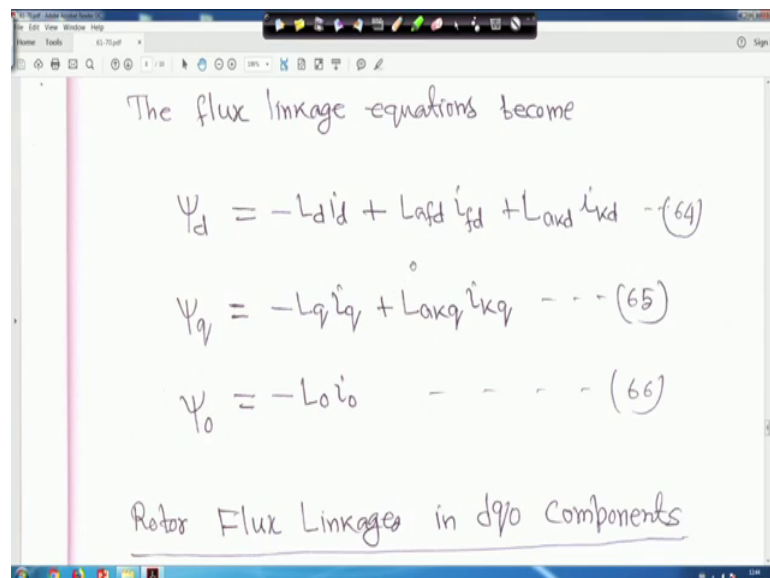


Power System Dynamics, Control and Monitoring
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Indian Institute of Technology, Kharagpur

Lecture - 07
Power System Stability (Contd.)

So, we are back again. So, whenever starting this your next lecture, just starting from the beginning of the previous one, right. So, some just said that things will be easier for all of us.

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The flux linkage equations become

$$\Psi_d = -L_d i_d + L_{afd} i_{fd} + L_{akd} i_{kd} \quad (64)$$
$$\Psi_q = -L_q i_q + L_{akq} i_{kq} \quad (65)$$
$$\Psi_0 = -L_0 i_0 \quad (66)$$

Rotor Flux Linkages in d/q/0 Components

So, the flux linkage equation then psi d, psi q and psi 0 this already we have seen in the previous lecture, but just we have to move further. So, psi d will be minus L d i d plus L afd i fd plus L akd i kd after your transformation. Then psi q also will be minus L q i q plus L akq i kq and psi 0 will be minus L 0 i 0, right.

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Substitution of the expressions for i_d, i_q in eqn. (53) to (55) gives

$$\Psi_{fd} = L_{ffd} i_{fd} + L_{fkd} i_{kd} - \frac{3}{2} L_{afd} i_d \quad (67)$$
$$\Psi_{kd} = L_{fkd} i_{fd} + L_{kkd} i_{kd} - \frac{3}{2} L_{akd} i_d \quad (68)$$
$$\Psi_{kq} = L_{kkq} i_{kq} - \frac{3}{2} L_{akq} i_q$$

So, now next we will see the rotor flux linkages in d q 0 component, right. So, substitution of the expression for i_d, i_q in equations 53 to 55 if you make everything then we will find the ψ_{fd} will become $L_{ffd} i_{fd}$ plus $L_{fkd} i_{kd}$ minus $\frac{3}{2} L_{afd} i_d$.

What I suggest that whenever you will listen to the lecture you please keep the list of nomenclature video because, so many terms are there, right. Like you are mutual inductant between stator and rotor, then self inductance of stator self inductance of the rotor circuit, right. So, similarly your ψ_{kd} will be $L_{fkd} i_{fd}$ plus $L_{kkd} i_{kd}$ minus $\frac{3}{2} L_{akd} i_d$ and similarly ψ_{kq} will be $L_{kkq} i_{kq}$ minus $\frac{3}{2} L_{akq} i_q$.

Now, question is that if we put everything in the 3 into 3 matrix form and simplify you will get like this, but as to save the time directly I have written all these things, right otherwise it will take long time. But, as an exercise you please try yourself you will find hardly it will take few minutes to derive that.

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The screenshot shows a presentation slide with handwritten mathematical equations and text. The equations are:

$$\Psi_{kd} = L_{fkd} i_{fd} + L_{kkd} i_{kd} - \frac{3}{2} L_{akd} i_d \quad \text{--- (68)}$$
$$\Psi_{kq} = L_{kq} i_{kq} - \frac{3}{2} L_{akq} i_q \quad \text{--- (69)}$$

Below the equations, the text reads: "Again, all the inductances are seen to be constant, i.e., they are independent of the rotor position." A small video inset of a man is visible in the bottom right corner of the slide.

So, again all the inductance is are seen to be constant that is they are independent of the rotor position. Here this ψ_{fd} , ψ_{kd} and your ψ_{kq} they are independent of θ , right.

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The screenshot shows a presentation slide with handwritten text. The text reads: "It should, however, be noted that the saturation effects are not considered here. The variations in inductances due to saturation are of a different nature and this will be treated separately." Below this, it says: "It is interesting to note that i'_d does not appear in the rotor flux equations. This is because zero seq" (partially cut off). A small video inset of a man is visible in the bottom right corner of the slide.

So, it should however, we noted that the saturation effects are not considered. We will not consider the saturation effects throughout right.

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It is interesting to note that i_0 does not appear in the rotor flux linkage equations. This is because zero sequence components of armature current do not produce net mmf across the air-gap.

While the dq0 transformation has resulted in constant inductances in eqns. (64) to (69), the mutual inductances between stator and rotor quantities are not reciprocal.

So, now it is interesting to note that i_0 does not appear in the rotor flux linkage equation. This is because 0 sequence component of armature current do not produce the mmf across the air gap this should you this you should keep it in your mind.

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While the dq0 transformation has resulted in constant inductances in eqns. (64) to (69), the mutual inductances between stator and rotor quantities are not reciprocal.

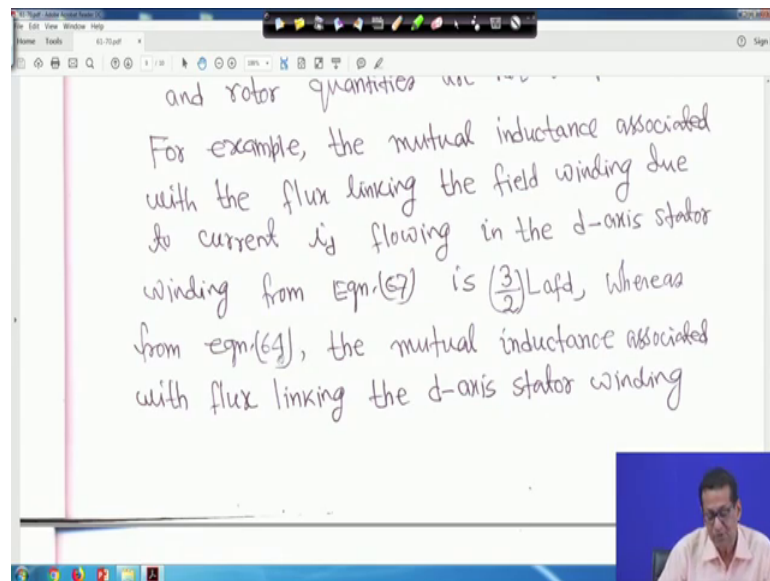
For example, the mutual inductance associated with the flux linking the field winding due to current i_d flowing in the d-axis stator winding from eqn. (67) is $\frac{3}{2}L_{afd}$, whereas from eqn. (64), the mutual inductance with flux linking the d-axis stator winding is $\frac{2}{3}L_{afd}$.

Now, while the dq0 transformation has resulted in constant in that sense equation 64 to 69, right they are independent of theta. The mutual inductances between stator and rotor quantities are not reciprocal. Reciprocal means for example, when you have studied say transformer when you make per unit values then either it is refer to primary or refer to secondary.

secondary it is same per unit values, but in this case it is not, right. So, later you will see you will see some mechanism such that both can be you are made it same I mean either to the stator side or to the rotor side, right when you make, the per unit system analysis.

For example, the mutual inductance associated with the flux linking the field winding due to the current i_d flowing in the d-axis stator winding from equation 67 is $3/2 L_{afd}$, right.

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Whereas, if you look into equation 64 you will find the mutual inductance associated flux linking the d-axis stator winding due to field current is L_{afd} , it is $3/2 L_{afd}$ and it is L_{afd} . So, it is not reciprocal, reciprocal means they are not same, right.

(Refer Slide Time: 03:56)

(Z0)

due to field current is L_{afd} . This problem is overcome by appropriate choice of the per unit system for the rotor quantities.

Stator Voltage Equations in dq0 Components

stator voltage

A small video inset of a man in a pink shirt is visible in the bottom right corner of the slide.

So, this problem is overcome by appropriate choice of the per unit system for the rotor quantities that we will see later.

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per unit system for the rotor quantities.

Stator Voltage Equations in dq0 Components

~~stator voltage~~

Eqn. (26) and (28) are basic equations for phase voltages in terms of phase flux linkages and currents.

By applying the dq0 transformation of Eqn

A small video inset of a man in a pink shirt is visible in the bottom right corner of the slide.

Now, next is this already we have seen in the previous lecture, but started with this only. So, stator voltage equation in dq 0 component now equation 26 to 28 are basic equation for phase voltages, right that we have seen earlier in terms of phase flux linkages and current.

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voltages in terms of phase flux linkages and currents.

By applying the dq0 transformation of Eqn(59), the following expressions in terms of transformed components of voltages, flux linkages and currents result:

$$e_d = p\psi_d - \psi_q p\theta - R_a i_d \quad \dots (70)$$

$$e_q = p\psi_q + \psi_d p\theta - R_a i_q \quad \dots (71)$$

By applying the dq 0 transformation of equation 59 the following expressions in terms of the transformed components of voltage flux linkage and currents results. I mean all the derivations not putting it here because it is say it will I told you again and again that it will take long time. I suggest you try to make yourself and certain things you should keep it in your mind. For example, this should you keep it in your mind. In this case what will happen e d will become p psi d minus psi q p theta minus R a i d, right.

So, this is p ddt so, it is d of dt psi d, right. So, this is later it is explain. This is actually one term called psi or transformer induced voltage this is p theta that is d theta by d t that is speed voltages, right. And similarly e q is p psi q plus psi d p theta minus R a i q and e 0 will be p psi 0 minus R a into i 0 this is equation 70, 71 and 72. This is actually keep it in your mind but try to derive of your own. All derivations are not given here because it will consume then many hours, right.

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$$e_d = p\psi_d - \psi_q p\theta - R_a i_d \dots (E0)$$
$$e_q = p\psi_q + \psi_d p\theta - R_a i_q \dots (E1)$$
$$e_o = p\psi_o - R_a i_o \dots (E2)$$

The angle θ , as defined in Fig.9, is the angle between the axis of phase a and the d-axis. The term $p\theta$ in the above equations represents

So, the angle theta as defined in figure 9 earlier I told you, right. And you see the figure 9 also everything is drawn there, that angle between the axis of phase a and d-axis the term $p\theta$ in the above equation actually it will represent I will come to that; I will come to that, right.

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(71)

the angular velocity ω_p of the rotor. For a 60 Hz system under steady-state conditions

$$p\theta = \omega_r = \omega_s = 2\pi \times 60 = 377 \text{ electrical rad/sec.}$$

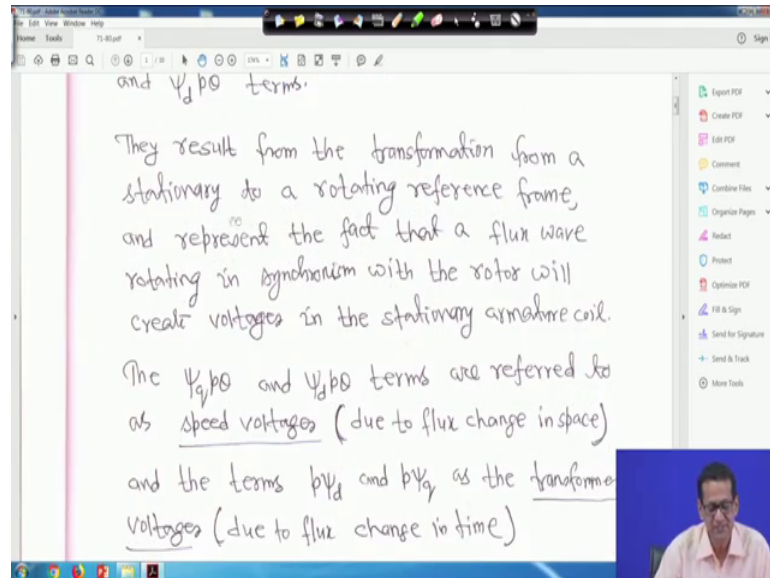
The above equations have a form similar to those of a static coil, except for the $\psi_q p\theta$ and $\psi_d p\theta$ terms.

They result from the transformation from a stationary to a rotating reference frame

That your, the angular velocity ω of the rotor that is $b\theta$ upon dt is equal to ω_r and ω_r will be basically ω_s because it is synchronous machine, right. So, for a 60 hertz system under steady state condition $p\theta$ will be ω_r is equal to

omega s because it is your both are same so, it will be 377 electrical radian per second. Now, the above equation have a form similar to those of a static coil except for the psi q p theta and psi d p theta terms, right.

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The result from the transformation from a stationary to a rotating reference frame and your, what you call the represent the fact that a flux wave rotating in synchronism with the rotor will create voltages in the stationary armature coil, right.

Similarly, this for example, the psi q p theta and psi d p theta psi q p theta mean psi q into d theta dt that similarly psi d p theta means psi d into d theta dt, right. A term referred to as a speed voltages because it is d theta dt in general, right due to the flux changes in phase voltages. And that is your due to the flux and your p psi d and p psi q equal transformer voltages due to the flux change in time because we know d psi by dt. So, it is p psi d means your d psi d dt and this is d psi t and dd psi q dt, right.

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and the terms $p\psi_d$ and $p\psi_q$ as the transformer voltages (due to flux change in time)

The speed voltage terms are the dominant components of the stator voltage.

Under steady-state conditions, the transformer voltage terms $p\psi_d$ and $p\psi_q$ are in fact equal to zero;

So, the speed voltage terms are the dominant components of the stator voltage that $d\theta$ by $d\theta$ under steady state condition the transformer voltage terms $p\psi_d$ and $p\psi_q$ are in fact, equal to 0, right.

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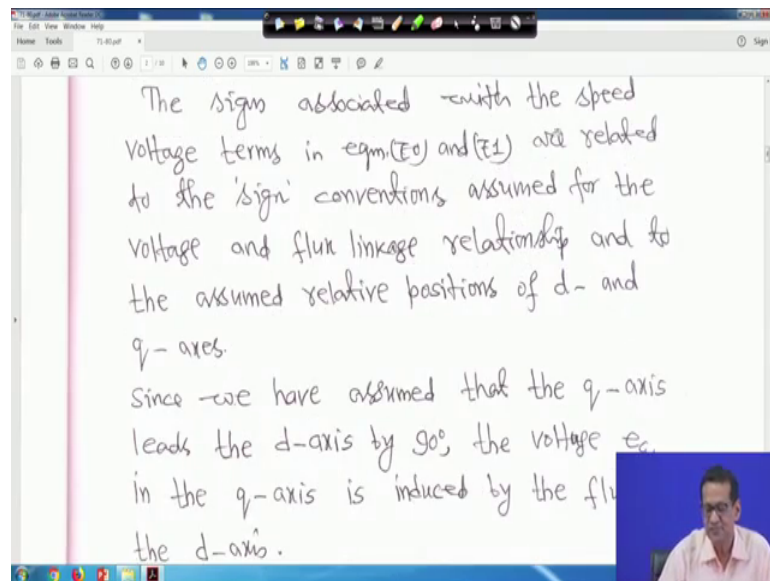
there are many transient conditions where the transformer voltage terms can be dropped from the stator voltage equations without causing errors of any significance. However, in other situations they could be important (sections 3-7 & 5-1),

The sign associated with the speed voltage terms in eqn.(E0) and (E1) are related to the 'sign' conventions assumed for

So, there are many transient conditions where the transformer voltage terms can be dropped from the stator voltage equation without causing error of any significance, right. However, in other situations they could be important, right. This is actually just hold on so, just hold on; just hold on. So, in this case what happen that you are however, in other

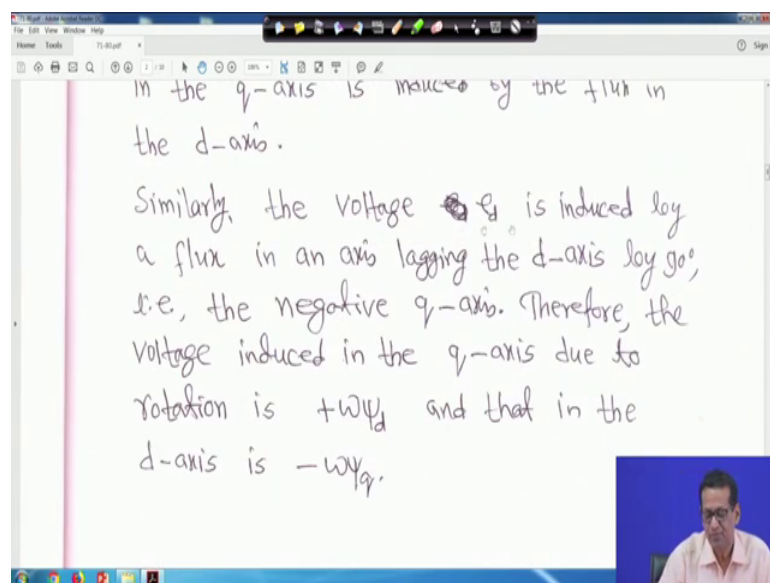
situations that is that is this is something written section 3.7 5.1 it is not for you, right.

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Now, signs associated with the speed voltage terms in equation 70 and 71 are related to the sign conventions assumed for the voltage and flux linkage relationships and to the assumed relative position of d and q axis, right. We have seen the q-axis actually leading the d-axis by 90 degree, the voltage e_q in the q-axis is induced by the flux in the d-axis, right.

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Similarly, the voltage e_d is induced by a flux in an axis lagging the d-axis by 90 degree

that is the negative q-axis. Therefore, the voltage induced in the q-axis due to rotation will be plus omega psi d and that in that d-axis will be minus omega psi q this should you keep it in your mind.

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73

Electrical Power and Torque.

The instantaneous three-phase power output of the stator is.

$$P_t = e_a i_a + e_b i_b + e_c i_c$$

Eliminating phase voltages and currents terms of dq0 components, we have

Now, electrical power and torque; now, the instantaneous 3 phase power output of the stator is that generally we know P_t is equal to $e_a i_a$ plus $e_b i_b$ plus $e_c i_c$. Now, if you if you make this all these things your transforms are d and e q at your dq 0 transformation eliminating phase voltage and currents in terms of dq 0 components I mean you substitute, all right.

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Eliminating phase voltages and currents in terms of dq0 components, we have

$$P_t = \frac{3}{2} (e_d i_d + e_q i_q + 2e_0 i_0) \quad \text{--- (73)}$$

Under balanced operation, $e_0 = i_0 = 0$ and the expression for power is given by

$$P_t = \frac{3}{2} (e_d i_d + e_q i_q)$$

Wherever, $e_a i_a$, $e_b i_b$, $e_c i_c$ all the in terms of your dq component and if you simplify then you will get P by 2 will be 3 by P_t will be 3 by 2 into $e_d i_d$ plus $e_q i_q$ plus $2 e_0 i_0$, 73. This also you should keep it in your mind it will be 3 by 2 into $e_d i_d$ plus $e_q i_q$ plus $2 e_0 i_0$. After making all these all the you substitute all and you simplify then only you will get this one, right. And under balance operation $e_0 i_0$ is equal to 0 therefore, the expression per power is given by 3 by 2 $e_d i_d$ plus $e_q i_q$, right.

Now, using equation 70 to 72, right to express the voltage component in terms of flux linkage and currents by recognising ωr as the rotor speed is $d\theta$ by $d t$ and rearranging we have.

(Refer Slide Time: 10:31)

$$v_t = \frac{1}{2} (\psi_d i_d + \psi_q i_q) \quad \dots \quad (73)$$

Using eqn (70) ~~and~~ (72) to express the voltage components in terms of flux linkages and currents, by recognizing ω_r as the rotor speed $\frac{d\theta}{dt}$, and rearranging, we have

$$P_t = \frac{3}{2} \left[(i_d p \psi_d + i_q p \psi_q + 2 i_0 p \psi_0) + (\psi_d i_q - \psi_q i_d) \omega_r - (i_d^2 + i_q^2 + 2 i_0^2) R_a \right] \quad \dots$$

Now, here what you do that e d e q expressions are known to you put everything in this expression because that is equation your 70, 71 and 72 put every everything and you substitute here.

(Refer Slide Time: 10:34)

by recognizing ω_r as the rotor speed $\frac{d\theta}{dt}$, and rearranging, we have

$$P_t = \frac{3}{2} \left[(i_d p \psi_d + i_q p \psi_q + 2 i_0 p \psi_0) + (\psi_d i_q - \psi_q i_d) \omega_r - (i_d^2 + i_q^2 + 2 i_0^2) R_a \right] \quad \dots \quad (74)$$

If you do so, you will get this term 3 by 2 into your $i_d p \psi_d$ plus $i_q p \psi_q$ plus $2 i_0 p \psi_0$ plus $\psi_d i_q$ minus $\psi_q i_d$ into ω_r minus in bracket i_d^2 plus i_q^2 plus $2 i_0^2$ bracket close into R_a this is equation 74. This expression you will get, but each term has separate meaning. For example, the first term that is your $i_d p$

$\psi \frac{di}{dt} + i \frac{d\psi}{dt} + 2i_0 \frac{d\psi_0}{dt}$ this is actually rate of change of armature magnetic energy, right. This is the 1st term.

Now, 2nd term is more important second term is $\psi \frac{di}{dt} - \psi \frac{di}{dt}$ into ω_r this is actually second term is the power transferred across the air gap. This transfer this term is the power transfer your transform across the air gap. And third one that is $i^2 r_a$ this is nothing but the armature resistance loss.

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(71)

$= \left\{ \begin{array}{l} \text{Rate of change of armature magnetic} \\ \text{energy} \end{array} \right\}$
 $+ \left\{ \text{power transferred across the air-gap} \right\}$
 $- \left\{ \text{armature resistance loss} \right\}$

The air-gap torque T_e is obtained

So, this is the first term has some significant, second term also, third term also that is why it is written first term is rate of change of armature magnetic energy plus second term is power transferred across the air gap of the machine and third term is the armature resistance loss, right.

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The air-gap torque T_e is obtained by dividing the power transferred across the air-gap [i.e., power corresponding to the speed voltages] by the rotor speed in mechanical radians per second,

$$T_e = \frac{3}{2} (\psi_d i_q - \psi_q i_d) \times \frac{\omega_r}{\omega_{mech}}$$

Now, therefore, the air gap torque we know p in general power is equal to torque into angular speed that we know, right. Therefore, the air-gap torque T_e is obtained by dividing the power transferred across the air-gap that is power corresponding to the speed voltages that means, this term this is the second term is the power; second term is the power.

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mechanical

$$T_e = \frac{3}{2} (\psi_d i_q - \psi_q i_d) \times \frac{\omega_r}{\omega_{mech}}$$

$$\therefore T_e = \frac{3}{2} (\psi_d i_q - \psi_q i_d) \cdot \frac{p_f}{2} \quad \dots (7.5)$$

The flux-linkage equations (64) to (69) associated with the stator and rotor circuits, together with the voltage equations (70) to (72) for stator the voltage equations (50) to (52)

Then if you want to make it torque, so it will be T is equal to 3 by 2, $\psi_d i_q$ minus $\psi_q i_d$ into ω_r upon your ω_{mech} because this is your power and power

generally we know torque into speed. So, T into ω mechanical is equal to $3/2$ $\psi_d i_q$ minus $\psi_q i_d$ ω that means, this term; this term this second term, this second term, right this second term, so that means, this is my torque equation.

So, or you can simplify ω_r upon ω_{mech} earlier we have seen this is nothing but the number of field poles by 2. So, it will be $3/2$ $\psi_d i_q$ minus $\psi_q i_d$ into p_f by 2, right this is equation 75.

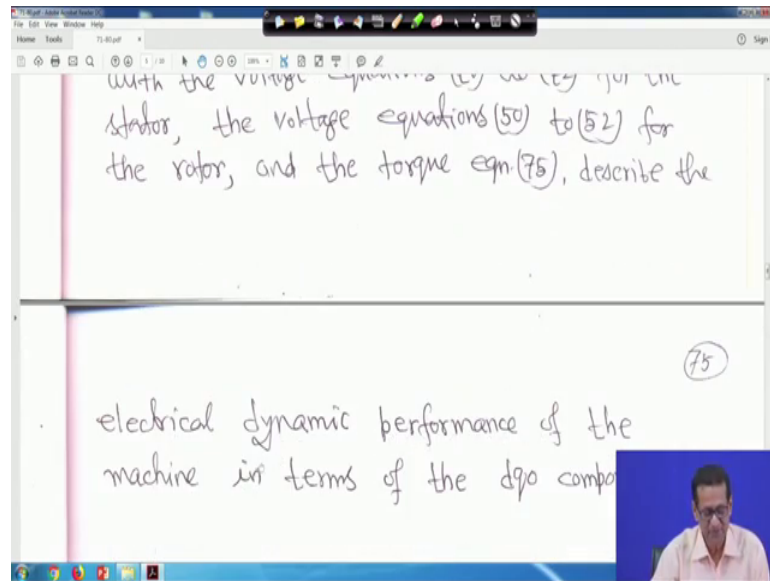
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$$T_e = \frac{3}{2} (\psi_d i_q - \psi_q i_d) \cdot \frac{p_f}{2} \quad \dots (75)$$

The flux-linkage equations (64) to (69) associated with the stator and rotor circuits, together with the voltage equations (70) to (72) for the stator, the voltage equations (50) to (52) for the rotor, and the torque eqn (75), describe the

Now, the flux linkage equation that is 64 to 69 associated with the stator and rotor circuit. Together with the voltage equation that is 70 to 72 for the stator and the voltage equation 50 to 52 for the rotor and the torque that is equation 75 actually describes the electrical dynamic performance of the machine in terms of dq0 component.

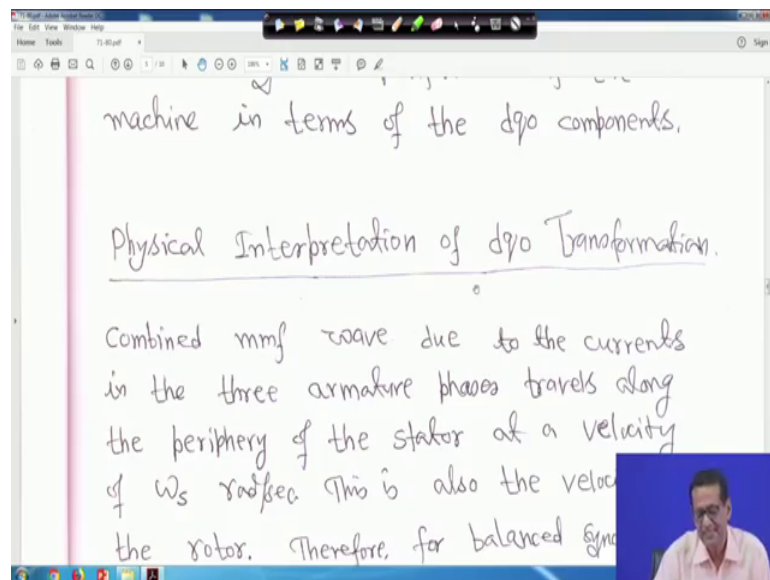
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With the voltage equations (50) to (52) for the stator, the voltage equations (50) to (52) for the rotor, and the torque eqn. (75), describe the electrical dynamic performance of the machine in terms of the dq0 compo

So, all these equation basically 64 to your 75 all these equations actually represent the dynamics of the equation related to the dynamics of the machine much more detail we will see later, all right.

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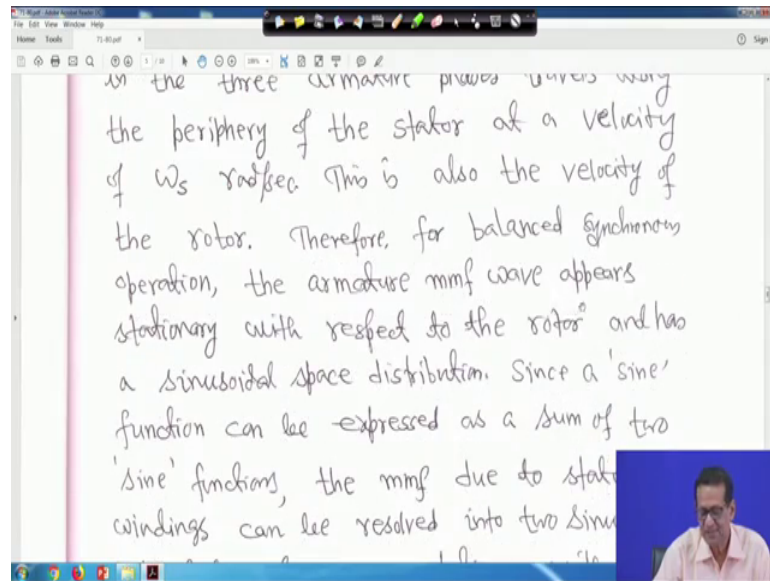
machine in terms of the dq0 components.

Physical Interpretation of dq0 Transformation.

Combined mmf wave due to the currents in the three armature phases travels along the periphery of the stator at a velocity of ω_s rad/sec. This is also the velocity of the rotor. Therefore, for balanced sym

Now, physically interpretation of dq 0 transformation, why you do so; that is the physical interpretation. Now, combine mmf wave due to the currents in the 3 armature phases travels along the periphery of the stator at a velocity of ω_s radian per second, right.

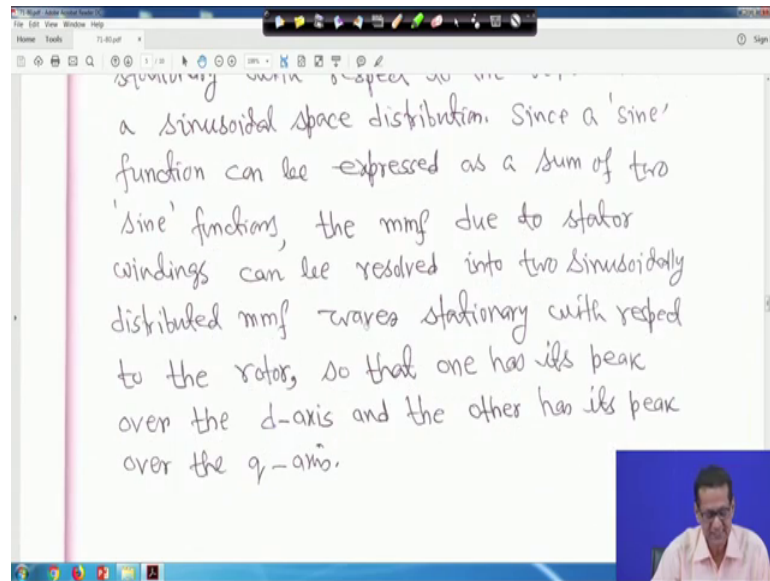
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This is also the velocity of the rotor because the rotor in synchronous phase. Therefore, for balanced synchronous operation the armature mmf wave appears stationary with respect to the rotor and has a sinusoidal phase distribution this also we have discussed before, right.

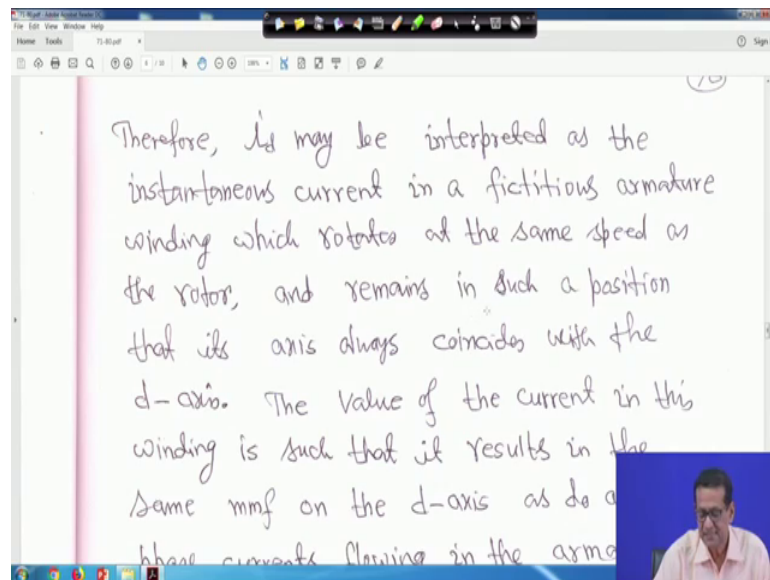
Since, a sine function can be expressed as a sum of 2 sine functions, right the mmf this also we have seen the mmf wave mmf d-axis, mmf q-axis, right the mmf due to stator windings can be resolved into 2 sinusoidally distributed mmf wave stationary with respect to the rotor.

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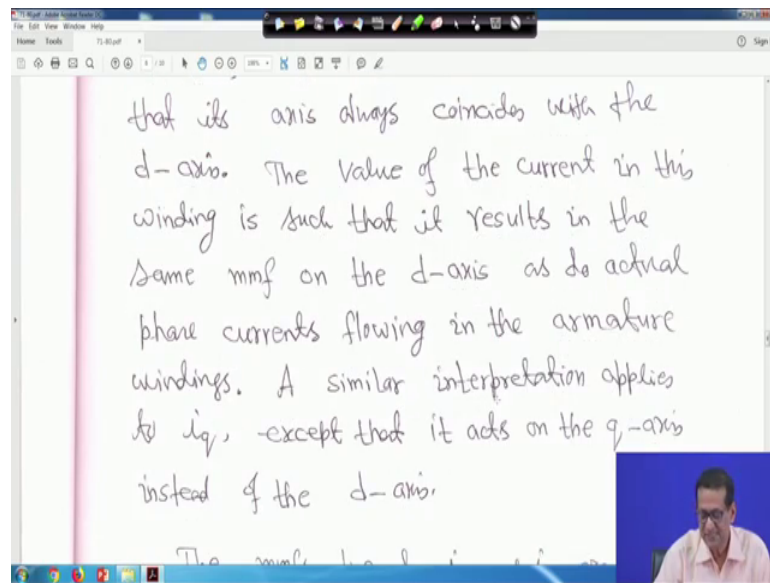
So, the one has its peak over the d-axis and the other has its peak over the q-axis this diagram we have seen earlier also, right.

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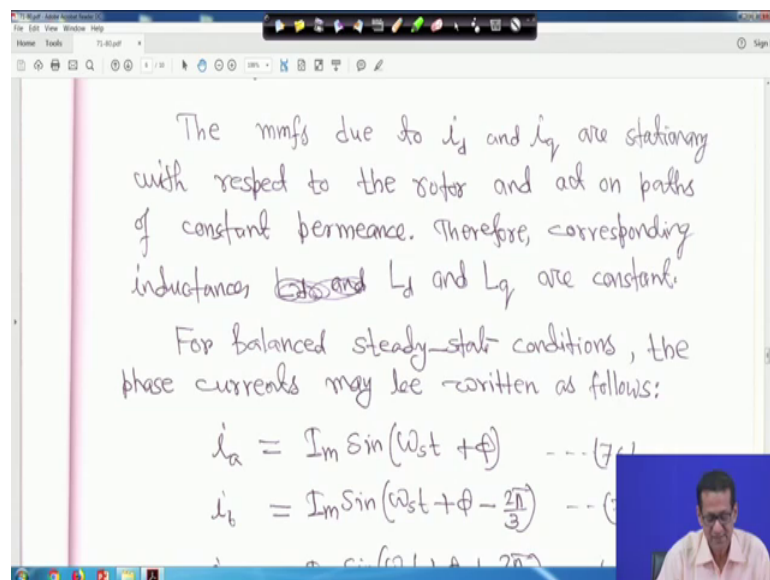
Therefore, it may be interpreted as the instantaneous current in a fictitious armature winding which rotates at the same speed as the rotor and remains in such a position that its axis always coincides with the d-axis. The value of the current in this winding is such that it results in the same mmf on the d-axis as do actual phase current flowing in the armature windings.

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A similar interpretation applies also to i_q except that it acts on the q-axis instead of the d-axis. This certain thing this is the physical interpretation.

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The mmf due to i_d and i_q are stationary with respect to the rotor, right and act on paths of constant permeance. Therefore, your corresponding in your it inductances L_d and L_q then d-axis and q-axis are constant. For balance steady state conditions the phase currents may be written as follows, right.

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For balanced steady-state conditions, the phase currents may be written as follows:

$$i_a = I_m \sin(\omega_s t + \phi) \quad \dots (76)$$
$$i_b = I_m \sin(\omega_s t + \phi - \frac{2\pi}{3}) \quad \dots (77)$$
$$i_c = I_m \sin(\omega_s t + \phi + \frac{2\pi}{3}) \quad \dots (78)$$

We can write i_a is equal to $I_m \sin \omega_s t + \phi$, i_b is equal to $I_m \sin \omega_s t + \phi - 2\pi/3$ and i_c is equal to $I_m \sin \omega_s t + \phi + 2\pi/3$. This is 76, this is 77 and this is 78, right.

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Where $\omega_s = 2\pi f$ is the angular frequency of stator currents.

Using the dq0 transformation,

$$i_d = I_m \sin(\omega_s t + \phi - \theta) \quad \dots (79)$$
$$i_q = -I_m \cos(\omega_s t + \phi - \theta) \quad \dots (80)$$
$$i_0 = 0 \quad \dots (81)$$

Now, ω_s is, ω_s is equal to $2\pi f$ that is that angular frequency of the stator current, right using the dq0 transformation. So, this one if you go for a dq0 transformation that this equation i_d can be written as $I_m \sin \omega_s t + \phi - \theta$ because here also your $\omega_s t + \phi$, right.

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$$i_d = I_m \sin(\omega_s t + \phi - \theta) \quad \dots (79)$$

$$i_q = -I_m \cos(\omega_s t + \phi - \theta) \quad \dots (80)$$

$$i_0 = 0 \quad \dots (81)$$

For synchronous operation, the rotor speed ω_r is equal to the angular frequency ω_s of the stator currents. Hence,

$$\theta = \omega_r t = \omega_s t.$$

Now, even you go for d q transformation i_d will be $I_m \sin \omega_s t + \phi - \theta$ similarly i_q will be $-I_m \cos \omega_s t + \phi - \theta$ this is equation 80 and i_0 will be 0 that is equation 81. Now, we know that because it your what you call machine rotates in say synchronous speed. So, basically θ is equal to $\omega_r t$ is equal to $\omega_s t$.

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$$i_q = -I_m \cos(\omega_s t + \phi - \theta) \quad \dots (80)$$

$$i_0 = 0 \quad \dots (81)$$

For synchronous operation, the rotor speed ω_r is equal to the angular frequency ω_s of the stator currents. Hence,

$$\theta = \omega_r t = \omega_s t.$$

Therefore,

$$i_d = I_m \sin \phi = \text{constant}$$

That means, for synchronous operation the rotor speed ω_r is equal to the angular frequency ω_s of the stator current, right same thing. Therefore, θ is equal to

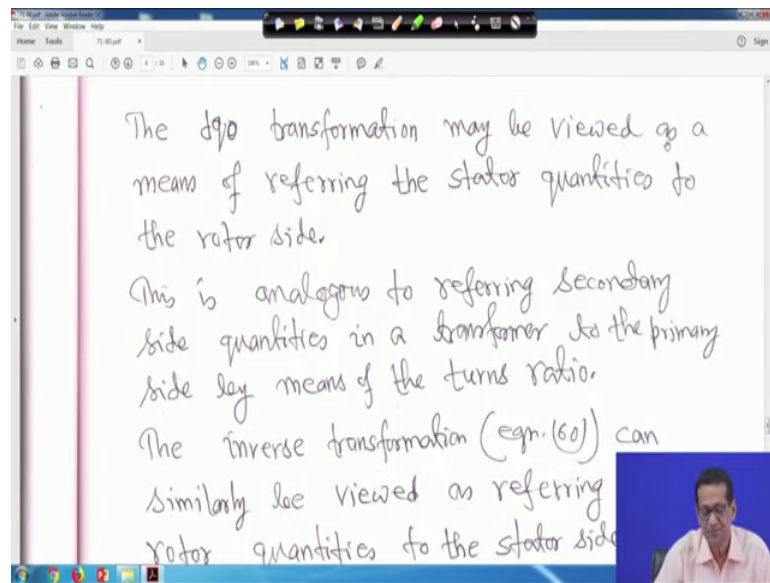
$\omega_r t$ is equal to $\omega_s t$ that means, here if you put θ is equal to $\omega_s t$ then $\omega_s t$ will be cancel only $\sin \phi$ will be there and $\cos \phi$ will be there.

(Refer Slide Time: 16:56)

The image shows a whiteboard with handwritten mathematical equations and text. At the top, the equation $i_d = I_m \sin \phi = \text{constant}$ is written. Below it, the equation $i_q = -I_m \cos \phi = \text{constant}$ is written. The text below these equations states: "For balanced steady-state operation, i_d and i_q are constant." followed by "In other words, alternating phase currents in the abc reference frame appear as direct currents in the dq0 reference frame." A small video inset in the bottom right corner shows a man speaking.

So, i_d is equal to $I_m \sin \phi$. So, it is a constant and similarly i_q is equal to minus $I_m \cos \phi$ that is also constant as if their DC current; as if their DC current, right. For balanced steady operation i_d and i_q are constant, right. So, in other words alternating phase current in the abc reference frame appear as direct currents in the dq0 reference frame, right. This is your, this is that physical interpretation of your dq0 transformation, right.

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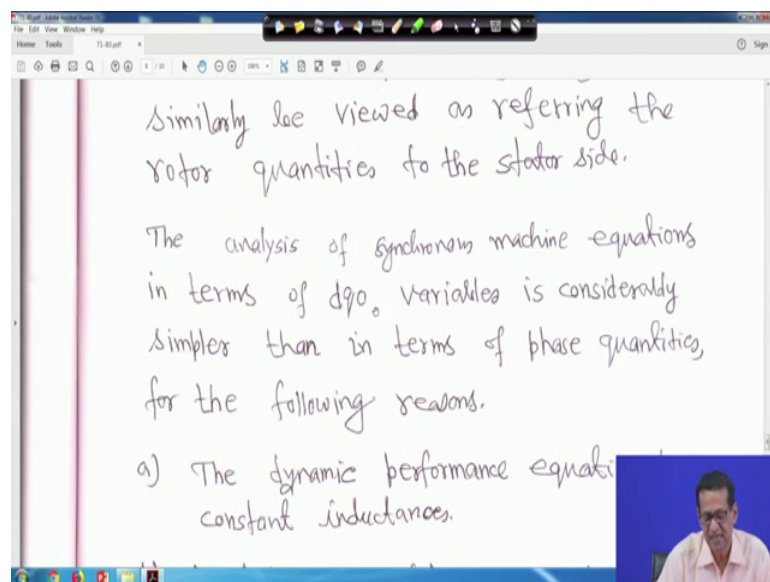
The dq0 transformation may be viewed as a means of referring the stator quantities to the rotor side.

This is analogous to referring secondary side quantities in a transformer to the primary side by means of the turns ratio.

The inverse transformation (eqn. (60)) can similarly be viewed as referring rotor quantities to the stator side.

Now, the dq0 transformation may be viewed as a means of referring the stator quantities to the rotor side. This is advantageous to referring secondary side quantities in a transformer to the primary side by means of the trans ratio.

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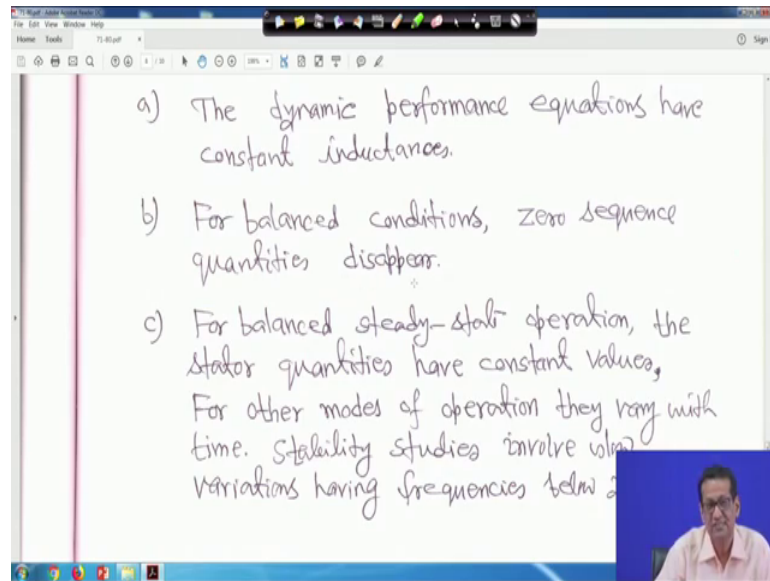
Similarly be viewed as referring the rotor quantities to the stator side.

The analysis of synchronous machine equations in terms of dq0 variables is considerably simpler than in terms of phase quantities, for the following reasons.

- a) The dynamic performance equations with constant inductances.

So, the inverse transformation that is equation 60 can similarly be viewed as referring to the rotor quantities to the stator side, right. The analysis of synchronous machine equations in terms of dq0 variables is considerably simpler than the terms of phase quantities in the following reason. Some advantages are there for dq0 transformation.

(Refer Slide Time: 18:04)



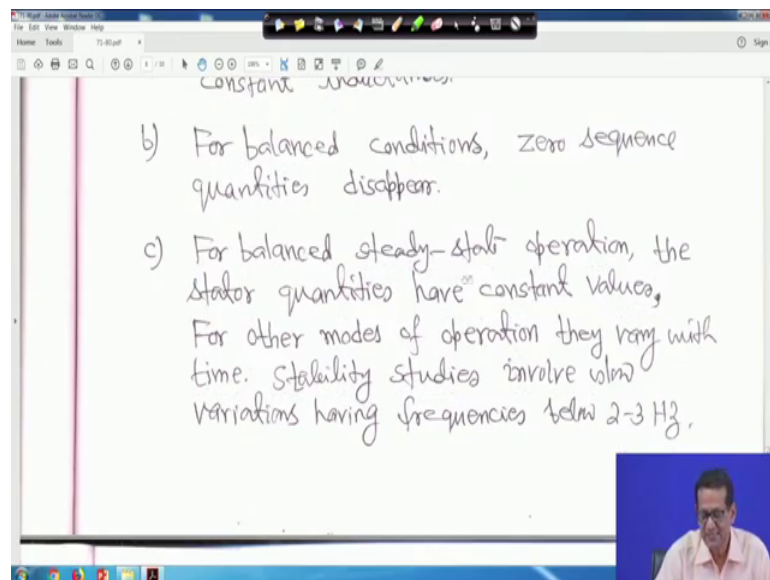
The screenshot shows a presentation slide with three handwritten points:

- a) The dynamic performance equations have constant inductances.
- b) For balanced conditions, zero sequence quantities disappear.
- c) For balanced steady-state operation, the stator quantities have constant values, For other modes of operation they vary with time. Stability studies involve slow variations having frequencies below 2-3 Hz.

A small video inset in the bottom right corner shows a man speaking.

First one is the dynamic performance equation have constant inductances this is the major advantage, right. Next is for balanced conditions 0 sequence quantities disappear, right.

(Refer Slide Time: 18:16)



The screenshot shows a presentation slide with two handwritten points:

- b) For balanced conditions, zero sequence quantities disappear.
- c) For balanced steady-state operation, the stator quantities have constant values, For other modes of operation they vary with time. Stability studies involve slow variations having frequencies below 2-3 Hz.

A small video inset in the bottom right corner shows a man speaking.

For balanced steady state operation the stator quantities have constant values for other modes of operation they may vary with time and stability studies involved your slow variations having frequencies below your 2 or 3 hertz, right.

(Refer Slide Time: 18:35)

d) The parameters associated with d and q-axes may be directly measured from terminal tests.

PER UNIT REPRESENTATION

And fourth one the parameters associated with d and q-axis may be directly measured from your terminal tests, right. So, these are the 4 major advantages for dq 0 transformation.

(Refer Slide Time: 18:49)

quantity in per unit = $\frac{\text{Actual Quantity}}{\text{Base Value of Quantity}}$

Per unit system for the stator quantities

Let us choose the following quantities for the stator (denoted by subscript s)

$E_{\text{base}} = \text{peak value of rated line-to-voltage. Volt}$

Next is the per unit representation. In power system analysis we have already studied the per unit representation, right, but here for synchronous machine although will follow the same thing, but per unit representation will be slightly different, right. And whole heartedly we will try to understand this certain things I will tell you certain things small

derivation I leave up to you also to save some time, but just see how things are the per unit representation. Now, your quantity in per unit can be defined as actually quantity by base value of the quantity this you know. Now, per unit system for the stator quantities, right.

(Refer Slide Time: 19:28)

The following quantities for the stator (denoted by subscript s)

$E_{s\text{ base}}$ = peak value of rated line-to-neutral voltage, Volt.

$I_{s\text{ base}}$ = peak value of rated line current, Amp

f_{base} = rated frequency, Hz.

The base values of the remaining quantities are automatically set and depend on the above

$\omega_{\text{base}} = 2\pi f_{\text{base}}$, elect. radians/sec.

Now, let us choose the following quantities for stator denoted by subscript s, right. So, we have to choose some your what you call your base quantities. For example, $e_{s\text{ base}}$ this is actually peak value of rated line to neutral voltage in volt, right. Then $i_{s\text{ base}}$ it is peak value of rated line current that is ampere and f_{base} that is rated frequency in hertz, right, so this base value. Now, the base values of the remaining quantities are automatically set and depend on the above as follows.

(Refer Slide Time: 20:04)

I_{sbase} = peak value of rated line current, Amp
 f_{base} = rated frequency, Hz.
The base values of the remaining quantities are automatically set and depend on the above as follows:
 ω_{base} = $2\pi f_{base}$, elect. radians/sec.
 ω_{mbase} = $\omega_{base} \left(\frac{2}{p_f} \right)$, mech. radians/sec.

For example omega base here also we need some omega base it will be 2 pi f base that is electrical radian per second then omega mechanical base, right that is omega base into 2 by field force, right p f. So, that is your mechanical radian per second.

(Refer Slide Time: 20:21)

(80)
 Z_{sbase} = $\frac{E_{sbase}}{I_{sbase}}$, Ohms
 L_{sbase} = $\frac{Z_{sbase}}{\omega_{base}}$, Henrys
 Ψ_{sbase} = $L_{sbase} \cdot I_{sbase} = \frac{E_{sbase}}{\omega_{base}}$

Now, if it is so then Z s base will be now e s base upon i s base that basically Z is equal to e by i you know. So, Z s base is equal to e s base upon i s base this is ohm, right. Now, L s base L s base will be your Z s base upon omega base Henry's, right. So, actually when you find the actually we are going to making the base quantities when you try to find out

say reactants of a line, right.

What we do? We make $L \omega$, right $L \omega$ is the reactants. So, here also if you multiply L s base and ω base that actually in per unit we want to represent. So, you represent L s base and into ω base is equal to Z s base, right. Instead of your reactance type we will this is actually Z s base, right dimensionally we have to see that correct thing. So, L s base will be Z s base upon ω base this is in Henry's.

(Refer Slide Time: 21:14)

$$\psi_{sbase} = L_{sbase} \cdot i_{sbase} = \frac{e_{sbase}}{\omega_{base}}, \text{ W-T}$$

$$3\text{-Phase } VA_{base} = 3 E_{RMSbase} \cdot I_{RMSbase}$$

$$= 3 \cdot \frac{e_{sbase}}{\sqrt{2}} \cdot \frac{i_{sbase}}{\sqrt{2}}$$

$$= \frac{3}{2} e_{sbase} i_{sbase}, \text{ Volt}$$

Now, ψ s base that is the flux linkages; we know ψ is equal to Li , right that means, ψ s base will be L s base into i s base, right and this is your L s base and another thing is that can be written as e s base upon ω s base, right. So, this is also this is also your known to known to you, right because you know e is equal to your $d\psi$ by dt in general dimension wise, right. So, this one we make e s base upon ω s. So, L s base into i s base is equal to e s base upon ω this is ω turn, right.

Now, next is 3 phase volt ampere base. So, 3 phase volt ampere base means we will take 3 into E_{RMS} base into I_{RMS} base that is the RMS value of the voltage and RMS value of the current this capital RMS, right. So, this can be written as 3 into E_{RMS} base will be e s base upon $\sqrt{2}$ because we have chosen this your e s base the peak value of the rated line to neutral voltage. Similarly, for the current peak value of rated line current.

(Refer Slide Time: 22:24)

The screenshot shows a whiteboard with the following handwritten equations:

$$\begin{aligned}
 \text{3-Phase } VA_{\text{base}} &= 3 E_{\text{RMSbase}} \cdot I_{\text{RMSbase}} \\
 &= 3 \cdot \frac{E_{\text{base}}}{\sqrt{2}} \cdot \frac{I_{\text{base}}}{\sqrt{2}} \\
 &= \frac{3}{2} E_{\text{base}} I_{\text{base}}, \text{ Volt-Amperes}
 \end{aligned}$$

$$\text{Torque base} = \frac{\text{3-Phase } VA_{\text{base}}}{\omega_{\text{mbase}}}$$

So, when you take capital E RMS, right, basically this is this E RMS base we can write e_{base} upon root 2 and I RMS we could write i_{base} upon root 2. So, root 2 root 2, 2 2 will be 2. So, 2 2 will be your what you call it will be then 3 by 2 e_{base} into i_{base} volt ampere. This is your 3 phase volt ampere base, right.

(Refer Slide Time: 22:45)

The screenshot shows a whiteboard with the following handwritten equations:

$$\begin{aligned}
 &= \frac{3}{2} E_{\text{base}} I_{\text{base}}, \text{ Volt-Amperes} \\
 \text{Torque base} &= \frac{\text{3-Phase } VA_{\text{base}}}{\omega_{\text{mbase}}} \\
 &= \frac{3}{2} \left(\frac{p_f}{2} \right) \psi_{\text{base}} I_{\text{base}}, \text{ Newton-meters.}
 \end{aligned}$$

Now, torque base, we know torque base is equal to in general we know power is equal to speed into torque, right, but we are trying to represent the base quantities. So, torque base will be 3 phase volt ampere base divided by omega mechanical base, right. So, if you and

So, this is your what you call this is your base, right here it is in your what you call here it is given that your this is your just hold on. This is your psi s base is equal to given that is L s i s base into i s base and it is equal to e s base suppose omega base.

So, here also when you write torque is equal to your 3 phase VA base by omega mechanical base. So, in this case this 3 phase volt ampere base is 3 by 2 e s base i s base you substitute here 3 by 2 e s base i s base and this is your omega mechanical base. Just simplify this one, just simplify this one you will get 3 by 2 p f by 2 psi s base into i s base that is Newton your meters, right.

So, here already it is given something is given that your what you call psi s base is equal to L s base i s base everything is given, right. Just you please just to substitute and just to bring this one that it will be psi s base into i s base just. Just little bit you manipulate you put this one, you put this one e s base and i s base and just to meet this one omega base your what you call omega r upon omega s base relationship you put it here and then little bit you simplify you will get 3 by 2 p f by 2 psi s base psi s base Newton per meter. This is your what you call an exercise small exercise for you, right. So, this is your torque base.

(Refer Slide Time: 24:29)

Per Unit Stator Voltage Equations.

From eqn(70),

$$e_d = p\psi_d - \psi_q\omega_r - R_a i_d$$

Dividing throughout by e_{sbase} , and noting that $e_{sbase} = i_{sbase} Z_{sbase} = \omega_{base} \psi_{sbase}$, we get,

$$\frac{e_d}{e_{sbase}} = \frac{p\psi_d}{e_{sbase}} - \frac{\psi_q\omega_r}{e_{sbase}} - \frac{R_a i_d}{e_{sbase}}$$

Next is, per unit stator voltage equation we will come now after making all these small thing we will come to this one. So, from equation 70 this is the equation e_d is equal to $p\psi_d$ minus $\psi_q\omega_r$ minus $R_a i_d$. Now, we have to make the per unit stator voltage

equation, right you have to convert it. So, you what you can do it dividing throughout e by e s base and noting that e s base is equal to i s base into Z s base is equal to omega a base into psi s base.

(Refer Slide Time: 24:59)

we get,

$$\frac{e_d}{e_{sbase}} = p \frac{\psi_d}{\psi_{sbase}}$$

$$\frac{e_d}{e_{sbase}} = p \left(\frac{1}{\omega_{sbase}} \cdot \frac{\psi_d}{\psi_{sbase}} \right) - \frac{\psi_q}{\psi_{sbase}} \cdot \frac{\omega_r}{\omega_{sbase}} - \frac{R_a}{Z_{sbase}} \cdot \frac{i_d}{i_{sbase}} \quad \text{--- (82)}$$

Expressed in per unit notation,

So, if you divide both side e d by e s base. So, this side also you can divide both side that e s base is equal to you know that i s base Z s base. So, that is your what you call that your this is my p psi d. So, both side we are dividing by e s base. So, e d by e s base that this term when you are taking p psi d we are making it as i s base your what you call omega base then psi s base. So, this term divided by omega s base here I have made e is a s here right. So, your this thing is omega s base psi s base.

And for the second term; second term psi q omega r here also we are making psi s base omega s base and last term it is a voltage drop R a i d, right what we are doing is we are making your z s base i s base because e s base is equal to Z s base. So, this term also e s base this term also e s base, but we have to make it per unit form and this relationship, we are using this relationship we are using, right.

(Refer Slide Time: 26:08)

$$\bar{E}_d = p \left(\frac{1}{\omega_{sbase}} \cdot \bar{i}_d \right) - \frac{\omega_r}{\psi_{sbase}} \bar{i}_q - \frac{R_a}{Z_{s-base}} \bar{i}_d \quad \text{--- (82)}$$

Expressed in per unit notation,

$$\bar{i}_d = \frac{1}{\omega_{base}} p \bar{\psi}_d - \bar{\psi}_q \bar{\omega}_r - \bar{R}_a \bar{i}_d \quad \text{--- (83)}$$

The unit of time in the above equation is seconds. Time can also be expressed in per unit notation.

So, with this with this if you do, so then what we will see? We will see that this \bar{e}_d , \bar{e}_d upon \bar{e}_s base now \bar{e}_d upon it is a per unit. So, we are making \bar{e}_d . So, this is actually \bar{e}_d . After making all the per units again we will remove the bar and from that onwards we will understandable that everything is in per unit, that I tell you later \bar{e}_d . And this is your $\bar{\psi}_d$ upon $\bar{\psi}_s$ base. So, this is $\bar{\psi}_d$ bar p into $\bar{\psi}_d$ bar then, right and into 1 upon ω_{base} . And this one $\bar{\psi}_q$ upon $\bar{\psi}_s$ base this is basically $\bar{\psi}_q$ bar and ω_r upon ω_s base that is basically $\bar{\omega}_r$ and this 1 minus R_a upon Z_s base, so per unit, so R_a per unit is \bar{R}_a and \bar{i}_d upon \bar{i}_s base this is \bar{i}_d .

So, the all these things your what you call \bar{e}_d is equal to this equation your what you call the per unit. Whenever putting bar means these are per unit values, right because ultimately you will find at the latest stage things are very simple and we will make it all everything in per unit.

(Refer Slide Time: 27:09)

The screenshot shows a whiteboard with the following content:

$$\bar{E}_d = \frac{1}{\omega_{base}} p \bar{\Psi}_d - \bar{\Psi}_q \bar{\omega}_r - \bar{R}_a \bar{i}_d \quad \dots (83)$$

The unit of time in the above equation is seconds. Time can also be expressed in per unit (or radians) with the base value equal to the time required for the rotor to move one electrical ~~radian~~ radian at synchronous speed.

The unit of time in the above equation is second, right, but most literature also we will find we will generally use time in second, but I will show you if you want per unit also time can be represented by per unit time, all right. So, time also can be expressed in per units with the base value equal to the time required for the rotor to move one electrical radian at synchronous speed, right.

(Refer Slide Time: 27:34)

The screenshot shows a whiteboard with the following content:

$$t_{base} = \frac{1}{\omega_{base}} = \frac{1}{2\pi f_{base}} \quad \dots (84)$$

With time in per unit, Eqn(83) may be written as:

$$\bar{E}_d = p \bar{\Psi}_d - \bar{\Psi}_q \bar{\omega}_r - \bar{R}_a \bar{i}_d \quad \dots (85)$$

Comparing Eqn.(70) and Eqn.(85), we see the form of the original equation is

So, what we can do is that t base actually 1 upon omega base because generally omega is a radian per second radian is a dimensionless quantity so, it is radian per second. So, t

base we will take ω_1 upon ω_{base} that is nothing but your second, right that is $\frac{1}{2\pi f_{base}}$, right, this is equation 84. With the time in per unit equation 83 it may be written as $\bar{e} = \bar{p} \bar{\psi} - \bar{q} \bar{\omega} - \bar{r} - \bar{a} \bar{i}$. Even time also $\bar{p} \frac{d}{dt}$ that is also we are converting in per unit how I will show you. Next lecture I am coming back, ok.

Thank you very much, we will back.