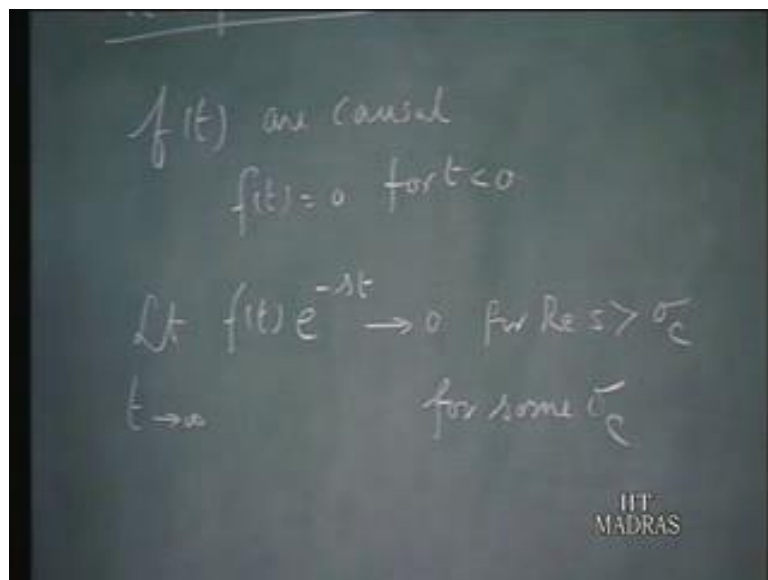


Networks and Systems
Prof V G K Murti
Department of Electrical Engineering
Indian Institute of Technology, Madras
Lecture – 21
Laplace Transforms (2) (Time 01:03)
Properties Of L-Transforms.
Differentiation Integration And Multiplication By T In Time Domain.

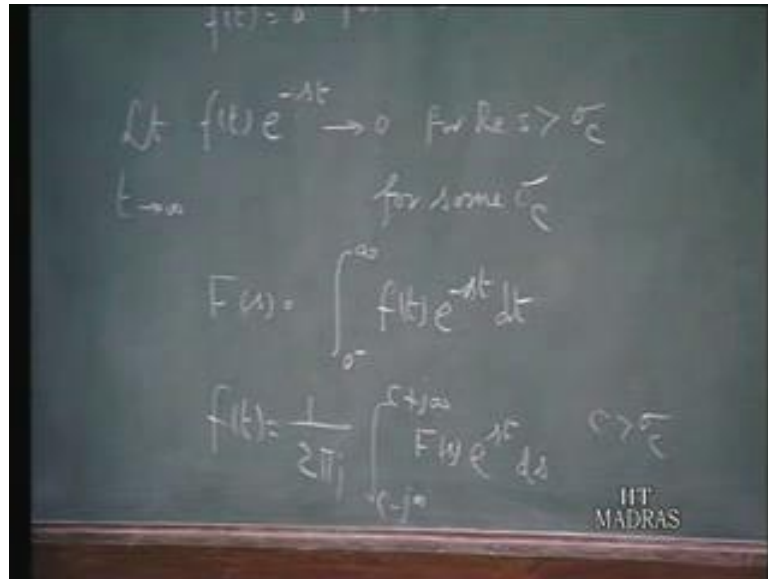
We had introduced ourselves to the concept of Laplace, transformation of the function of time. In the last class and we had a look at the Laplace transforms of some important functions of time like the unity impulse function the unit step function and the 2 sinusoidal functions $\cos \omega t$ and $\sin \omega t$. Let us, first recapitulate some of the important points that was discussed in the last class.

(Refer Slide Time: 01:43)



We are concerned with functions of time which are causal. So, the $f(t)$ which are concerned with are causal that is $f(t)$ is 0 for t less than 0. And further we said their exponential order; that means, limit as t goes infinite of $f(t) e^{-\sigma t}$ close to 0 for real part of s greater than some number σ_c for some σ_c which is the function of $f(t)$. So, if the function of time is the exponential order that is the t goes to infinite the e to the power.

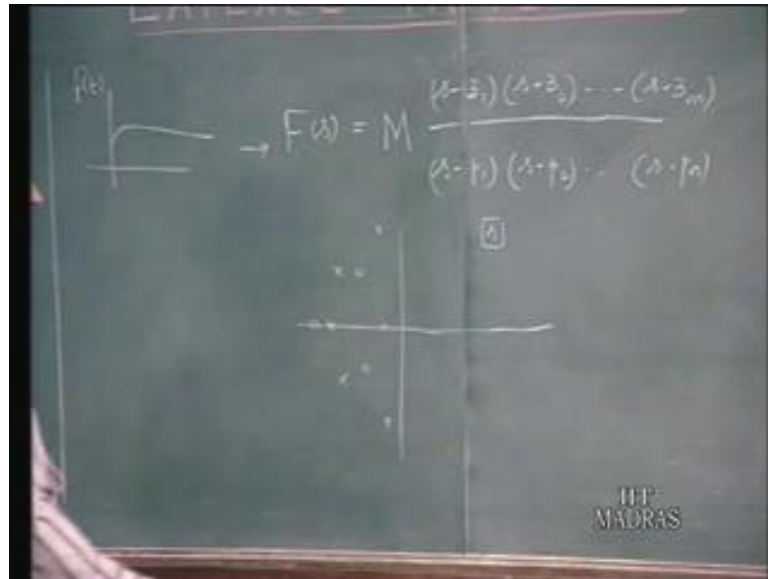
(Refer Slide Time: 3:06)



Of minus st is able to pull thus out the negligible proportion. Then for such time functions we said we have a Laplace transformation f off s . Which is given by the integral sub 0 minus to infinite of f off t e to the power of minus st dt . That is; the forward Laplace transformation formula and the inverse Laplace transform is given by 1 over ft is 1 over $2\pi j$ c minus j infinite to c plus j infinite of f off s e to the power of st ds this is the inverse Laplace transformation formula.

This contour of integration which extends from c minus j infinite to c plus j infinite is the vertical line in the region of the convergence of the Laplace transformation; that means, c must be larger than σ_c . And this contour of integration as i mentioned is referred to as the ground which contour. Now, as the result of this the Laplace transformation that we get is usually a rational function for the type of functions of time that we consider.

(Refer Slide Time:4:36)

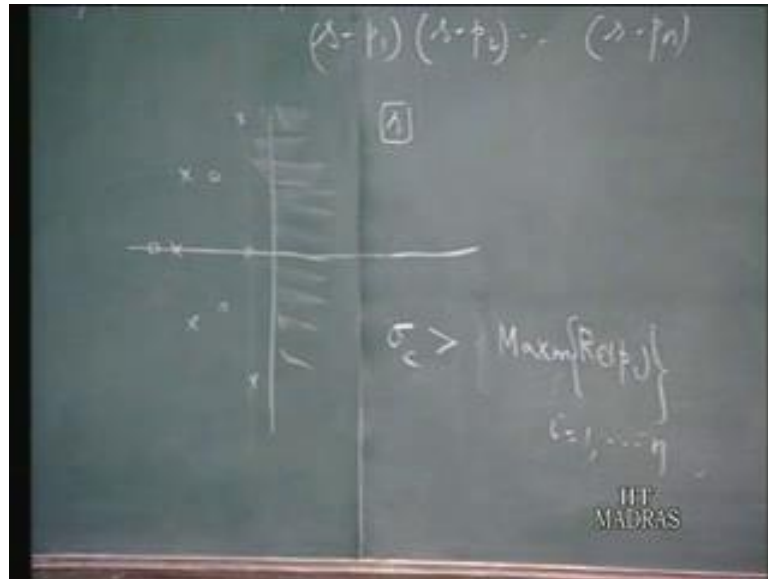


So, for the given $f(t)$ that we have we have an $f(s)$ which is the rational function. And this rational function by means, the ratio of 2 polynomial and this could be the 2 polynomial could be put in the form m is the numerator is factorized as $s - z_1$ $s - z_2$ down the line $s - z_m$. And the denominator likewise the factorized as $s - p_1$ $s - p_2$ $s - p_n$, then the Laplace transformation $f(s)$ is defined by the values of the z value z_1, z_2, z_3, z_m and the values p_1 up to p_n apart from the constant multiplied factor m .

So, this $f(s)$ can be represent in the complex plane by the location of the zeros of $f(s)$ and poles of $f(s)$. So, zeros are normally marked by means of small circles. So, this may be 0 locations poles may be are marked by crosses. So, there may be a pole here pair a poles here extra. So, the pole in 0 locations are marked in the s plane.

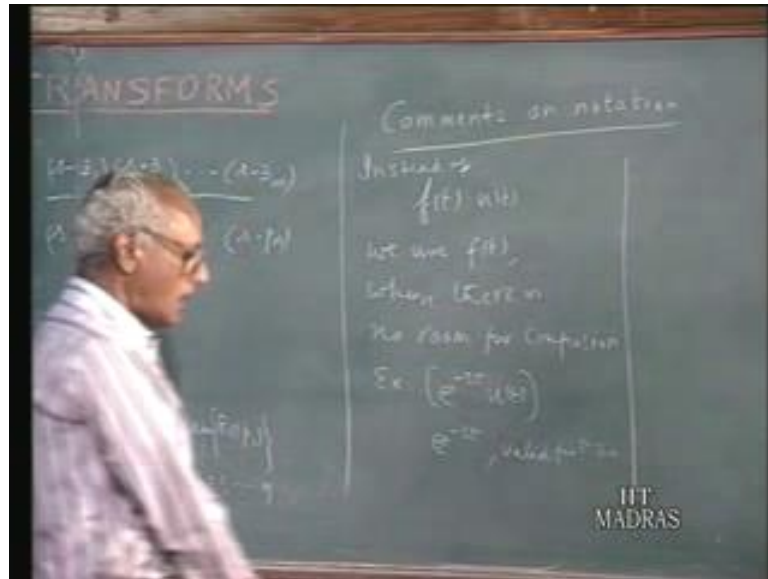
So, the pole of 0 locations of $f(s)$ completely defined $f(s)$ expect for the multiplied factor m . And we also lot that as long as coefficients in the 2 polynomials are real then early complex pole is accompanists by its conjugate any complex 0 is accompanists by its conjugate and the absence of convergence is.

(Refer Slide Time: 06:36).



σ_c the absence of convergence is larger than the maximum value of the real part of p_i . So, whatever the p_i you are having is the maximum value or the real part of p_i i from 1 to n . So, that will be define the absence of convergence; that means, the region of convergence starts from the right most pole. So, if this is the right most pole anything beyond that if there is in the conversions. So, the maximum value of the real part of the poles defines the reason of convergence. And we have seen this in the case of simple functions like the unit stuff function which as the σ_c as 0 of the e to the power of minus $2t$ where σ is minus 2 and so an so for, as that is what we have seen earlier a few comments about.

(Refer Slide Time: 07:36).

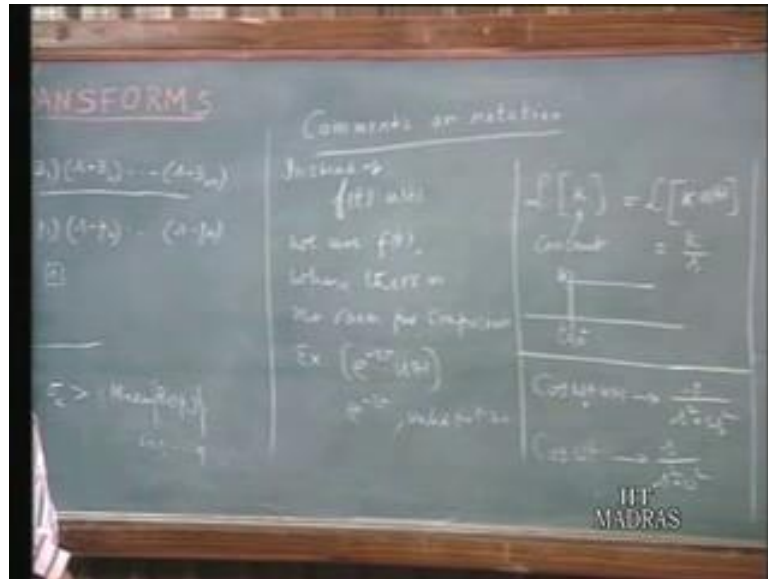


The notation as I mentioned the Laplace transforms exist only for unilateral Laplace transform that we are talking about exist only for causal time functions. So, if $f(t)$ by definition has the value for negative values of time as well, to make sure that we are talking about the truncated part of that function of t we normally write $f(t)u(t)$. To make sure that $f(t)$ the composite function is 0 for negative values of time. Often instead of $f(t)u(t)$ we use simply $f(t)$ then there is no cause for confusion when there is no room for confusion.

In other words example, suppose e^{-2t} this is $f(t)$ that you are talking that you can't find Laplace transform such a function. We have to find the Laplace transform and $e^{-2t}u(t)$. So, when you are got $1/(s+2)$ the inverse Laplace transform of that $e^{-2t}u(t)$. Instead of that we write e^{-2t} itself, with the understanding that we are talking about e^{-2t} valid for t greater than or equal to 0.

So, as long as we understand that whatever time function we are having is valid only for t greater than or equal to 0 and it is 0 for negative values for time. Instead for $e^{-2t}u(t)$ we offer by e^{-2t} to save or write. Another point is suppose we have a some constant.

(Refer Slide Time: 9:41)



Laplace transform of some constant k again by definition by constant k exists for all values of time, but in the context of Laplace time form. We assume the distance function is the value k starting from t equal 0 minus. Because we can take only functions which are 0 for negative values of time in the Laplace transformation consideration; therefore, even if you take the going to Laplace transformation constant. We regard that constant that will have the function will have the value k starting from t equal to 0 minus and 0 for values of t less than this.

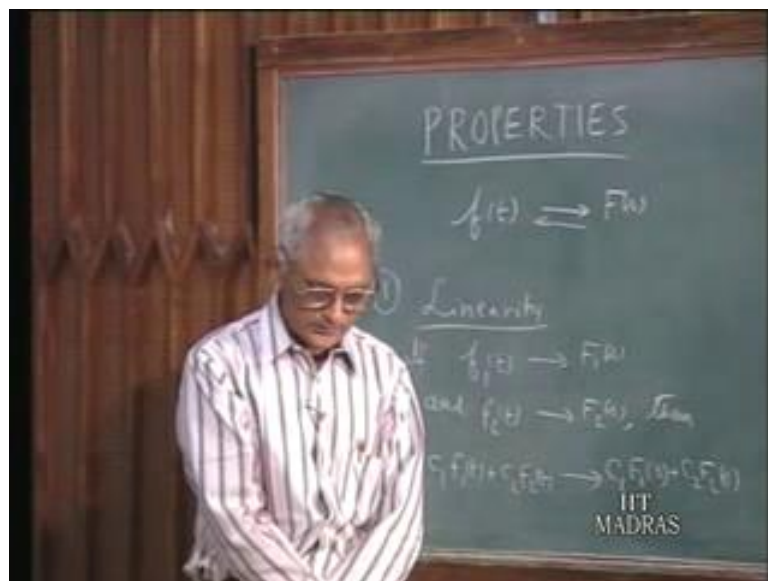
So, when we are talking about the Laplace transform the constant. We treat this as Laplace transform of $k u(t)$ at say this is k apart s . So, we do not sometimes exhibit specifically this u off t which understood just to say, a right. Another feature of common notation is you recall that I said $\cos \omega t$ has Laplace transformation s over $s^2 + \omega^2$. We have taken a particular angular frequency ω and truncate it at t equals for negative values of time and arrive get this Laplace transformation. Now, s is equal to $\sigma + j\omega$. The imaginary part of s is also given the simple ω it is a running variable and to distinguish that running variable ω with this ω .

So, we have given as a special simple as a ω not. But in our manipulation we very rarely decompose s $\sigma + j\omega$ looks at the imaginary part of s . So, where there is no confusion we even right simply $\cos \omega t$ having a Laplace transform. As s

square plus omega square as this is return in this manner in this notation we understand that this omega is not to be confused with the imaginary part of the s. This is the omega a particular value of omega associate a trigonometric function this should not be confused with the imaginary component of s. And secondly, as we mentioned earlier actually we are talking about $\cos \omega t$.

But this additional function $u(t)$ is often dropt. Assuming that, we are talking about a function which has this functional notation only $t \geq 0$ plus onwards 0 minus onwards. So, often instead of righting like this we simply said $\cos \omega t$ is the laplace transform s over $s^2 + \omega^2$. These are minor deviations in the notation we often find to simplify the righting of the various expressions this type of simplification is distorted 2. But basically, we must note that whatever function we are talking about is assume to be 0 for values of t less than 0 minus.

(Refer Slide Time: 13:15)



Let us now, take a look at some of the important properties of the Laplace transforms. We use, the notation that $f(t)$ has the Laplace transform $f(s)$ and $f(t)$ inverse the Laplace transform of $f(s)$ $F(t)$. Most of the properties that we are going to discuss parallel goes which we have already discussed in the context fourier a transform. So, as we go along, you compare the properties that we are having discussing here with those that we discussed earlier in the context of fourier transform. The more or less are similar

except for minor variations which arise as a result of the fact that the Laplace transform concerns itself essentially with causal time functions.

Whereas, the Fourier transform deals with functions which can exist from $t = -\infty$ to $t = +\infty$. So, basically, that is the difference and that creates some differences otherwise the properties are quite similar. First property is something which we already assumed is the linearity property. What we mean is; if $f_1(t)$ has the Laplace transform $F_1(s)$ if $f_2(t)$ has the Laplace transform $F_2(s)$ then a combination of $c_1 f_1(t)$ and $c_2 f_2(t)$ where c_1 and c_2 are arbitrary constants the linear combination of this is having Laplace transform $c_1 F_1(s) + c_2 F_2(s)$ is nothing much poorer about this.

You plug this in to find out the Laplace transform of this plus this in to the expression for the Laplace transform defining integral break it up into 2 parts and you can show it. There is no difficulty doing this. As a matter of fact we assume this relationship, if you recall when we are trying to find out the Laplace transform of $\cos \omega t$. We said $\cos \omega t$ is $\frac{1}{2} (e^{j\omega t} + e^{-j\omega t})$ and therefore, there we are already making use of this area to property.

So, we read that discuss it any further for them this is quite obvious and it can be really demonstrated.

(Refer Slide Time: 15:51)

(a) $\mathcal{L}\left[\frac{df}{dt}\right] = ?$

$$\int_0^{\infty} \frac{df}{dt} e^{-st} dt = \left[f(t) e^{-st} \right]_0^{\infty} - \int_0^{\infty} f(t) (-s) e^{-st} dt$$

$$= -f(0) + s \int_0^{\infty} f(t) e^{-st} dt$$

(b) $\mathcal{L}\left[\frac{df}{dt}\right] = sF(s) - f(0)$

IFT
MADRAS

The second property is going to talk about is the differentiation in time domain we ask the question what is the Laplace transform of df by dt ? This is the question which we like to also. Given that f of t as the Laplace transform capital f of s you will like to find out what is the differential what is the Laplace transform of df by dt . Assuming that the Laplace transform of the differentiated function exists. Then we set up the differential equation defining the integral 0 minus to infinite of df by dt e to the power of minus st dt could be the Laplace transform of this derivative of this function of time.

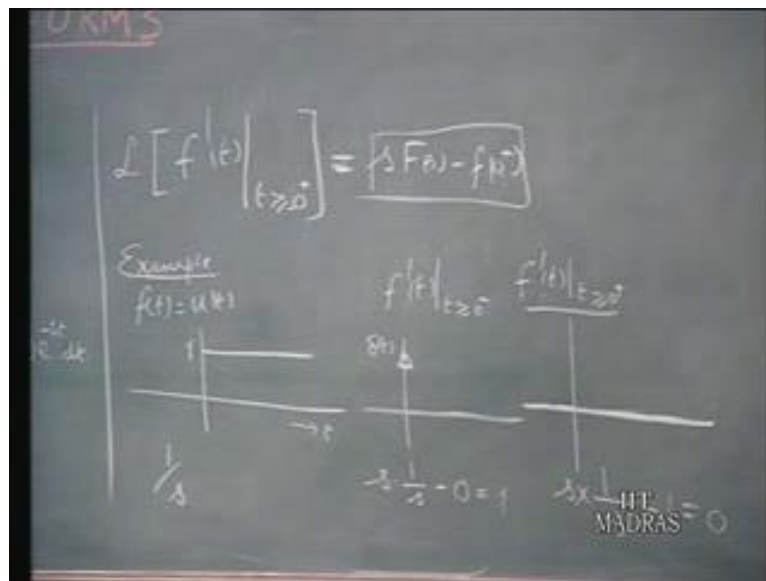
Now, this can be integrated by parts can be integrated by parts, you take the integral of df by dt which is f of t e to the power of minus st evaluated with in the limits 0 minus to infinite minus integral from 0 to infinite of f of t and take the derivative of e to the power of st with reference to time.

So, minus s e to the power of minus st dt . So, when we substitute this 2 limits in the first expression in the regional convergence as t goes to infinite limit as s t goes to infinite of f of t e to the power of minus st that goes to 0 . That is, the definition of the value of s that we are going to take in the regional converters. So, at the upper limit this is 0 at the lower limit e to the power of minus st becomes 1 and therefore, you have got f of 0 minus. Because it is the lower limit minus of f of 0 minus. And here the integration reference to s ; therefore, s times 0 minus to infinite of f of t e to the power of minus st dt . This

integral is capital f off s is Laplace transform of f off t itself therefore, this is s times f off s.

So, this will have s times f off s minus f 0 minus. That is, the laplace transform of f off t which is the derivative of f off t. This is s 2 link that f off t talking about we want to take the behavior of the f off t for t greater than are equal to 0 minus; that means, you are taking the derivative of the function of time.

(Refer Slide Time: 19:17)



We take into account the variation of f off t from 0 minus onwards often we would like to find out the Laplace transform of f prime t for t greater than or equal to 0 plus; that means, any jump that this function has got in transiting from the 0 minus to 0 plus would like to exclude. That the derivative you want to take only from 0 plus onwards we would like to this will got whatever is happening before that. In such case this will be the laplace transform of that will be s f s minus f 0 plus. That means, you are now taking about the Laplace transformation of the derivative. Starting from 0 plus instead of 0 minus dealings of this will become clear as we take an example.

So, in the literature find the Laplace transform of f prime t is write in this manner or in this manner both are equal both are correct depends up on whether you want to consider the derivative starting from 0 minus or 0 plus. Let be illustrate what we meant by this by example, let us take f off t to be a unit step function of t. So, you have this is f off t. Now let me say, that I want to find out to derivative f prime t for t greater than are equal to 0

minus. So, I would like to take the derivative of this function starting from 0 minus. So, as take into account the variation of f off t starting from 0 minus at from 0 minus to 0 plus it goes through jump a 1 0 minus to 0 plus.

So, the derivative of this function at t equal to 0 is an impulse. Because we jumping from 0 to 1. So, derivative is infinite and we integrate with impulse we get arise a step 1 that the derivative desired the behavior of the function 0 minus to 0 plus is an impulse. And from t equals 0 plus onwards there is no derivative the 0 because it is a flat function of time. So, f' of t , t greater than taking into account our variation where t equal 0 minus is in delta function delta t . On the other hand, if you want to consider the derivative only starting from t equals 0 plus onwards. From 0 plus onwards the variation of this function there is no variation; therefore, it is 0 and from all negative values of time we assume the function to a 0 value; therefore, f' t is identically 0.

So, for this that is the function. Now, let us see the Laplace transform of this we know this 1 over s . The Laplace transform the unit f function is 1 over s . What is the Laplace transform of this if you ask this question f' of t is greater than equal to 0 minus? You must use this formula $s f$ off s use this formula $s f$ off s minus $f(0^-)$. This is the formula that we have to use we want to take the derivative that starting from t equal 0 minus; therefore, you write s times 1 over s minus. What is the 0 minus value of this f off t is 0 this is equal to 1 and we know that, the Laplace transform of the unit impulse function which is derivative of this is equal to 1.

On the other hand, if you want to consider the derivative to be 1 which the value from t equal 0 plus onwards. Then we use this form which is $s f$ off s minus $f(0^+)$. So, that will be s times, the Laplace transform of f off t 1 over s minus $f(0^+)$ plus $f(0^+)$ are the unit time function is equal to 1; therefore, we substitute 1 here and that will be 0 and that is should be because our f' t in this case is identically 0 if Laplace transform as well will be identically 0. So, that is the distinction between this 2 formulas; depending up on how you want to take the derivative whether you want to consider the derivative from t equal 0 minus or 0 plus. So, you can use either of this the difference between this 2 will be the Laplace transform of impulse if there is any existing the jobs. Now, let us now proceed further and consider the second derivative of this.

(Refer Slide Time: 24:13)

(b) $\frac{d^2 f}{dt^2} = \frac{d}{dt} \left(\frac{df}{dt} \right)$

$\rightarrow s [sF(s) - f(0^-)] - f'(0^-)$

$= s^2 F(s) - s f(0^-) - f'(0^-)$

(c) d^n

IT
MADRAS

That means; what is the Laplace transform of this $d^2 f dt^2$. Now, the second derivative $d^2 f dt^2$ required is differential, the derivative of d by dt of df by dt . Since we know, the Laplace transform of this we have to apply the formula which we already derived. So, this will be s times the Laplace transform of df by dt is $s f$ off s minus $f(0^-)$ minus i will continue writing 0^- because that more general, minus the value of the derivative df by dt $t = 0^-$ I will write this $f'(0^-)$. So, that will be $s^2 f$ off s minus $s f(0^-)$ minus $f'(0^-)$.

So, the Laplace transform the second derivative of the time function in terms of, the Laplace transform the original time function is given in this and this can be continued.

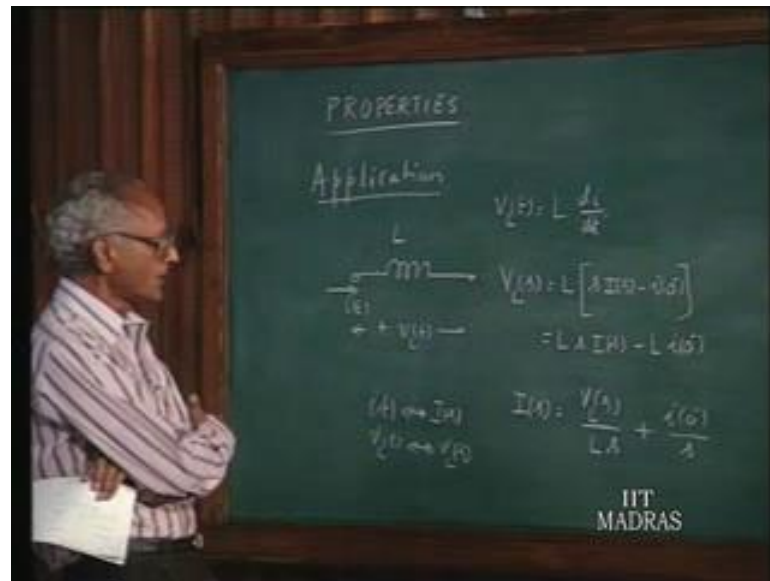
(Refer Slide Time: 25:40)

$$\mathcal{L}\left\{\frac{d^n f}{dt^n}\right\} = s^n F(s) - s^{n-1} f(0) - s^{n-2} f'(0) - \dots - f^{(n-1)}(0)$$

So, a general formula of the Laplace transform of the n 'th derivative $\frac{d^n f}{dt^n}$ can be written as, s to the power of n f off s we taking about the n 'th derivative of time function. minus $s^{n-1} f(0)$ minus like that you take next of the $s^{n-2} f'(0)$ minus down the line to $f^{(n-1)}$ the derivative evaluated 0 minus that all would be, you don't have to remember this formula. All we know is, all we should know is the formula for the derivative and then I can extrapolate this in this fashion that will be. In fact, you can easily get this by repeating the application of the derivative formula.

We have seen how differentiation in time domain carries over in the transform domain by multiplication of f off s by s and adding on to get the information according the initial value of the quantities being differentiated.

(Refer Slide Time: 26:55)



Let us look at the application of this. Application is important in the sense, that the result that are going to get here, will be used r and r again instead froude analysis. Let us consider, an inductor of 1 henries carrying a current i t that the voltage across the inductance be VL. So, this are the 2 terminals of the inductor and you have the current i t passing through 1 generating a voltage VL. The fundamental rule relating VL and it domain this is VL off t off course, VL t equals 1 times di dt.

Now, if you know the Laplace transform for the current i t. Then we should be able to find out the Laplace transform of the voltage VL off t using the differentiation rule that was just now study. So let us, Laplace transform the left hand side and the right hand side assume that i off t has the Laplace transform I off s. So, if you make the Laplace transform for the left hand side and right hand side respectively. And let VL off t have the Laplace transform VL off s then we have VL off s equals 1 time. The Laplace transform of di up on dt will be s times i off s minus i 0 minus.

We can write this alternately as Ls times i off s minus 1 i 0 minus. Now, 2 points have emoted. 1 time domain quantities arte given lower case letters VL and i when you make the Laplace transform of that we use the capital letters VL off s and i off s this is the convention. Secondly, we are interesting to finding out VL off t starting from 0 minus onwards. There is the possibility that i becomes to 0 minus to 0 plus. So, if you want to take the derivative of the current including the transition from 0 minus to 0 plus. Then

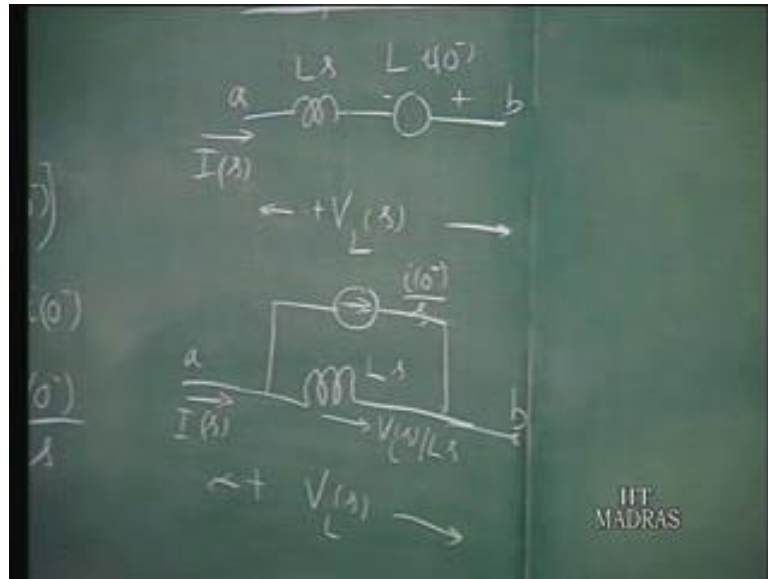
we must take the initial condition here a 0 minus. On the other hand, you are interested in the behavior of the VL off t from 0 plus onwards only and you would like to consider all values of VL from before 0 minus then we have take this for 0 plus. That we are going to illustrate this taking 0 minus values because that is more general situation and optionally you can just take 0 plus as well.

So, this equation tells as that if you know the Laplace transform of current i off s because we calculate VL off s in this manner. But to do this must be the realist value to be current we can inward this expression and find out i off s in terms of VL off s as 1. So, if you do that then I off s can be written as VL off s divided by Ls plus you are going to transfer this to the other side get i 0 minus. So, that gives you an alternative expression relating the terminal current and terminal voltage in the Laplace transform domain i off s is VL off s over Ls plus i 0 minus over s . So, both this equations now are the terminal equations bring that in the transform domain. This the terminal equation bring that in the time domain, this are the equations transform domain both are equal.

Notice 1 thing that, the derivative operator that remove this are purely algebra equations; however, the extra term that you need to have here initial conditions in the invert. Now, provide the initial conditions are 0 suppose i 0 minus is 0. There is the proportional relationship between VL off s and i off s VL off s is Ls i off s i off s is VL off s over Ls ; that means, there is a kind of generalized impedance to the current flow in the transform domain. If the current i off s passes through inductor the voltage is Ls times i off s . So, Ls can be regarded at the generalized impedance of the inductor in the transform the way. same thing here also, VL off s over Ls .

So, 1 over Ls can be thought of the generalized remittance. So, the induct the impedance concept the proportional relationship between voltage and current whatever domain we are taking about transform domain or j omega domain, sinusoidal circuit take situation for all valid provided you don't have to the extra term; that means, if there are no initial condition there is a proportional relationship between voltage and current in the transform away. These initial conditions that spoil the correct. But will see how to take care of that relates, but now, this particular equation can be represented by means of circuit representation like this we have an inductance.

(Refer Slide Time: 32:34)



We can think of that having a generalization impedance Ls and we have a current passing through it not the raw current, but the transformed current Laplace transform variable and across the terminals of inductor we must away among voltage of V_L off s . So, this is the voltage of the terminals of that we know, that V_L off s is the Ls times i off s minus $L i(0^-)$. So, this current is passing through Ls will developing voltage which is Ls times i off s in transform domain. But to complete the picture we must have traditional term here, minus $L i(0^-)$ and this quantity is independent of i off s .

So, $L i(0^-)$ is independent of i off s therefore, there must be extra voltage here which is independent of i off s that is equivalent to voltage source. So, you must have a voltage source in the transform domain which is L times $i(0^-)$ like this. So, this description on such the sequence that is this circuit V_L off s can be thought of as i off s passing through Ls developing a voltage Ls times i off s to as minus $L i(0^-)$ because this is the drop followed by arise. Similarly, if you want to model this equation by means; of equivalent circuit representation. We have a element having a generalized impedance Ls with the terminal voltage V_L off s .

So, naturally the current here will be V_L off s over Ls . But the terminal current is larger than this by additional value; therefore, you should have a current source which is $i(0^-)$ over s . So, the sum of this 2 currents will be your terminal current I off s .

So, this then is another equivalent representation of an inductor in the transform domain. We have, an inductor having generalized impedance Ls in parallel with a current source of $i(0^-)/s$ in the transform domain this is the terminal relation. So, you have 2 equivalent circuits for an inductor, which incorporates initial value in the transform domain.

Both these are called transform diagrams, transform diagram for a particular inductor. One incorporates the voltage source the other incorporates a current source. So, depending upon your convenience when a circuit analysis problem. You favor the useable voltage source we use yes and the other hand, if you are a loud with the current source you can use this in particular if you have a network in which the load equations are preferred in whether of solution may be you need to have like have a current source. But both are equivalent the important thing to keep in mind is that if the terminals of the inductor in the actual circuits are a and b here, the equivalent the terminals in the transform diagram are of this a and b .

We should not think that the terminals of the inductor in the transform diagram are this 2. Because the totality of the structure here represents the inductor carrying an initial current. So, that is be kept in mind what we have done here is representation of an inductor in the transform domain. Where the variables in the circuit are the transformed variables are the functions of time where this is the Laplace transform of the voltage this is the Laplace transform the current. So, the variables are the transformed variables and the element Ls this is called generalized impedance of an inductor. In the transform domain the generalized impedance of inductor transform domain is Ls .

In a similar fashion, we can give a similar representation of the capacitor with initial charge on that we will postpone a latter time we will not do this in greater detail at this point of time. Just as we have generalized impedance Ls for an inductor for a capacitance out of the generalized impedance is $1/s$ of our Cs . We will do that later, but our point in presenting this at this stage is to show an application of the derivative rule in the Laplace transform theory. How you can apply that to evolve circuit models of this time this are called transform diagrams. So, we have a transform diagram of a whole circuit we have similarly generalized impedance for inductance capacitors so on. A whole assembly is for the transform diagram this is the transform diagram, but 1 element which is an inductor.

So, after having consider this who will now go on to the study of the third property of the Laplace transformation multiplication by t.

(Refer Slide Time: 38:04)

The image shows a chalkboard with the following handwritten text:

$$\mathcal{L}[t f(t)] = ?$$

$$\int_0^{\infty} t f(t) e^{-st} dt = -\frac{d}{ds} F(s)$$

$$\frac{d}{ds} F(s) = \frac{d}{ds} \left[\int_0^{\infty} f(t) e^{-st} dt \right] = \int_0^{\infty} f(t) \frac{d}{ds} (e^{-st}) dt$$

$$= - \int_0^{\infty} t f(t) e^{-st} dt$$

ITT
MADRAS

So, we will like to ask the question what the Laplace transforms of $t f(t)$. This is the question that we like to ask. So, the Laplace transform of $t f(t)$ from 0 minus to infinite $t f(t) e^{-st} dt$, this is what we like to ask. Now, to do this let me take d by ds of f off s . Suppose, I want to take the derivative of f off s which is Laplace transform of f off t this is d by ds off $\int_0^{\infty} f(t) e^{-st} dt$ this is f . The integration is to the reference to t and we are taking the derivative reference to another variable.

So, we can introduce this inside the integration time and this can be written as $\int_0^{\infty} f(t) d$ by ds of $e^{-st} dt$. And this when you take the derivative of this with reference to s that is $-t$ is sin outside. So, $\int_0^{\infty} f(t) (-t) e^{-st} dt$ a input a minus sign because of that, $t f(t) e^{-st} dt$. And that is exactly what we are having here you wrote $-\int_0^{\infty} t f(t) e^{-st} dt$. And that integral is equal to $-\frac{d}{ds}$ of f off s as we can see. So, this integral is equal to $-\frac{d}{ds}$ of f off s because is a negative sign; therefore, this will be $-\frac{d}{ds}$ of f off s .

So, when you multiply a function of time by t as for the transform is concerned. It is equivalent to the negative of the derivative of the f off s to s . So, you can see that there is

a nice arrangement here, you multiply by t you get minus d by ds this. And the other hand, when you take the derivative of f off t df by dt you are multiplying that s f off s . In the derivative formula you get d by dt of f off t has the Laplace transform s times of s . On the other hand, when you multiply t off t is d by ds of f off s except for the negative sign is somewhat a dual relationship exists.

(Refer Slide Time: 41:00)

$$t^2 f(t) \rightarrow -\frac{d}{ds} \left[-\frac{d}{ds} F(s) \right]$$

$$= (-1)^2 \frac{d^2 F(s)}{ds^2}$$

$$t^n f(t) \rightarrow (-1)^n \frac{d^n F(s)}{ds^n}$$

IIT
MADRAS

Now, based upon this let me extend this result. Suppose, I take t square f off t then t f off t is minus d by ds of f off s ; therefore, I am straight minus off d by ds off the Laplace transform of t f t which is minus d by ds of f off s . So, this will be minus 1 square d squared by ds square f off s that is all. So, you have t^n f off t the terms to be if you take t^n you have minus n to the power of n n 'th derivative of f off s . All the time we are assuming that the causal functions 1 a verity an f off t we assume that t^n f off t will be 0 for negative values of n this is understood.

(Refer Slide Time: 42:07)

The image shows a chalkboard with the following handwritten text:

$$\frac{d}{ds}$$

Example

$$u(t) \rightarrow \frac{1}{s}$$

$$t u(t) \rightarrow -\frac{d}{ds} \left(\frac{1}{s} \right)$$

$$\downarrow$$

$$r(t)$$

$$= \frac{1}{s^2}$$

MIT MADRAS

So, let me take simple example. We know that, u off t as the Laplace transform 1 over s . Now what happens, if I take the Laplace transform of $t u$. I multiply this by t , then you must take the negative of the derivative of this with reference to s minus d by ds off 1 over s that is 1 over s square and $t u$ is indeed r . So, that Laplace transform of a ramp function is 1 over s square.

(Refer Slide Time: 43:08)

The image shows a chalkboard with the following handwritten text:

$$t u(t) = r(t)$$

$$t^2 u(t)$$

$$\frac{2}{s^3}$$

MIT MADRAS

So, we can write the extend over Laplace transform implies the table, we have already consider $u(t)$, t , $\sin \omega t$, $\cos \omega t$ we can also say $\mathcal{L}\{t^n\}$ Laplace transform is $\frac{n!}{s^{n+1}}$.

(Refer Slide Time: 41:00)

$t^2 u(t)$	$\frac{2}{s^3}$
$t^3 u(t)$	$\frac{3 \times 2}{s^4}$
\vdots	
$t^n u(t)$	$\frac{n!}{s^{n+1}}$

IIT
MADRAS

Now, suppose I take $t^2 u(t)$ this is $t^2 u(t)$ of course, $t^2 u(t)$; that means, state this derivative of these ones again and data's are negative sign. So, if you do that it is minus 2 over s^3 will be derivative and you attach the negative sign it is a 2 over s^3 . Continuing in this fashion suppose I have $t^3 u(t)$ then it will be again you take the derivative of this is 3 times 2 divided by s to the power of 4. And if you proceed further in this fashion, if you have $t^n u(t)$ then you have n factorial over s to the power of $n + 1$.

(Refer Slide Time: 41:00)

$$\begin{array}{l|l} t^3 u(t) & \frac{3 \times 2}{s^4} \\ \vdots & \\ t^n u(t) & \frac{n!}{s^{n+1}} \end{array}$$

That's how it goes as we take higher and higher powers of t the Laplace transform goes in this manner. Let us now consider another example, suppose I have e to the power

(Refer Slide Time: 44:15)

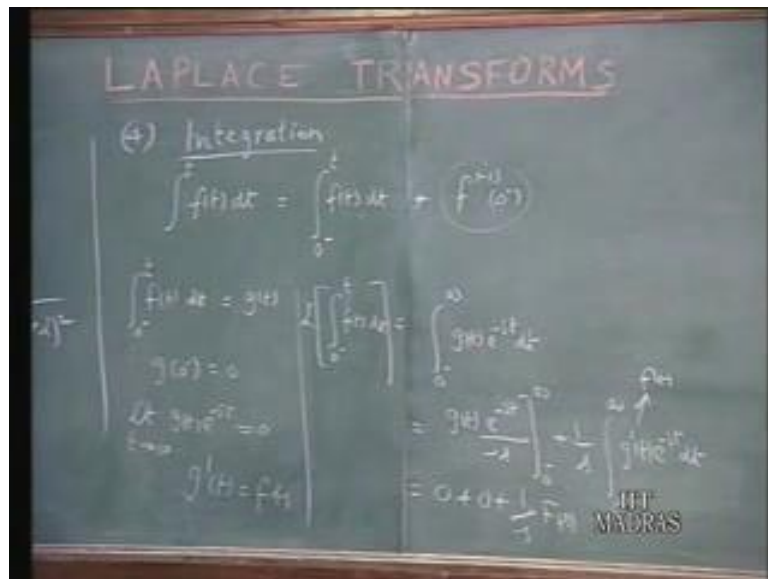
$$\begin{array}{l} e^{-dt} u(t) \rightarrow \frac{1}{s+d} \\ t e^{-dt} u(t) \rightarrow -\frac{d}{ds} \frac{1}{s+d} = \frac{1}{(s+d)^2} \\ t^2 e^{-dt} u(t) \rightarrow \frac{2}{(s+d)^3} \\ \vdots \\ t^n e^{-dt} u(t) \rightarrow \frac{n!}{(s+d)^{n+1}} \end{array}$$

of minus alpha t $u(t)$, then we know the Laplace transform is 1 over s plus alpha. Now, I suppose multiply this like $t e$ to the power of minus alpha t $u(t)$ then its Laplace transform is minus d by ds off 1 over s plus alpha. And that is indeed 1 over s plus alpha Whole Square 1 over s plus alpha Whole Square. And suppose I have t square e to the power of minus alpha t $u(t)$ than it can be shown this is indeed 2 over s plus alpha to the power of 3.

And in fact, as before you have t to the power of n e to the power of minus αt . You continue in the same fashion you have n factorial over s plus α to the power of n plus 1.

So in other words, in we need not remember the Laplace transforms for t multiplied answers like this either e to the power of minus αt or in terms of u off t . It's enough you know, the basic formula for u then when ever is multiplied by t to the power p square t cube extra than you can easily derive them without any difficult. Let me now take the next property which is integration.

(Refer Slide Time: 45:51)



Which is again very important and which commonly occurs. Let me consider, f off t dt in infinite form which we can always regard this as 0 minus to t of f off t dt plus. Suppose, this f off t is a current let us say then what we are doing this by integrating only go for the charge. So, we find it convenient to consider the charge at a function of point at a time t . As the initial charge plus the additional charge that is carried by the current the point 0 minus t the interval 0 minus to t . So, this is the value of integral to limit 0 minus to t to this the initial value of the integral of this write this as f inverse minus 1 with in bacl at 0 minus. That is, this is the initial value of this integral plus the addition to the integral in the value in the interval from 0 minus to t .

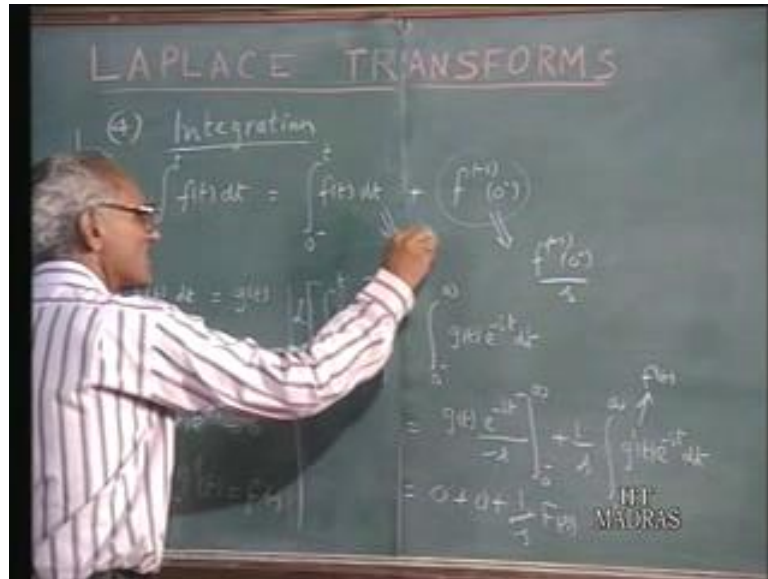
So, this is how we can pit the interfered integral in the form of definite integral because they initialize. Now, you not to find out the Laplace transform of that. Let us find out the

Laplace transform of the 2 individual quantities. So, $\int_0^t f(t-\tau) g(\tau) d\tau$ suppose I called the $g(t)$. Then, we know it straight away that $g(0) = 0$ because like in the integral between the limits 0 to 0 that must specify. And further limit as t has going to infinite of $g(t) e^{-st}$ equal 0 . Because, we are taking the value of $g(t)$ to be Laplace transformable and you are taking the value of real part of s is that this tends to 0 . That is value of st is the region of convergence of the Laplace transform of $g(t)$.

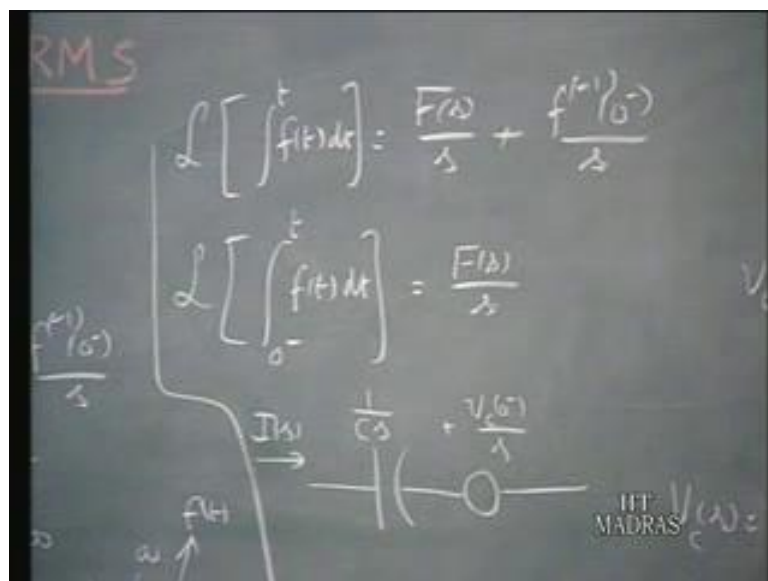
With these conditions and further we also know that $g'(t)$ the derivative of $g(t)$ $g'(t)$ is $f(t)$ because this the integral of this the fundamental role of the integration tells as that the derivative of this is going to be $f(t)$. Using this the Laplace transform $\int_0^t f(t-\tau) g(\tau) d\tau$ Laplace transform of that equals 0 to infinite of this quantity which is $g(t) e^{-st}$ to the power of $-\infty$. And that would be obtained as, let us do the integration by the parts. I will take $g(t) e^{-st}$ to the power of st take the integral of this divided by $-\infty$ between 0 to $-\infty$ plus $1/s$ 0 to $-\infty$ of $e^{-st} g'(t)$ the derivative of $g(t)$ times e^{-st} dt. And this will be at the upper limit this will be 0 because limit as t tends to infinite of $g(t) e^{-st}$ this is 0 at the lower limit this is also 0 or $g(0) = 0$ therefore, this is 0 and you have $1/s$ of $g'(t)$ is the same as $f(t)$.

Therefore, $1/s$ times $f(s)$. So, we have that the Laplace transform of this is $f(s)/s$ and this is a constant f . The initial value of the integral is going to a constant that will be the Laplace transform of this is f initial value divided by s and the Laplace transform of this is f/s .

(Refer Slide Time: 50:00)



(Refer Slide Time: 50:03)



So, we have Laplace transform of the integrate integral $f t dt$ as f off s over s plus the initial value of the integral at 0 minus over s . On the other hand, if you want to take the Laplace transform of the definite integral 0 minus to t of f off $t dt$; that means, the initial value of the integral is ignored. Where this is simply f off s over s , that this are the 2 important results relating to integral. And we can make use of this, when suppose have a capacitor value c and this is the voltage across the capacitor vct and the current the capacitor is i then we can write vct is 1 over c the integral of the current which is the

charge on the capacitor. I can write this further as $\frac{1}{c} \int_0^- t i dt$ plus the initial charge on the capacitor $q(0^-)$.

So, if I take the Laplace transform of the various quantities. I have V_c as the Laplace transform of the capacitor voltage is $\frac{1}{cs}$ this is I over s plus a constant this is the constant $q(0^-)$ over s . So, I can write this as $\frac{I}{cs} + \frac{q(0^-)}{cs}$ which is $\frac{I}{cs} + \frac{q(0^-)}{cs}$ is the initial value voltage across the capacitor; therefore, I can write this as $\frac{V_c(0^-)}{s}$.

(Refer Slide Time: 52:07)

$$= \frac{1}{c} \left[\int_{0^-}^t i dt + q(0^-) \right]$$

$$V_c(s) = \frac{1}{c} \left[\frac{I(s)}{s} + \frac{q(0^-)}{s} \right]$$

$$= \frac{I(s)}{cs} + \frac{q(0^-)}{cs} = \frac{I(s)}{cs} + \frac{V_c(0^-)}{s}$$

So, this equation can be represented nicely in these fashions. This is the Laplace transform the current this is the Laplace transform of the voltage across the capacitor and this is the generalized impedance $\frac{1}{cs}$ I s passing through $\frac{1}{cs}$ create this voltage in addition, I have the initial value of the capacitor voltage divided by s a voltage source.

So, this is then the transform diagram of the capacitor in the same way as we derived it for the inductor. An equivalent in the term in the parallel form the capacitor parallel with the current source can also be derived, I will not do that it leave it the exercise for you that it can be drawn up in similar fashion as we did in the case of an inductor is very important. Now, if you see that you must distinguish between that 2 formulas and when you have the initial charge in the capacitors we have that is equivalent to a source which

is the represents the initial charge in the capacitor or the initial voltage of the capacitor. So, will stop this discussion of the properties of the Laplace transform of this point.

So, in this lecture what we have covered is we started of with revive of the definition of the Laplace transformation and type of functions which are Laplace transformable then we went to discuss the various properties of the Laplace transform in particular what we talked about with linearity property then we talked about the derivative how the derivative time function carry over in the transform domain. Then we discussed how multiplication of $f(t)$ by various powers of t how such functions are transformed in the Laplace transform domain.

In particular we said $t^n f(t)$ becomes as the Laplace transform minus n times $f(s)$ then we went to the integration of the time functions and we distinguish between the definite integral indefinite integral and in both the derivative case and the integral case we saw how they can be applied to the situation of the inductor than the capacitor how the relation between the transformed variables of currents and voltages across inductor and capacitors can be represented by means of such a diagrams which we called transform diagrams. We will continue the discussion of the properties of the Laplace transforms in the next lecture.