$ need not be coincident with $x y z$ of the principle axis system of the crystal; this is some arbitrary $x y z$ system.

So, let me write the P non-linear term, P non-linear is now in non-linear optics notation. There are different kinds of notations used by different people, but what I want to do is, use the following notation in which P non-linear, remember I had written as, $\epsilon_0 \chi^2 E^2$; I want to write this as $2 \epsilon_0 d E^2$, where χ^2 is represented by $2 d$. what will be the units of d in SI units? Look at the dimensions of P non-linear electric field, **sorry**, 1 by electric field, because P has a same dimension ϵ_0 times E.

So, 1 by electric field means, d will be in units of meters per volt, because E has volts per meter. So, let me tell you the numbers, d is typically of the order of 10 to the minus 12 meters per volt. For most media, the coefficient d has this magnitude about 10 to the minus 12 , 10 times 10 to the minus 12 , 3 times 10 to minus 12 , something like this, these are some numbers of media that are used in non-linear optics.

So, let me substitute now this, what is E? E is the total electric field which consists of $E_1 e^{i(k_1 z - \omega t)}$ plus $E_2 e^{i(k_2 z - 2\omega t)}$. So, P non-linear will be $2 \epsilon_0 d$, so when I add $E_1 e^{i(k_1 z - \omega t)}$ given by this equation and $E_2 e^{i(k_2 z - 2\omega t)}$ given by this equation and take square, so this factor half will give me 1 by 4 outside; so, I will have $E_1^2 e^{i(2k_1 z - 2\omega t)}$ plus complex conjugate plus $E_2^2 e^{i(2k_2 z - 4\omega t)}$ plus complex conjugate whole square; so, this is E^2 .

So, from here, let me pick up the non-linear polarization terms at frequency ω and 2ω . So, first, let me polar the term with non-linear polarization term at frequency 2ω ; P non-linear at 2ω , so can you tell me what will I get from here, which terms when I take the square, I will have square of this, square of this, square of this, square of this, 2 times this into this, 2 times this into this, you have a plus b plus c plus d whole square (Refer Slide Time: 35.02).

(Refer Slide Time: 35:26)

$$P_{NL}^{(2\omega)} = \frac{1}{2} \left[\epsilon_0 d E_1^2 e^{2i(k_1 z - \omega t)} + cc \right]$$

$$P_{NL}^{(\omega)} =$$

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So, can you tell me, which are the terms which will have a frequency 2 omega; this square and this square. Please remember, because the total function is real, for every complex number I get, I will have a complex conjugate. So, I will have, this will be simply, so this factor of 2 and 1 by 4 gives me half, so half; let me write inside as epsilon 0 d E 1 square e to the power 2 i k 1 z minus omega t plus complex conjugate. And there will be no other term ; none of the mixing terms in this square will give me a frequency 2 omega.

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$$E^{(2\omega)} = \frac{1}{2} \left[E_2 e^{i(k_2 z - 2\omega t)} + cc \right]$$

$$P_{NL} = \epsilon_0 \chi^{(2)} E^2$$

$$= 2 \epsilon_0 d E^2 \quad d \approx 10^{-12} \text{ m/V}$$

$$E = E^{(\omega)} + E^{(2\omega)}$$

$$P_{NL} = 2 \epsilon_0 d \frac{1}{4} \left[E_1 e^{i(k_1 z - \omega t)} + cc + E_2 e^{i(k_2 z - 2\omega t)} + cc \right]^2$$

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What about P non-linear at omega? So, let me look at this equation again. So, which product will give me omega frequency?

(C)

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$$P_{NL}^{(2\omega)} = \frac{1}{2} [\epsilon_0 d E_1^2 e^{2i(k_1 z - \omega t)} + cc]$$

$$P_{NL}^{(\omega)} = \frac{1}{2} [2\epsilon_0 d E_2 E_1^* e^{i(k_2 - k_1)z - \omega t} + cc]$$

This into the complex conjugate of the first term, because this complex conjugate has exponential plus i omega t multiplied by this, will give me exponential minus i omega i. Similarly, this into its complex conjugate (Refer Slide Time: 36:26), so I will have essentially half of 2 epsilon 0 d E 2 E 1 star; star represents complex conjugate; and what will be exponential term? I will have exponential i k 2 minus k 1 z minus omega t plus complex conjugate.

See, E 1 is the coefficient which multiplies E into the minus i omega t; so, E 1 square will be the coefficient multiplying exponential minus 2 i omega t. So, E 1 square is sitting here for the second harmonic polarization; E 2 is the coefficient of e to the power s minus 2 i omega t; E 1 star is the coefficient of exponential plus i omega t.

So, when I multiply these two terms, I will get exponential minus i omega t, which is what gives me the polarization at frequency omega. Because later on, by looking at this product, you can sort of guess the product term that will appear here and to which frequency that will correspond. For example, E 2 square would correspond to non-linear polarization at 4 omega; E 1, E 2 will give me 3 omega.

So, if you expand this, but you need to worry about these factors which are coming out in the squaring operation here. And so, this will be the non-linear polarization 2 omega frequency; this is a non-linear polarization at omega frequency.

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The image shows handwritten mathematical derivations on a grid background. The equations are as follows:

$$E^{(2\omega)} = \frac{1}{2} [E_2 e^{i(k_2 z - 2\omega t)} + cc]$$

$$P_{NL} = \epsilon_0 \chi^{(2)} E^2$$

$$= 2 \epsilon_0 d E^2 \quad d \approx 10^{-12} \text{ m/V}$$

$$E = E^{(\omega)} + E^{(2\omega)}$$

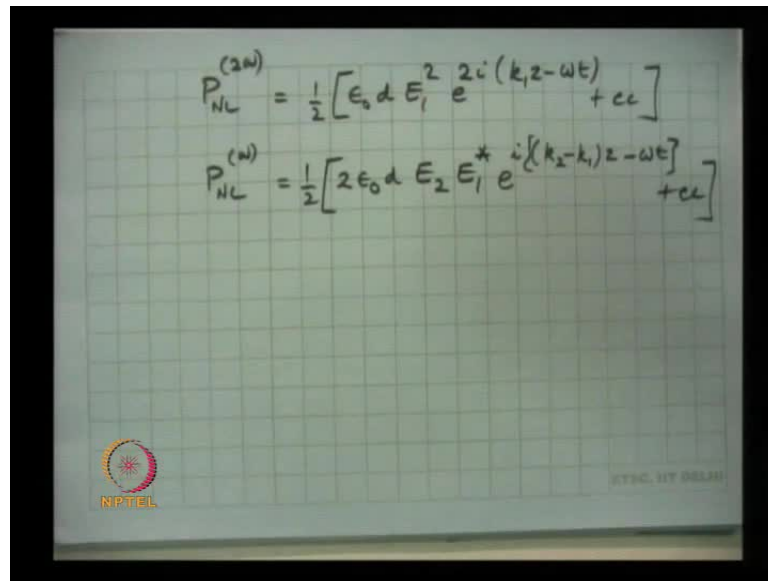
$$P_{NL} = 2 \epsilon_0 d \frac{1}{4} [E_1 e^{i(k_1 z - \omega t)} + cc + E_2 e^{i(k_2 z - 2\omega t)} + cc]^2$$

In the bottom left corner, there is a logo for NPTEL (National Programme on Technology Enhanced Learning) and in the bottom right corner, it says "IIT DELHI".

So, first, let me derive the equation corresponding to the frequency 2 omega; so, this is my equation of the electric field at 2 omega. Now, please note now, that because of this generation, the amplitude - complex amplitude - of the second harmonic will now be a function of z; z is the propagation direction, because at the input to the crystal, there was only frequency omega present, but as the wave propagates, the 2 omega frequency gets generated. So, E 2 must have been 0 at the entrance phase of the crystal; and as the wave propagates, E 2 will start to get generated, which means E 2 must be a function of z; z is the propagation direction. And because of this nonlinearity, E 2 will become a function of z.

And if E 2 becomes a function of z, E 1 also has to become a function of z, because you cannot generate second harmonic, unless you lose energy from the fundamental. You are only converting electromagnetic energy at frequency omega, to energy at 2 omega frequency. So, if E 2 is a function of z, E 1 is also a function of z; so, in these equations, I must remember E 1 and E 2 are now functions of z, because of this interaction. because of this nonlinearity.

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Handwritten equations on a grid background:

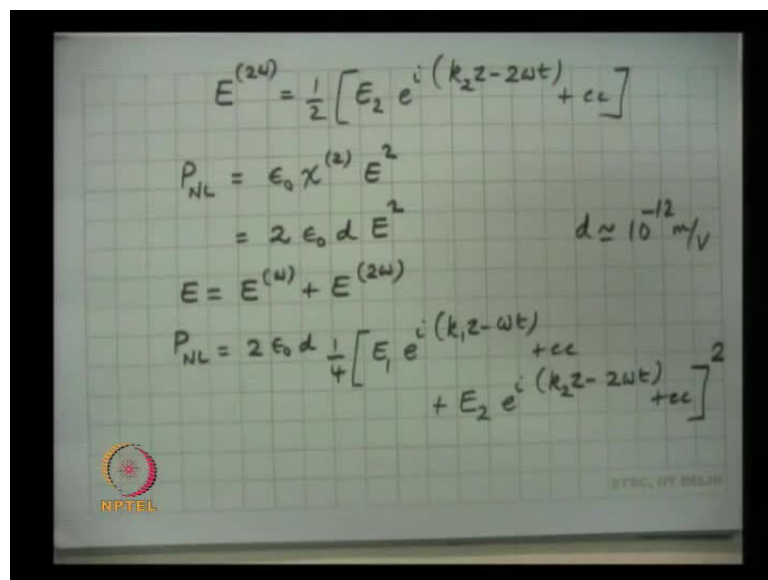
$$P_{NL}^{(2\omega)} = \frac{1}{2} \left[\epsilon_0 d E_1^2 e^{2i(k_1 z - \omega t)} + cc \right]$$

$$P_{NL}^{(\omega)} = \frac{1}{2} \left[2\epsilon_0 d E_2 E_1^* e^{i(k_2 - k_1)z - \omega t} + cc \right]$$

NPTEL logo and "IITC, HY DELHI" are visible at the bottom of the slide.

So, the problem is very simple now. This is the expression for E_2 at 2ω ; E_2 is a function of z . This is the expression for P non-linear at 2ω and this is the equation at 2ω (Refer Slide Time: 40:00).

(Refer Slide Time: 30:32)



Handwritten equations on a grid background:

$$E^{(2\omega)} = \frac{1}{2} \left[E_2 e^{i(k_2 z - 2\omega t)} + cc \right]$$

$$P_{NL} = \epsilon_0 \chi^{(2)} E^2$$

$$= 2\epsilon_0 d E^2 \quad d \approx 10^{-12} \text{ m/V}$$

$$E = E^{(\omega)} + E^{(2\omega)}$$

$$P_{NL} = 2\epsilon_0 d \frac{1}{4} \left[E_1 e^{i(k_1 z - \omega t)} + cc + E_2 e^{i(k_2 z - 2\omega t)} + cc \right]^2$$

NPTEL logo and "IITC, HY DELHI" are visible at the bottom of the slide.

So, if I forget about the vectors, so I have expression for electric field, I have an expression for P non-linear at 2ω , I must substitute into this equation and simplify. To get an equation for the development of E_2 as the function of z , I will get a differential equation for E_2 of z .

So, let me do this, so remember in this equation - wave equation, there is del square; so, this is differential, del square by del x square plus del square del y square plus del square del z square. So, there is no x dependence, there is no y dependence, so I will be lonely left with the z dependence. So, there will be a del square del z square here, there is a time derivative here, time derivative here.

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$$P_{NL}^{(2\omega)} = \frac{1}{2} \left[\epsilon_0 d E_1^2 e^{2i(k_1 z - \omega t)} + c.c. \right]$$

$$P_{NL}^{(\omega)} = \frac{1}{2} \left[2\epsilon_0 d E_2 E_1^* e^{i(k_2 - k_1)z - \omega t} + c.c. \right]$$

$$\nabla^2 E^{(2\omega)} = \frac{\partial^2}{\partial z^2} [E^{(2\omega)}]$$

$$= \frac{\partial^2}{\partial z^2} \frac{1}{2} \left[E_2(z) e^{i(k_2 z - 2\omega t)} + c.c. \right]$$

$$= \frac{1}{2} \left[\frac{d^2 E_2}{dz^2} + 2i k_2 \frac{dE_2}{dz} - k_2^2 E_2 \right] e^{i(k_2 z - 2\omega t)} + c.c.$$

So, let me now calculate, what is the z differential. So, del square E 2 omega will be actually del square by del z square of E 2 omega, which is equal to del square by del z square of half, now let me write E 2 of z e to the power i k 2 z minus 2 omega t plus complex conjugate.

Now, there is a z dependence in E 2, there is a z dependence in exponential i k 2 z; so, this is simple calculation, so let me write down the final expression, I will have d square E 2 by d z square plus 2 i k 2 d E 2 by d z minus k 2 square E 2 into e to the power i k 2 z minus 2 omega t plus complex conjugate; it is a second differential of a product of two functions, you can just differentiate that and you get this these three terms here.

Now, usually, the rate of growth of the second harmonic with z is slow, wavelengths out of the order of micron to 1 micron, but this E 2 will grow in maybe hundreds of micron slowly - it grows slowly as a function of z.

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$$\nabla^2 \vec{E}^{(\omega)} - \mu_0 \epsilon^{(\omega)} \frac{\partial^2 \vec{E}^{(\omega)}}{\partial t^2} = \mu_0 \frac{\partial^2 \vec{P}_{NL}^{(\omega)}}{\partial t^2}$$

$$\nabla^2 \vec{E}^{(2\omega)} - \mu_0 \epsilon^{(2\omega)} \frac{\partial^2 \vec{E}^{(2\omega)}}{\partial t^2} = \mu_0 \frac{\partial^2 \vec{P}_{NL}^{(2\omega)}}{\partial t^2}$$

$$\begin{aligned} E^{(\omega)} &= \tilde{E}_1 \cos(k_1 z - \omega t + \phi_1) \\ &= \frac{1}{2} \left[\tilde{E}_1 e^{i\phi_1} e^{i(k_1 z - \omega t)} + c.c. \right] \\ &= \frac{1}{2} \left[E_1 e^{i(k_1 z - \omega t)} + c.c. \right] \end{aligned}$$

So, I will neglect the second derivative of E^2 with respect to z . Assuming that the growth of the second harmonic slow and so, this is neglected; that means, in one wavelength or a few wavelengths, E^2 does not change significantly, and that is simply neglecting the second derivative. So, I substitute this $\nabla^2 E^{(\omega)}$, E^2 into this equation, I also calculate this second derivative with respect to time and on the right hand side here.

So, what I need to do is the following, I substitute for this expression into this first term here, then the time derivative is very simple, because it is exponential minus $2i\omega t$. So, I will get minus $4\omega^2$, when I differentiate twice; and P_{NL} also will give me minus $4\omega^2$, with the amplitude term sitting here.

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$$P_{NL}^{(2\omega)} = \frac{1}{2} \left[\epsilon_0 d E_1^2 e^{2i(k_1 z - \omega t)} + cc \right]$$

$$P_{NL}^{(\omega)} = \frac{1}{2} \left[2\epsilon_0 d E_2 E_1^* e^{i(k_2 - k_1)z - \omega t} + cc \right]$$

$$\nabla^2 E^{(2\omega)} = \frac{\partial^2}{\partial z^2} [E^{(2\omega)}]$$

$$= \frac{\partial^2}{\partial z^2} \left[\frac{1}{2} \left[E_2(z) e^{i(k_2 z - 2\omega t)} + cc \right] \right]$$

$$= \frac{1}{2} \left[\frac{d^2 E_2}{dz^2} + 2i k_2 \frac{dE_2}{dz} - k_2^2 E_2 \right] e^{i(k_2 z - 2\omega t)} + cc$$

So, now, let me write this equation neglecting this second derivative; so, what you will have? Also another thing, you have every term has exponential minus 2 i omega t and also an exponential plus 2 i omega t. If I need to satisfy this equation for all times, you can show that the coefficient of exponential minus 2 i omega t must be the same everywhere. And similarly, the coefficient of exponential plus 2 i omega t will also automatically be the same on both sides. So, I just do not have to write the complex conjugate term; I only pick up the coefficient of exponential minus 2 i omega t term from each of these terms and write down the equation.

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$$\left(2i k_2 \frac{dE_2}{dz} - k_2^2 E_2 \right) e^{i k_2 z} - \mu_0 \epsilon_0 (2\omega)^2 (-4\omega^2) \times E_2 e^{i k_2 z} = \mu_0 (-4\omega^2) \epsilon_0 d E_1^2 e^{2i k_1 z}$$

$$k_2^2 = \frac{4\omega^2 \mu_0 \epsilon_0 d E_1^2}{E_2} = 4\omega^2 \mu_0 \epsilon_0 n_2^2$$

$$= \frac{4\omega^2}{c^2} n_2^2$$

$$k_2 = \frac{2\omega}{c} n_2$$

$$2i k_2 \frac{dE_2}{dz} = -4\mu_0 \epsilon_0 \omega^2 d E_1^2 e^{-i(k_2 - 2k_1)z}$$

So, what will I get? And remember, there is the factor of half in every term, half also I forget. So, I will get $2 i k^2 d E^2$ by $d z$ minus $k^2 \text{ square } E^2$ into exponential $i k^2 z$, that is the first... From the del square term, I am not writing the exponential minus $2 i \omega t$, because that is I am picking the coefficients of exponential minus $2 i \omega t$ from every term and substituting.

So, this minus $\mu \text{ naught } \epsilon$ of 2ω into, second differential with respect to time, I will get minus $4 \omega \text{ square}$ into $E^2 e$ to the power $i k^2 z$, $\text{del square } E^2 \omega$ by $\text{del } t \text{ square}$ gives me minus $4 \omega \text{ square}$ into this, is equal to $\mu \text{ naught}$ into, $\text{del square } \text{del } t \text{ square}$ of this term factor of half is cancelled off, so I will get, what will I get? Minus $4 \omega \text{ square } \epsilon_0 d E^2$ into e to the power $2 i k^2 z$.

Let me recall, I am substituting the expressions for $E^2 \omega$ and $P_{NL}^2 \omega$ in this wave equation; in this wave equation, I am substituting $E^2 \omega$ and $P_{NL}^2 \omega$. And I am equating terms, which are coefficients of exponential minus $2 i \omega t$ in each of those expressions and I land up with this equation.

Now, can you tell me, what is the relationship between k^2 and ϵ of 2ω ; k^2 is the propagation constant at frequency 2ω . So, $k^2 \text{ square}$ is equal to actually $\omega \text{ square}$ $4 \omega \text{ square } \mu_0$ into ϵ at 2ω ; k is the frequency times $\mu \text{ naught}$ into ϵ of the medium. This is actually, if you substitute this, you will get $4 \omega \text{ square } \mu_0 \epsilon_0$ into $n^2 \text{ square}$. Remember, the ϵ of 2ω is ϵ_0 into $n^2 \text{ square}$ of refractive index; and what is $\mu_0 \epsilon_0$? $1/c^2 \text{ square}$. So, $4 \omega \text{ square}$ by $c^2 \text{ square}$ and 2^2 square ; so, this implies k^2 is equal to 2ω by c into n^2 , which is what we have assumed.

So, k^2 is related to ϵ at 2ω by frequency square 2ω whole square, because the frequency at which the propagation constant is k^2 is 2ω ; so, $2 \omega \text{ square } \mu \text{ naught } \epsilon$ at 2ω ; so this term and this term cancel off. $E^2 e$ to the power $i 2 k^2 z$ - that is common, $k^2 \text{ square}$; so, you will have, there is a minus into minus is plus; so, $\mu \text{ naught } 4 \omega \text{ square } \mu \text{ naught } \epsilon$ into 2ω ; so, these two terms cancel off and this equation will get further simplified.

Now, also let me take this e to the power $i k^2 z$ on the right hand side and simplify this equation; so, what I will get is the followings. **Let me write it ,write here itself**. So, I get $2 i k^2 d E^2$ by $d z$ is equal to minus $4 \mu \text{ naught } \epsilon \text{ naught } \omega \text{ square } d E^2$

square exponential i minus i k 2 minus 2 k 1 z; minus 4, epsilon 0, mu 0, omega square d, E 1 square, e to the power, ok

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A photograph of a handwritten equation on a grid background. The equation is $\frac{dE_2}{dz} =$. In the bottom left corner, there is a circular logo with a star and the text 'NPTEL'. In the bottom right corner, there is the text 'IIT DELHI'.

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A photograph of a handwritten derivation on a grid background. The equations are as follows:

$$\begin{aligned} (2i k_2 \frac{dE_2}{dz} - k_2^2 E_2) e^{i k_2 z} &= -\mu_0 \epsilon_0 (2\omega)^2 (-4\omega^2) \\ &\times E_1 e^{i k_1 z} = \mu_0 (-4\omega^2) \epsilon_0 d E_1^2 e^{2i k_1 z} \\ k_2^2 &= 4\omega^2 \mu_0 \epsilon_0 (2\omega)^2 = 4\omega^2 \mu_0 \epsilon_0 n_2^2 \\ &= \frac{4\omega^2}{c^2} n_2^2 \\ k_2 &= \frac{2\omega}{c} n_2 \\ 2i k_2 \frac{dE_2}{dz} &= -4\mu_0 \epsilon_0 \omega^2 d E_1^2 e^{-i(k_2 - 2k_1)z} \end{aligned}$$

In the bottom left corner, there is a circular logo with a star and the text 'NPTEL'. In the bottom right corner, there is the text 'IIT DELHI'.

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$$\frac{dE_2}{dz} = i \frac{\omega^2}{c^2} \frac{4}{2k_2} dE_1^2 e^{-i\Delta k z}$$

$$= i \frac{\omega^2}{c^2} \frac{2}{\frac{2\omega}{c} n_2} dE_1^2 e^{-i\Delta k z}$$

$$\boxed{\frac{dE_2}{dz} = i \frac{\omega d}{c n_2} E_1^2 e^{-i\Delta k z}}$$

$$\Delta k = k_2 - 2k_1$$

So, let me take this $2i k_2$ on the other side also; so, what I will get is, dE_2 by dz . So, when the i goes to the right hand side, it goes to denominator and because of the minus sign, I take the i in the numerator with the plus sign here. So, I get I , so now $\omega^2 \epsilon_0 \mu_0$ is $1/c^2$, so I get $4/2k_2 dE_1^2$ minus I , let me call this $\Delta k z$.

Now, I substitute for k_2 , so i , so this factor of 2 goes off, so $i\omega d$ by c into 2 by, now k_2 is $2\omega/c$ into $n_2 dE_1^2$ exponential minus $i\Delta k z$. So, this is i times $\omega d - \omega^2$ by c^2 – so, ωd by c into E_1^2 square exponential minus $i\Delta k z$ - a much simplified equation starting from the equation for the second harmonic field.

This tells me, how the electric field of the second harmonic varies with z direction. Please notice that, E_1 is also a function of z in general. So, I cannot solve this equation without writing the equation for E_1 .

And Δk is defined as $k_2 - 2k_1$. Now, E_1 is the electric field of the incident wave at frequency ω ; E_2 is the electric field of this generated second harmonic field. Now, usually what happens is, in this process of second harmonic generation, the efficiencies are usually very weak; few percent, less than a percent, that is the kind of efficiency, we will calculate some numbers.

So, in most situations, the efficiency for second harmonic generation is not very high. What is approximation I can make? That means, E_1 is almost a constant. If there was a lot of conversion from ω to 2ω , the energy at ω would have dropped down significantly. If that is the case, E_1 would have varied significantly, but usually the efficiencies are very low. And so, as a first approximation, I can assume E_1 as a constant; and then, I can immediately solve this equation.

So, for example, it may be possible that I take a medium, I launch 1 Watt of frequency ω , and I generate 10 milli Watts of frequency 2ω . So, the fundamental power will go down from 1 Watt to 990 milli Watts; 10 milli Watts I have generated at second harmonic, this power must have come from the ω frequency. So, the electric field of 1 Watt and 990 milli Watt, which is 0.99 Watts is almost equal; so, that is approximation which I am making here. If the efficiency becomes large, then I cannot make this approximation and I need to solve this equation along with the coupled equation for E_1 .

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$$P_{NL}^{(2\omega)} = \frac{1}{2} \left[\epsilon_0 d E_1^2 e^{2i(k_1 z - \omega t)} + cc \right]$$

$$P_{NL}^{(\omega)} = \frac{1}{2} \left[2\epsilon_0 d E_2 E_1^* e^{i(k_2 - k_1)z - \omega t} + cc \right]$$

$$\nabla^2 E^{(2\omega)} = \frac{\partial^2}{\partial z^2} [E^{(2\omega)}]$$

$$= \frac{\partial^2}{\partial z^2} \left[\frac{1}{2} \left[E_2(z) e^{i(k_2 z - 2\omega t)} + cc \right] \right]$$

$$= \frac{1}{2} \left[\frac{d^2 E_2}{dz^2} + 2ik_2 \frac{dE_2}{dz} - k_2^2 E_2 \right] e^{i(k_2 z - 2\omega t)} + cc$$

So, before we solve this equation, let me look at this term here and tell you what is the meaning of this. **Where is the non-linear polarization term?** Look at this equation, this equation tells me the non-linear polarization is also a travelling wave; this is also travelling wave, $kz - \omega t$. Now, what is the velocity of this wave?

ω by k .

Omega by k 1, it is 2 omega by 2 k 1, which is omega by k 1 and that is the velocity of the wave at frequency omega. So, omega frequency is propagating, it is generating non-linear polarization, which is also travelling along with it. So, omega by k 1 is the velocity of the wave at frequency of the non-linear polarization term.

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$$\text{NLP: } v_{nl} = \frac{\omega}{k_1}$$

$$\text{EMW at } 2\omega: v_{em} = \frac{2\omega}{k_2}$$

$$k_1 = \frac{k_2}{2} \Rightarrow k_2 = 2k_1$$

$$\Rightarrow \Delta k = 0$$

PHASE MATCHING CONDITION

So, non-linear polarization term is propagating a velocity, so polarization is equal to omega by k 1. What is the velocity of the electromagnetic wave at 2 omega? Frequency divided by k 2. The velocity of the electromagnetic wave is the frequency, which is 2 omega divided by the propagation constant at second harmonic. Now, you see this non-linear polarization is the source of this electromagnetic wave, and if the source should keep on continuously feeding energy into the electromagnetic wave, these two velocities must be equal.

If the source weak velocity and the velocity at which it is generating the wave are not at the same speed, they are not travelling together; what will happen is, it will not feed continuously, the energy that is being fed into the omega frequency or 2 omega frequency will not keep on adding constructively.

What is the condition implies? 2 k 1 must be equal to k 2, because if these two velocities must be equal, this implies k 1 is equal to k 2 by 2, which implies k 2 is equal 2 k ,1 which implies delta k is equal to 0; this is very important condition called the phase matching condition.

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$$\frac{dE_2}{dz} = i \frac{\omega^2}{c^2} \frac{4}{2k_2} dE_1^2 e^{-i\Delta k z}$$
$$= i \frac{\omega^2}{c^2} \frac{2}{\frac{2\omega}{c} n_2} dE_1^2 e^{-i\Delta k z}$$
$$\boxed{\frac{dE_2}{dz} = i \frac{\omega d}{c n_2} E_1^2 e^{-i\Delta k z}}$$
$$\Delta k = k_2 - 2k_1$$

(Refer Slide Time: 54:52)

$$\text{NLP: } v_{ph} = \frac{\omega}{k_1}$$
$$\text{EMW at } 2\omega: v_{em} = \frac{2\omega}{k_2}$$
$$k_1 = \frac{k_2}{2} \Rightarrow k_2 = 2k_1$$
$$\Rightarrow \Delta k = 0$$

PHASE MATCHING CONDITION

So, what I will do is, we will discuss this phase matching condition little more detail. In the next class, we will solve this equation, when delta k is not equal to 0. And I will show you that this equation tells me, that for maximum conversion, for maximum efficiency, delta k must be 0. And the origin is one of the origins, physically what is happening is the source of electromagnetic field, which is the non-linear polarization. And the electromagnetic field that it is generating must be travelling at the same speed, for the energy to constructively keep on being fed from the polarization to the electromagnetic field.

So, do you have any questions, yes.

Sir, possible to use the conservation, when I assume these equations right (()) because you ignore 1 term of E omega

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$$\frac{dE_2}{dz} = i \frac{\omega^2}{c^2} \frac{4}{2k_2} dE_1^2 e^{-i\Delta kz}$$

$$= i \frac{\omega^2}{c^2} \frac{2}{\frac{2\omega}{c} n_2} dE_1^2 e^{-i\Delta kz}$$

$$\boxed{\frac{dE_2}{dz} = i \frac{\omega d}{c n_2} E_1^2 e^{-i\Delta kz}}$$

$$\Delta k = k_2 - 2k_1$$

So, the question is, is the conservation energy can be used. Yes, I can use the equations provided, I write this equation and the corresponding E 2 d E 1 by d z equation. Because actually right from the beginning, I am neglecting the generation of all other frequencies; so, I am consistent, I am assuming that is only frequency omega and 2 omega present and those two equations will satisfy the conservation of energy equation.

So, whatever energy I have generated at 2 omega, must have come from omega frequency. So, the sum of the energies of omega and 2 omega must remain constant, as they propagates; that means, I am assuming no other absorption or reflection or scattering; so, the energy conservation is still valid.