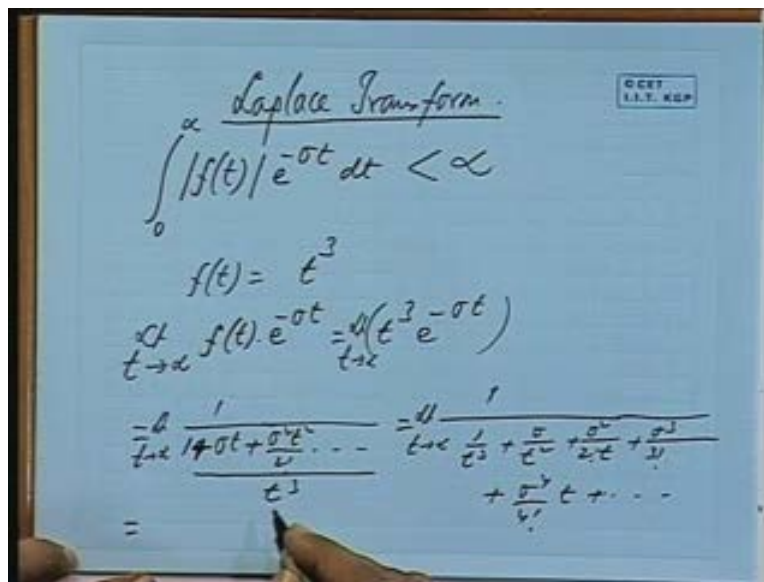


**Networks Signals and Systems**  
**Prof. T. K. Basu**  
**Department of Electrical Engineering**  
**Indian Institute of Technology, Kharagpur**  
**Lecture - 08**  
**Laplace Transform (contd...)**

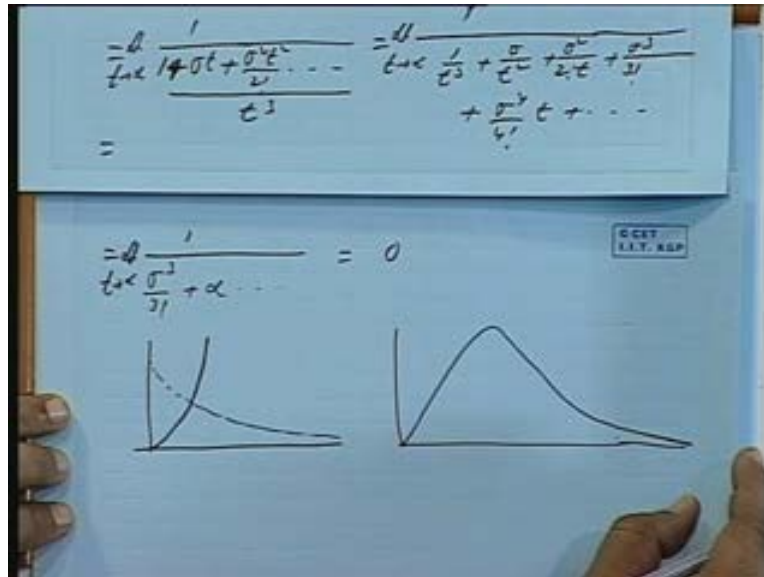
Okay Good morning friends, we shall continue our discussions on Laplace transform. Yesterday, we derived a transforms of some standard functions will continue with that before we go to the set of other properties I just like to mention there are function  $f(t)$  is Laplace transformable if  $\int_0^{\infty} |f(t)| e^{-\sigma t} dt < \infty$  mod e to the power minus sigma t dt, 0 to infinity is a finite quantity that is it is less than infinity if you have a function say  $f(t)$  equal to say  $t$  squared or  $t$  cubed will this property hold good for this  $f(t)$ .

(Refer Slide Time: 01:06)



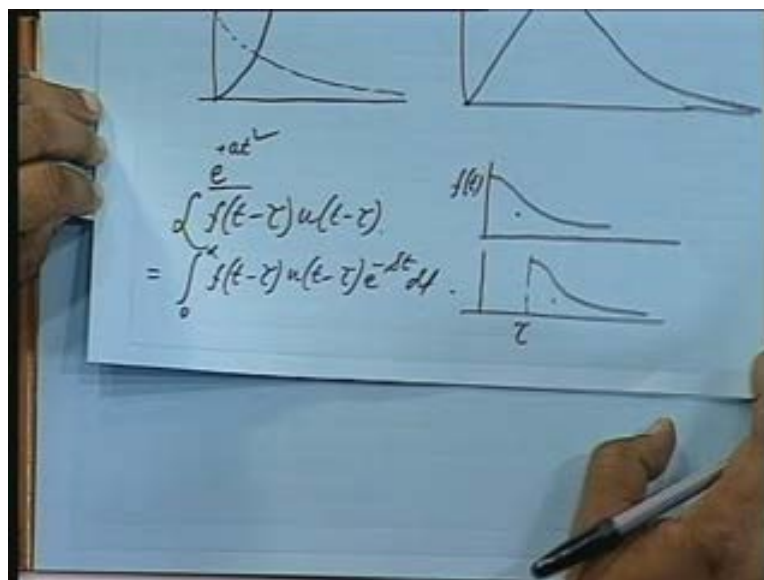
Let us see,  $f(t)$  in to e to the power minus sigma t as  $t \rightarrow \infty$ , how much is this product  $t$  cubed e to the power minus sigma t okay limit  $t \rightarrow \infty$ , so what will be the limit. I can always write this as e to the power sigma t, I can write as  $1 + \sigma t + \frac{\sigma^2 t^2}{2!} + \frac{\sigma^3 t^3}{3!} + \dots$  so  $1 + \sigma t$  sorry,  $1 + \sigma t + \frac{\sigma^2 t^2}{2!} + \frac{\sigma^3 t^3}{3!} + \dots$  the  $t$  cube on the numerator can be brought back here, it can be brought down here. So this will be  $1 + \sigma t + \frac{\sigma^2 t^2}{2!} + \frac{\sigma^3 t^3}{3!} + \dots$  by limit  $t \rightarrow \infty$   $1 + \sigma t + \frac{\sigma^2 t^2}{2!} + \frac{\sigma^3 t^3}{3!} + \dots$  all right like that plus sigma cubed by factorial 3  $t$  cube will get cancelled and after that you will get sigma 4 by factorial 4  $t$  and so on okay. Now how much is this coming to  $1 + \sigma t$  cube it will be 0, this will be 0, this will be 0, this will be finite. So it would be equal to  $1 + \sigma t$  cubed by factorial 3 plus this will be infinite terms multiplied by  $t$  so it will be equal to 0 this is may be  $t \rightarrow \infty$ , so this will be 0.

(Refer Slide Time: 03:23 min)



so a function now when  $t$  tends to 0, when tends to 0 this is a infinity, this is a infinity, this is a infinity, these are all 0's so 1 by infinity that is again 0. So what will this function look like  $t$  cubed is a function which is going up like this to be parallel alright and  $t$  cubed in to  $e$  to the power minus sigma  $t$ ,  $e$  to the power minus sigma  $t$  is a function going like this. So the product will be starting from 0 ending at 0 so somewhere it may have a peak and then it will go like this.

(Refer Slide Time: 04:43)



So the area will be finite this is going to 0, so area is finite. So I have just taken an example with t cube it can be show on that for any t to the power n it will be always converging to finite value and hence the integration from 0 to infinity will also give you some finite sum it will be less then infinity, so it is transformable. Functions of this kind, so e to the power minus or plus a t squared if f (t) is of this kind then you will find again by taking the limits this will not tend to a finite value. So this is not transformable but any other functions set e to the power n are functions of this kind they will be transformable.

So for any transformable functions you have to check whether this property holds good or not, f (t) may go to infinity but f (t) in to e to the power minus sigma t this integration this entire product if it goes to a finite value then and the integration if it gives you a finite value then it is transformable. Next will take up some properties of Laplace transform, Laplace transform of shifted function. Last time we discussed about some shifted functions how to represent them suppose f (t) is a function like this then f (t) minus tau in to u t minus tau is basically this same function starting after the interval tau okay.

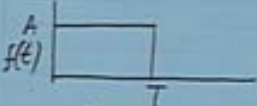

So what will be the Laplace transform of this in terms of Laplace transform of this function okay, what will be the Laplace transform of this function in terms of the Laplace transform of this function, you would like to find out that. Now Laplace transform of f (t) minus tau u t minus tau will be equal to integration 0 to infinity f (t) minus tau u t minus tau e to the power minus s t dt okay, by definition. So that gives me equal to 0 to infinity now between 0 and infinity if I take a function now the function starts after tau that means it is 0 here. So the integration from 0 to infinity means tau to infinity so I can break it up in to 0 to tau f t minus tau, u t minus tau e to the power minus s (t) dt plus tau to infinity f (t) minus tau, u t minus tau e to the power minus s (t) dt and this part is 0, 0 to tau is equal to 0 plus tau to infinity t minus tau u t minus tau, e to the power minus s t dt.

(Refer Slide Time: 06:49)

$$\begin{aligned}
 &= \int_0^{\infty} f(t-\tau) u(t-\tau) e^{-st} dt \\
 &\quad + \int_{\tau}^{\infty} f(t-\tau) u(t-\tau) e^{-st} dt \\
 &= 0 + \int_{\tau}^{\infty} f(t-\tau) u(t-\tau) e^{-st} dt \\
 &= \int_{\tau}^{\infty} f(t_1) u(t_1) e^{-s(t_1+\tau)} dt_1, \quad \begin{array}{l} t-\tau = t_1 \\ dt = dt_1 \\ t = t_1 + \tau \end{array} \\
 &= e^{-s\tau} \int_{\tau}^{\infty} f(t_1) u(t_1) e^{-st_1} dt_1 \\
 &= e^{-s\tau} F(s)
 \end{aligned}$$

Now  $t$  minus  $\tau$  I can put another variable  $t_1$  so  $dt$  is equal to  $d(t_1)$  and  $t$  is equal to  $t_1$  plus  $\tau$ , so replace  $t$  by  $t_1$  plus  $\tau$ , so when  $t$  is  $\tau$  when  $t_1$  is 0, so this limit become 0 to infinity  $f(t_1) u(t_1)$  actually  $f(t_1) u(t_1)$  from 0 it does not mean much here I can break it up in to  $e$  to the power minus  $s(t_1)$ ,  $e$  to the power minus  $s\tau$ ,  $e$  to the power minus  $s\tau$  can be taken out because the integration is with respect to  $t_1$ , so thank you very much 0 to infinity  $f(t_1) u(t_1)$ ,  $e$  to the power minus  $s(t_1) d(t_1)$  and what is this  $f$  of  $s$ . So in terms of the original transform the shifted function has a transform that is  $e$  to the power minus  $s\tau$  in to  $f(s)$  okay. So now let us derived some of those the Laplace transform of some of the shifted functions, we discussed yesterday.

(Refer Slide Time: 09:27)

$f(t)$   
 $=$    $+$    
 $= \frac{A}{s} - \frac{A}{s} e^{-sT}$   
 $= \frac{A}{s} [1 - e^{-sT}]$   
 For  $\delta(t)$   
 $= \frac{A}{s} [1 - (1 - sT + \frac{s^2 T^2}{2} \dots)] = \frac{A}{s} (sT - \frac{s^2 T^2}{2} \dots)$   
 $AT=1$   
 $T \rightarrow 0, A \rightarrow \infty$   
 $f(t) = \delta(t)$

Let us take a simple example, suppose this is  $A$ ,  $T$  this we wrote as equal to a function like this plus function like this minus  $A$  and plus  $A$ , is it not? So what will be the Laplace transform of the first function  $A$  by  $s$  this one  $A$  by  $s$  shifted by  $T$ , so multiplied by  $e$  to the power minus  $s t$  just now we have seen that any function if it is shifted by  $\tau$  so the original Laplace transform multiplied by  $e$  to the power minus  $s\tau$ , this will be the result. So this is  $A$  by  $s$  in to  $1$  minus  $e$  to the power minus  $s t$  okay. Now suppose we take  $T$  very **very** small  $T$  very **very** small  $10$  in to  $0$  and  $A$  is  $10$  in to infinity such that  $A$  in to  $T$  is  $1$  we get a unit impulse, is it not? When  $T$  tends to  $0$  and  $A$  tends to infinity but the product is finite and in this case if it is  $1$  then what will be this this will tend to for this  $f(t)$  is nothing but  $\delta t$  okay.

So this value are  $f(t)$  so Laplace transform of  $\delta t$  we want to compute will substitute that condition here  $A$  by  $s$  in to  $1$  minus what is  $e$  to the power minus  $s t$ ,  $1$  minus  $s t$  plus  $s$  squared  $T$  squared by factorial  $2$  and so on okay. Now let me simplify here itself  $1$  will go  $A$  by  $s$  in to  $s$  in to  $T$  then will find  $1$   $s$  will go  $s t$  squared and so on okay,  $A$  by  $s$ ,  $s t$  minus  $s$  squared  $T$  squared by factorial  $2$  and so on. So that gives me equal to  $AT$  minus sorry  $As t$  squared by factorial  $2$  plus  $A s$  squared  $T$  cube by factorial  $3$  and so on and this  $AT$  is  $1$  and this is  $AT$  in to  $T$ ,  $T$  tends

to 0 means this may be all 0 okay T tends to 0. So Laplace transform of delta T is unity okay what would be the Laplace transform of a periodic function. So let us take periodic function like this and so on so this is of magnitude 10 this is 1, 2, 3, 4 and so on.

(Refer Slide Time: 12:23)

$$= A T - \frac{A s T^2}{2!} + \frac{A s^2 T^3}{3!} \dots \quad T \rightarrow 0$$

$$= 1 - 0 + 0 \dots$$

$$\mathcal{L}\{\delta(t)\} = 1.$$

Graph of  $f(t)$  showing a periodic function with pulses of height 10 and width 1, repeated every 2 units.

$$F_1(s) = \frac{10}{s} (1 - e^{-s})$$

$$F(s) = F_1(s) + F_1(s)e^{-2s} + F_1(s)e^{-4s} + \dots$$

$$= F_1(s) [1 + e^{-2s} + e^{-4s} + \dots] = F_1(s) \frac{1}{1 - e^{-2s}}$$

So let us take first of all the the first block which is repeated after every period, so what is the Laplace transform of this first block, could you please tell me? It will be 10 by s in to 1 minus e to the power minus 1 s is it not tau is 1. Let us call it  $F_1(s)$  then what would be  $f(s)$  the same thing being repeated, so it would be  $F_1(s)$  plus you  $F_1(s)$  is repeated after 2 seconds so  $F_1(s)$  in to e to the power minus 2 s plus  $F_1(s)$  in to e to the power minus 4 s and so on up to infinity okay. So  $F_1(s)$  if I take common will be 1 plus e to the power minus 2 s plus e to the power minus 4 s and so on which means  $F_1(s)$  in to its a geometric series common ratio is e to the power minus 2 s. So 1 minus e to the power minus 2s, now  $F_1(s)$  is already obtained substitute that what you get  $F_1(s)$  equal to already known so just repeating it here therefore  $f(s)$  will be 10 by s 1 minus e to the power minus s, 1 minus e to the power minus 2 s okay e to the power 1 minus e s goes out, is that all right.

Let us take another example, periodic function once again like this this is 10 this is minus 10 okay 1, 2, 3, 4 and so on what would be the Laplace transform of this function? Can you write in 1 row, what is this?  $F_1(s)$  is see if it is a same block being repeated with alternate signs then there is no problem I can take this as the primary block, first initial block whose Laplace transform is already known then what will be  $f(s)$  will be  $F_1(s)$  in to 1 then minus e to the power it is shifted by 1 second and then its sign is changed so e to the power minus s plus e to the power minus 2 s minus e to the power minus 3 s plus e to the power minus 4 s and so on, is that all right. So how much is this common ratio is now minus e to the power minus s so it will be  $F_1(s)$  by 1 plus e to the power minus s, so that is 10 by s 1 minus e to the power minus s by 1 plus e to the power minus s very familiar form know what is it?

(Refer Slide Time: 15:17)

$$F_1(s) = \frac{10}{s} (1 - e^{-s})$$

$$F(s) = \frac{10(1 - e^{-s})}{s(1 - e^{-2s})} = \frac{10}{s(1 + e^{-s})}$$

$$F(s) = \frac{10}{s} (1 - e^{-s})$$

$$F(s) = F_1(s) [1 - e^{-s} + e^{-2s} - e^{-3s} + e^{-4s} - \dots]$$

$$= \frac{F_1(s)}{1 + e^{-s}} = \frac{10}{s} \frac{1 - e^{-s}}{1 + e^{-s}}$$

(Refer Slide Time: 18:11)

$$F(s) = \frac{10(1 - e^{-s})}{s(1 + e^{-s})} = \frac{10}{s(1 + e^{-s})}$$

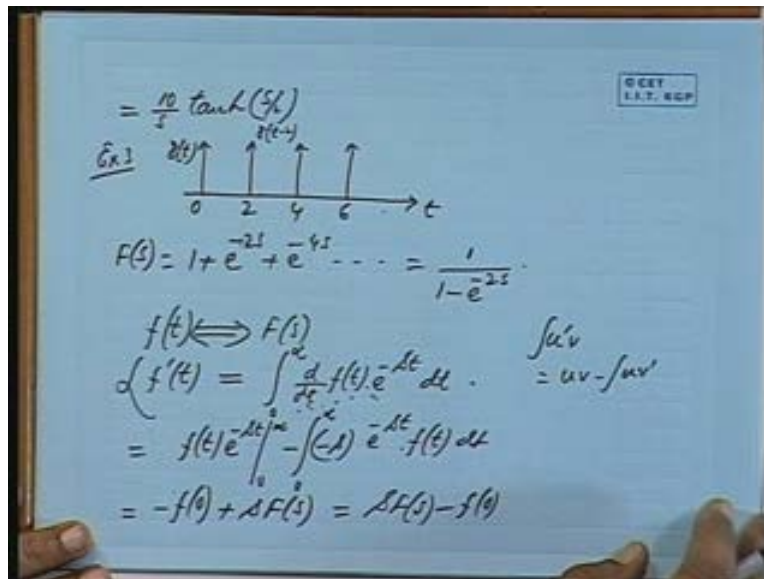
$$\frac{10}{s} (1 - e^{-s})$$

$$F_1(s) [1 - e^{-s} + e^{-2s} - e^{-3s} + e^{-4s} - \dots]$$

$$\frac{F_1(s)}{1 + e^{-s}} = \frac{10}{s} \frac{1 - e^{-s}}{1 + e^{-s}} = \frac{10}{s} \frac{e^{s/2} - e^{-s/2}}{e^{s/2} + e^{-s/2}}$$

Can I can I not take e to the power s by 2 multiplied throughout by e to the power s by 2, so it will become e to the power s by 2 minus e to the power minus s by 2 divided by e to the power s by 2 plus e to the power minus s by 2. So this will be not 2 j why it j I am not putting j omega and all that this is simply cos this is sin or sin and cos, so it will be 10 so it will be 10 by s tan hyperbolic s by 2.

(Refer Slide Time: 18:53)



Okay another example, if I have a pulse train 0, 2, 4, 6 and so on this is delta t, this is delta t minus 2 and so on. So what would be f s for this for the first one it is 1 for the next one it is 1 in to e to the power minus 2 s plus e to the power minus 4 s and so on so it will be 1 by 1 minus e to the power minus 2 s is that alright if I have alternately coming say plus 1 minus 1 delta t minus delta t plus delta t minus delta t, we shift then only change will be here okay plus, very good. Now you have another important property suppose Laplace transform of f (t) is given as f (s) what would be the Laplace transform of f dash t, where f dash denotes the first derivative what will it be. So by definition this will be 0 to infinity d dt f (t) e to the power minus st dt.

So if you integrate by parts alright this you can take as integration of u dashed v is u v minus integration u v dashed okay. So if I take this as this as v okay this can be differentiated this can be integrated so it will be f (t) if I integrate this in to v, e to the power minus s (t), 0 to infinity minus derivative of this will become minus s in to e to the power minus s t and integration of this is f (t) dt so this 1 will give me if I put infinity this becomes 0 minus if I put 0 it will be minus f (0) minus and minus will make it plus and integration of f (t) e to the power minus s t dt is f (s). So it is s in to f (s) where f (s) is the Laplace transform of original function f t minus f at 0 okay.

Now we shall derive the Laplace transform of integrals of functions from here. You can go from the basic definitions there is another way of looking at it. Let us take f (t) as some f<sub>0</sub> (t), df (t) by dt as df<sub>0</sub> by dt as some f<sub>1</sub> (t) therefore d square f (t) by dt squared is df<sub>1</sub> t by dt as f<sub>2</sub> (t) okay. So f<sub>0</sub> (t), f<sub>1</sub> (t), f<sub>2</sub> (t) mean they are all successive derivatives and corresponding Laplace transforms we denote as capital F<sub>0</sub> (s), F<sub>1</sub> (s), F<sub>2</sub> (s) and so on. So just now we have derived F<sub>1</sub> (s) is what derivative F<sub>1</sub> (t) its Laplace transform is F<sub>1</sub> (s), so it will be in terms of F<sub>0</sub> (s), f<sub>0</sub> (s) minus f<sub>0</sub>, f<sub>0</sub> at 0 is that alright.

(Refer Slide Time: 22:50)

$$\begin{aligned}
 f(t) &= f_0(t) && \Leftrightarrow F_0(s) \\
 \frac{df(t)}{dt} &= \frac{df_0(t)}{dt} = f_1(t) && \Leftrightarrow F_1(s) \\
 \frac{d^2f(t)}{dt^2} &= \frac{d^2f_0(t)}{dt^2} = f_2(t) && \Leftrightarrow F_2(s) \\
 F_1(s) &= sF_0(s) - f_0(0) \\
 F_2(s) &= sF_1(s) - f_1(0) \\
 &= s^2F_0(s) - sf_0(0) - f_1(0) \\
 \dots \\
 F_n(s) &= s^n F_0(s) - s^{n-1} f_0(0) - s^{n-2} f_1(0) - \dots - f_{n-1}(0)
 \end{aligned}$$

Similarly,  $F_2(s)$  will be  $s$  in to  $F_1(s)$  minus  $f_1$  at 0 if I substitute  $F_1(s)$  here it will be  $s$  square  $f_0(s)$  minus  $s$  minus  $f_{10}$  that means if I take the  $n$ th derivative and its Laplace transform will be in terms of the original Laplace transform  $s$  to the power  $n$  sorry this  $n$  minus  $s$  to the power  $n$  minus 1  $f_{00}$  minus  $s$  to the power  $n$  minus 2  $f_{10}$  and so on minus  $s$  to the power 0 when I am calculating  $f_n$  then  $f_n$  minus 1 0 okay is just by induction. We have gone downward from  $f_0$  derivative is  $f_1$  next derivative is  $f_2$  etcetera if you go up it will be integration all right. So these equations can be just a little bit of manipulation I can write on this side  $F_1 s$  is  $s f_0$ , so  $F_0(s)$  is how much in terms of  $F_1$  it will be  $F_1(s)$  by  $s$  plus  $f_{00}$  by  $s$  okay.

So sorry, if you are given the Laplace transform of  $F_1$  in terms of that that is  $f_1(t)$  is known in terms of that can you compute the Laplace transform of its high order function by high order I mean the integral lower order means derivative, so  $F_0 s$  that is the integral of  $f_1(t)$  what will be its Laplace transform. So  $F_0(s)$  will be the Laplace transform of the original function divided by  $s$ . Now plus what is  $f_{00}$  the integral evaluated at  $t$  equal to 0 what does it mean? So it is like this if you have a current expression for example, current is for example, in a capacitor  $1$  by  $c$  integral  $I dt$ ,  $0$  to  $t$  is the voltage at any instant if I take if I have to take the Laplace transform of this given the Laplace transform of  $I(t)$  as  $I(s)$  what will be  $v(s)$ ,  $v(s)$  is integral of  $I$  is it not. So if I have to compute  $v(s)$  it will be  $1$  by  $c$  in to integral of this means original Laplace transform divided by  $s$  then plus plus what is it  $I_0$  what is  $f_0$  it is integral of that so integral of  $I dt$  okay divided by  $s$  and this is to be evaluated at 0 from where do we start, so 0 minus to 0 plus or basically whatever has been there before the counting of time before  $t$  equal to 0 so that gives me integral of  $I dt$  is what charge  $q$ , is it not?



(Refer Slide Time: 25:42)

Handwritten mathematical derivations on a blue board:

$$f'(t) = f_0(t) \iff F_0(s)$$

$$f'(t) = \frac{df_1(t)}{dt} = f_1(t) \iff F_1(s)$$

$$f'(t) = \frac{df_2(t)}{dt} = f_2(t) \iff F_2(s)$$

$$f(t) = \int F_0(s) - f_0(t) \quad \left| \quad F_0(s) = \frac{f_1(s)}{s} + \frac{f_0(t)}{s} \right.$$

$$= \int F_1(s) - f_1(t)$$

$$= \int F_2(s) - f_2(t) - f_1(t)$$

$$= \int F_0(s) - \int F_1(s) - \int F_2(s) - \dots - \int F_n(s)$$

(Refer Slide Time: 26:58)

Handwritten mathematical derivations on a blue board:

$$\frac{1}{c} \int_0^t i dt \rightarrow v(t)$$

$$I(s) \quad V(s) = ? \quad \int i dt = q_0$$

$$\frac{1}{c} \left[ \frac{I(s)}{s} + \frac{\int i dt}{s} \right]$$

$$= \frac{1}{cs} I(s) + \frac{q_0}{cs} = V_0$$

•

So integral  $\int i dt$  means this is to be evaluated at  $t$  equal to 0 means the initial charge  $q_0$  alright mind you it is not initial value of  $i$  initial value of integration of  $i$ , when you take the integration so initial value of not  $f_1$   $f_0$  means integration of  $f_1$  alright. If the initial condition is given as 0 then there is no problem, so  $\frac{1}{cs}$  in to  $I(s)$  plus if I call it  $q_0$  by  $cs$ ,  $q_0$  by  $c$  is nothing but initial voltage across the capacitor, so  $V_0$  by  $s$  okay. So when we evaluate for example when we evaluate the voltage across a capacitor you are energizing by this that can be a resistance also you want to measure the voltage across this capacitor then its initial voltage plus at any instant if

current is  $I_t$  if you want to deal with this problem in the Laplace domain then  $I(s)$  in to 1 by  $c s$  this total sum will be the voltage across the capacitor, initial voltage across the capacitor plus the drop due to this is the total voltage at any instant.

(Refer Slide Time: 30:39)

$$\begin{aligned} \mathcal{L}^{-1}\left\{\frac{1}{s(s^2+\omega^2)}\right\} \\ \mathcal{L}^{-1}\left\{\frac{1}{s^2+\omega^2}\right\} &= \frac{\sin \omega t}{\omega} \\ \frac{1}{s(s^2+\omega^2)} &= \int_0^t \frac{\sin \omega t}{\omega} dt \\ &= \frac{1}{\omega^2} (1 - \cos \omega t) \end{aligned}$$

Now we will take up 1 or 2 small examples, then again will come to some other properties what would be Laplace inverse of say 1 by  $s$  in to  $s$  squared plus  $\omega$  squared 1 by  $s$  square plus  $\omega$  squared what is the Laplace inverse, if you remember this is  $\sin \omega t$  what is the Laplace transform of  $\sin \omega t$   $\omega$  by  $s$  square plus  $\omega$  squared, so  $\omega$  is not there so I will just write  $\sin \omega t$  by  $\omega$  okay and just now we have seen if the initial conditions are not given otherwise mention say it will be treated as 0. So what would be therefore Laplace inverse of 1 by  $s$  in to  $s$  squared plus  $\omega$  squared 1 by  $s$  in to  $s$  square plus  $\omega$  squared it will be integration 0 to  $t$ ,  $\sin \omega t$  by  $\omega$  sorry  $dt$ , is it not? So that will give me 1 minus  $\cos \omega t$  by  $\omega$  okay 1 comes because you are putting limit on  $\sin \cos \omega t$ . So this there is another  $\omega$  so while integrating you will get  $\omega$  squared  $\sin \omega t$  will give me another  $\omega$ . So 1 by  $\omega$  squared 1 minus  $\cos \omega t$  is that okay one may do it by partial fractions I can write this as  $A$  by  $s$  plus  $B$  by  $s$  square plus  $\omega$  squared. Let us see how much is  $A$  multiply by  $s$  put  $s$  equal to 0.

Let it be 1 by  $\omega$  squared how much is  $B$  multiply by  $s$  squared plus  $\omega$  squared divide by  $s$ , so will be and then put  $s$  square plus  $\omega$  squared equal to 0. So if I am multiply by  $s$  square plus  $\omega$  squared this will go on the left hand side will be 1 by  $s$  then this will be  $s$  square plus  $\omega$  squared in the numerator and here it will be  $B$  by  $s$  and then if I divide by  $s$  this will be  $s$  square this will be  $s$  square. So it will be 1 by  $s$  square on this side on this side it will be  $A$  in to  $s$  square plus  $\omega$  squared plus  $B$  and now, I am making  $s$  square plus  $\omega$  squared equal to 0.

(Refer Slide Time: 32:32)

Handwritten mathematical derivation on a blue board:

$$\frac{1}{s^2 + \omega^2} = \frac{\sin \omega t}{\omega}$$

$$\frac{1}{s(s^2 + \omega^2)} = \int_0^t \frac{\sin \omega t}{\omega} dt$$

$$= \frac{1}{\omega^2} (1 - \cos \omega t)$$

$$\frac{1}{s(s^2 + \omega^2)} = \frac{A}{s} + \frac{Bs}{s^2 + \omega^2} = \frac{1}{\omega^2} \left( \frac{1}{s} - \frac{Bs}{s^2 + \omega^2} \right)$$

$$A = \frac{1}{\omega^2}, \quad B = -\frac{1}{\omega^2}$$

(Refer Slide Time: 34:28)

Handwritten mathematical derivation on a blue board:

$$\frac{1}{s^2(s^2 + \omega^2)} = \left( \frac{1}{s^2} - \frac{1}{s^2 + \omega^2} \right) \frac{1}{\omega^2}$$

$$\Leftrightarrow \frac{1}{\omega^2} t \cos t - \frac{1}{\omega^2} \frac{\sin \omega t}{\omega}$$

$$F(s) = \int_0^{\infty} f(t) e^{-st} dt$$

$$\frac{dF(s)}{ds} = - \int_0^{\infty} t \cdot e^{-st} \cdot f(t) dt$$

$$= - \mathcal{L} \{ t f(t) \}$$

$$\mathcal{L} \{ t f(t) \} = - \frac{dF(s)}{ds}$$

So that means 1 by minus 1 by omega square all right that will be B. So this one will be 1 by omega squared in to 1 by s minus B means 1 by omega square. So s by s square plus omega square. Again, if you take the Laplace inverse it get the same result, so you can do it by either method alright making use of the property of integrating a function or making partial fractions. Similarly, you can solve for 1 by s square in to s square plus omega square. So if you know the result of this once again divide by s means once again an integral function, so you integrate this one again you will get the Laplace transform partial fraction will be simpler, is it not? will be 1

by  $s^2 - 1$  by  $s^2 + \omega^2$ . So that will give me  $s^2 + \omega^2$  square minus **minus**  $s^2$ .

So I will divide by  $\omega^2$  so corresponding inverse will be  $1/s^2$  is Laplace transform form of what ramp function  $t$  in to  $t$  okay and this  $1/s^2 + \omega^2$   $\sin \omega t$  by  $\omega$  so it will be  $1/\omega^2 t$   $t - 1/\omega^3 \sin \omega t$  okay. I need not write  $u(t)$  all the time I told you if you write  $t$  so long as you understand if it is a ramp function it is okay. Another important property will derive now so by definition we have  $f(t) e^{-st} dt$ . The common term is  $s$  common variable is  $s$  if I differentiate with respect to  $s$  what do you get if you differentiate with respect  $s$  it will be  $t$  in to  $e^{-st}$  with a negative sign  $f(t) dt$  and what is this Laplace transform of  $t$ ,  $f(t)$  with a negative sign.

(Refer Slide Time: 37:21)

The image shows a handwritten derivation on a blue board. The text is as follows:

$$\frac{d}{ds} \left( \frac{1}{(s+\beta)^2} \right) = ?$$

$$\frac{d}{ds} \left( \frac{\beta}{(s+\beta)^2} \right) \Rightarrow \frac{-2s\beta}{(s+\beta)^3}$$

$$\mathcal{L}\{t \sin \beta t\} = \frac{2\beta s}{(s+\beta)^3} \quad \frac{d}{ds} \left( \frac{1}{(s+\beta)^2} \right)$$

$$\mathcal{L}\{t \cos \beta t\} = -\frac{d}{ds} \left( \frac{s}{(s+\beta)^2} \right)$$

$$= -\frac{(s+\beta)^2 - 2s^2}{(s+\beta)^3} = \frac{s-\beta}{(s+\beta)^3}$$

So Laplace transform of  $t$  in to  $f(t)$  is  $-\frac{d}{ds} f(s)$  okay  $-\frac{d}{ds} f(s)$  by  $ds$  again we take a few examples, what will be Laplace inverse of this  $1/s^2 + B^2$  if I take the derivative with respect to  $s$  what is it, derivative of this  $1/s^2 + \beta^2$  **beta** squared yes, **could you please tell me  $2s$  minus  $2s$  okay**. So if I put a  $\beta$  here all right okay  $t$  into this is corresponding to  $\sin \beta t$  is it not  $t$  in to  $\sin \beta t$ ,  $t$  in to  $\sin \beta t$  Laplace transform of  $t$  in to  $\sin \beta t$  is how much then  $t$  in to  $\sin \beta t$ ,  $t$  in to  $f(t)$  is  $-\frac{d}{ds} f(s)$  by  $ds$  with a negative sign alright.

So this is corresponding to  $\sin \beta t$ , so  $t$  in to  $\sin \beta t$  should be equal to derivative of this with respect with an negative sign, so derivative is this. So this is nothing but  $2\beta s$  by  $s^2 + \beta^2$  whole square is it not is that alright  $t$  in to  $\cos \beta t$  what will the Laplace transform  $\cos \beta t$  is  $s/s^2 + \beta^2$  if you take the derivative of this how much is it? So  $-\frac{d}{ds}$  of  $s$  by  $s^2 + \beta^2$  how much is that

derivative of  $\frac{\beta}{s^2 + \beta^2}$  with respect to  $s$  is  $-\frac{2s\beta}{(s^2 + \beta^2)^2}$ .

So that gives me  $\frac{s^2 - \beta^2}{(s^2 + \beta^2)^2}$ . Okay the other 1 is  $t$  in  $\sin \beta t$ ,  $\beta s$  by  $(s^2 + \beta^2)^2$  okay. So I can make any combination  $t$  in to cosine  $\beta t$  plus minus  $t$  in to sine  $\beta t$ . Laplace transform of this will be  $\frac{s^2 - \beta^2}{(s^2 + \beta^2)^2}$  plus  $\frac{2\beta s}{(s^2 + \beta^2)^2}$  okay.

(Refer Slide Time: 40:47)

Handwritten derivations on a blue notepad:

$$\frac{d}{ds} \left( \frac{\beta}{s^2 + \beta^2} \right) \Rightarrow -\frac{2s\beta}{(s^2 + \beta^2)^2}$$

$$\mathcal{L}\{t \sin \beta t\} = \frac{2\beta s}{(s^2 + \beta^2)^2}$$

$$\mathcal{L}\{t \cos \beta t\} = -\frac{d}{ds} \left( \frac{s}{s^2 + \beta^2} \right)$$

$$= -\frac{(s^2 + \beta^2) - 2s^2}{(s^2 + \beta^2)^2} = \frac{s^2 - \beta^2}{(s^2 + \beta^2)^2}$$

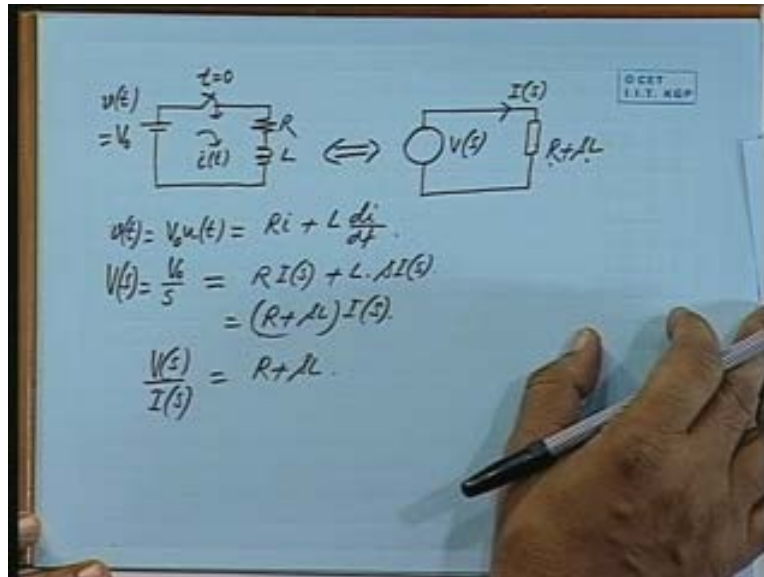
$$\mathcal{L}\{t \cos \beta t \pm t \sin \beta t\} = \frac{s^2 - \beta^2}{(s^2 + \beta^2)^2} \pm \frac{2\beta s}{(s^2 + \beta^2)^2}$$

Other notes on the notepad:

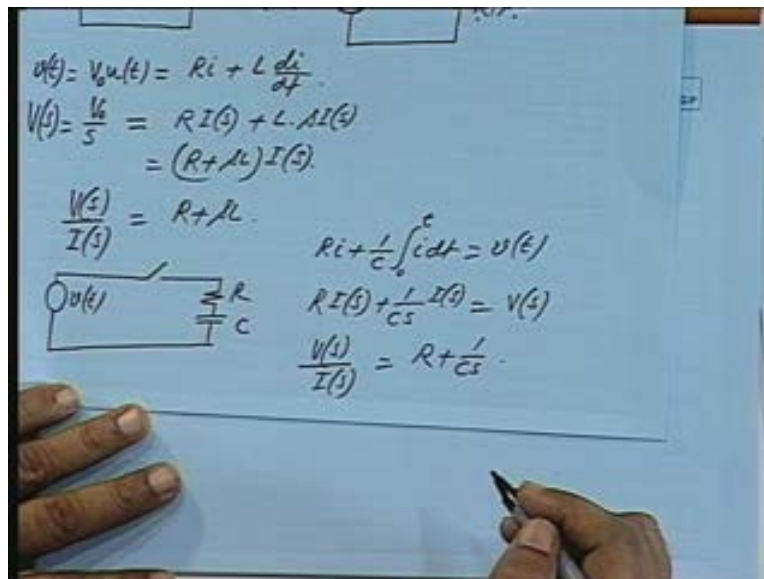
- $\frac{d}{ds} \left( \frac{1}{s^2 + \beta^2} \right) \Leftrightarrow \mathcal{L}\{\beta t\}$
- $\frac{\beta}{s^2 + \beta^2} \Rightarrow \sin \beta t$

Now you already know  $\beta$  by  $s^2 + \beta^2$  corresponds to  $\sin \beta t$  and this one corresponds to  $\cos \beta t$ . So by making manipulations there are 4 results now 1 is this plus this this minus this so I have already got 4 such Laplace transforms by manipulating you can find out the Laplace transform of this because I have break it up in to see, if I make partial fractions or manipulate these I can get the inverse of this I leave it as an exercise you do it yourself okay, not very difficult. Next we take up application of Laplace transform in simple network problems, suppose you consider a simple RL circuit there is a DC source of voltage  $V$  and the switch is put on at  $t$  equal to 0 initially this is uncharged what would be the expression for the current  $i(t)$  you write  $V$  you are suddenly switching on. So it will be  $v u(t)$ ,  $V$  is a magnitude of voltage you are applying a step voltage and if  $I$  is the current when it is  $Ri$ , I am not writing  $i(t)$  it is understood  $i$  is a time varying quantity  $I \frac{di}{dt}$  if I take Laplace transform on both sides it will be  $V$  by  $s$  on this side it will be  $r$  is plus  $I \frac{di}{dt}$  and what is the Laplace transform of the derivative function  $s$  times  $I(s)$  minus  $I(0)$  if I assume the initial condition to be 0 then  $I_0$  is 0 so that will not come.

(Refer Slide Time: 42:11)



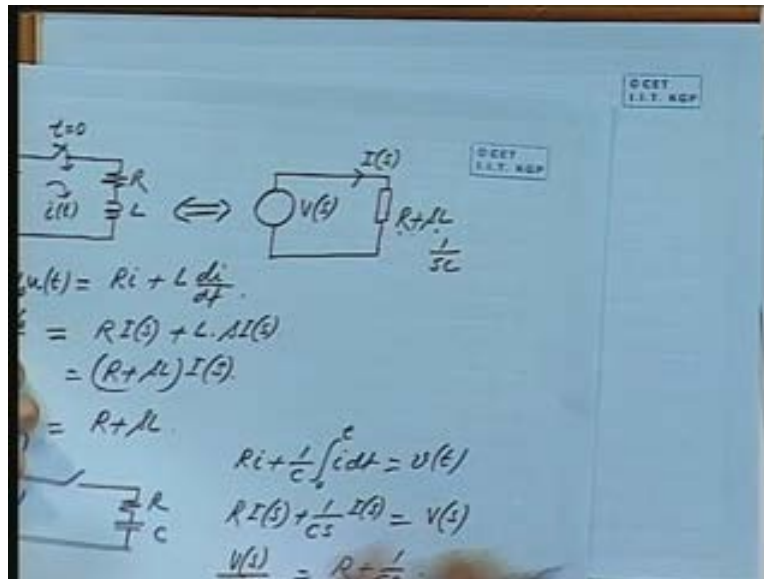
(Refer Slide Time: 45:45)



So it is  $R + sL$  in to  $I(s)$  alright so like you are having voltage equal to impedance in to current in simple AC circuit. Here, also the applied voltage in the transform domain this is equal to  $R + sL$  in to current in the time domain Laplace domain. So in the transform domain I will write  $V$  is 0 excuse me  $v_0$  mean a specific value and  $v(s)$  is the Laplace transform of  $v(t)$  in this case it is  $V_0$  in to  $u(t)$  okay. So  $v(s)$  by  $I(s)$  that means in the transform domain voltage function by current function is equal to  $R + sL$  in the transform domain. So a resistance therefore these circuit I will write as  $v$  s the voltage source and an impedance which is  $R + sL$

and the current is  $I(s)$  so an impedance is given by  $R$  plus  $sL$  for  $R$  and  $L$  in a similar manner we can show if there is an if the there is a capacitance, if there is a voltage source here  $v(t)$   $Ri$  plus  $1$  by  $c$  integral  $I dt$  is equal to the applied voltage if I take the Laplace transform will be  $R$  in to  $I s$  plus  $1$  over  $c$  in to  $s$  in to  $I(s)$  once again if I assume the capacitor to be initially uncharged then that  $q_0$  by  $c$  ( $s$ ) will be  $0$ , so this side it will be  $v(s)$  or  $v(s)$  by  $I(s)$  will turn out to be  $R$  plus  $1$  by  $c$  ( $s$ ).

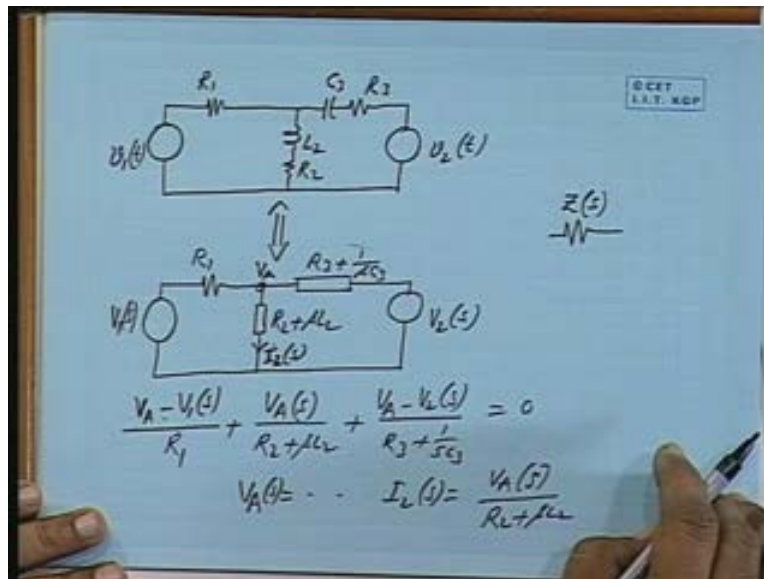
(Refer Slide Time: 46:46)



So if it is an inductive element it will be  $R$  plus  $sL$  if there is a capacitive element present here then it will be equivalently represented by  $1$  by  $sc$ . So if you have RLC wherever there are resistances you replace it by you return it that is resistance should be returned as it is inductance will be replace by  $sL$  capacitance will be replace by  $1$  by  $s c$  and in all network configurations next part will be using the transform domain variables are  $s_{11}$  by  $s c$  sorry if you do this then any network problem can be solved. Let us take 1 or 2 more examples, you have a source here a resistance here. Okay therefore, we shall replace this entire circuit by an equivalent  $v_1(s)$ ,  $R_1$  you can show it like this or a symbol like this can also be used for a generalized impedance  $z(s)$  it can be  $R$  it can be  $1$  by  $sL$ ,  $1$  by  $sc$  it can be  $sL$  or any combinations this is  $R_2$  plus  $sL_2$ .

Similarly, this  $1$  is  $R_3$  plus  $1$  by  $sc_3$  and this side you have got  $V_2(s)$ . Now for calculating the current through any branch or potential at any node you can apply supposition theorem, nodal analysis, theros theorem, any theorem you want by treating these as simple algebraic operational impedances all right, treating  $s$  as an algebraic operator and treat this as it is and then when you get the expression for current  $I_s$  take the inverse you get the current  $I(t)$ . So if I am interested in the current through this  $I_2$ ,  $I_2(s)$  what will be  $I_2(s)$ , we can go for nodal analysis.

(Refer Slide Time: 47:46)



So let this potential be  $V$  okay  $V_A$ , so  $V_A$  minus  $V_1(s)$  by  $R_1$  plus  $V_A$  by  $R_2$  plus  $sL_2$  plus  $V_A$  minus  $V_2$  s by  $R_3$  plus  $1$  by  $sC_3$  equal to  $0$ . Now  $V_A$ ,  $V_1(s)$  and  $V_2(s)$  will be given to you it may be stay function ramp function or any other function so put those values solve for  $V_A$ ,  $V_A$  means  $V_A(s)$  okay once you know  $V_A(s)$ , then  $V_A(s)$  divided by  $R_2$  plus  $sL_2$  therefore  $I_2(s)$  will be  $V_A(s)$  by  $R_2$  plus  $sL_2$  okay. So far you have got the solutions in terms of  $s$  now you make partial fractions expand it partial fractions and then take the inverse you will get the current  $I_2$  okay while making partial fractions sometimes you may come across multiple poles at a particular point say

You want to find out the partial fractions of say the denominator is  $s$  plus  $\alpha_1$  in to  $s$  plus  $\alpha_2$  and so on but  $s$  plus some  $\alpha_k$  is having  $n$  number of poles that is  $3, 4$  multiple roots then this can be written as  $A_1$  by  $s$  plus  $\alpha_1$  plus  $A_2$  by  $s$  plus  $\alpha_2$  and so on valuation of  $A_1, A_2$  etcetera is very simple straight forward multiplied by  $s$  plus  $\alpha_1$  then make  $s, s$  plus  $\alpha_1$  equal to  $0$  will get  $A_1$  and so on but this  $1$  will be some  $ak_1$  by  $s$  plus  $\alpha_k$  plus  $ak_2$  by  $s$  plus  $\alpha_k$  minus  $1$  and so on. Evaluation of this becomes a little difficult straight forward if you apply that technique then you will be able to get only a  $k_1$ .

So for evaluation of a  $k_1$  you multiplied by  $s$  plus  $\alpha_k$  to the power  $n$  both sides, so  $f(s)$  is multiplied by this and then put  $s$  equal to minus  $\alpha_k$  okay. This product let me call it some  $q(s)$  evaluated minus  $\alpha_k$  then  $A_{k2}$  will be derivative of this  $q$  s and then evaluate at  $s$  equal to minus  $\alpha_k$  similarly  $A_{k3}$  will be  $1$  by factorial  $2$ ,  $d^2$  square  $q(s)$  by  $ds^2$  square, second derivative we take and then substitute  $s$  equal to minus  $\alpha_k$  and so on third derivative it will be factorial  $3$  and so on so you keep on taking the derivatives of this product and then evaluate the function at  $s$  equal to minus  $\alpha_k$  that will give you successive residues. Once you know that you can expand you can find out the inverse okay so will stop here for today. Thank you



very much you try some other problems at home and will take up some numerical problems in the next class.

(Refer Slide Time: 51:46)

$$F(s) = \frac{f(s)}{(s+\alpha_1)(s+\alpha_2) \dots (s+\alpha_k)^n}$$

$$= \frac{A_1}{s+\alpha_1} + \frac{A_2}{s+\alpha_2} + \frac{A_{k1}}{(s+\alpha_k)^1} + \frac{A_{k2}}{(s+\alpha_k)^2} \dots$$

$$A_{k1} = \frac{(s+\alpha_k)^n \cdot f(s)}{(s+\alpha_k)^{n-1}} \Big|_{s=-\alpha_k}$$

$$= Q(s) \Big|_{s=-\alpha_k}$$

$$A_{k2} = \frac{d}{ds} Q(s) \Big|_{s=-\alpha_k}, \quad A_{k2} = \frac{1}{2!} \frac{d^2 Q(s)}{ds^2} \Big|_{s=-\alpha_k}$$

**Preview**

**Lecture - 09**

**Tutorial on Laplace Transform-Application to Circuit Problems**

(Refer Slide Time: 55:08)

Ex 1 TUTORIAL PROBLEMS

$$\frac{1}{2a^2} [\sinh(at) - \sin(at)]$$

$$= \frac{1}{2a^2} \left[ \frac{e^{at} - e^{-at}}{2} - \sin at \right]$$

$$= \frac{1}{4a^2} \left[ \frac{1}{s-a} - \frac{1}{s+a} \right] - \frac{1}{2a^2} \cdot \frac{a}{s^2+a^2}$$

$$= \frac{1}{2a} \cdot \frac{1}{s^2+a^2} - \frac{1}{2a} \cdot \frac{1}{s^2+a^2}$$

$$= \frac{1}{2a} \left[ \frac{1}{s-a} - \frac{1}{s+a} \right] = \frac{a}{s^2-a^2}$$

Okay Good afternoon friends, will continue with Laplace transform. Today will have some tutorial session, will be solving some problems employing Laplace transform. Let us take simple functions what be the Laplace transform of  $1 - \sin at$  by  $2a$  squared  $\sin at$  minus  $\sin at$ , what be the Laplace transform of this? So this  $1 - \sin at$  can be written as  $1 - \sin at$  what is  $\sin at$  e to the power  $at$  minus e to the power  $-at$  divided by  $2$  minus  $\sin at$ . So it is pretty simple  $1 - \sin at$  will be  $1 - \frac{e^{at} - e^{-at}}{2}$  is that all right minus  $1 - \sin at$  squared  $\sin at$  will give you  $1 - \sin at$  divided by  $s^2 + a^2$ , is that okay? So if I add these 2 it will be  $s + a$  minus  $s + a$  so twice  $a$ , so that will be  $1 - \sin at$  into  $s^2 - a^2$  minus here  $1 - \sin at$  goes, so  $1 - \sin at$  by  $s^2 + a^2$ .

(Refer Slide Time: 57:31)

The image shows handwritten mathematical work on a blue board. The work is divided into two main sections. The top section deals with Laplace transforms of shifted functions. It starts with  $V_p(t)$  and shows the decomposition of a function into two parts:  $\frac{a}{10} [u(t) - e^{-10t}] - \frac{a}{10} [u(t-5) - e^{-10(t-5)} u(t-5)]$  and  $+\frac{a}{10} [u(t-10) - e^{-10(t-10)} u(t-10)]$ . Below this, it defines  $V_L(t) = V_p(t) - V_p(t-10)u(t-10) + V_p(t+20)u(t-20)$ . The bottom section, labeled 'Ex 4', shows a differential equation  $\frac{d^2 y(t)}{dt^2} + 3 \frac{dy(t)}{dt} + 2y(t) = 5u(t)$ . This is transformed into the s-domain as  $s^2 Y(s) + 3sY(s) + 2Y(s) = \frac{5}{s}$ . The equation is then rearranged to  $(s^2 + 3s + 2)Y(s) = \frac{5}{s}$  and solved for  $Y(s)$  using partial fraction decomposition:  $Y(s) = \frac{5}{s(s+1)(s+2)} = \frac{A}{s} + \frac{B}{s+1} + \frac{C}{s+2}$ .

So  $1 - \sin at$  I can take out it will become  $1 - \sin at$  by  $s^2 - a^2$  minus  $1 - \sin at$  by  $s^2 + a^2$  plus  $a^2$  and that gives me  $s^2 - a^2 + a^2$  minus  $1 - \sin at$  by  $s^2 + a^2$ . So twice  $s^2$  twice  $a$  will go, so  $a$  by  $s^2 - a^2$  minus  $a$  to the power  $4$  okay plus  $2$  times  $y(s)$  is equal to  $5$  by  $s$  so it is  $s^2 + 3s + 2$  in to  $ys$  which is  $5$  by  $s$ . So how much is  $y(s)$   $5$  divided by  $s$  in to  $s^2 + 3s + 2$  can be written as  $s + 1$  in to  $s + 2$ . Okay I can write it as  $A$  by  $s$  plus  $B$  by  $s + 1$  plus  $C$  by  $s + 2$  and taking the inverse of these 3, I will get the final  $y(t)$  as  $A$  into  $u(t)$  plus  $B$  in to  $u(t)$  the power minus  $t$  plus  $C$  in to  $u(t)$  the power minus  $2t$  where  $A B C$  residues can be calculated from here by standard partial fraction technique, okay. So we will stop here for today we will take it up in the next class with further examples using Laplace transform for network problems. Thank you very much.