

Electromagnetic Theory
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Module - 5
Lecture - 36
Waveguides

Till now we have talked about electromagnetic waves rather in particular uniform electromagnetic waves, its propagation in free space and in isotropic homogeneous medium. We have also seen that electromagnetic waves carry energy and momentum. Therefore, the energy of the electromagnetic waves, it should be possible to transfer one place to another. When you are working with low frequencies, typically less than about 200 megahertz, sending electric signals through either parallel transmission lines or coaxial cable is fairly common place.

However, once the frequencies become higher, one needs special devices to send electromagnetic waves from one place to another. And the method by which these are sent they are known as wave guides, and in this lecture we will be talking about the waveguides, which transfer electromagnetic energy or power from one place to another.

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ELECTROMAGNETIC THEORY

What are waveguides?

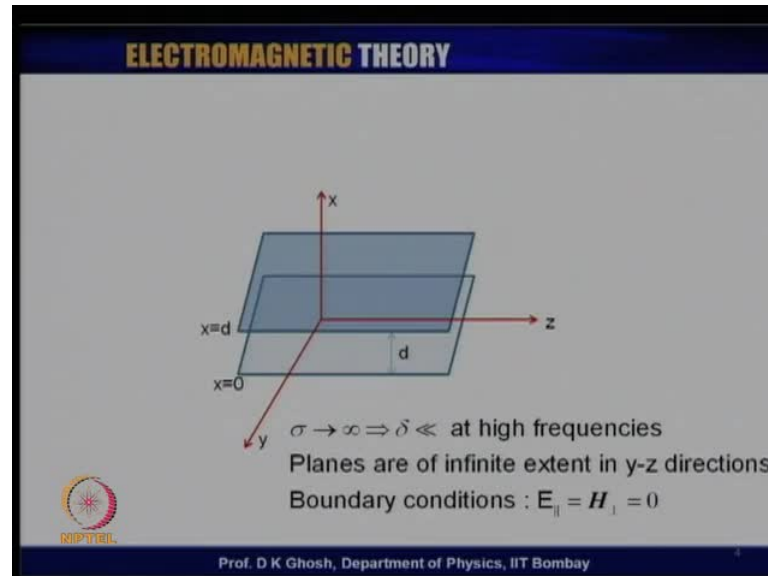
1. They are devices to transmit electromagnetic power from one point to another.
2. For low frequency transmission, one can use parallel transmission lines or coaxial cables. For higher frequencies one uses waveguides.
3. Waveguides can be hollow metal tubes or optical fibers.

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So, as we have pointed out a typical waveguide consists of a hollow metal tube, which is one of the ways of having transferring power. The other more common place today for

example, optical fibers, which are widely used today to carry signals, the light signals for example, from one place to another.

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We will be in this lecture talking primarily about the first part and there in order to explain the basic principle what I have taken are two infinite metal plates. And basically what we have done is to, have one plate over the other separated by a distance d along the x axis, and the, well I have not shown the transmission section, but the power is being sent in the hollow space between these two planes. And it propagates along the z axis. The, between the metal plates there is basically empty space or here and the, so far as the metal plates are concerned, I will take their conductivity to be infinite, theoretically or in practice. What we require is that this skin depth in the material should be much lower, at very high frequencies.


Now, notice that what we have done is that the planes are of infinite extent in the y z direction. And I need to specify the boundary conditions. As we know that the parallel component of electric field and the normal component of the magnetic field, they must be 0 on the metal plates. So, basically we will start with equations governing the 6 components and we will put in these boundary conditions as you go along.

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ELECTROMAGNETIC THEORY

Waveguide

$$\vec{\nabla} \times \vec{H} = \vec{J} + \epsilon \frac{\partial \vec{E}}{\partial t}; \quad \vec{\nabla} \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t}$$
$$\nabla^2 \vec{E} = i\omega\mu(\sigma + i\omega\epsilon)\vec{E}$$
$$\nabla^2 \vec{H} = i\omega\mu(\sigma + i\omega\epsilon)\vec{H}$$


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So, let us let us repeat the standard equations, that we have been taking about namely the Maxwell's equation. We will be primarily taking about the two Curl equation, so one is del cross H equal to J plus epsilon d E by d t and del cross E, that is the Faraday's law is minus mu d H by d t of course, del dot of H equal to del dot of E equal to 0.

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$$\vec{\nabla} \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t}$$
$$\vec{\nabla} \times (\vec{\nabla} \times \vec{E}) = \nabla(\vec{\nabla} \cdot \vec{E}) - \nabla^2 \vec{E} \sim e^{i\omega t}$$
$$= -\mu(i\omega)(\vec{\nabla} \times \vec{H})$$
$$= -\mu(i\omega)[\sigma \vec{E} + \epsilon i\omega \vec{E}]$$
$$\nabla^2 \vec{E} = i\mu\omega(\sigma + i\epsilon\omega)\vec{E}$$
$$\nabla^2 \vec{H} = i\omega\mu(\sigma + i\epsilon\omega)\vec{H}$$
$$\sigma = 0$$

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Now, the thing is that if I have this, let me start with any one of them. Del cross E is equal to minus mu d H by d t, the waves since I am interested in electromagnetic waves the time dependence is typically E to the power i omega t. Therefore, d by d t is the same

as multiplying it with $i\omega$. And let us take a del cross of this equation, we have seen that del cross del cross of a quantity is del of del dot of E, which of course, I know is 0 because of divergence of electric field is 0 minus del square of E and that is equal to minus $\mu \frac{d}{dt}$ which I have seen, shown is $i\omega$ times H.

So I got $i\omega$ times del cross H and that is equal to minus $\mu i\omega$ and so I have got this del cross H is equal to J, which is σE and plus $\epsilon \frac{dE}{dt}$ is $i\omega$ times E. So, this is what I get for del square of E, so if you take care of this minus sign you get del square of E is equal to $i\mu\omega$ times σ plus $i\epsilon\omega$ times E. You can check that the magnetic field del square H also satisfies essential and identical equation, $i\omega\mu\sigma$ plus $i\epsilon\omega$ times H.

So, these two equations and of course, these equations are what we are going to be talking about. However, way you will be looking at the propagation between the planes and therefore, I will take this σ to be equal to 0. So, between the planes σ will be taken to be equal to 0, so let us then write down these equations, that we have. So, let us look at what we have got. So, let us start with for example, the equation on del cross H.

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The image shows a whiteboard with the following handwritten equations:

$$\nabla \times \vec{H} = i\omega \epsilon \vec{E}$$

$$\frac{\partial}{\partial y} H_z - \frac{\partial}{\partial z} H_y = i\omega \epsilon E_x$$

$$\frac{\partial}{\partial z} H_x - \frac{\partial}{\partial x} H_z = i\omega \epsilon E_y$$

$$\frac{\partial}{\partial x} H_y - \frac{\partial}{\partial y} H_x = i\omega \epsilon E_z$$

The whiteboard also features the NPTEL logo in the bottom left corner and a small number '2' in the top right corner.

So, del cross H was σE , but then σE is 0 and I have got $i\omega\epsilon E$, this is my equation. And let me write this down in component form, so which means I will take x y and the z component. So, far as the x component is concerned, so I write del

cross H s x component which is d by d y of H z minus d by d z of H y is equal to i omega epsilon E x, y component that is d by d z of H x minus d by d x of H z, that is equal to i omega epsilon E y. And z component is d by d x of H y minus d by d y of H x that is equal to i omega epsilon times E z. now, these are three equations and I have a set of parallel equations corresponding to the Faraday's law, namely del cross E is equal to minus mu d h by d t.

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$$\nabla \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t}$$

$$\frac{\partial E_z}{\partial y} - \frac{\partial E_y}{\partial z} = -i\mu\omega H_x$$

$$\frac{\partial E_x}{\partial z} - \frac{\partial E_z}{\partial x} = -i\mu\omega H_y$$

$$\frac{\partial E_y}{\partial x} - \frac{\partial E_x}{\partial y} = -i\mu\omega H_z$$

So, look at the corresponding x y and z component, I am going to write down three more equations. So, which is d by d y x components d by d y of E z minus d by d z of E y and this is i omega, so I get minus i mu omega H x d by d z of H x E z, sorry E x minus d by d x of E z equal to minus i mu omega H y. And finally, z component d by d x of E y minus d by d y of E x is equal minus i mu omega H z. So, I have got six equations, this is this is pair, first one pair and this is the second pair, and of course, I have the equation governing the del square of E and del square of H.

So, the what we need to do now, is to try to solve the set of equations, there are huge number of equations, that we will see what is the method of doing a you know systematic study of them. And then put in the boundary conditions namely the tangential component of the electric field and the normal component of the magnetic field, they should be 0 on the plates. Now, let me talk about some classifications. Now, what we will find, will show it later, that the set of equations that we have written down.

This is in terms of x y and z component of the electric field and the x y and z component of the magnetic field. We can express these or rather classify them into three distinct categories, they are done by defining, what are known as the longitudinal or the transverse modes. So, for example, in the present case my direction of propagation is the z direction. So, the z component of electric or the magnetic field is my longitudinal electric or the magnetic field components. The x and the y components will be known as the transverse components, in this case transverse meaning that that that directions are perpendicular to the direction of propagation.

So, let us give some name, now what will see is this, that if you can define the z components. That is I can have a group of solutions, for which the longitudinal electric field component is 0, that is $E_z = 0$. Now, what it means is electric field is transverse. Now, such a mode will be called transverse electric mode called t E mode now remember $E_z = 0$, will also imply $H_z \neq 0$. Now, if $H_z \neq 0$, so t E mode is also known as the H mode, that is H, longitudinal H exists. Similarly, I can have a situation for which the longitudinal component of the magnetic field is 0, that is magnetic field is transverse and that is $H_z = 0$, but $E_z \neq 0$. The corresponding mode is called the transverse magnetic mode, alternatively t m mode, alternatively.

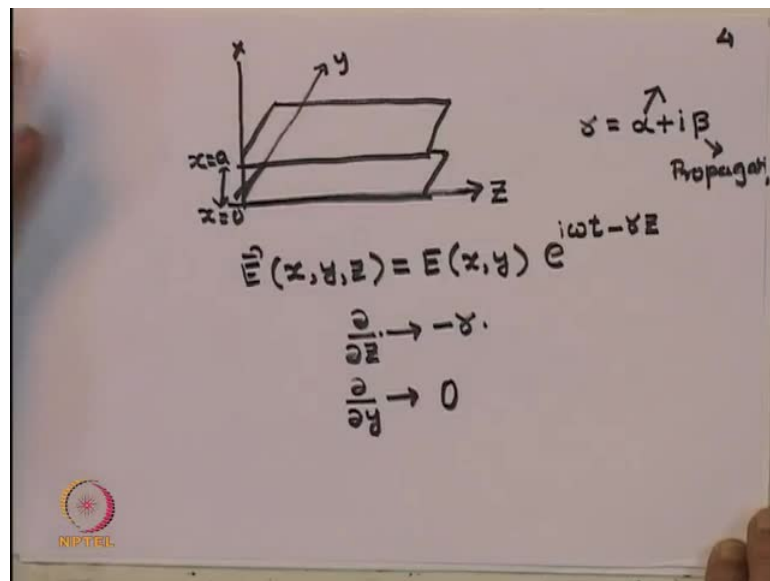
Since, in this case the longitudinal electric field is not equal to 0, it is also called the E mode or the E wave. Now, in a in very special cases not always it is possible to have both E_z and H_z to be equal 0. This is a case which is very similar to propagation of uniform electromagnetic waves through space, where we have seen that the direction of propagation the electric field and the magnetic field that mutually perpendicular to each other. Now, remember because of the fact that you are confining or constraining, the electromagnetic wave to move in a particular direction, subject to certain bounded condition. A typical electromagnetic wave, which is been guided is not a transverse wave, in the sense that we have learnt about.

However, even in guided case, there is there are occasions, where you can have completely transverse wave that is both $E_z = 0$ and $H_z = 0$. Such a wave or a mode is known as t e m mode, namely transverse electromagnetic mode. Of course, there is a possibility that neither $E_z = 0$ nor $H_z = 0$. This will be called a

mixed mode, which you can write either as an E H mode or an H E mode, so these will be the basic way of classifying the modes, which are being guided.

Now, what will do is this, to illustrate our mathematical technique, we will take a simple problem first. And this is a parallel plate waveguide, the parallel plate waveguide meaning, that I have 2 semi-infinite plates, which are parallel to each other with a separation in the x direction equal to a, this is the picture that I had shown you earlier but let me write again.

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So, this is one plate and this the second plate and this distance is of course, this is the x direction. So, this is, I will take this to the x equal to 0, this as x equal to a, so this is my x axis, this should be my y axis and the direction of propagation, which is runs along the plates, these will be my z axis. Now, since the wave is being constrained to propagate along the z direction and the plates are infinite, so the only way the z dependence and come in in the electromagnetic field will be through a for example, take electric field which is a, x y z dependent and so there is a part which depends upon x and y, which we need to finalize, and there is a part which stands for the propagation as we have seen.

So, this I will take as i omega t as we have talked about earlier, minus gamma times z. This gamma is actually in general, if in the space between the media is a general space then this will be essential equal to alpha plus i beta. So, this is that attenuation part and this my propagation part, but since we will assume that this space between the two plates

is essentially vacuum or sigma equal to 0, we will see that this is either propagating namely beta not equal to 0 alpha equal to 0 or attenuating that beta equal to 0 alpha not equal to 0. So, this is the structure, that is the z component of the wave is a propagating solution and of course, that leaves us with x y.

Now, this next thing that we do is this, This is basically done by observing that whenever we have a d by d z, we replaces this with a minus gamma times whatever you are differentiating on. So, d by d z of any component of E for instance will be minus gamma times that component of E. The second thing is that since the plates are of infinite extent in the y direction, there is no boundary condition to be met in the y direction. So, all y must be the same therefore, d by d y of anything is equal to 0. So, there is no variation with respect to y direction because of the nature of infinite extent of the plates.

If this is so then let me come back to the set of equations that I had and write down what does it actually imply? So, let us look at the first set if equations that we had for the electric field, these these were the set of equations that we had here. So, this d by d y of H z that is equal to 0 because I do not have a derivative with respect to y, I do not take. d by d z of H y will be minus minus gamma, so I will a plus gamma times H y.

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$$\left. \begin{aligned} \gamma H_y &= i\omega \epsilon E_x \\ -\gamma H_x - \frac{\partial}{\partial x} H_z &= i\omega \epsilon E_y \\ \frac{\partial}{\partial x} H_y &= i\omega \epsilon E_z \end{aligned} \right\}$$

$$\left. \begin{aligned} \gamma E_y &= -i\omega \mu H_x \\ -\gamma E_x - \frac{\partial}{\partial x} E_z &= -i\omega \mu H_y \\ \frac{\partial}{\partial x} E_y &= -i\omega \mu H_z \end{aligned} \right\}$$

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So, this equation gives me plus gamma times H y is equal to i omega epsilon times E x. Likewise I have d by d z of H x, so which gives me a minus gamma times H x minus d by d x of H z equal to i omega epsilon E y. The third one is d by d x of H y minus d by d

y that term is 0. So, this is equal to $i\omega\epsilon E_z$, so this is one set of equations that I have. The other set which came from the corresponding $\nabla \times E$ equation, I will write down with the very similar way, that is using $\frac{d}{dy}$ is equal to 0 $\frac{d}{dz}$ is minus gamma.

So, that gives me gamma times E_y , the equations are very symmetrical, excepting that you have to replace epsilon with a minus mu, that is because the Faradays law has a minus sign there. So, minus $i\omega\mu$ and E with an H , H_x minus gamma times E_x minus $\frac{d}{dx}$ of E_z equal to minus $i\omega\mu H_y$ and finally, $\frac{d}{dx}$ of E_y is equal to minus $i\omega\mu H_z$. So, this is the second set of equations that I have got. Finally, the wave equation that I have written down, which was $\nabla^2 E$ here, this was $\nabla^2 E$ is $i\mu\omega\sigma$ is zero.

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The image shows a whiteboard with handwritten equations. At the top right, there is a small number '5'. The equations are as follows:

$$\left. \begin{aligned} \alpha H_y &= i\omega\epsilon E_x \\ -\alpha H_x - \frac{\partial}{\partial x} H_z &= i\omega\epsilon E_y \\ \frac{\partial}{\partial x} H_y &= i\omega\epsilon E_z \end{aligned} \right\}$$

$$\left. \begin{aligned} \alpha E_y &= -i\omega\mu H_x \\ -\alpha E_x - \frac{\partial}{\partial x} E_z &= -i\omega\mu H_y \\ \frac{\partial}{\partial x} E_y &= -i\omega\mu H_z \end{aligned} \right\}$$

$$\nabla^2 E = -\omega^2\mu\epsilon \vec{E} \cdot \left(\frac{\partial^2}{\partial x^2} + \alpha^2 \right) \left(\frac{\vec{E}}{H} \right) = -\omega^2 \left(\frac{\mu\epsilon}{H} \right)$$

$$\nabla^2 H = -\omega^2\mu\epsilon \vec{H}$$

There is an NPTEL logo in the bottom left corner of the whiteboard image.

So, let us write it down, $\nabla^2 E$ or $\nabla^2 H$ does not matter, is i well... So, this is $\nabla^2 E$, so I get $i\mu\epsilon$, so there is no ω^2 there. So, i and i is minus 1, so that I am left with a minus $\omega^2\mu\epsilon E$. And of course, a very similar equation for the $\nabla^2 H$ as well, so $\nabla^2 H$ is minus $\omega^2\mu\epsilon H$. Now, remember that these are actually equations each one of is actually three equations because this is similar equation for E_x , E_y and E_z and this for H_x , H_y and H_z .

So, basically this whole set of equations that we have got are our workhorses. And what we will do now is to see, how to symmetrically or systematically solve this separate equations. So, here again since del square of E is d square by d x square, let me just write down for one of them, plus d square d y square which is equal to 0 plus d square by d z square, since d by d z is minus gamma. So, I get gamma square either E or H, I do not care, that is equal to minus omega square mu epsilon times E or H.

So, the this is among the set of equations, so let us now try to find out, how to solve this set of equations? So, once we have said that we have a classification and the classification that will work out while we for example, I will work it out in detail for that T E mode.

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TE-Mode $E_z = 0$;

$$\frac{\partial^2 E_y}{\partial x^2} + \gamma^2 E_y = -\omega^2 \mu \epsilon E_y.$$

$$k^2 = \gamma^2 + \omega^2 \mu \epsilon$$

$$\frac{\partial^2 E_y}{\partial x^2} + k^2 E_y = 0$$

$$E_y = A \sin kx + B \cos kx.$$

At $x=0$ $E_y = 0$

$$E_y = E_y^0 \sin\left(\frac{n\pi}{d} x\right) e^{-\gamma z} \quad n=1,2,3..$$

Remember T E mode, implies that the electric field is transfers the direction of propagation, which implies E z equal to 0. And what will now do is to write down, the equations corresponding to the six equations; that we have wrote down here. So, I have said E z equal to 0 and let us look at what it implies? This implies, let us look at here. So, if E z equal to 0, you can talk about this or this equation also has an E z. So, what we will do is this, that we will write down first the solution for E y, so notice that if E z is equal to 0, then I get H y equal to 0.

It is actually constant because d by d x of H y is equal to 0, but this is a constant and since we have said that it is independent therefore, this is, this will turn out to be

constant. So, let us look at what it implies? So, we will write down $\frac{d^2 E_y}{dx^2}$. If $\frac{d^2 E_x}{dx^2}$ is equal to 0, then I get... So, we will this is equal this plus $\gamma^2 E_y$ is equal to minus $\omega^2 \mu \epsilon E_y$. So, I can now define a new constant. For example, I will define k^2 equal to $\gamma^2 + \omega^2 \mu \epsilon$.

So, this equation gives me $\frac{d^2 E_y}{dx^2} + k^2 E_y = 0$. This of course a very well know equation. So, I write down E_y as equal to some $A \sin kx + B \cos kx$. So, this is, but but you remember that, what we actually mean when we do this, is that the actually y is this quantity, multiplied by E to the power minus γz . Now, I also know that when x is equal to 0, that is on the plate my E_y is equal to 0. So, at x equal to 0 E_y must be equal to 0 because it is tangential component of the electric field.

Therefore, this constant B is equal to 0 and as a result my solution then becomes, full solution let me write down, is some constant which is $E_y(0) \sin kx$, but let me keep it tentative for the moment times E to power minus γz . Now, this $\sin kx$ I have to determine by realizing that at x equal to d , this field must also be 0 because x equal to d is the upper plate. Now, which means $\sin kd = 0$. $\sin kd = 0$ means, k is given by some number $n\pi$ by d times x . So, n is a number which can be 1, 2, 3, 4 etcetera. It cannot be 0 because that would make the E_y also identically vanish.

So, this is my solution for E_y and using these equations, which we have written down earlier. I can now write down for example, what H_x is, so H_x is given by this is what I have got here. So, H_x is γ by minus $i\omega\mu$ times E_y and so therefore, I can write down how much is H_x .

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$$H_x = \frac{1}{i\omega\mu} \frac{\partial E_y}{\partial z} = -E_y^0 \frac{\gamma}{i\omega\mu} \sin\left(\frac{n\pi}{d}x\right) e^{-\gamma z}$$

$$H_z = -E_y^0 \frac{n\pi}{i\omega\mu d} \cdot \cos\left(\frac{n\pi}{d}x\right) e^{-\gamma z}$$

So, 1 over i omega mu and d by $d E_y$, this is what I have determined by $d z$. And this is simply equal to minus E_y^0 remember d by $d z$ is just gamma. So, gamma by i omega mu times sine of $n \pi$ by $d x$ times e to the power minus gamma z . I can write down H_z , remember H_z is not equal to 0, this is very trivially. This is $E_y^0 n \pi$ by i omega mu d times cosine of $n \pi$ by $d x$ E to the power minus gamma z . So, these are the components of H and of course, E i have determined. So, let us look at a specific case.

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ELECTROMAGNETIC THEORY

**Parallel Plate Waveguide
TE₁₀ Mode**

$$E_y = E_y^0 \sin\left(\frac{n\pi}{d}x\right) e^{-\gamma z}$$

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So, look at in this picture, I, we are showing E_y equal to $E_y^0 \sin n\pi x/d e^{-\gamma z}$ to the power minus γz . now, this is the cross section, this is the upper plate at x equal to d , this is the lower plate at x equal to 0 . Let me look at, what is the distribution of E_y ? So, notice that it says E_y must be 0 at both the plates and supposing I take n is equal to 1 , then this is the way the field looks like, this is the way the field looks like that E_y is this. Now, if your to look at the propagation direction, right?

Remember that this is E_y and so if you are looking at this section, then of course, the fields sort of get crowd at the center and the fields sort of becomes, because they get crowd at this center of because the strength is much more here, and then they becomes past towards them. And this is the way TE_{10} mode look like. A word about the nomenclature, this number n its number, I have told you it cannot be 0 , so it can be $1, 2, 3, 4$ etcetera. So, TE_{10} is the lowest transverse electric mode.

The 0 is a written without any specific meaning at this movement, but that is because this is a notation, which is also used for the rectangular waveguides, that will be talking shortly. So, this 0 is to be essentially ignored and this is the one which stands for the value of n .

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ELECTROMAGNETIC THEORY

**Parallel Plate Waveguide
TE₁₀ Mode**

$x=d$

$x=0$

$$H_x = -E_y^0 \frac{\gamma}{i\omega\mu} \sin\left(\frac{n\pi}{d}x\right) e^{-\gamma z}$$

$$H_z = -E_y^0 \frac{n\pi}{i\omega\mu d} \cos\left(\frac{n\pi}{d}x\right) e^{-\gamma z}$$

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If you looked that, so we had drawn H_x and if you looked at H_z it is sort of looks like this, remember it was a cosine function. So, this is, these were the quantities, which we had derived jut now. So, this is the wave the H_x and H_z look like.

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$$H_x = \frac{1}{i\omega\mu} \frac{\partial E_y}{\partial z} = -E_y^0 \frac{\gamma}{i\omega\mu} \sin\left(\frac{n\pi}{d}z\right) e^{-\gamma z} \dots$$

$$H_z = -E_y^0 \cdot \frac{n\pi}{i\omega\mu d} \cdot \cos\left(\frac{n\pi}{d}z\right) e^{-\gamma z}$$

The diagram shows a cross-section of a parallel plate waveguide with two horizontal plates. A square is drawn in the center, containing a circle with a dot, representing the direction of the magnetic field.

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If you want to sort of plot along the axis, so what you will find will be that, this will sort of become... Remember that these are H field, so they become normal near this and out of this you will be closed to in its section like this.

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ELECTROMAGNETIC THEORY

Parallel Plate Waveguide cutoff frequency

$$\gamma = \sqrt{\left(\frac{n\pi}{d}\right)^2 - \omega^2 \mu\epsilon} \equiv i\beta \quad \text{if } \omega > \omega_c = \frac{1}{\sqrt{\mu\epsilon}} \frac{n\pi}{d}$$

$$\text{As } \omega \rightarrow \infty \quad \beta = \omega \sqrt{\mu\epsilon}$$

$$v_\varphi = \frac{\omega}{\beta} = \frac{\omega}{\sqrt{\omega^2 \mu\epsilon - \left(\frac{n\pi}{d}\right)^2}} \rightarrow \infty \text{ as } \omega \rightarrow \omega_c$$

$$\text{For } \omega \gg \omega_c \quad v_\varphi = \frac{1}{\sqrt{\mu\epsilon}} \rightarrow c \text{ for vacuum}$$

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So, let us look at is such a transmission always possible. The first thing that to understand is this, that we had shown that gamma square, when we define k square in terms of gamma square.

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$$\gamma = \sqrt{\left(\frac{n\pi}{d}\right)^2 - \omega^2 \mu \epsilon} = i\beta$$

$\text{If } \omega > \omega_c = \frac{1}{\sqrt{\mu \epsilon}} \frac{n\pi}{d}$ Propagating Soln.
 $\omega < \omega_c \rightarrow$ Attenuating

$$v_\phi = \frac{\omega}{\beta} = \frac{\omega}{\sqrt{\omega^2 \mu \epsilon - \left(\frac{n\pi}{d}\right)^2}} \rightarrow \infty \text{ as } \omega \rightarrow \omega_c.$$

$\omega \gg \omega_c \quad v_\phi = \frac{1}{\sqrt{\mu \epsilon}} \rightarrow c \text{ for vacuum}$



So, we had shown that gamma square or gamma square root of k square, which we have shown just now as n pi by d whole square minus omega square mu epsilon. Now, I want this to be a propagating solution. Now, if I wanted it to be a propagating solution, it tells me that the omega square mu epsilon term, must be greater than this term. So, this can be written as equal to i beta. Beta is the propagation constant, if omega is greater than some critical omega and this critical omega I define, when this quantity equals that quantity. So, in other words this is equal to 1 over square root of mu epsilon n pi by d.

So, if omega becomes greater than this quantity propagation takes place, so I have got propagating solution. If omega falls below that, I have attenuating solution, that is because in this case my gamma becomes real and E to the power minus gamma z. So, I can say evanescent or a decaying solution. The phase velocity if you like is, for the propagating solution is omega over beta. The incidentally, if you just find out the corresponding normal frequencies f and that would, that is then called the cut off frequency. So, omega by beta is given by, omega divided by square root of omega square mu epsilon minus n pi by d whole square.

You notice that as cut off frequency is approached, this phase velocity approaches infinity, as omega goes to omega C. And if omega is very large approaching infinity, then you find the velocity phase velocity is 1 over square root of mu epsilon because in that case this term can be neglected, and I will get 1 over square root of mu epsilon. If

the space between the plates is vacuum or air then this will simply go to the velocity of light, this is for vacuum. So, basically what we have seen in parallel plate capacitor are two things, one that the modes can be classified into transverse electric or transverse magnetic waves.

There is a special case of T e m, which is same as T e 0 0 or T e m 0 0 is possible. We have seen that the frequency of the electromagnetic waves must be greater than a cut frequency in order that such a transmission can take place. I will not be working out the, T e m mode in detail because the method is identical. You have to take H z is equal to 0, impose this similar boundary condition and we have done, and you can write down these components.

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The slide displays the following equations for TM modes in a parallel plate waveguide:

$$H_y = H_y^0 \cos\left(\frac{m\pi}{d}x\right) e^{-i\beta z}$$

$$E_x = H_y^0 \frac{\beta}{\omega\epsilon} \cos\left(\frac{m\pi}{d}x\right) e^{-i\beta z}$$

$$E_z = H_y^0 \frac{im\pi}{\omega\epsilon d} \cos\left(\frac{m\pi}{d}x\right) e^{-i\beta z}$$

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You can see that in this case your E z becomes a cosine function and in fact these are all cosine functions there.


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ELECTROMAGNETIC THEORY

Parallel Plate Waveguide – TEM Modes

TM mode with $m=0$ does not identically vanish

$$H_y = H_y^0 e^{-i\beta z}$$
$$E_x = \frac{\beta}{\omega\epsilon} H_y^0 e^{-i\beta z}$$
$$\frac{E_x}{H_y} = \frac{\beta}{\omega\epsilon} = \sqrt{\frac{\mu}{\epsilon}}$$

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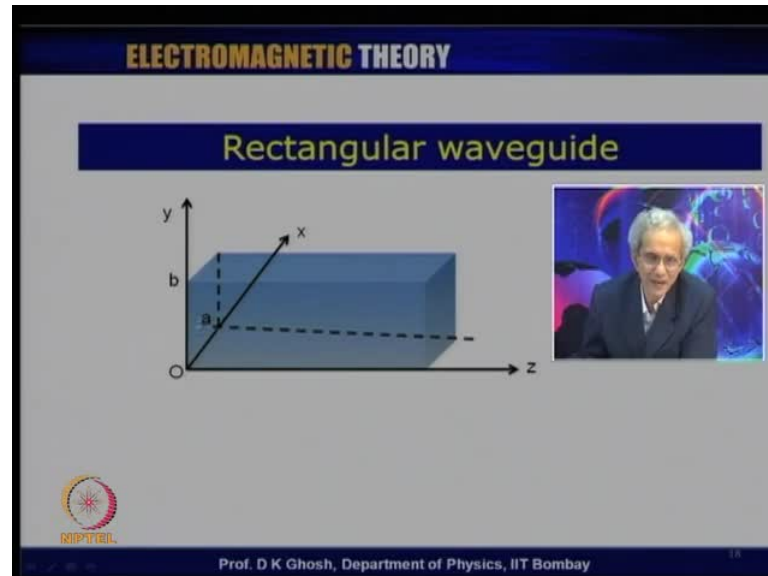
Incidentally we notice that in case of transverse electric wave because these were proportional to the sine function. The H_z was proportional to the sine function, so I could not put n is equal to 0 there because that would have made all the field components vanish. Now, in this case I find that my solutions are in terms of the cosine function. Therefore, in principle I can have a m is equal to 0 here. Now, in such a case if m is equal to 0, then we look at this H_y . So, H_y becomes H_y^0 and cosine is equal to 1, so $H_y^0 E$ to the power of minus $i\beta z$.

E_x becomes H_y^0 into β by $\omega\epsilon$ and this 1 again and this is E to the power minus βz . Now, if you take the ratio of the two and since m is equal to 0, E_z is equal to 0. Now, E_z is equal to 0, H_z is equal to 0 because it is a TM mode, but not all components of the electric and the magnetic fields vanish. I notice that the magnetic field is along the y direction and the electric field is along the x direction. Now, if you take the two ratios, you find E_x by H_y become β by $\omega\epsilon$, which is root of μ by ϵ . Root of μ by ϵ , if you remember is the intrinsic impedance, that we have talked about earlier.

Now, this mode the TM mode corresponding to m is equal to 0, which we would have normally written down as TM_{00} . This is identical to what we have been calling as TEM mode, namely transverse electromagnetic mode. Having done the parallel plate waveguide, this was done primarily to establish the notation and the method of doing.

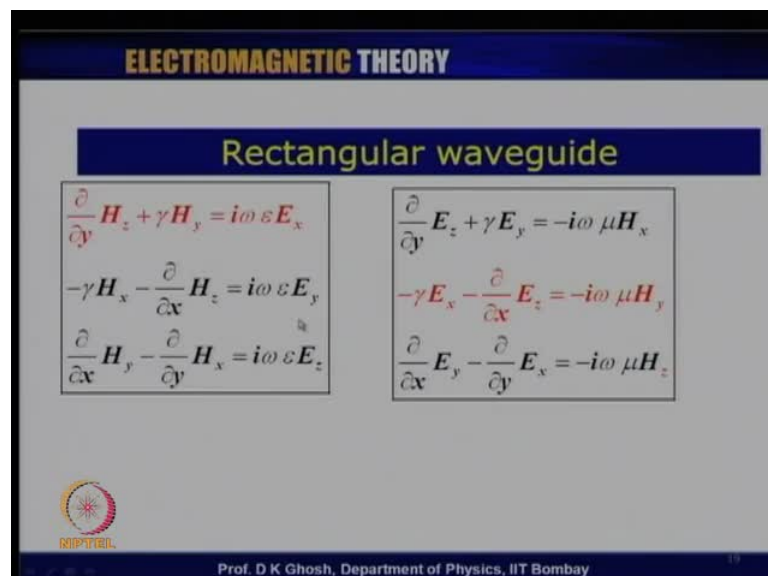
Let me go over to a more practical waveguide. Here I have metal tube, but for convenience I have taken the cross section to be the...

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So, it is not a infinite waveguide. So, basically I have got the metal tube, but this propagation direction I will assume it to be very large. So, what I have got is this, that there are two plates there, separated along the x axis direction by an amount a separated along the y axis direction by an amount b. So, this is the origin and at x equal to a there is 1 and at y equal to b there is the other plate.

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Now, we return back to the same set of equations, that we have talked about earlier, but earlier we had said $\frac{\partial}{\partial y} H_z$ is equal to 0. Now, $\frac{\partial}{\partial y} H_z$ is not equal to 0. So, that term which we had put to be equal to 0, that is still there. $\frac{\partial}{\partial y} H_z$ of for example, this is the set of equations which we had, were $\frac{\partial}{\partial y} H_z$ minus $\frac{\partial}{\partial x} H_y$ and $\frac{\partial}{\partial x} H_z$ still is minus gamma. So, this is one equation.

There is a reason, why I have marked it with red, not anything special but I will be handling these pair of equations separately, then the same with the remaining components, so these are the equations that I have got. But let us look at this pair, that I have written down.

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$$\begin{aligned} \frac{\partial}{\partial y} H_z + \gamma H_y &= i\omega\epsilon E_x \\ -\gamma E_x - \frac{\partial}{\partial x} E_z &= -i\omega\mu H_y \\ i\omega\epsilon E_x &= \gamma H_y + \frac{\partial}{\partial y} H_z \\ &= \frac{\gamma}{i\omega\mu} [E_x + \frac{\partial}{\partial x} E_z] + \frac{\partial}{\partial y} H_z \\ (i\omega\epsilon - \frac{\gamma}{i\omega\mu}) E_x &= \frac{\gamma}{i\omega\mu} \frac{\partial}{\partial x} E_z + \frac{\partial}{\partial y} H_z \\ E_x &= \frac{-\gamma}{k^2} \frac{\partial}{\partial x} E_z - \frac{i\omega\mu}{k^2} \frac{\partial}{\partial y} H_z \\ k^2 &= \gamma^2 + \omega^2\mu\epsilon \end{aligned}$$

$\frac{\partial}{\partial y} H_z$ plus gamma times H_y is equal to $i\omega\epsilon E_x$. Here the other equation is minus gamma times E_x minus $\frac{\partial}{\partial x} E_z$ is equal to minus $i\omega\mu H_y$. Now, I am to use this pair of equations to express things in terms of only E_z and H_z and let us see how it is done? So, here I have got $i\omega\epsilon E_x$, so let me write this. I got gamma times H_y plus $\frac{\partial}{\partial y} H_z$ and this is equal to gamma by... Let me go back to the set of equation, so what I am going to do is I am going to plug in here the expression for H_y , which has E_x and $\frac{\partial}{\partial x} E_z$.

So, this is gamma by $i\omega\mu$, this is this equation. I have got E_x plus $\frac{\partial}{\partial x} E_z$ and of course, I still have that second term $\frac{\partial}{\partial y} H_z$. So, this term E_x , I bring it to that side, so I got $i\omega\epsilon$ minus gamma by $i\omega\mu$ of E_x . This is written in

terms of gamma by i omega mu d by d x of E z plus d by d y of H z. Now, notice what has happened, that E x has been written, in terms of derivatives of E z and H z. Now, you can take the remaining pairs and show that all the components, you should be, it should be possible to write it in terms of the derivatives of the E z and H z alone.

In other words E z and H z completely specify all the derivative, all the components that are there. So, let us simplify this, so this is, you can write this as E x equal to minus gamma by, just do a little bit of simplification, k square. I simply multiply this with this, you can see that this will become omega square mu epsilon plus gamma square with a common minus sign d by d x of E z minus i omega mu by k square of d by d y of h z, were I have defined k square as from here. That is and equal to gamma square plus omega square mu epsilon. So, this is my expression for E x in terms of derivatives of E z and H z. Now, I could do the same thing with other components.

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The slide displays the following equations for a rectangular waveguide:

$$E_x = -\frac{\gamma}{k^2} \frac{\partial}{\partial x} E_z - \frac{i\omega\mu}{k^2} \frac{\partial}{\partial y} H_z$$

$$E_y = -\frac{\gamma}{k^2} \frac{\partial}{\partial y} E_z + \frac{i\omega\mu}{k^2} \frac{\partial}{\partial x} H_z$$

$$H_x = \frac{i\omega\epsilon}{k^2} \frac{\partial}{\partial y} E_z - \frac{\gamma}{k^2} \frac{\partial}{\partial x} H_z$$

$$H_y = -\frac{i\omega\epsilon}{k^2} \frac{\partial}{\partial x} E_z - \frac{\gamma}{k^2} \frac{\partial}{\partial y} H_z$$

$$k^2 = (\gamma^2 + \omega^2 \mu\epsilon)$$

The slide also includes the IIT Bombay logo and the text 'Prof. D K Ghosh, Department of Physics, IIT Bombay'.

Get this whole set of E x, E y, E z, H x, H y and H z written in this fashion. So, I have got these the remaining things. Now, so what we have said is, that things are written in terms of E z H z or rather their derivatives and the remaining 4 quantities namely H x H y and E x E y, we have seen, can be written in terms of that one pair. I have shown one, I have shown the remaining. You should be in a position to work it out therefore, even in this case it seems reasonable to classify things, in terms of solutions for which, E z equal to 0, which you call has the transverse electric mode.

The solutions for which T_m equal to 0, if H_z equal to 0, namely the transverse magnetic waves. The first thing is the solutions, for which the transverse wave solutions, for which transverse electric mode exists, transverse magnetic mode exists. But in this case unlike the case of parallel plate waveguides, no transverse electromagnetic wave is possible, because we have seen that if both E_z and H_z are 0, then since all the components, remaining 4 components are return purely in terms of, the derivatives of E_z and H_z , all the field components will be 0.

So, in other words a rectangular wave guide of the type we have talked about, does not support $T_e m$ mode. So, the restriction is coming in now. Now, so will be talking now about, let us say T_e mode. So, we will do exactly the same as, what we were doing earlier, namely we will write down the Helmholtz equation, remember this was del square of E .

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$$\left(\frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + k^2\right) \tilde{H}_z(x,y) = 0$$

$$H_z(x,y,z) = \tilde{H}_z(x,y) e^{-\gamma z}$$

$$\tilde{H}_z(x,y) = X(x)Y(y)$$

$$Y \frac{\partial^2 X}{\partial x^2} + X \frac{\partial^2 Y}{\partial y^2} + k^2 XY = 0$$

$$\frac{1}{X} \frac{\partial^2 X}{\partial x^2} + k^2 = -\frac{1}{Y} \frac{\partial^2 Y}{\partial y^2} = k_y^2$$

So, I get d^2 by $d x^2$ plus d^2 by $d y^2$. Now, d^2 by $d z^2$ gave you γ^2 , of a component let us say H_z because my transverse electric mode I am doing, plus k^2 because on the other side I had $\omega^2 \mu \epsilon$. So, I bring the $\omega^2 \mu \epsilon$ minus $\omega^2 \mu \epsilon$ to this side and since k^2 is $\gamma^2 + \omega^2 \mu \epsilon$, this is what I get. This times your H_z , I will write it only as a function of $x y$ because the z variation is simply $e^{-\gamma z}$, this quantity is equal to 0, this is our Helmholtz equation.

Now, actual solution is H_z of $x y z$, which is equal to this H_z , that I have written down. Then let us just put a tilde here, H_z of $x y$ times e to the power minus γz . So, when I have next job is to solve this equation, or if I am looking at that transverse magnetic mode, I will solve for the electric field. Now, the way to solve this equation is known as separation of variable method. So, what is done is, to realize that H of $x y$, I should be able to write it the product of something, which I have written as capital X which is the function of x alone and a capital Y which is a functional y alone.

Now, if you plug it in here, you find $Y \frac{d^2 x}{dx^2} + X \frac{d^2 y}{dy^2} + k^2 x y = 0$. Now, notice if I divide this equation by $x y$ all over, I get $\frac{1}{X} \frac{d^2 X}{dx^2} + \frac{1}{Y} \frac{d^2 Y}{dy^2} + k^2 = 0$. I take this to other side minus one over $Y \frac{d^2 Y}{dy^2} = -1$. Since, this depends upon X and this depends upon Y , each one of them must be constant, because otherwise such an equation cannot be valid, for all values of x and y . So, we will write this as a constant and which I will call them call it as k_x^2 . Now, what I will do next is to take this set of separation of variable equation, obtain the solution for capital X capital Y plug it in here and then substitute the necessary boundary conditions.