

Networks and Systems
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Lecture- 13
Magnitude and Phase Spectra
Fourier Transforms - I

In this lecture, we introduce ourselves to the concept of the Fourier integral and the Fourier transform. Let us recall our motivation in discussing the Fourier series. Recapitulate, that the characteristic property of a linear time and variant system was, that if it is driven by sinusoidal excitation. The response also be sinusoidal in character. This particular property is valid only for sinusoidal. No, other periodic function will have the such a characteristic, will provide such a characteristic.

And once, this particular property is satisfied. We brought in additional concept, like frequency response function, impedance and admittance concepts, which will simplify the analysis of linear systems, driven with sinusoidal excitation functions. Now, this naturally let us to enquire, how we can analyze the linear system. When, it is driven by a periodic non sinusoidal excitation function. And the Fourier series came in very handy in the contest.

In the sense that, very non sinusoidal periodic function can be thought of as composed of several sinusoidal of different frequencies. And by virtue of linearity, principle, we can find out the response to each one of the excitation functions. And superpose the responses to get the total response, under steady state conditions. When, the linear system is driven by a periodic, but non sinusoidal excitation function.

Naturally, you like to see, whether this analysis can be carried forward to a case. When, the excitation function is no longer periodic. We will call such excitation functions, a periodic function. So, when a function is periodic, is it possible to for us to consider, that a periodic function has composed of different sinusoidal functions of different frequencies.

Surprisingly, the answer is yes. And this constitutes the discussion of the subject matter of our discussion under the Fourier integral, Fourier transform methods. I use

the words surprisingly not without justification. Because, when Fourier proposed that any a periodic function can also be a thought of as composed of several sinusoids. You did not gain immediate acceptance from all the mathematicians.

And actually, the fact that the periodic function can be composed of several trigonometric functions of different frequencies. Appears have to be known, even before Fourier, even though, the earlier work is did not carried out this analysis to a logical limit logical extent. But, it is the singular and unique contribution of Fourier to extended this kind of analysis to a periodic functions as well.

What he really said was, that if you have the periodic function, it can be thought of, as the sum of several sinusoidal with infinite number of frequencies. Infinite number of such small components, each with vanishingly the small amplitude. And if you put all of them together, you will get the given the periodic function. So, let us see, how we go about splitting up a given a periodic f of t into various sinusoidal components of the description, which we gave just now.

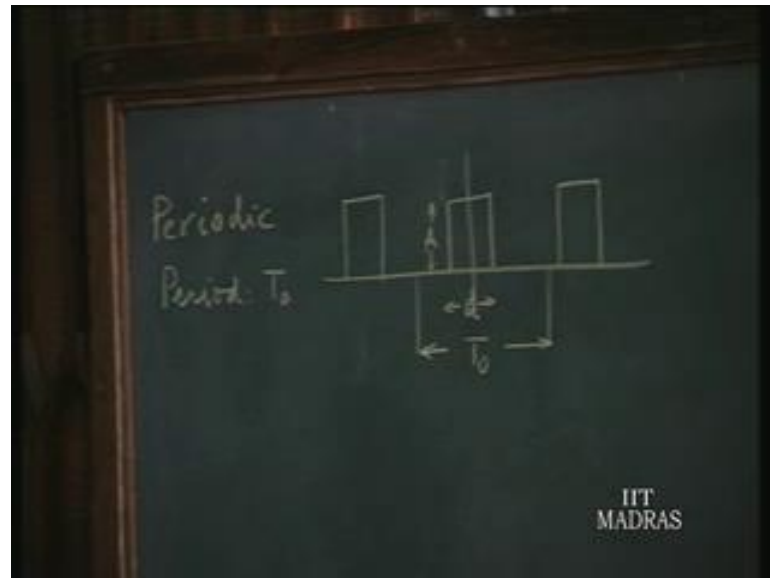
The key to the whole idea is, that if a particular periodic function is given. We can assume that the period is infinity. We are extending from minus infinity to plus infinity. And what happens, beyond plus infinity on one side and what happens, beyond minus infinity, we will not bother about it. So, any periodic function can be a thought of as a periodic function with period extending from minus infinity to plus infinity it occurs just ones.

So, that means, the period of this is going to infinity. What is the consequence? The consequence is, that the frequency of this is 0. And so, if the frequency of this 0 is 0, then how are we going to think of various harmonic components. Because, in the Fourier series, various frequency components, that are presented integrally related, integral multiple of the fundamental frequency.

Here, the fundamental frequency is 0. And in the Fourier series the only term, which has the fundamental frequency, whose frequency is 0 is the dc term and after all, the dc term can split, any periodic function. So, must look for a different approach all together. So, let us see, how we develop this. So, basically the key idea as I mentioned is, that we think of the period of this periodic phenomenon to be infinite extending from minus infinity to plus infinity.

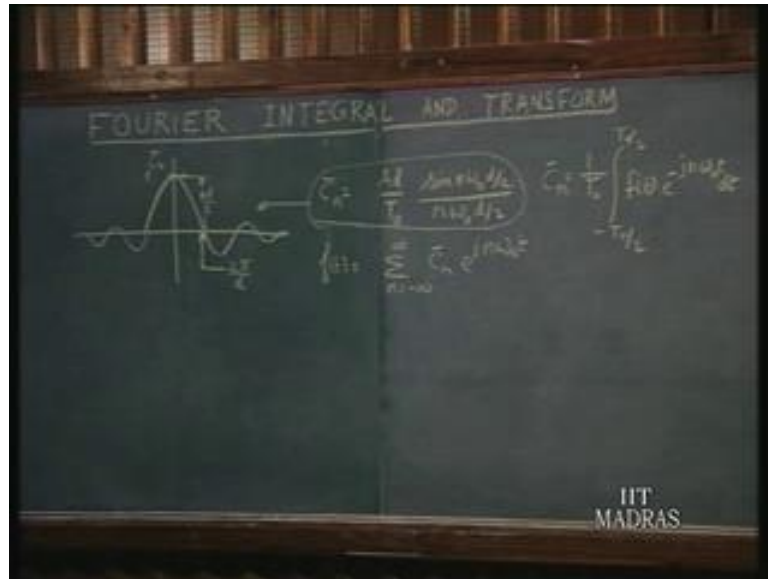
So, what we will do is, will start with the Fourier series of a periodic function. And then see, what happens, when the period becomes larger and larger, till it goes out to infinity. So, the limiting process, we will find out, how the various frequency components arise and what their amplitude will be and so on and so forth.

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So, let me start with a function, which is periodic, period T_0 . Just give a concrete example; I will take a specific wave form. But, all over discussion will be in completely general terms. So, we have periodic pulse train of amplitude A and pulse duration d and the periodicity being the T_0 .

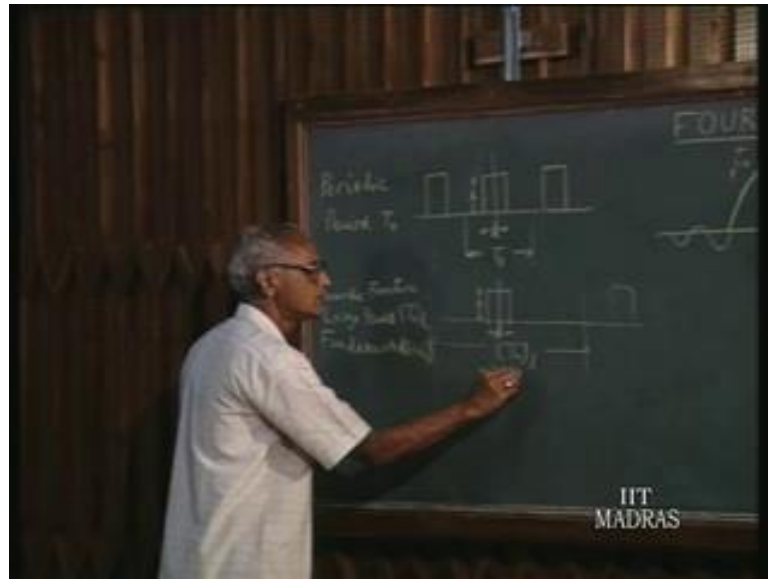
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We know the spectrum for this, would be something like this. Where, this is C_n and we know that, C_n is $\frac{A}{T} \frac{\sin(n\omega d/2)}{n\omega d/2}$. This is something which we have talked about several times. So, let us look at the spectrum. And since, this is going to be a real number, because the function is even. So, it is going to be a real number.

I am not plotting the phase and magnitude separately; I am plotting the entire C_n itself directly. So, it is either way. So, it is always either positive or negative. The height of this is $\frac{A}{T}$ and the first 0 occurs, when $n\omega d/2$ happens to be π . Therefore, the frequency corresponding to this 2π upon d . So, this is the nature of the spectrum, for a periodic pulse train like this.

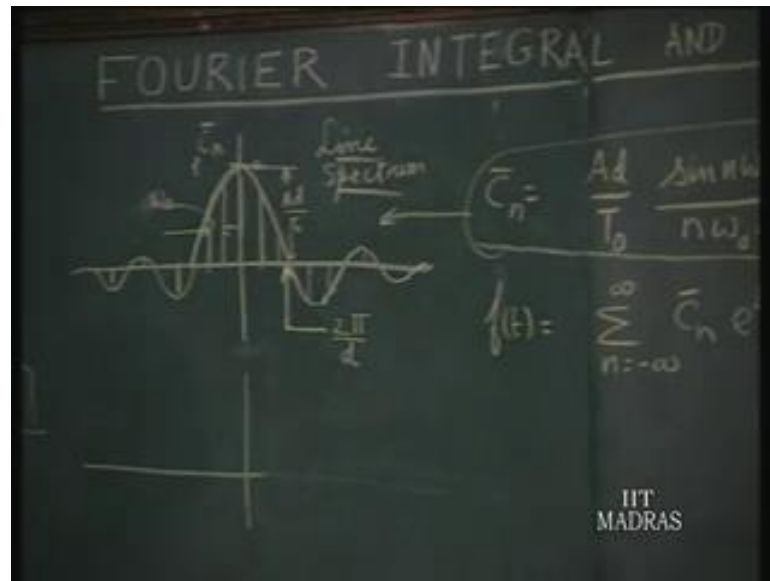
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And if we wanted to find out, $f(t)$ from this, we certainly write this as n from minus infinity to plus infinity of $C_n e^{jn\omega_0 t}$. That is what gets. This is the general that we are getting. This particular C_n ; that we are getting apply to this specific wave form, this is the general formulation. The general formulation for C_n ; in general case would be $\frac{1}{T_0} \int_{-T_0/2}^{T_0/2} f(t) e^{-jn\omega_0 t} dt$ as we already know.

So, these are the two general expressions for C_n and this is the $f(t)$, which is the Fourier series for this wave form. So, these are two general expressions. In the particular case of a periodic pulse train, this would be the particular value of C_n . That would derive for our special case that we have considered.

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Now, let us see, what happens, when we have a periodic pulse train, a periodic function again. But, the period becomes very large, large period. I will simply say T instead of L will indicate as the large period. We have the same waveform, but the period becomes very large. So, I can put it in this fashion. I have the same pulse of amplitude A , duration d . But, the period is very large. That means, the next pulse occurs after long period, very large period.

So, consequently, I will find the angular frequency corresponding to this. Angular frequency here is fundamental period angular frequency is ω_0 . So, we will say, fundamental of this is ω_0 , I am saying. So, it includes emphasize, there is the small angular frequency ω_0 . So, T is 2π over ω_0 .

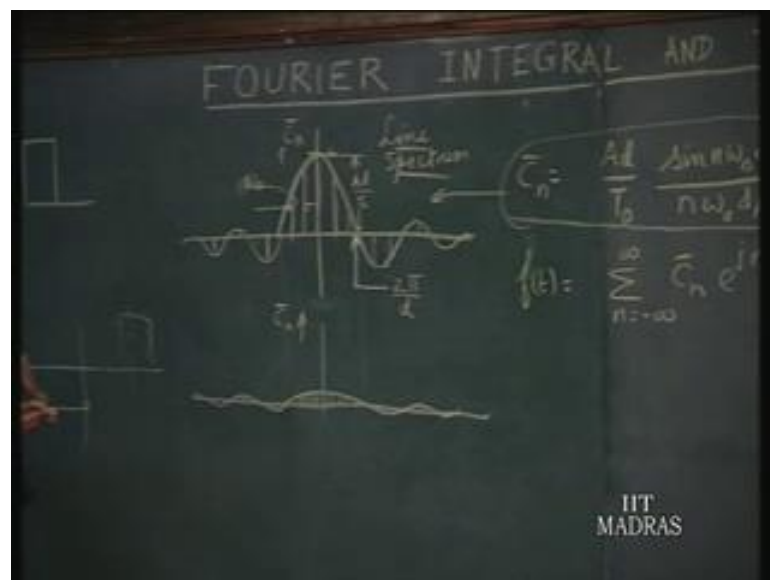
So, what would be the consequence as far as the spectrum is concerned. Now, look at this, this is the formula for C_n . Now, what happens, if T becomes larger and larger, Ad upon T becomes smaller. That means, the initial height here, the amplitude will become smaller. And if you have very large T , this becomes almost negligible smaller and smaller.

I am sorry, what we should say is, C_n is a continuous thing. C_n will be, this is the envelope, what we have drawn is the envelope. The actual C_n will be at discrete points on this. Because, the spacing between two lines being ω_0 . The

fundamental frequency, this is what we had in the first place. What I have to draw here is, the envelope of C_n .

So, the actual C_n will be presented only at discrete points. And this is the line spectrum and the spacing between two adjacent components is ω_0 . Now, when we make the period larger and larger, keeping the pulse with same d ; what you have is, d upon T_0 is smaller, very, very small. So, the result is, that after all $2\pi d$ remains invariant. Because, the 0 occurs here, when $n\omega_0 t$ by 2 is π and that corresponds to 2π upon d .

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And therefore, what you are having here therefore is, that you will have, if you plot C_n , the envelope will be like this. And what happens to the lines, previously the separation between two lines is ω_0 . Now, since the fundamental frequency is $\Delta\omega_0$, which is smaller, the lines come closer and closer. The lines become closer and closer, the whole amplitude comes down.

Now, what we really want to know is, what happens when T_0 close to infinity. Because, we want to treat a periodic function as function of infinite period. So, if T_0 becomes infinitely large. Then, obviously, the whole thing squashes down to the horizontal line, because the amplitude becomes 0. So, that will not give us any useful information. So, must see some other way of handling this information.

So, what we should like to see is, you have some kind of new quantity, which we should plot, which will not become 0, which will be meaningful. And which will give us an idea of the relative changes of the various wave harmonic components. So, to do this, what we will think of is, instead of C_n , after all the whole problem has come because, C_n has 1 over T naught upon this.

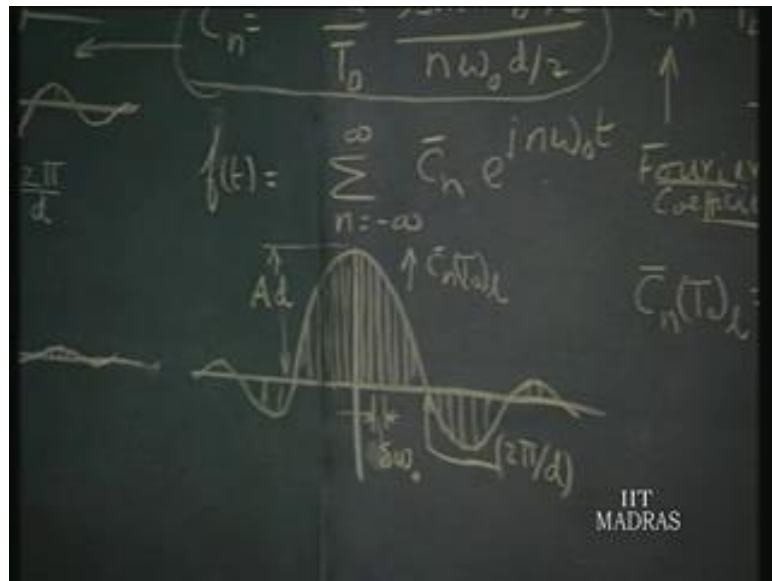
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Handwritten mathematical notes on a chalkboard. The top part shows the Fourier coefficient formula: $C_n = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} f(t) e^{-jn\omega_0 t} dt$. Below this, it says "FOURIER Coefficient". At the bottom, it shows the relationship: $C_n(T_0) = \frac{C_n}{(1/f_0)} = \frac{C_n}{(\Delta\omega_0/2\pi)}$. The IIT MADRAS logo is in the bottom right corner.

So, C_n is the Fourier coefficient. It represents to us the amplitude in a sort of fashion of e to the power of $j n \omega_0$ naught t term. Instead of that, if I plot C_n time T naught, so in our case, C_n stands T naught of 1 . This I can write it as, C_n times Δf naught. Because, the $\Delta \omega_0$ naught is the fundamental angular frequency, Δf not is the corresponding frequency. So, I can write this as C_n upon $\Delta \omega_0$ naught upon 2π .

So, you plot this quantity, we are going to multiply this C_n by T naught of 1 . That means, as far this particular wave form is concerned. Here, multiplying C_n with T naught, then you are going to plot $A_d \sin \omega_0$ naught t by two upon $n \omega_0$ naught d upon 2 . Therefore, the spectrum will now be restored to its original side.

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So, what you are plotting now will be having this envelop, which is exactly the same as this. But, the lines will become closer and closer. And what you are plotted here is, C_n time T_0 of 1 large T_0 . And the amplitude will be, since A_d is multiplied by T_0 . This will be $A_d T_0$. And the spacing between two particular lines becomes $\Delta\omega_0$. And this frequency corresponds to $2\pi/d$, because that is not changing, d is the same.

So, we now see, that if we do not want to deal with C_n , but C_n times T_0 of 1 . Then, we can think of a meaningful way of representation the various harmonic components and what is the relative size.

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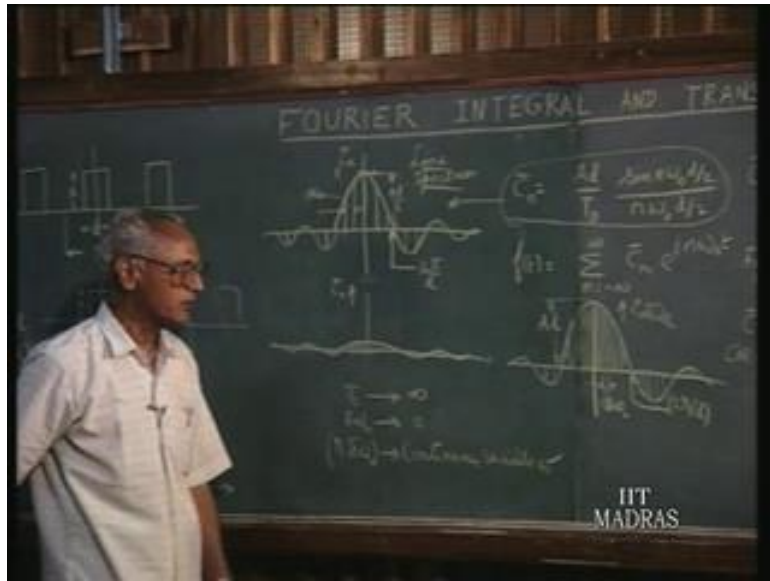
$$f(t) = \frac{1}{2\pi} \sum_{n=-\infty}^{\infty} \frac{C_n}{(\Delta\omega_0/2\pi)} e^{jn\omega_0 t}$$

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Since, this is C_n . The coefficient per unit frequency, this is called coefficient density. Whereas, this is Fourier coefficient, this is called Fourier coefficient density. This is called Fourier coefficient density and once, you have the concept of coefficient density. Then, f of t can be written as further. I will write here, f of t can be written as, you recall f of t is the Fourier series expansion is this, n minus infinity to plus infinity C_n , e to the power of $j n \omega_0 t$.

But, instead of C_n , I would like to talk in terms of the coefficient density. So, I will write this as, C_n upon $\Delta\omega_0/2\pi$; e to the power of $j n \omega_0 t$, sum from n minus infinity to plus infinity. And since, I have divided by $\Delta\omega_0/2\pi$. I have to multiply by $\Delta\omega_0/2\pi$ and then, I have to write here $1/2\pi$. So, that would be the form of the Fourier series now, in terms of coefficient density rather than the coefficients.

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Now, let us see what happens, when you have a periodic function. In the periodic function, we still take this same function of same pulse of amplitude A , duration d . But, there is no other pulse involved, the period is infinity. So, infinite period, so imagine that we have this situation here, where now $\Delta\omega$ tends to 0. So, what this consequence of this T goes to infinity, $\Delta\omega$ become smaller and smaller.

So, what happens the spectrum now, $A d$ is independent of your T . So, that remains the same. This $2\pi d$ remains the same. All that happens now is, as you increase the period, $\Delta\omega$ become smaller and smaller. These lines come closer and closer will be more lines. And the limit, you will find a particular line at every point on the x axis.

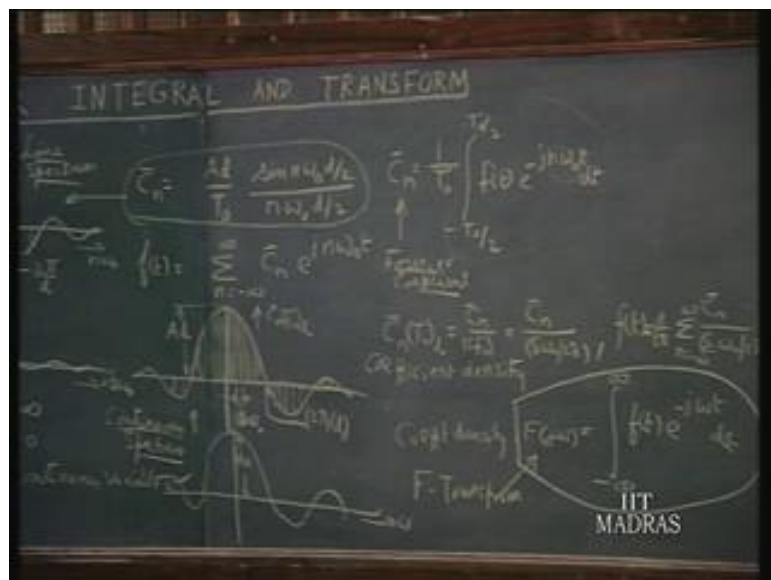
So, that in other words, you will get a frequency component at every point on the x axis rather than the discrete point along the x axis. Because, $\Delta\omega$ become smaller and smaller. So, the line gets crowding almost next each other at every point on the curve. And therefore, we can say $n\Delta\omega$, which $n\Delta\omega$ what is significant $n\Delta\omega$ is the frequency at which a line exists in this spectrum now a line theoretically exists at every point on the spectrum.

So, $n\Delta\omega$ can be replaced by continuous running variable ω . So, that is how the limiting process occurs as the fundamental frequency become smaller and

smaller the harmonic exists at every point along the omega axis. So, in the limit you will have a frequency component at every point along the omega axis. So, the spectrum now would be just like this. But, a continuous spectrum, because you have at every point, there is a meaning for this amplitude.

So, this is the continuous spectrum and the dc value here it is A d. And x axis in terms of omega as before, every where is the omega, this is n omega naught of course here. This is n delta omega naught and here, the running variable is the omega. So, what we have here is, when arrangement where you have every frequency presents. But, the coefficient density corresponding to each frequency is different. That is dictated that is given by the spectrum.

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So, let us see what happens, the coefficient density is now, C_n times T naught, wherever we are having is called now F of j omega. That is F of j omega. How do we get now F of j omega, here you observe, that in the C_n is given by this formula, the coefficient density is obtained by multiplying C_n by T naught. So, once, you multiply C_n by T naught, what you are having is, the integral relation minus T naught upon 2 plus T naught upon 2, f of t , e to the power of minus j n omega naught t , $d t$.

So, in our case it happens, because T naught is multiplied by this. So, what you here is, it is this formula, applicable to the periodic case. And what is the period T naught by minus T naught upon 2 and plus T naught upon 2 are infinity. So, you are integrating F

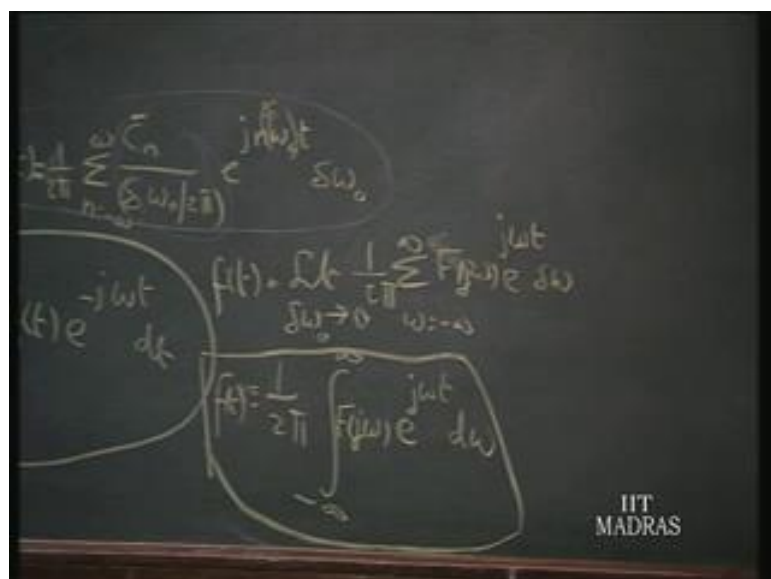
of t multiplied by something over a complete period, which is extended from minus infinity to plus infinity, f of t and e to the power of minus $j \omega t$.

$N \omega$ now, in the first case, it is replaced by $n \Delta \omega$ here. And when you go to a periodic case, $n \Delta \omega$ is a continuous running variable ω . Therefore, this will be the f , e to the power of minus $j \omega t$. So, the coefficient density in the A periodic case will be f of ω , which is minus infinity to plus infinity of f of t , e to the power of minus $j \omega t$ $d t$. And this is called the Fourier transform of f of t .

Now, what is the corresponding result for the Fourier series, that means, Fourier transform is an equivalent of what we had for c of n . The formula to getting the Fourier coefficient, we had some formula. The corresponding formula here is, the Fourier transform, instead of C_n , we deal with F of ω . Why I will explain for a moment later, about the notation.

Now, what is the result we have corresponding to this, f of t in the periodic case is expressed as the sum of different exponential functions. Now, what is that we have here, f of t as you recall in the case of large period is this. So, now, we have to take the limit of this as $\Delta \omega$ becomes goes to 0. So, limit of f of t , limit of this series as $\Delta \omega$ goes to 0, means we have to take the integration. After all, $C_n \Delta \omega$ by 2π , f of ω , this is F of ω .

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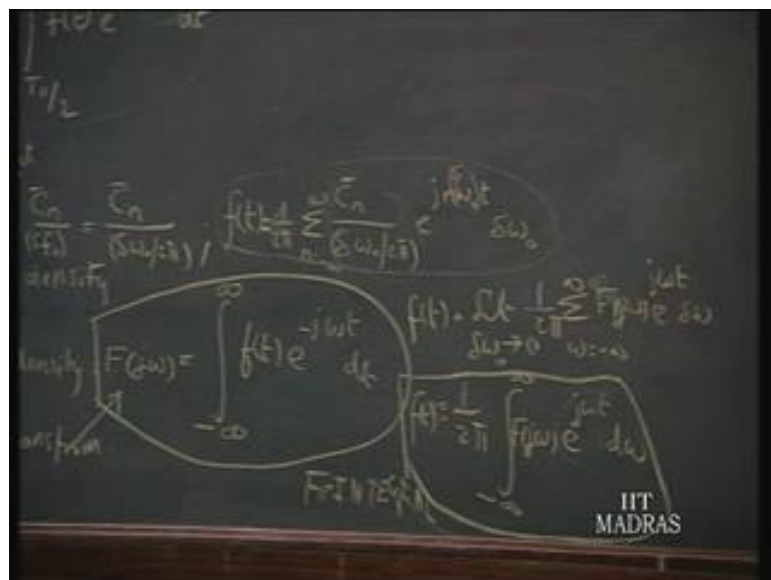


Therefore, we have to write here as the f of t, write it here. Limit as delta not omega naught goes to 0 of 1 over 2 pi, the summation of C n over delta omega naught over 2 pi, we call that f of j omega. So, f of j omega, e to the power of j omega t, because n omega naught becomes, n delta omega naught becomes. Here, also delta omega naught in the large case, n delta omega naught becomes omega and then, this is delta omega.

And the frequency omega, because n from minus infinity to plus infinity. That means, omega is minus infinity to plus infinity. So, thus really becomes 1 over 2 pi. So, that summation in the limit becomes integration from minus infinity to plus infinity and omega f of t e to the power of j omega t d omega, this is f of t. So, this is f of omega, this is must be F of j omega the same.

So, instead of Fourier series for f of t, instead of f of t is being thought of the summation of several exponential functions like this. What we are really having is, f of t is consisting of similar exponential functions, e to the power of j omega naught t, where omega can take any value from minus infinity to plus infinity. Each of these is multiplied by a coefficient, which is F j omega d omega by 2 pi.

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And you are integrating that over the complete omega axis from minus infinity to plus infinity, this is f of t. And this is called Fourier integral, this is called Fourier integral. So, instead of series, you have an integral and instead of a formula like this for

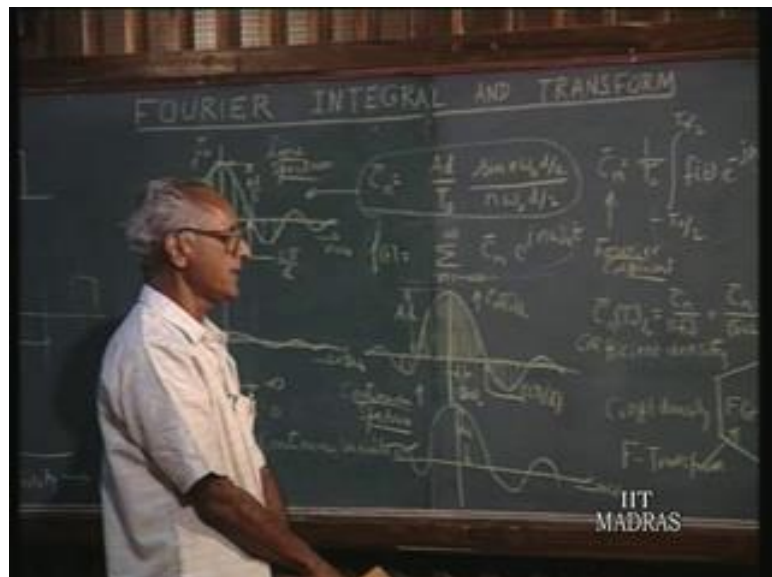
calculating the Fourier coefficient. You are calculating the Fourier coefficient density, which is referred to as a Fourier transform.

Now, a few words about this, why we get this 2π here. This 2π arises, because we are calculating the Fourier density, coefficient density in terms of frequency rather than omega. This is, how it is conventionally defined, therefore, because we are talking about the coefficient density. In terms of having the denominator frequency rather than in hertz rather than radius per second, we get this 2π term.

And secondly, we call this F of $j\omega$ instead of f of ω . Because, later on, we like to relate the Laplace transform, where this f of $j\omega$ can be a thought of as a special case f of s , we deal with Laplace transform domain. Therefore, it would be convenient for us to put this f as $j\omega$, because after all, when you talk about the exponential terms here.

Every j is accompanied by every frequency is accompanied by j . So, $j\omega$ would be all right. You can think of f of ω as well, but f of $j\omega$ would be more convenient for us to use.

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Now, let me quickly summaries, what we have done up to this stage. We started with a periodic function. The formula for finding the Fourier coefficients for the periodic function is this, we know general function. And once, we have the various coefficients,

we can construct the f of t by combining all such exponential terms from minus infinity to plus infinity. This is the Fourier series and this is the integral; that we use to calculating the Fourier coefficient.

Taking a particular case of a periodic pulse strain, we showed that it has the spectrum like this and the characteristic of the spectrum is, that is the line spectrum. You have coefficients only at integral multiples of the basic frequency. Therefore, this spacing between two adjacent lines is the fundamental frequency ω_0 in radians per second.

Now, then we enquired as to what happens, if we keeps this particular pulse constant at the center around the origin and allow this T_0 to become larger and larger. So, that is the transition, before we go make T_0 equal to infinity to see the trends. So, as T_0 becomes larger and larger, we observe that the amplitude of this various C_n coefficients become smaller and smaller.

So, the spectrum, line spectrum we had becomes shallower and shallower. It will be of this order. It will be like this and the height will diminish as T_0 increases for them. But, the another fact, that we noticed was, the spacing between the lines now. Instead of ω_0 , it was the fundamental frequency, earlier becomes smaller and smaller.

Because, as T_0 increases ω_0 becomes smaller and smaller. But, this is not meaningful to us in the limit, when T_0 becomes infinity, because when we are moving towards the direction in which the period is considered to be infinitely large. So, if the period is going to infinity large, this whole thing become 0. And this C_n will become 0. And therefore, we will now, no way of knowing, what is the relative sizes of two different frequency components.

So, to beat this problem, what we said was, instead talking about the Fourier coefficient C_n , let us talk about Fourier coefficient density, which is C_n divided by Δf . So, if you request this new quantity, which is C_n times T_0 the period. Then, immediately the trouble that arose, because T_0 being the denominator will vanish. And then, we had $C_n T_0$ will be $\frac{1}{2\pi} \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$. And immediately this spectrum would be like this, corresponding to this periodic pulse strain.

So, you have depending upon the pulse width and amplitude, you have certain value A_d . And this no longer, as period becomes larger and larger. This no longer has to become smaller, A_d will remain the same, the line will become closer. And the spacing between two adjacent sides, $\Delta\omega$, which is the new fundamental frequency, which is quite small.

And we can put $f(t)$ as $\frac{1}{2\pi}$ times, this coefficient density times e to the power of $j n \Delta\omega t$. I am writing this as $\text{times } \Delta\omega$. So, the Fourier series now will be in this form. But, instead of C_n , we are talking about, we are talking in terms of coefficient density. Now, at this stage, we take the limit as T becomes infinitely large.

So, if T becomes infinitely large, we observed that $\Delta\omega$ become smaller and smaller. That means, the line, the spacing between the lines becomes 0 in the limit; you have a frequency component at every point on the ω axis. So, what we started as a line spectrum now becomes continuous spectrum. That means, almost every single frequency is presented.

Of course, almost everywhere, because at this point, we can say this is 0. But, some other, nearly everywhere, you have the frequency component present. And what we are plotting now is not any more the Fourier coefficient C_n , but the coefficient density. And we give the new name for the coefficient density, we call that of $j\omega$ the Fourier transform, which is given by the formula minus infinity to plus infinity of $f(t)$, e to the power of minus $j\omega t$ dt . That is $F(j\omega)$.

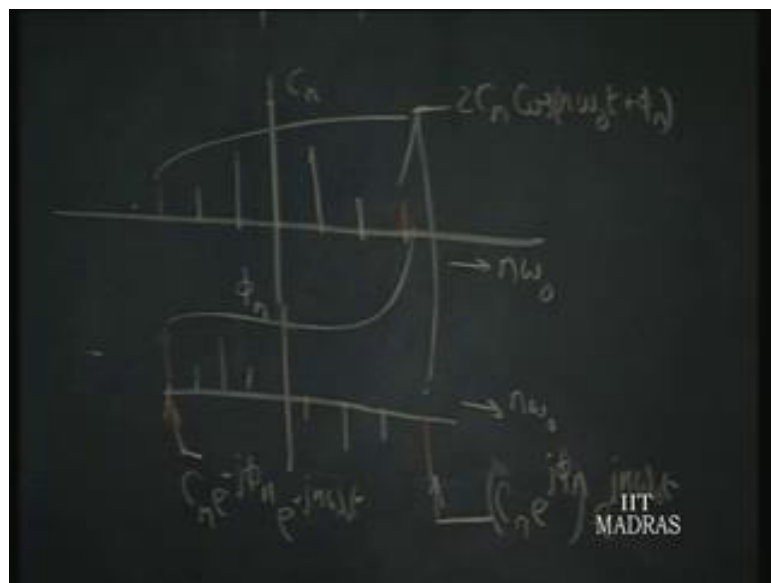
To recover the original function $f(t)$ from the $f(j\omega)$. Instead of the Fourier series, we will have now, take the limit as the summation as $\Delta\omega$ tends to 0 of $\frac{1}{2\pi}$ times the Fourier coefficient density, times e to the power of $j\omega t \Delta\omega$. And that leads us to the formula like this, which is called the Fourier integral. So, basically, then we have when we are dealing with Fourier transform for a periodic function.

We have two basic integral to deal with, from the given $f(t)$, we can generate this Fourier transform, which is really Fourier transform coefficient density as explained here. And once, we have the Fourier transform, we can recover the Fourier function of

time through this integral. These are the two basic integrals for characteristic of the Fourier integral theory.

Now, we will look into the meaning of the coefficient density and the similarities. And the contrast between the periodic function and a periodic function in a greater detail at this point of time. After introduced ourselves to the concept of Fourier integral and Fourier transform. Now, let us look at the points of comparison of the spectra of a periodic signal and a periodic signal.

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So, if you had a periodic function of time, say period T naught. Then, you have an amplitude spectrum magnitude spectrum C_n . And a phase spectrum, as you recall the magnitude is the human function of n and the phase spectrum is not function of n . So, this x axis can be graduated in terms of frequency $n \omega_0$ naught or in terms of n . Now, in particular C_n here and the corresponding phase together these two components together will give me.

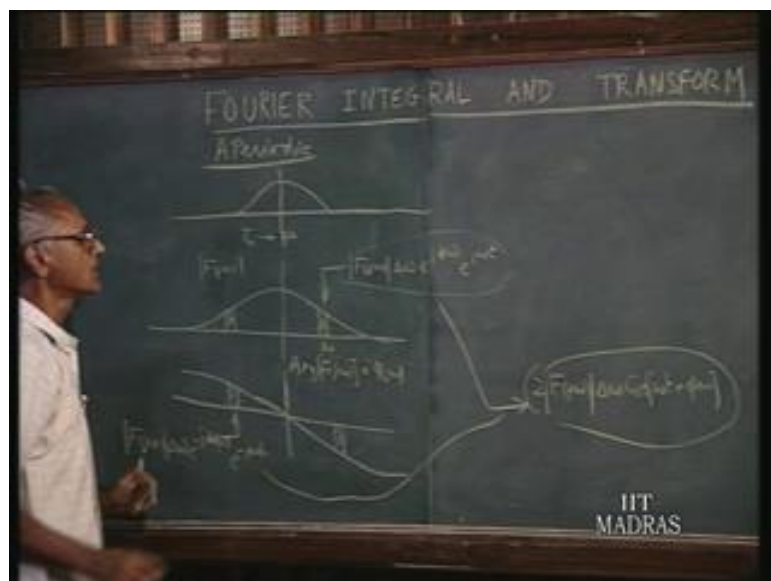
That means, this specifies the magnitude, this specifies the phase. So, it is $C_n e^{j \phi_n}$, $e^{j n \omega_0 t}$. So, the significance of these two terms, one specifies the magnitude. The other is specifying the phase is $C_n e^{j \phi_n}$ is the coefficient of this exponential term, $e^{j n \omega_0 t}$.

Likewise, for the negative value of n , $e^{-jn\omega_0 t}$ will have these two terms together will give me the conjugate of this. Because, when C_n and C_{-n} are related by conjugate relationship. So, this will become C_n same magnitude $e^{j\phi_n}$, e to the power of $j\phi_n$, e to the power of $-j\omega_0 t$. So, all these four put together. That means, the magnitude and phase together will give me $2C_n \cos(n\omega_0 t + \phi_n)$.

So, the meaning of this negative frequency $-n\omega_0$ is only, it means that, this is the exponential. This is the coefficient of this $e^{j\omega t}$ terms in this exponential representation. Even though, the frequency is negative. These two terms together combine to provide you with a real frequency of $n\omega_0$, $n\omega_0$.

Now, this is the line spectrum and there will be frequency component present only at discrete value, along the frequency axis. And they are all integral multiples of fundamental frequency and as I said the phasing between two lines is $\omega_0 t$. Whereas, when we have got the, this is the periodic case.

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In the A periodic case, we have some phenomena, which lose, only one it does not repeat itself. The period T tends to infinity and we said this will lead to a spectrum, which is continuous. And we call that coefficient density F of $j\omega$, magnitude. And we also have associate with after all F of $j\omega$ in general is a complex number. It is complex function; therefore, we have magnitude and sign. So, also an angle associated with that it is a continuous function.

This is argument of f of t ω , which we may call $\phi_j \omega$. Now, we said this coefficient density, what does it really mean. If you take any small interval $\Delta \omega$ and the corresponding here, this lies on the spectrum of the two spectra, the magnitude spectrum and the phase spectrum together. These two will imply a constituent of our f of t . What is that function, F of $j \omega$ magnitude, this is the coefficient density.

That means, over this interval of frequency. The density at the centre is F of $j \omega$. Therefore, the total magnitude is F of $j \omega$ magnitude times $\Delta \omega$. It has the phase e to the power of $j t \omega$. And then, the exponential representation of function of time is e to the power of $j \omega t$. So, this magnitude and this phase together will give rise function of time.

Naturally, on the other side, corresponding to, this is ω . This is minus ω and again, if you take two sections of the spectrum of width $\Delta \omega$. These two together will give me, F of $j \omega$ magnitude, times $\Delta \omega$ times e to the power of minus $j \phi_j \omega$, e to the power of minus $j \omega t$. So, this portion of the spectrum combined with portion of the spectrum. These two together, if you combine these two, you will get 2 times F of $j \omega \Delta \omega \cos \omega t$.

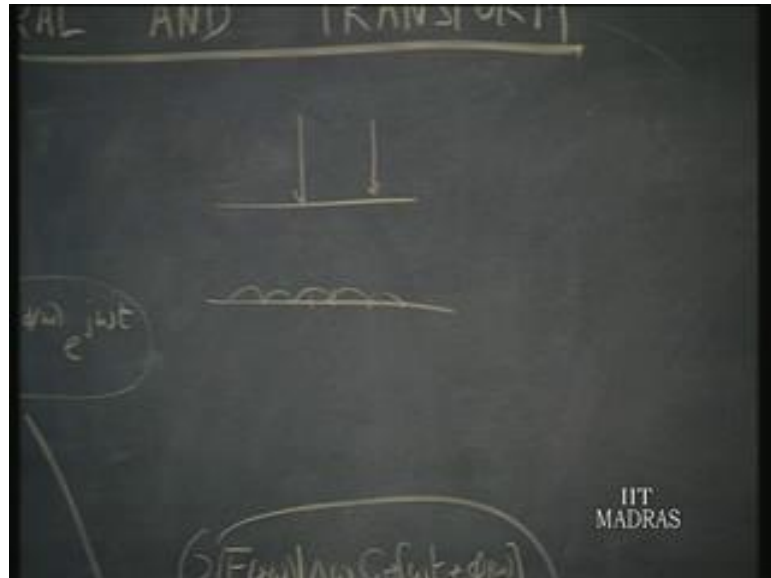
So, even though you have components, corresponding to negative frequency, what it really means. For every negative frequency here, you have positive frequency and all together will lead a real function of a cosine term. This is magnitude of course. And you have frequency at every point. In other words, what we are having here is not and at any given, if you have a periodic case at any given point of time, at any given point of time at point of frequency, if you want to know, what is the amplitude, the amplitude is 0 , vanishingly small.

But, if you take a small along the ω axis and if you consider this entire slice to be representative of one single frequency term at the centre. Then, the amplitude of that particular exponential term will be given by this. So, this is called coefficient density. So, in other words, f of t is in volts. If f of t is in volts the dimension of F of $j \omega$ is in volts per cycle per second or dimensions will be volt seconds.

Whereas, in the periodic case, if f of t is in volts, C_n will be in volts. It gives the amplitude of each individual component, whereas, here it gives not the amplitude, but

the magnitude density. It is something similar to, what we you know in your applied mechanics courses and so on.

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You have beam here, we have concentrated loads at point. This is one way of looking at it. And then, sometimes you have distributed loads. In case of distributed loads what you say is, so much kilogram per unit length. That means, at given point, there is no load. But, if you take a small interval and the total load in the interval will be the density times the small elemental distances.

So, it similarly here, you have situation where, each individual frequency component is almost negligible amplitude. But, you have over a small interval of along the omega axis. You can think of that the sections of the spectrum to represent a function of time, which has the amplitude, which is $F_j \omega$ times $\Delta \omega$. This is the density, this is the total amount.

So, what we have done is, to demonstrate by taking the limiting case of the Fourier series as the frequency becomes smaller and smaller and as the period becomes larger and larger. That even in A periodic case can be a thought of, as composed of number of different frequency components. In contrast to the Fourier series, which concerns itself, with the periodic case.

Here, we have a frequency component at every conceivable frequency all along the omega axis from minus infinity to plus infinity in the exponential representation. But, their amplitudes are vanishingly small. So, the amplitude at any given frequency is 0, but over a small width, you can think of an amplitude. So, what we are really talking about is, the amplitude density of the magnitude density, otherwise called coefficient density.

And that coefficient density is, we term it is F of j omega, because function of omega and you like associate j with the term omega. Therefore, it is F of j omega. So, we have basically these two relations, which are important, which will use to our further work.

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The image shows a chalkboard with the following content:

- At the top, the words "Fourier Integral" are written in chalk.
- The main equation is $f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(j\omega) e^{j\omega t} d\omega$.
- Below the main equation, there is a smaller equation: $\int_{-\infty}^{\infty} F(j\omega) e^{j\omega t} d\omega = 2\pi f(t)$.
- In the bottom right corner, the logo "IIT MADRAS" is visible.

The coefficient density F of j omega is obtained from your given f of t as the integral from minus infinity to plus infinity of f of t, e to the power of minus j omega t d t. This is the coefficient density, which we call the Fourier transform. And once, we have the Fourier transform. We can recover the time function f of t by writing the integral relation from minus infinity to plus infinity of F of j omega, e to the power of j omega t d.

Both integrals from minus infinity to plus infinity, in one case, we get 1 over 2 pi here. The other case, we do not. And as I said, we get this 2 pi, because we are defining the coefficient density as the amplitude per cycle per second, f not instead of omega

naught. And this is called the Fourier integral. That means, you always look at this as f of t is being composed of an exponential term the function of time like this.

And watch its amplitude, F of $j\omega$ this is the coefficient density $\Delta\omega$ over 2π . So, this is the amplitude, F of $j\omega$ is the coefficient density. So, at this frequency, if you want to look this, you take a small interval along the ω axis. So, $\Delta\omega$ over 2π this is the base, this is the amplitude density. Therefore, this is the amplitude. And take the limits as $\Delta\omega$ goes to 0 overall possible frequencies and summation overall possible frequencies and that leads to this integral.

So, what we have done in this lecture is to say that, the any periodic function can be thought of as an infinite summation of elementary frequency components. Each with vanishingly small amplitude and with all frequencies present along the ω axis. We continue next class, next lecture discussion of the properties of the Fourier transforms.