

Introduction to Time – Varying Electrical Networks

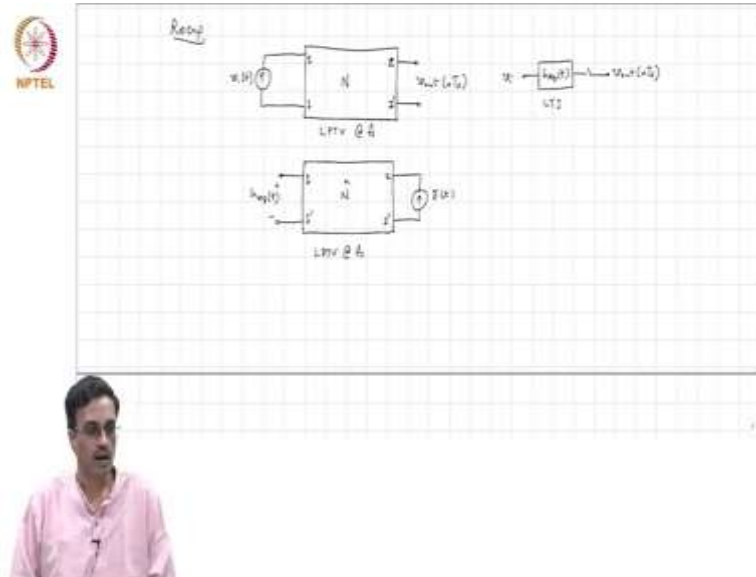
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Lecture No. 74

Finding the equivalent LTI filter of a sampled LPTV system with offset sampling

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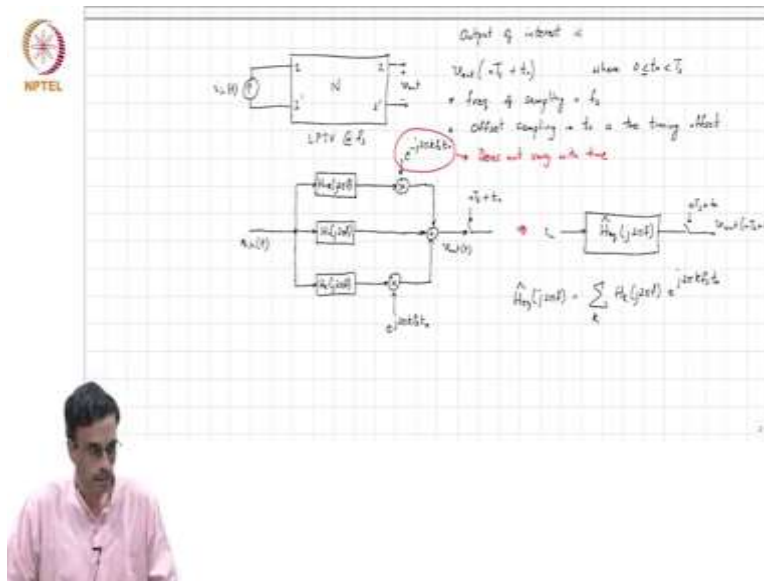


A quick recap what we were doing the last class. So, we did a few examples where we applied the principle that we have been discussing so far. Namely, that if you are interested only in the output samples taken at n times T_s when your drive and LPTV network with an input.

So, we saw that this is equivalent to taking that input v_i of t and exciting an equivalent LTI system with v_i of t and sampling the output of that at n times T_s . And how to find the equivalent of t ? We basically said, we form the adjoint network \hat{N} LPTV \hat{f}_s and excite the output port with δt and the voltage that is developed is h equivalent of t .

So, today, and we saw the example of how one can use this to advantage in a switched RC network as well as the first order continuous time delta sigma converter. If you have not seen a continuous time delta sigma converter before, then you just think of that system as simply an LPTV system and we saw h equivalent of t .

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Today we will look at a different situation where for instance, again, concerning sampled LPTV networks. And I would say it is simply an extension of what we have seen already. So, we have a network, again, we have an input current and an output voltage. And we are, let us say we are not interested in v out of nT_s , but we are interested in v out of nT_s plus some t_0 where t_0 is greater than equal to 0, less than or equal to, less than T_s .

So, in other words the frequency of sampling is the same as that at which the network is varying. Except that earlier we were interested in sampling.

Student: Sampling at adjoints.

Professor: Sample at multiple. I mean, at $0 T_s$ and so on. Now, a reasonable question to ask is what happens when you sample with an offset? So, this is called offset sampling and t_0 is the timing offset. And as usual, we have seen I mean, so remember that the LTI system or LPTV system is $h_{sub k}$ of $j 2 \pi k f_s$, e to the $j 2 \pi k f_s$ times t , $h_{sub -k}$, $j 2 \pi k f_s$, e to the minus $j 2 \pi k f_s$ into t .

So, this is let us call this some in of t , this is v out, but we are only interested in, v out of nT_s plus t_0 . So, in other words, we are sampling this output, this is v out of t and this is being sampled nT_s plus t_0 . So, as usual as we have done before we will simply move the sampler inside. And so, the moment you sample this you basically will now get $j 2 \pi k f_s$ times t_0 .

And likewise, here also you will get this times t_{naught} . And the reason is that when you evaluate this factor at nT_s plus t_{naught} , that nT_s becomes an integral multiple of 2π and this is just what is remaining is just t_{naught} . Correct. So, now, does this depend on time or is this independent of time?

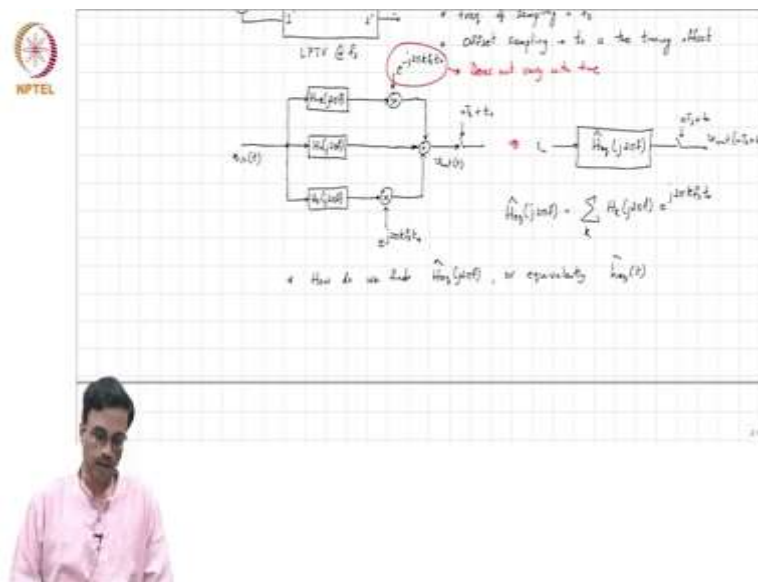
Student: It is a complex number.

Professor: It is a complex number, but this is independent of time. So, does not vary with time, and is therefore it is simply a fixed number, and therefore, you can think of this therefore, as taking in of t passing it through a h equivalent of h equivalent hat of $j 2\pi f$. I use the hat because it is different from what we obtained when t_{naught} was 0, and the output of this is a sampled at nT_s plus t_{naught} . So, you get v out of nT_s plus t_{naught} . And what is h equivalent hat of $j 2\pi f$ is simply the sum over all k of h sub k of $j 2\pi f$, e to the power $j 2\pi$.

Student: $k f_s$.

Professor: $k f_s$ times t_{naught} . Sanity check, if t_{naught} is 0 you should simply get some of the harmonic transfer functions.

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Now, the question is so, as usual the question is, how do we find h equivalent hat of $j 2\pi f$ or equivalently the time domain h equivalent hat of t .

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The slide contains the following content:

- Top Left:** A block diagram of an LPTV system. The input is $v_i(t)$ and the output is $v_o(t)$. The system is labeled "LPTV $\hat{h}(t)$ ".
- Top Right:** A block diagram of an equivalent LTI system. The input is $v_i(t)$ and the output is $v_o(t)$. The system is labeled "LTI $\hat{h}(t)$ ".
- Middle:** A table showing the relationship between input and output samples.

Input Applied at	Output at t_0
t_0	$\hat{h}_{eq}(t_0)$
$t_0 - \Delta t$	$\hat{h}_{eq}(t_0 - \Delta t)$
$t_0 - 2\Delta t$	$\hat{h}_{eq}(t_0 - 2\Delta t)$
- Bottom:** A graph showing the impulse response $\hat{h}(t)$ and its samples. The input is applied at t_0 and the output is measured at t_0 . The graph shows a blue curve for $\hat{h}(t)$ and red dots for the samples.

So, our situation is like this. So, we have an LPTV system. Let us say v_i of t and this is the network and we have v_o of t or let me just use since we are talking about currents and voltages let us do that. So, this is v_i of t and this is a sampled at.

So, we are interested in finding the samples. And we know that this is exactly equal into finding I mean equivalent to an LTI filter whose impulse response we want to find. So, if you excite if v_i of t is and so, the impulse response of the equivalent. So, let us say v_i of t is the input, you have some LTI system \hat{h} equivalent hat of t and the output is sampled at nT_s plus t_0 . So, this is v_o out of nT_s plus t_0 .

So, if I want to find, so if I apply an input $\delta(t - t_0)$ the output waveform here will be \hat{h} equivalent hat of t . So, the sample here will be, the first sample here will be \hat{h} equivalent hat of t_0 . But if I want \hat{h} equivalent of \hat{h} equivalent to $\hat{h}(0)$ what should I put in? At what time must excite the system?

Student: That is t_0 plus t_0 .

Professor: You cannot have t_0 plus t_0 not in the.

Student: Before we have done.

Professor: No, no, no. At what time should I apply the impulse so that I get \hat{h} . When I sample the output at t_0 I get \hat{h} equivalent of 0.

Student: Δt of t minus t_0 .

Professor: That is so if I apply a delta of $t - t_0$. In other words, I must apply the impulse at?

Student: After t_0 .

Professor: At t_0 . Correct. So, if I apply the impulse at t_0 then what will I get here? I will get h equivalent hat of t_0 .

Student: $T - t_0$.

Professor: $T - t_0$. So, if I sample it at T_0 I will get h equivalent hat of 0 . Do you understand this? So, alright. So, in other words so impulse applied and sample value at t_0 let me just do that. So, if apply the impulse at t_0 the sampled value of the output at t_0 will be.

Student: H equivalent of 0 . And h equivalent at 0 .

Professor: 0 . Now, if I want H equal. If I want the sample to be, if I wanted to sample to be h equivalent hat of Δt with Δt is some small number a small time at what time should I apply the input? Do you understand the question?

Student: I understood sir. Like, we have to apply that $t_0 + t_0 - \Delta t$.

Professor: Very good. We have to apply it at $t_0 - \Delta t$?

Student: Δt .

Professor: Δt . Correct. If you apply, if you want to see h equivalent hat $2 \Delta t$ you must apply $t_0 - 2 \Delta t$ and so on. So, therefore, so again, we see that we are applying we are changing the inputs, but our output measurement is always the same. So, that basically means that you can exploit the adjoint again, as expected. We did the same thing with the case when $t_0 = 0$ now we will do this, it is the same. So, this is basically n and this is n had and this is the adjoint.

This is also LPTV at f_s and this is the adjoin network. And this is 2 and this is $2'$. And what should I do? Now in the once you, I mean, we already know how to find the adjoint, you do to time reversal of all the control signals and so on. And so, remember, if you apply an input at the t_1 , so, regional or rather let me call this for a network n if you apply the impulse at t_1 , measure at t_2 , it is the same as you will get the same result if in the adjoint impulse at t_2 measure at t_1 .

And you must have applied the impulse at the output port of the adjoint. So, what should we do now? So, in this experiment, this is corresponding to n we are going on changing the instants at which the input is applied, but the measurement remains the same. So, in the adjoint the input has always to be applied at what time?

Student: T naught. Minus t naught.

Professor: Minus t naught. So, we will apply therefore a current of delta of t plus t naught. And if you measure the output of the adjoint at t naught what will you get at I mean, to get when will you see h hat equivalent of 0 in the adjoint?

Student: Minus t naught.

Professor: At minus t naught. So, what is the waveform that you will get there at the output?

Student: H hat equivalent of.

Professor: H hat equivalent of?

Student: T plus t naught.

Professor: T plus t naught. So, in other words in the adjoint what do you do is if I want to draw a picture what do you like to do is, you apply an input at minus t naught and what will happen? Well the output waveform will do something like this. So, what is the, to get h equivalent the hat of this is h equivalent hat of t plus t naught that basically means that h equivalent hat of t has been moved to the left by t naught, so h equivalent hat of t therefore, is you take this so, this is going to be you shift it to the right here is the h equivalent hat of t .

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NPTEL

freq of sampling = f_s

offset sampling \rightarrow is a timing offset

Does not vary with time

$x[n]$

$x[n+1]$

$H_{eq}(z)$

$H_{eq}(z) = \sum_k H_k(z) e^{j2\pi k n T_s}$

How do we find $H_{eq}(z)$, or equivalently $h_{eq}[n]$

So, this is how you determine the equivalent impulse response of the system, rather the impulse response to the equivalent LTI filter. If the.

Student: Sampling is offset.

Professor: The sampling is offset. That makes sense?