

**On Power System Dynamics Control and Monitoring**  
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**Lecture - 16**  
**Power System stability (Contd.)**

We are back again.

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Second, find  $p = \frac{d}{dt}$ ,  $t$  in sec.

Eqn. (224) can be written as:

$$p \Delta \omega_r = \frac{1}{2H} [\Delta T_m - \Delta T_e - K_D \Delta \omega_r]$$
$$\therefore p \Delta \omega_r = \frac{1}{2H} [\Delta T_m - K_S \Delta \delta - K_D \Delta \omega_r] \text{ --- (226)}$$

Where  $K_S$  is the synchronizing torque coefficient given by

→ 12:75

So, this is this looks like a little bit funny know what question is that is something like.

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The equations of motion in per unit are

$$p\Delta\omega_r = \frac{1}{2H} [T_m - T_e - K_D \Delta\omega_r] \quad \dots (224) \rightarrow 12.73$$

$$p\delta = \omega_0 \Delta\omega_r \quad \dots (225) \rightarrow 12.74$$

where  $\Delta\omega_p$  is the per unit speed deviation,  
 $\delta$  is rotor angle in electrical radians,  $\omega_0$  is

This so, actually we know that we know that your  $\omega_r$  delta  $\omega_r$  is equal to this we have seen earlier  $\omega_r$  minus  $\omega_0$  right this we know. So, what we do you put here that delta  $\omega_r$  is equal to  $\omega_r$  minus  $\omega_0$ . So, if you do so if you do. So, it will writing over it just see that it is  $\omega_r$  minus  $\omega_0$  this term is equal to 1 upon 2 H right in bracket it is  $T_m$  minus  $T_e$  minus your  $K_D$  into your  $\omega_r$  minus  $\omega_0$  right. Actually it is  $K_D$  into  $\omega_r$  minus  $\omega_0$ . Now, if you take small perturbation on both side then what will happen, then it will be  $p$  into delta  $\omega_r$  this is your reference  $p \omega_0$ .

So,  $p$  into delta  $\omega_r$  is equal to it will be 1 upon 2H right and it will be delta  $T_m$  because writing over it just have a look then it will be delta  $T_e$  and again it will be minus  $K_D$  and it is  $\omega_r$  minus  $\omega_0$  it takes a small perturbation into  $K_D$  delta  $\omega_r$  right. So, that is why this  $T_m$  and  $T_e$  it is becoming delta  $T_m$  minus delta  $T_e$  other thing remains same in that equation. So, this is the your already thing is that delta  $\omega_r$  is equal to  $\omega_r$  minus  $\omega_0$  right. So, that is why this equation is written like this right similarly next page we will see, but I making it here just hold on.

Similarly, this is nothing for you similarly equation 225 it is  $p$  delta delta  $p$  delta. So,  $p$  delta is nothing, but this is  $\omega_0$  and delta  $\omega_r$  actually  $\omega_r$  minus  $\omega_0$  right. Now, if you take small perturbation so, it will be  $p$  delta delta is equal  $\omega_0$

then  $\Delta \omega_r$ . So,  $p \Delta \delta$  will be  $\omega_0 \Delta \omega_r$  because this is  $\omega_r$  this is nothing minus  $\omega_0$  right.

So, if you take small perturbation so,  $p \Delta \delta$  is  $\omega_0$ ; that means, this equation can be written as left hand side only  $p \Delta \delta$  right hand side remain as it is  $\omega_0$  into  $\Delta \omega_r$  right. So, there is no confusion at all from this from this mathematical relationship it is coming right absolutely no confusion right. So, this is my equation 20, 225 you know  $\Delta \omega_r$  raise the per unit speed derivation  $\Delta \delta$  is the angle in electrical radian per second  $\omega_r$  reading is the base your base rotor electrical speed in radian per second and  $p$  is the defence law operator later we will replace  $p$  by  $s$  Laplace operator right and  $t$  in second.

Now, equation 224 can be written as that  $p$  it has that what I know  $p \Delta \omega_r$  is equal to  $1 \text{ upon } 2H \Delta T_m \text{ minus } \Delta T_e \text{ minus } \Delta \omega_r$  this I have explained already. Therefore,  $p \Delta \omega_r$  is equal to  $1 \text{ upon } 2H \text{ minus } \Delta T_m \Delta T$  we representing your  $K_S \Delta \delta$ ; that means, these equation this equation this equation later has been explained these equation can be written as it is  $K_S$  into  $\Delta \delta$  right.

So, that is my  $\Delta T_e$  and this  $K_S$  is called synchronising torque coefficient synchronising torque coefficient why it is called synchronising torque coefficient this is question to you right. Book you may not find it, but you have to while you listen this thing you just put this answer to your forum right this term is called synchronising torque coefficient right and it is  $K_S \Delta \delta$ . So, in that equation we replace  $\Delta T$  by  $K_S \Delta \delta$  right. So, therefore,  $\Delta T_e$  is  $K_S \Delta \delta$  and minus  $K_D$  into  $\Delta \omega_r$ .

This is equation 226 forget about this one this is for my own reference. Where  $K_S$  is the synchronising torque coefficient and this  $K_S$  I told you is nothing, but  $E' \text{ upon } X_T \cos \delta_0$  this is equation 227 right.

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$$K_S = \left( \frac{E'/E_B}{X_T} \right) \cos \delta_0 \dots (227) \rightarrow 12.76$$

Eqn. (224) can be written as:

$$p \Delta S = \omega_0 \Delta \omega_p \dots (228) \rightarrow 12.77$$

Eqn. (226) and (228) can be written as

Now equation 224 I told you can be written as I told you already I told you that it was  $p \Delta S = \omega_0 \Delta \omega_p$  then again  $p \Delta S = \omega_0 \Delta \omega_r$  this already I have told you previous page right just previous page from that only that equation 224 how it is coming right.

Therefore equation 226 and 228 can be written in that your state variable form that is your this is there  $p$  is nothing, but  $d/dt$  that is basically  $\Delta \omega$  I mean  $d/dt$  of  $\Delta \omega_r$  means  $\Delta \omega \dot{r}$  and  $d/dt$  of  $\Delta S$  means  $\Delta \dot{S}$ . You put this equation 226 and 228 in the matrix form then it will become a state variable form it will become that it is 2 into 2 matrix from slowly and un slowly dimensionally we will consider other thing.

So, dimension of the matrix will grows slowly and slowly. So, in this case it is minus  $K_D$  upon  $2H$  this is minus  $K_S$  upon  $2H$  this is  $\omega_0$  and this is 0 this is  $\Delta \omega_r$  this is  $\Delta \dot{S}$  plus 1 upon  $2H$  and 0 this is  $\Delta T_m$ .

So, basically your  $d/dt$  your  $p \Delta \omega_r$  is equal minus  $K_D$  upon  $2H$   $\Delta \omega_r$  that you are seen minus  $K_S$  upon  $2H$   $\Delta \dot{S}$  plus 1 upon  $2H$   $\Delta T_m$ . Similarly,  $\Delta \dot{S}$  is equal to  $\omega_0 \Delta \omega_r$  right and this is  $\Delta \omega \dot{S}$  this is  $\Delta T_m$  this is equation 229 right.

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The slide displays the following state-space equation:

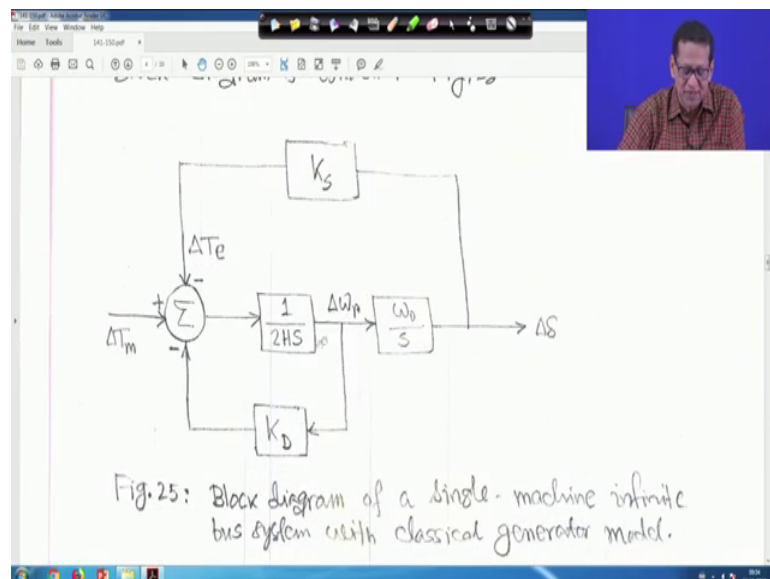
$$\frac{d}{dt} \begin{bmatrix} \Delta\omega_p \\ \Delta\delta \end{bmatrix} = \begin{bmatrix} -\frac{K_D}{2H} & -\frac{K_S}{2H} \\ \omega_0 & 0 \end{bmatrix} \begin{bmatrix} \Delta\omega_p \\ \Delta\delta \end{bmatrix} + \begin{bmatrix} \frac{1}{2H} \\ 0 \end{bmatrix} \Delta T_m$$

Below the equation, it is noted that  $\dot{x} = Ax + Bu$  and that the block diagram is shown in Fig. 25. A reference number (229) is also present.

This equation can be written as  $\dot{x}$  is equal to  $Ax$  plus  $Bu$ ; that means, that means your  $x$  actually this is your a matrix right this is a matrix this is a 2 into 2 matrix right and this is your  $B$  and this is your  $u$ . So, actually  $u$  we are just we are representing like this  $u$  is equal to  $\Delta T_m$  right.

And this is my  $d$  matrix and this is and state variable  $\dot{x}$  is equal to; that means,  $x_1$  is equal to here  $\Delta\omega_p$  and  $x_2$  will be is equal to  $\Delta\delta$  right and this is my state variable equation.

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Now if you represent this in block diagram then this is my  $\Delta T_m$  right and if you look into just see those equation I crossed like the block diagram this is  $\Delta T_m$  and this is minus  $\Delta T_e$  right. So, we have seen know and this is your  $\Delta \omega_r$ .

So,  $K_D \Delta T_m$  that is  $\Delta T_m$  minus  $K_S \Delta T_e$  minus your what you call  $K_D \Delta \omega_r$  right into  $1$  upon  $2HS$  is equal to  $\Delta \omega_r$  this you have seen right and this is your and that is also we have seen from the 2nd equation that is I think 226, I think this is to sorry 2 228 this is 228 it is 228 it is  $p \Delta$  is equal to your  $\omega_0 \Delta \omega_r$  right. So, when you are put in the Laplace transform actually  $p$  is represented by  $S$  right, I mean this one this one now  $p$  actually you replace by  $S$ .

Therefore, this equation yeah instead of look instead of  $S \Delta \Delta S$  then  $\Delta \omega_r$   $S$  we are not representing it is understandable only  $p$  is replace by  $S$ . So, this is  $S \Delta \Delta$  is equal to  $\omega_0 \Delta \omega_r$  right; that means, I am putting it here; that means, my  $\Delta \Delta$  this one is equal to your  $\omega_0$  by  $S$  right into  $\Delta \omega_r$  right.

So,  $\omega_0$  by  $S \Delta \Delta$  is equal to into  $\Delta \omega_r$  that is why your this one your this one your what you call it is  $\Delta \Delta$  is equal to  $\omega_0$  by  $S$  into  $\Delta \omega_r$ . Similarly, here also that your  $2HS$  that  $p$  1st equation, that is your equation your 220 226 right. So, this equation also you replace by you have what you call that  $p$  is equal to  $S$  let  $p$  is equal to  $d/dt$  just one or two lines have write in for you it is  $p$  is equal to you replace  $p$  by  $S$  we will got to the Laplace domain right.

So; that means, this is my  $S$  and this is  $\Delta \omega_r$  is equal to  $1$  upon  $2H$  in bracket  $\Delta T_m$  minus  $\Delta T$  sorry minus your  $K_S \Delta \Delta$  then minus  $K_D \Delta \omega_r$  right. So, what you do multiply cross multiply that is will be  $2HS \Delta \omega_r$  bring this term to this side so, it will be plus  $K_D \Delta \omega_r$ .

So, take  $\Delta \omega_r$  common it will be  $2HS$  plus  $K_D$  right. So, based on that only your this thing is coming right this thing is coming that your block diagram this one is coming one upon your  $1$  upon  $2HS$  and this is your feedback is taken that is your  $K_D$  right that this oneth coming like this. So, in this case in this case later we will see in this case your  $\Delta \omega_r$  this  $K_D \Delta \omega_r$  term you did not bring to the your left hand side only simply you have taken  $S \Delta \omega_r$ , I mean it is simply like this.

Later we will bring it these two that side if you replaced p is equal to S say p is equal to S then your delta omega r is equal to 1 upon 2 HS right and in bracket it is delta Tm minus KS delta delta minus KD your delta omega r right this is taken (Refer Time: 10:37) we did not bring to the left hand side later we will see when block diagram will slowly un slowly will go and a your number your order of the equation will increase. At that time we will bring this one to this side at that time we will find it will be 2HS plus KD omega r will represent step by step right. So, this is 1 upon 2 HS and this one right.

So, that is why your that is why this one this equation this block diagram is like this right it is your 1 upon 2 HS right into your delta Tm minus your KS delta delta minus KD delta omega r right is equal to delta omega r. And this is I told you that o delta delta is equal to omega 0 upon S you know delta omega r. Now this is the block diagram of a single machine infinite bus system with classical generator model.

Now, next is from figure 25, then from this figure you just simplify this one then because if I rest I am still in have to do is delta delta is equal to omega 0 upon S delta delta is equal to omega 0 upon S into delta omega r and delta omega r is equal to 1 upon 2 S 2 HS delta omega r is equal to 1 upon 2 HS that you have seen into delta Tm minus KS delta delta minus KD delta omega r. So, that is delta delta is equal to omega 0 upon S into 1 upon 2 S minus all this things right.

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From Fig. 25,

$$\Delta S = \frac{\omega_0}{s} \left[ \frac{1}{2HS} \left( -K_S \Delta S - K_D \Delta \omega_p + \Delta T_m \right) \right]$$

$$\therefore \Delta S = \frac{\omega_0}{s} \left[ \frac{1}{2HS} \left( -K_S \Delta S - K_D S \cdot \frac{\Delta S}{\omega_0} + \Delta T_m \right) \right]$$

--- (230) → 12:79

Now or delta delta it will be omega 0 upon S 1 upon 2 H S that minus K S delta and this delta omega r you replace delta omega r is equal to your S delta delta upon omega 0. Because, here from here from here your I am writing here just hold on from here that delta delta is equal to omega 0 by S into delta omega r right therefore, delta omega r is equal to S by omega 0 into delta delta right. So, this delta omega is equal to S by omega 0 delta delta is substitute. So, if you do so if you do so it is coming delta omega is equal to KD minus KDS delta delta upon omega 0 plus delta Tm right this is equation 230 this right hand side whatever is it in this for my own reference.

So, now if you simplify this one cross multiplication and simplify then it will be coming like this S square delta delta plus KD upon 2 H S into delta delta plus KS upon 2 H omega 0 delta delta is equal to omega 0 upon 2 H delta Tm right. This is simply quadratic equation and characteristic equation right that is you have start it also in your 3rd year control system engineering.

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$$\therefore s^2(AS) + \frac{K_D}{2H}s(AS) + \frac{K_S \omega_0}{2H}(AS) = \frac{\omega_0}{2H} \Delta T_m$$

Therefore, the characteristic equation is given by

$$s^2 + \frac{K_D}{2H}s + \frac{K_S \omega_0}{2H} = 0 \quad \dots (231) \rightarrow 12.80$$

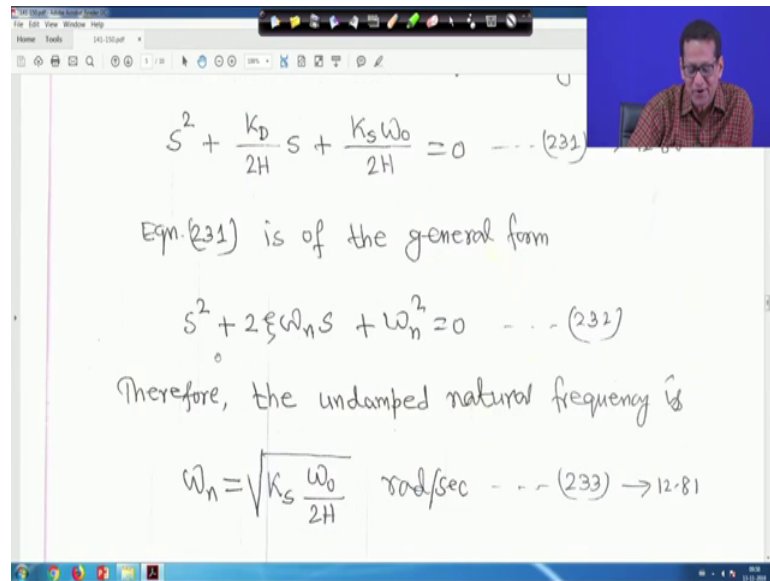
Eqn. (231) is of the general form

$$s^2 + 2\xi\omega_n s + \omega_n^2 = 0 \quad \dots (232)$$

Therefore the characteristics equation is given by second order system it is S square plus KD upon 2 H into S plus KS omega 0 upon 2 H is equal to 0 this is equation 231 right. So, this is the simply second order equation like characteristics equation.



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$$s^2 + \frac{K_D}{2H} s + \frac{K_S \omega_0}{2H} = 0 \quad \dots (231)$$

Eqn. (231) is of the general form

$$s^2 + 2\xi\omega_n s + \omega_n^2 = 0 \quad \dots (232)$$

Therefore, the undamped natural frequency is

$$\omega_n = \sqrt{K_S \frac{\omega_0}{2H}} \text{ rad/sec} \quad \dots (233) \rightarrow 12.81$$

So, equation 231 is of the general form, you have studied in your 3rd year control system engineering for second order system it is  $S$  square plus  $2 \xi \omega_n S$  right plus  $\omega_n$  square is equal to 0 right. Therefore,  $\omega_n$  square your is equal to having this is this equation this equation and this there analogues to each other therefore,  $\omega_n$  square is equal to  $K_S$  sorry just let me let me delete it once again.

So, these two equations are analogues to each other therefore,  $\omega_n$  square is equal to this one your  $K_S \omega_0$  by  $2 H$  right. Therefore,  $\omega_n$  is equal to root over  $K_S \omega_0$  upon  $2 H$  that is here radian per second  $\omega_n$  is equal to root over  $K_S$  upon  $\omega_0$   $2 H$  this is equation 233 this is nothing this for my own reference right. So, this is radian that is  $\omega_n$  right another thing is that  $2 \xi \omega_n$  let me clear it another thing is that this term is analogues to this term is equal to this one; that means,  $2 \xi \omega_n$  is equal to  $K_D$  upon  $2 H$  right. So, from which  $\omega_n$  is known and if you substitute  $\omega_n$  you will get this value of  $\xi$  right that I have written in the next page.

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(146)

and the damping ratio is

$$\zeta = \frac{1}{2} \cdot \frac{k_D}{2H\omega_n}$$

$$\therefore \zeta = \frac{1}{2} \frac{k_D}{\sqrt{2HK_S \omega_0}} \quad \text{--- (234) ---} \rightarrow 12.82$$

So, and the, that is and the psi is the damping ratio that is psi is equal to half into KD upon 2 H omega n right. So, therefore, the undamped natural frequency omega n is the this one right and your the damping ratio is psi is half into KD upon 2 H omega n. Now put omega n value and simplify you will get psi is equal to half KD root over 2 H KS omega 0 this is equation 234 forget about this one this is for my refer own reference.

As the synchronizing torque coefficient that is a KS increases the natural frequency increases and the damping ration decreases that as KS increases natural frequency your increases because this is this is actually omega n root over KS into omega 0 2 H. If KS increases then omega n natural frequency also increases and in the damping ratio KS KS term is root over 2 H KS omega 0 it is in the denominator. So, KS increases that psi decreases right that damping ration decreases.

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As the synchronizing torque coefficient  $K_s$  increases, the natural frequency increases and the damping ratio decreases.

An increase in damping torque coefficient  $K_D$  increases the damping ratio, whereas an increase in inertia constant decreases both  $\omega_n$  and  $\xi$ .

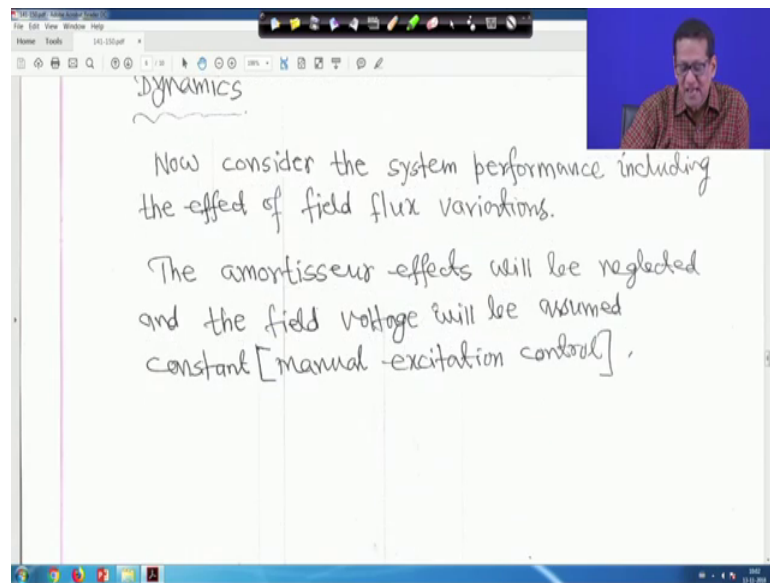
Effects of Synchronous Machine Field Circuit Dynamics

And increase in damping increasing damping torque coefficient  $K_D$  increases the damping ratio. Now  $K_D$  is the new meritory if  $K_D$  increases damping ratio increases right whereas, an increase in inertia constant decreases both  $\omega_n$  and  $\xi$ . So, if it is in the your denominator  $H$  increases means  $\xi$  will decrease right similarly here also it is in the root over  $K_S \omega_0$  upon  $2H$ .

So, each in the in the denominator so, if  $H$  increases. So, natural frequency will decrease right. So, next is that effect of I think this block diagram is understandable to later we will take some problem later right. So, next is effect of synchronous machines field circuit dynamics. So, now consider the system performance including the effect of field flux variations now will effect of synchronous machine field circuit dynamics all this equations for field circuit everything has been developed before.

So, what I will suggest that again and again I cannot go back to those equations which I have left much before. So, whenever you will listening to this lecture just open the those a those notes because all notes every hour every week you will get those notes keep it in point of view then things will be easier. With now consider the system performance including the effect of field flux variations.

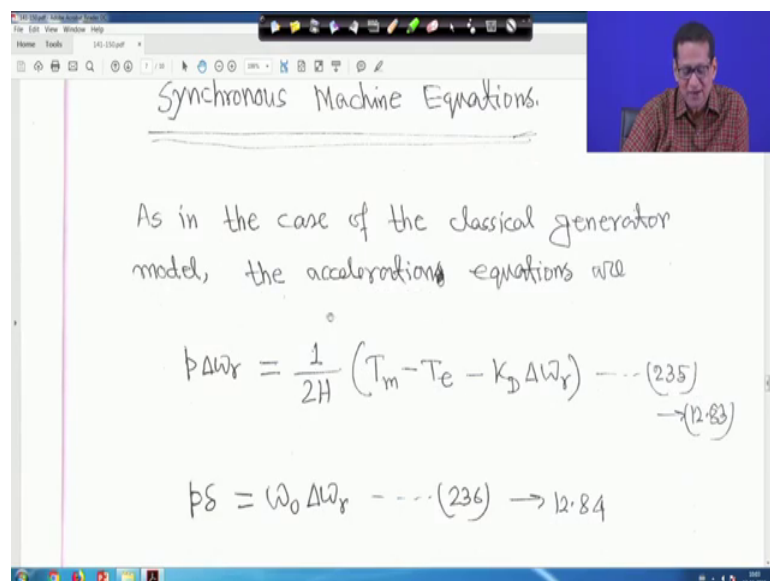
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The screenshot shows a whiteboard with the word "Dynamics" written at the top. Below it, the text reads: "Now consider the system performance including the effect of field flux variations. The amortisseur effects will be neglected and the field voltage will be assumed constant [manual excitation control].". A small video inset in the top right corner shows a man speaking.

The amortisseur effects will be neglected just for the simplicity and the analysis because you will think about the classroom exercises. So, will be neglected and the field voltage will be assumed constant that is the manual excitation system you will assume the field voltage will remain constant. Regarding excitation system I told you time permits and the end I will try, but if time does not permit it will be impossible right. So, amortisseur effects will be neglected and the field voltage will assume that is constant that is EFD if it is constant; that means, your perturbation  $\Delta EFD$  will be 0 right.

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The screenshot shows a whiteboard with the title "Synchronous Machine Equations." underlined. Below the title, the text reads: "As in the case of the classical generator model, the acceleration equations are". Two equations are written: 
$$p\Delta\omega_r = \frac{1}{2H} (T_m - T_e - K_D \Delta\omega_r) \dots (235) \rightarrow (12.83)$$
 and 
$$p\delta = \omega_0 \Delta\omega_r \dots (236) \rightarrow 12.84$$
. A small video inset in the top right corner shows a man speaking.

So, synchronous machine equations; this all that now and in the case of the classical generator model the accelerations equations are we have seen that  $p \delta = \omega_r$  is equal to  $\frac{1}{2H} (T_m - T_e - KD \omega_r)$  and  $p \delta = \omega_0 \delta$ . Or in other words small perturbation when you see  $p \delta = \omega_r$  upon  $\frac{1}{2H}$  we have seen  $\delta T_m - \delta T_e - KD \omega_r$  and  $p \delta$  will be  $\omega_0 \delta$  right.

Now  $\omega_0$  is equal to  $2\pi f_0$  electrical radian per second, that is your base frequency right. Now, in this case the rotor angle  $\delta$  is the angle that is electrical radian per second by the q axis lead the reference voltage  $E_B$  that also we have seen right.

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The image shows a video lecture interface. At the top, there is a toolbar with various drawing tools. Below the toolbar, the whiteboard contains the following handwritten text:

$p\delta = \omega_0 \Delta\omega_r \dots (236) \rightarrow 12.8$

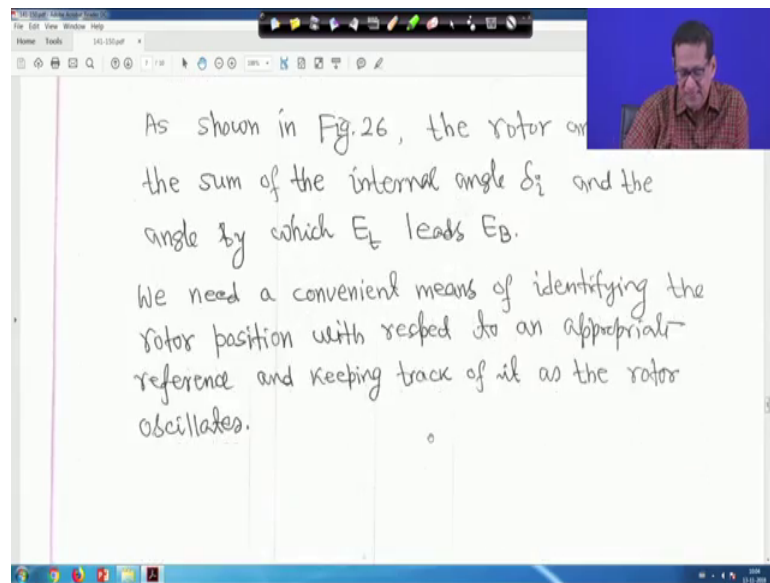
Where  $\omega_0 = 2\pi f_0$  elect. rad/sec.

In this case, the rotor angle  $\delta$  is the angle (elect. rad) by the q-axis leads the reference  $E_B$ .

As shown in Fig.26, the rotor angle  $\delta$  is the sum of the internal angle  $\delta_i$  and the angle by which  $E_t$  leads  $E_B$ .

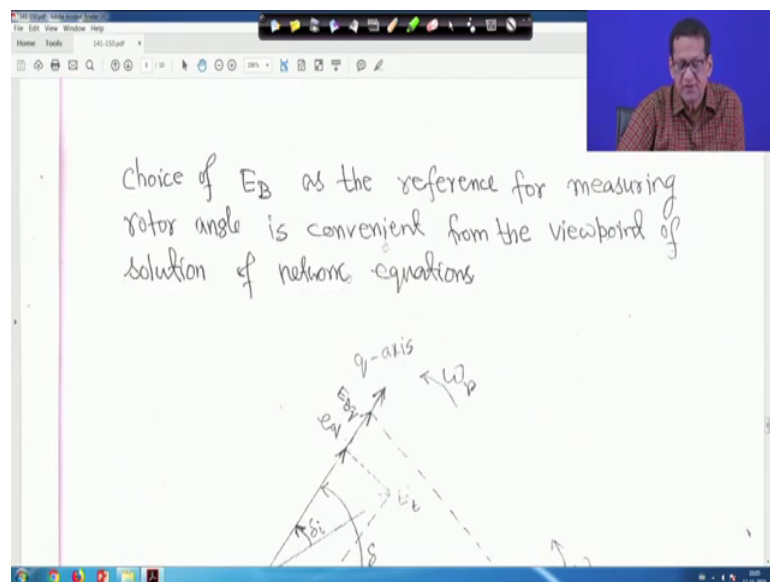
In the top right corner, there is a small video inset showing a man with glasses and a red shirt, presumably the lecturer.

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As shown in figure 26 the rotor angle  $\delta$  is the sum of the internal angle  $\delta_i$  and the angle by which  $E_t$  leads  $E_B$ .

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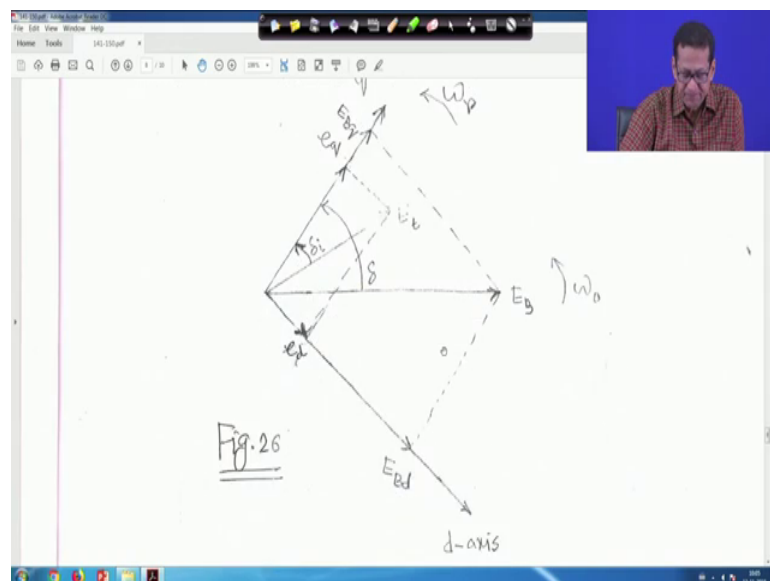
So, before going to this I will go to this figure this is figure 26 this is my  $E_B$  say right this is my  $E_B$  this is the reference we have taken and this is your q axis and this is d axis. So, on the q axis it will  $E_B$  that is capital  $E_B$  and on the d axis it will capital  $E_D$  right and this is my  $E_t$  that is your that is your terminal voltage  $E_t$  this is  $E_t$  hope it is readable right this is my  $E_t$  and this is the pressure diagram.

And its and its your on q axis component  $e_q$  and d axis component  $e_d$  and  $E_t$  this  $E_t$  actually your lagging from this  $e_q$  or capital  $E_B$  by your this thing by an angle  $\delta$  i this also you have seen earlier and this is the delta. Delta is equal to the  $\delta_i$  plus the angle between  $E_t$  and  $E_B$  and your delta is equal to  $\delta_i$  this angle plus the angle  $\beta$  in  $e_t$  and  $E_B$  right.

So, so, I will go back again. So, this is what I have told in that is figure 26 the rotor angle delta is the sum of the internal delta  $\delta_i$  just told you and the angle by which  $E_t$  leads  $E_B$ . Now, we need a conventional means of identifying the rotor position with respect to the appropriate reference and keeping track of it as the rotor oscillates right.

So, choice of  $E_B$  as the reference for measuring the rotor angle is convenient from the viewpoint of solution of network equation. So,  $E_B$  we have taken as a reference right and delta will be  $\delta_i$  plus the angle  $\beta$   $E_t$  and  $E_B$  right.

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Fig. 26

$E_{fd}$

d-axis

From eqn.(121), with time 't' in seconds instead of per unit, the field circuit dynamic equation is

$$p\Psi_{fd} = \omega_0 (e_{fd} - R_{fd} i_{fd})$$

$$\therefore p\Psi_{fd} = \frac{\omega_0 R_{fd}}{L_{adu}} E_{fd} - \omega_0 R_{fd} i_{fd} \text{ -- (237) } \rightarrow 12.55$$

So, from equation 121 right this all these equations had been given before you go to equation 121 with time t in second instead of per unit the field circuit dynamic equation can be given as your p psi fd omega 0 efd minus Rfdifd from equation 121. I suggest that this all the lecture notes will be available to you just open it point of view because if I want to go back because of 121 much time will be lost right.

So, or this efd also we have multiply omega 0 and efd also define report you put it here right. So, it will be and when you using u and s u means Unsaturated value and s means Saturated value with the inductance or reactance when you have put it right. So, p psi fd is equal to omega 0 Rfd upon Ladu Efd minus omega 0 Rfdifd this is equation 237 right.

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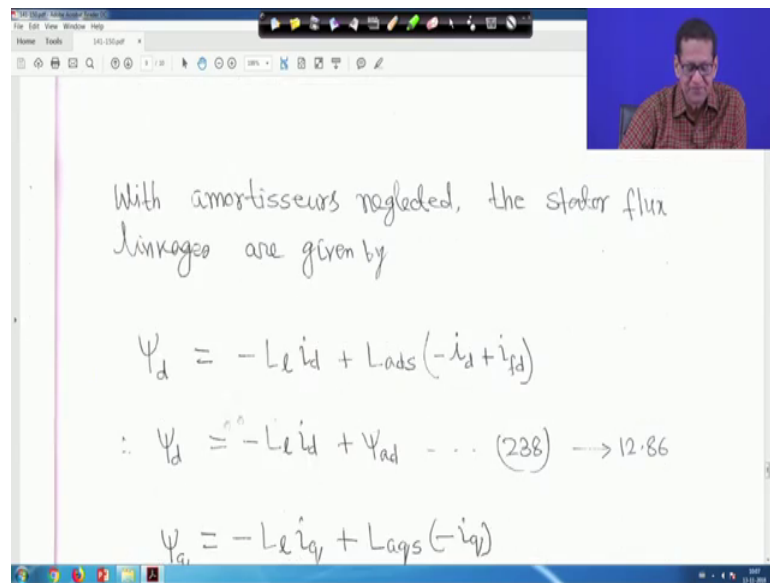
$$p\Psi_{fd} = \omega_0 (e_{fd} - R_{fd} i_{fd})$$

$$\therefore p\Psi_{fd} = \frac{\omega_0 R_{fd}}{L_{adu}} E_{fd} - \omega_0 R_{fd} i_{fd} \text{ -- (237) } \rightarrow 12.55$$

Where  $E_{fd}$  = exciter output voltage.



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With amortisseurs neglected, the stator flux linkages are given by

$$\Psi_d = -L_l i_d + L_{ads}(-i_d + i_{fd})$$
$$\therefore \Psi_d = -L_l i_d + \Psi_{ad} \dots (238) \rightarrow 12.86$$
$$\Psi_q = -L_l i_q + L_{aqs}(-i_q)$$

So, now, with amortisseurs neglected the stator flux linkage are given by that that go back to the stator circuit equation right this equation if I recall correctly go to the figure 28 right. So, from figure 28 that where analogous circuit is given flux then your inductance then your what you call that your current right. So, with amortisseurs neglected the stator flux linkage are given by this go to figure 20 which you have left out which we have left before right.

So, it will be minus  $L_l i_d$  plus  $L_{ads}$  in bracket minus  $i_d$  plus  $i_{fd}$  right or  $\psi_d$  is equal to minus  $L_l i_d$  plus  $\psi_{ad}$ .  $\psi_d$  is equal to this term this go back to the figure 28 if I recall correctly please go back to figure 20 this is my  $\psi_{ad}$  right this is nothing for you this is for my reference right. So, this is  $\psi_{ad}$   $\psi_d$  is equal to minus  $L_l i_d$  plus  $\psi_{ad}$  go to figure 20 I think figure 28 right. Similarly, figure 20 b the  $\psi_q$  is equal to minus  $L_l i_q$  plus  $L_{aqs}$  into minus  $i_q$  right or it is  $\psi_q$  is equal to minus  $L_l i_q$  plus  $\psi_{aq}$  this is actually  $\psi_{aq}$  this is equation 239 you will go back to figure 20 b right.

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$$\therefore \psi_q = -L_q i_q + \psi_{ag} \quad \dots (239) \rightarrow 12.87$$

$$\psi_{fd} = L_{ads} (-i_d + i_{fd}) + L_{fd} i_{fd}$$

$$\therefore \psi_{fd} = \psi_{ad} + L_{fd} i_{fd} \quad \dots (240) \rightarrow 12.88$$

$\psi_{ad}$  and  $\psi_{ag}$  are the air-gap (mutual) flux linkages, and  $L_{ads}$  and  $L_{ag}$  are the saturated values of the mutual inductances.

Now, therefore psi fd is equal to Lads bracket minus id plus ifd plus Lfd ifd all you back to figure 20 a or b right or psi fd is equal to psi ad plus Lfd ifd because these term psi ad that is equation 240 right. Now psi ad and psi q are the air gap that is mutual flux linkages and Lads and Laqs are the saturated values of the mutual inductances right.

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$\psi_{ad}$  and  $\psi_{ag}$  are the air-gap (mutual) flux linkages, and  $L_{ads}$  and  $L_{ag}$  are the saturated values of the mutual inductances.

From eqn. (240),

$$i_{fd} = \frac{\psi_{fd} - \psi_{ad}}{L_{fd}} \quad \dots (241) \rightarrow 12.89$$

From equation 240 I mean from this equation just previous equation we can write ifd is equal to psi fd minus psi ad upon Lfd right. So, this is equation 241 this is for my own reference right. Now that d axis mutual flux linkage can be written in terms of psi fd and id.

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The d-axis mutual flux linkage can be written in terms of  $\Psi_{fd}$  and  $i_d$  as follows:

$$\Psi_{ad} = -L_{ads} i_d + L_{ads} i_{fd}$$

$$\therefore \Psi_{ad} = -L_{ads} i_d + \frac{L_{ads}}{L_{fd}} (\Psi_{fd} - \Psi_{ad})$$

Therefore  $\psi_{ad}$  can be written as minus  $L_{ads} i_d$  plus  $L_{ads} i_{fd}$ . I mean just multiply that this equation this equation just a this equation is minus  $L_{ads} i_d$  plus your your  $L_{ads} i_{fd}$  right. I mean this is your  $\psi_{ad}$  this is actually somewhere this is actually your  $\psi_{ad}$  this is  $\psi_{ad}$  if  $i_{fd}$  is equal to minus  $L_{ads} i_d$  plus your  $L_{ads} i_{fd}$  right.

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$\Psi_{ad} = L'_{ads} \left( -i_d + \frac{\Psi_{fd}}{L_{fd}} \right) \dots (242) \rightarrow 12.90$

where

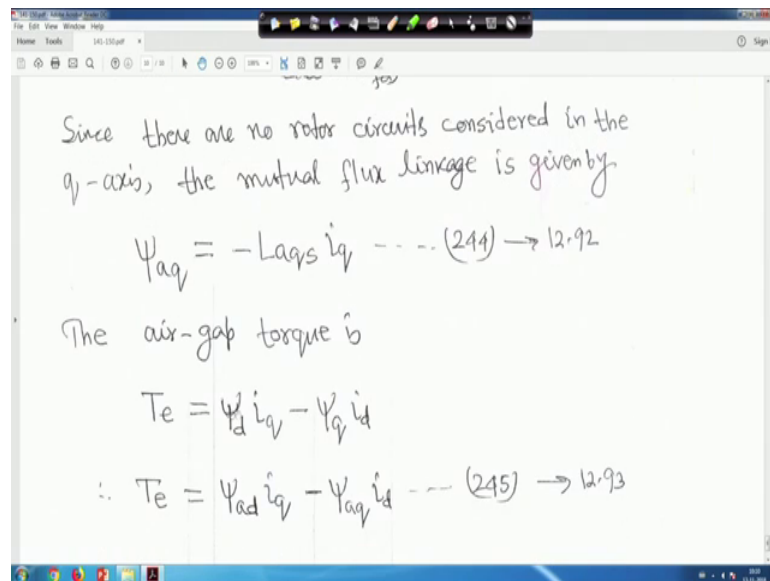
$$L'_{ads} = \frac{1}{\left( \frac{1}{L_{ads}} + \frac{1}{L_{fd}} \right)} \dots (243) \rightarrow 12.91$$

Since there are no rotor circuits considered in the q-axis, the mutual flux linkage is given by.

So, now, therefore, your this is minus  $L_{ads} i_d$  and  $i_{fd}$  from this equation this one you substitute here you substitute here right you put it here.

So; that means,  $\psi_{ad}$  is equal to  $L_{ad} i_d$  plus  $\psi_{fd}$  upon  $L_{fd}$  say this is equation 242 right. Where,  $L_{ad}$  is equal to  $\frac{1}{L_{ad} + 1}$  upon your denominator  $1$  upon numerator denominator will be  $1$  upon  $L_{ad} + 1$  upon  $L_{fd}$  this is the simplify you will get it just you simplify. Since there are no rotor circuits considered in the  $q$  axis the mutual flux linkage is given by  $\psi_{aq}$  is equal to minus  $L_{aq} i_q$  this we have seen this is equation 244 right.

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Since there are no rotor circuits considered in the  $q$ -axis, the mutual flux linkage is given by

$$\psi_{aq} = -L_{aq} i_q \quad \dots (244) \rightarrow 12.92$$

The air-gap torque is

$$T_e = \psi'_d i_q - \psi'_q i_d$$

$$\therefore T_e = \psi_{ad} i_q - \psi_{aq} i_d \quad \dots (245) \rightarrow 12.93$$

Therefore the air gap torque is we know this equation  $T_e$  is equal to  $\psi_d i_q$  minus  $\psi_q i_d$  this is the air gap torque or we can write  $T_e$  is equal to  $\psi_{ad} i_q$  minus  $\psi_{aq} i_d$  this is a small exercise for you that we are writing  $T_e$  is equal to  $\psi_d i_q$  minus  $\psi_q i_d$  this we have seen before is equal to a writing that  $\psi_{ad} i_q$  and  $\psi_{aq} i_d$ .

And just this is a this is a this is a small exercise for you that this one and this one right why why we are writing like. This is this is nothing for you and this also nothing for you right this is a small exercise for you put the answer in the forum if you cannot do it we will we will explain there right. So, 1 or 2, I am leaving upto you just all these derivations have been made before.

So, from that we are writing like this so, why we are writing like this right. So, so, just hold down we will go to the next page.

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(151)

With  $p\psi$  terms and speed variations neglected, the stator voltage equations are

$$e_d = -R_a i_d - \psi_q$$

$$\therefore e_d = -R_a i_d + (L_e i_q - \psi_{aq}) \quad \dots (246)$$

$$e_q = -R_a i_q + \psi_d$$

So, if the  $p\psi$  terms and speed variations neglected the stator voltage equations are. So, you go back to that your stator voltage equations whatever we have derived much before right there what you do you  $p\psi$  terms and speed variations you just neglect you neglect and rewrite those equation for  $e_d$  and  $e_q$  right.

So, you with  $p\psi$  terms and speed variations you neglect and the stator voltage equations then you will find simply will be  $e_d$  will be minus  $R_a i_d$  minus  $i_q$  just you go back to the stator voltage equations and drop those terms it is simply like this.

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the stator voltage equations are

$$e_d = -R_a i_d - \psi_q$$

$$\therefore e_d = -R_a i_d + (L_e i_q - \psi_{aq}) \quad \dots (246)$$

$$e_q = -R_a i_q + \psi_d$$

$$\therefore e_q = -R_a i_q - (L_e i_d - \psi_{od}) \quad \dots (247)$$

Network Equations

And your  $e_d$  will be minus  $R_a i_d$  and  $\psi_q$  will be  $L i_q$  minus  $\psi_a q$  this is again it is coming from figure twenty b this  $\psi_q$  is equals to  $L i_q$  minus  $\psi_a q$  this is figure your 20 b if I recall correctly I will just see the previous equations whatever I have been made right. So, this is actually equation 246 or similarly  $e_q$  is equals to minus  $R_a i_q$  plus  $i_d$  and this is again this  $\psi_d$  is equals to  $L i_d$  minus  $\psi_a d$  this equation again from figure 20 a just go back to that right therefore,  $e_q$  is equals to minus  $R_a i_q$  minus  $L i_d$  minus  $\psi_a d$  this is equation 247 right.

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The screenshot shows a presentation slide with the following content:

Network Equations

Referring to Fig. 26, the machine terminal and infinite bus voltages in terms of the d and q components are:

$$\vec{E}_t = e_d + j e_q \quad \dots (248) \rightarrow$$

$$\vec{E}_B = E_{Bd} + j E_{Bq} \quad \dots (249) \rightarrow$$

The network constraint equation for the system of Fig. 23(b) is

So, now with referring to figure 26 the missing terminal and infinite bus voltage in terms of that  $d_x$  and  $q_x$  is component r.

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q components are:

$$\tilde{E}_t = e_d + j e_q \quad \dots (248)$$

$$\tilde{E}_B = E_{Bd} + j E_{Bq} \quad \dots (249)$$

The network constraint equation for the system of Fig. 23(b) is

This we have seen in figure 26  $\tilde{E}_t = e_d + j e_q$  this is 248 and  $\tilde{E}_B$  that you are you are what you call they are a 2 components that they are direct axis and quadrature axis components. So, it is  $E_{Bd} + j E_{Bq}$  please go back to figure 26 and you can get it and you can easily see this already I have told they are need to be constant equations for the system of figure 23 b is right that is  $\tilde{E}_t = E_B + r_e + j X E$  it this is actually you go back to figure 23 b. Figure 23 was that your generator than a transmission system then a last system.

And thevenin equivalent was that  $r_e + j X E$  and in addition to that and the infinite thevenin base voltage that the thevenin voltage  $E_B$  that is given on your figure 23 b. So, from that from that from that figure figure you can easily write  $\tilde{E}_t = E_B + r_e + j X E$  it tilde right or  $e_d + j e_q$   $\tilde{E}_t = e_d + j e_q$   $\tilde{E}_t = e_{Bd} + j e_{Bq}$  and  $r_e + j X E$  it tilde is equals to  $i_d + j x_q$  right and this is my equation 250 right.

Now, if you multiply and you separate real and imaginary part then you will get you just multiply and simplify real and imaginary parts you will get.

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$$\tilde{E}_t = \tilde{E}_B + (R_E + jX_E) \tilde{I}_t$$

$$\therefore (E_d + jE_q) = (E_{Bd} + jE_{Bq}) + (R_E + jX_E)(i_d + j i_q) \text{ --- (250)}$$

$$\therefore E_d = R_E i_d - X_E i_q + E_{Bd} \text{ --- (251)}$$

$$E_q = R_E i_q + X_E i_d + E_{Bq} \text{ --- (252)}$$

where

$$E_{Bd} = E_B \sin \delta \text{ --- (253)}$$

ed is equals to Re id minus XEiq plus EBd capital EBd just you put it on and eq is equals to Re iq plus Xe id plus EB q this is equation 252 right; that means, this one you put all these things you multiply this one and then separate real and imaginary parts right you will get this two equation.

Thank you very much I will be back again.