

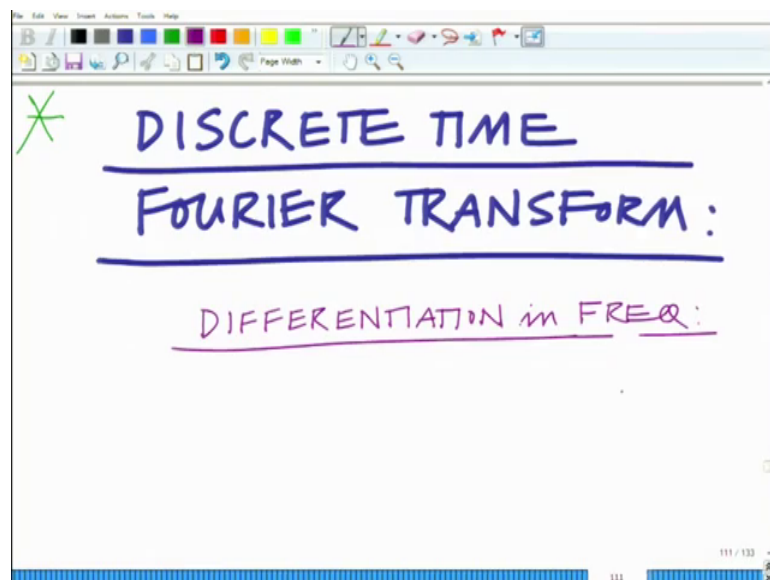
Principles of Signals and Systems
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Lecture – 58

Discrete Time Fourier Transform: Properties of DTFT - Differentiation in Frequency, Difference in Time, Convolution, Multiplication, Parseval's Relation

Hello welcome to another module in this massive open online course. So, we are looking at the Fourier analysis for discrete time a, periodic signals through the DTFT the discrete time Fourier transform alright. And we are looking at a, at the properties of the DTFT.

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So, let us continue our discussion on that so we are looking at the discrete time Fourier transform. And the next property so we have looked at several properties, we have looked at the duality, the next property that we want to look at is differentiation in frequency of course, in time it is a discrete signal. So, we cannot talk about differentiation in time so this is differentiation in frequency what happens when you differentiate in frequency.

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DIFFERENTIATION WITH RESPECT TO Ω

$$x(n) \leftrightarrow X(\Omega)$$
$$X(\Omega) = \sum_{n=-\infty}^{\infty} x(n)e^{-j\Omega n}$$
$$\frac{dX(\Omega)}{d\Omega} = \sum_{n=-\infty}^{\infty} x(n) \frac{d}{d\Omega} e^{-j\Omega n}$$

So, we have x of n let us say as the DTFT, capital X of ω alright which means basically your X of ω equals summation n equals minus infinity to infinity x n e raised to minus j ω n . Then if you differentiate this d x ω or d ω that is n equals minus infinity to infinity if you take the differentiation sign inside x n is a constant that is d of d or different derivative d or d ω of e raised to minus j ω n with respect to ω .

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$$X(\Omega) = \sum_{n=-\infty}^{\infty} x(n)e^{-j\Omega n}$$
$$\frac{dX(\Omega)}{d\Omega} = \sum_{n=-\infty}^{\infty} x(n) \frac{d}{d\Omega} e^{-j\Omega n}$$
$$= \sum_{n=-\infty}^{\infty} x(n) (-jn) e^{-j\Omega n}$$
$$\frac{dX(\Omega)}{d\Omega} = -j \sum_{n=-\infty}^{\infty} n x(n) e^{-j\Omega n}$$

DTFT of $n x(n)$.

So, or let me just write it that is d over $d\omega$ e raised to minus $j\omega n$ which is summation n equals minus infinity to infinity $x[n] e^{-j\omega n}$ which is minus j summation. So, d over $d\omega$ that is n equals minus infinity to infinity $x[n] e^{-j\omega n}$. So, this is basically if you look at it this is basically the DTFT of $x[n]$ of the signal $x[n]$ ok.

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The image shows a whiteboard with handwritten mathematical derivations. At the top, there are some annotations: $d\omega$ and $n = \omega$. Below these, the text reads "DTFT of $nx[n]$ ". The main derivation is:

$$\Rightarrow j \frac{dX(\omega)}{d\omega} = \sum_{n=-\infty}^{\infty} nx[n] e^{-j\omega n}$$

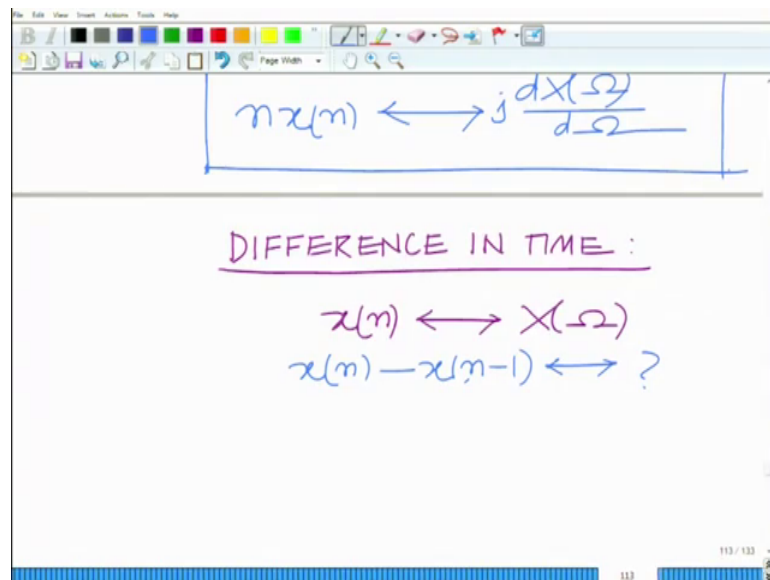
Below this, a boxed relationship is shown:

$$nx[n] \longleftrightarrow j \frac{dX(\omega)}{d\omega}$$

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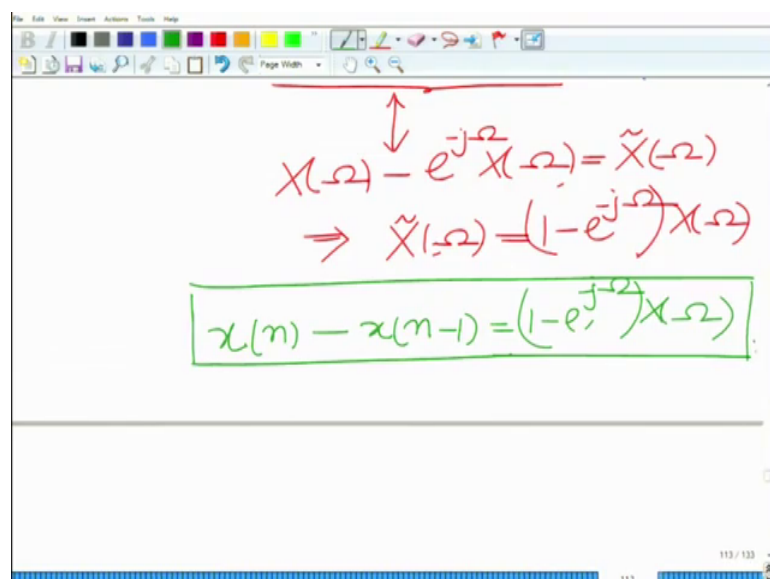
And therefore, this implies if you bring the j to the other side this means $j d$ over $d\omega$ $X(\omega)$ equals summation n equals minus infinity to infinity $x[n] e^{-j\omega n}$ ok. And therefore, we can conclude that $nx[n]$ is basically has the DTFT which is $j d$ over $d\omega$ $X(\omega)$ alright.

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So, next property is the differencing property since we cannot differentiate in time one can difference the differencing in time which means if x of n has the d t f t x of omega what can we say about the difference signal x n minus x n minus one. What can we say about this is the difference signal or the differential signal alright x n the successive differences sometimes also called as a differential signal alright so that is your basically your delta x in time.

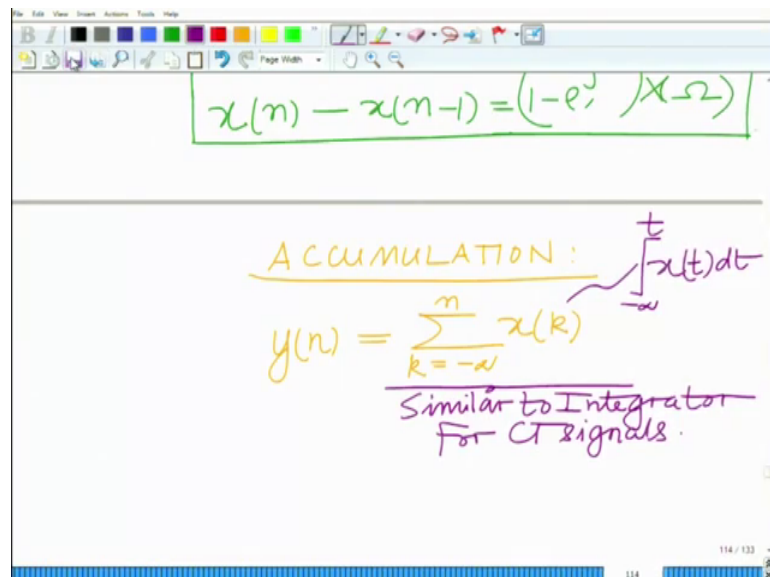
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And now if you look at this if you take the Fourier transform of this you can see it is very simple this is x of ω minus the Fourier transform of $x[n-1]$ that is $e^{-j\omega}$ raised to that is the delay in time leads to modulation in frequency $e^{-j\omega}$ raised to x of ω equals $X(\omega)$ which implies $X(\omega)$ that is the DTFT of $x[n-1]$ equals $(1 - e^{-j\omega}) X(\omega)$.

So, basically $x[n]$ minus $x[n-1]$ so if you look at this you have $x[n] - x[n-1]$ has the discrete time Fourier transform $(1 - e^{-j\omega}) X(\omega)$ so this is basically the discrete time Fourier transform.

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Let us now look at another property. So, all these are small properties, but fairly useful in the manipulation and analysis of DTFT of signals a lot of these properties can be used together to derive the final DTFT or to analyze the system or signal under consideration. So, let us consider the accumulation n equals or k equals minus infinity to n that is you are accumulating all the samples of the signal $x[k]$ until sample n .

You can see this is very simpler similar to the integrator which basically integrates the input signal $x(t)$ alright in analog time the analog alright the corresponding system in analog for an analog signal or for a continuous time signal is an integrator ok. So, basically it integrates or accumulates the signal at a certain time in the discrete time we are basically accumulating all the signal samples.

So, basically if you look at the continuous time analog that is minus infinity to t x t d t. So, this is similar to the integrator, integrator similar to the integrator for continuous time signals alright.

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$$y(n) = \sum_{k=-\infty}^n x(k)$$

Similar to Integrator for CT signals.

$$\sum_{k=-\infty}^n x(k) \longleftrightarrow \pi X(0) \delta(\omega) + \frac{X(\omega)}{1 - e^{-j\omega}}$$

And now the Fourier transform of this is given as pi x of 0 you can show that pi x of 0 into delta omega plus x of omega over 1 minus e power minus j omega, so that is basically your accumulation that is your integral n equals that is your summation k equals minus infinity to n x k that has the DTFT that is given as pi x 0 delta omega plus x of omega, divided by 1 over 1 x of omega divided by 1 minus e raised to minus j omega alright.

So, that is basically the result that is basically the result for accumulation of a discrete time signal ok. Let us now proceed to another important property that is a convolution because we have seen this several times our convolution is always a very important relation very important property in analysis of signals and systems because the convolution describes the output of a linear time invariant system to any input either continuous time or discrete time ok.

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CONVOLUTION:

$$y(n) = x_1(n) * x_2(n)$$
$$= \sum_{m=-\infty}^{\infty} x_1(m) x_2(n-m)$$

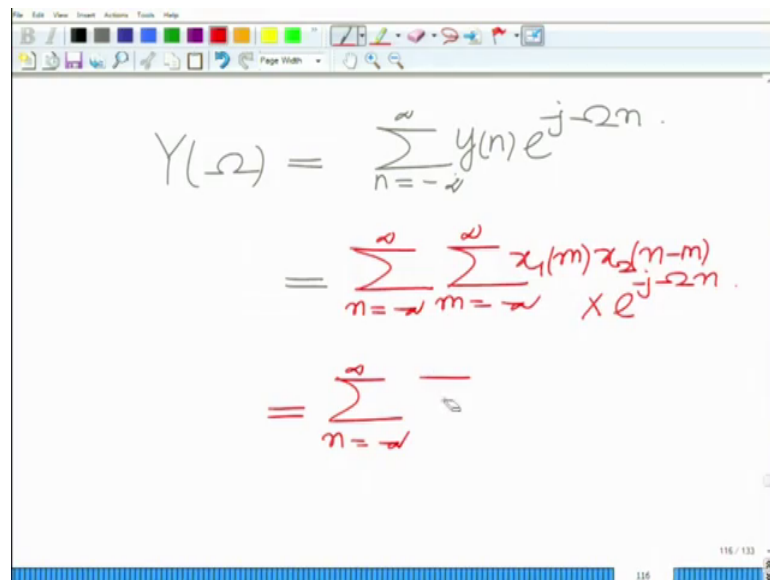
So, the next property that we want to look at is basically the convolution and now the convolution of 2 discrete time aperiodic signals. And convolution is always very important as I said because it is very closely related to the properties of LTI systems ok. So, $y(n)$ which is the convolution of $x_1(n)$ into $x_2(n)$ which means this is equal to the summation m equals minus infinity to infinity $x_1(m) x_2(n - m)$.

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$$y(n) = x_1(n) * x_2(n)$$
$$= \sum_{m=-\infty}^{\infty} x_1(m) x_2(n-m)$$
$$x_1(n) \longleftrightarrow X_1(-\omega)$$
$$x_2(n) \longleftrightarrow X_2(-\omega)$$

And now we want to find what is the DTFT of $y(n)$ given of course the DTFT's of $x_1(n)$ has a DTFT that is $X_1(\omega)$ and $x_2(n)$ has the DTFT that is $X_2(\omega)$.

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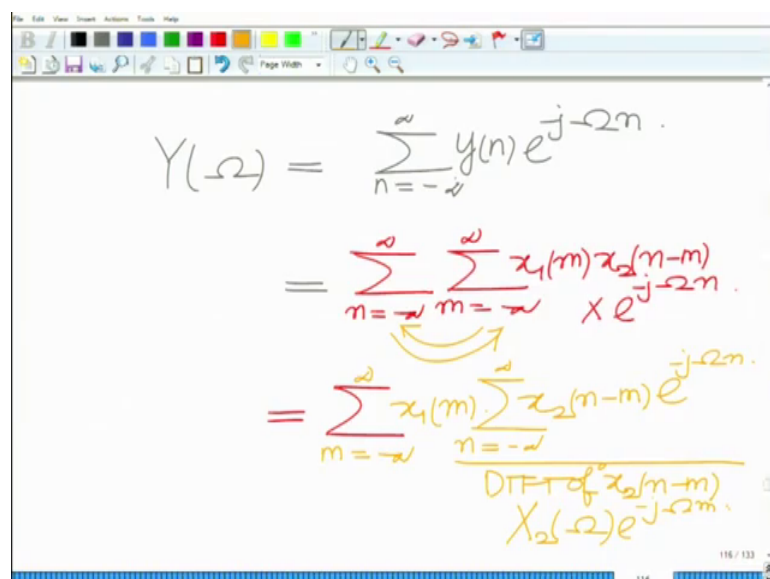
$$Y(\omega) = \sum_{n=-\infty}^{\infty} y(n) e^{j\omega n}$$

$$= \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} x_1(m) x_2(n-m) e^{j\omega n}$$

$$= \sum_{n=-\infty}^{\infty} -$$

And now you can see this DTFT Y of omega can be obtained as follows. So, Y of omega equals summation that is y of n, n equals minus infinity to infinity y of n e raised to minus j omega n which is basically I can always write this as summation n equals minus infinity to infinity substitute the expression for y of n that is m equals minus infinity to infinity x 1 m x 2 n minus m into e raised to minus j omega n. And now I can always write this as summation n equals minus infinity to infinity summation.

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$$Y(\omega) = \sum_{n=-\infty}^{\infty} y(n) e^{j\omega n}$$

$$= \sum_{n=-\infty}^{\infty} \sum_{m=-\infty}^{\infty} x_1(m) x_2(n-m) e^{j\omega n}$$

$$= \sum_{m=-\infty}^{\infty} x_1(m) \sum_{n=-\infty}^{\infty} x_2(n-m) e^{j\omega n}$$

DTFT of $x_2(n-m)$
 $X_2(\omega) e^{j\omega n}$

Now I can bring I can interchange the summation with respect to n and m so interchanging the summation I am going to describe this by this arrow. So, first you have the summation m equals minus infinity to infinity then x 1 m will come out because it depends only on m n equals minus infinity to infinity x 2 n minus m e raised to minus j omega n this is DTFT of x 2 n minus m that is x 2 n delayed by m samples; x 2 n delayed by m samples. Hence the corresponding DTFT is naturally x 2 omega e raised to minus j omega m that is modulation in frequency ok.

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The image shows a whiteboard with handwritten mathematical equations. At the top, the expression $X_2(-\Omega)e^{j\Omega m}$ is written in orange. Below it, the derivation proceeds as follows:

$$= \sum_{m=-\infty}^{\infty} x_1(m) e^{j\Omega m} \cdot X_2(-\Omega)$$

$$= X_2(-\Omega) \sum_{m=-\infty}^{\infty} x_1(m) e^{j\Omega m} = X_2(-\Omega) X_1(\Omega)$$

The final result is underlined in red. The whiteboard interface includes a toolbar at the top and a status bar at the bottom showing '117 / 133'.

So, that basically gives me summation m equals minus infinity to infinity of x 1 m e raised to minus j omega m into x 2 omega and now if you look at x 1 m into e raised to so I can write this as x 2 omega m equals minus infinity to infinity x 1 m e raised to minus j omega which is nothing but basically x 1 omega.

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$$X_1(\Omega)$$
$$Y(\Omega) = X_1(\Omega)X_2(\Omega)$$
$$x_1(n) * x_2(n) \leftrightarrow X_1(\Omega)X_2(\Omega)$$

And therefore, what we get is y of ω which is a convolution y n which is the convolution of x $n \times 1$ and x 2 n has the Fourier DTFT that is x 1 ω into x of ω . So, the net result is that x 1 n convolved with x 2 n has the DTFT x 1 ω into x 2 ω . So, convolution in time implies multiplication and frequency similar to what we have seen several times if equation of 2 discrete time aperiodic signals leads to a multiplication of their discrete time Fourier transform.

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$$x_1(n) * x_2(n) \leftrightarrow X_1(\Omega)X_2(\Omega)$$

CONVOLUTION IN TIME ↔ MULTIPLICATION IN FREQ DOMAIN

So, basically I can summarize this as convolution in time multiplication in the multiplication in the frequency domain ok. So, when you convolved to signals in time the corresponding DTFT's correct in these case are basically convolved in the frequency alright.

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MULTIPLICATION:

$$x_1(n) \cdot x_2(n) \longleftrightarrow \frac{1}{2\pi} X_1(\omega) \otimes X_2(\omega)$$

$$= \frac{1}{2\pi} \int_{-\pi}^{\pi} X_1(\theta) X_2(\omega - \theta) d\theta$$

Any period 2π

Let us look at similarly for multiplication in time again the dual of that is basically it have it has convolution in the; similarly let us consider multiplication in time that is if you look at 2 signals x 1 n into x 2 n.

Now remember the x 1 omegas are periodic hence this is the periodic convolution 1 over 2 pi x 1 omega periodically convolved with because remember the net DTFT, also has to be periodic correct when you convolve to a periodic thing remember the DTFT, is always a periodic signal so you cannot perform a general convolution, but you have to convolve in such a way that the resulting output is periodic and that is given by the periodic convolution ok.

So, this is basically the periodic convolution of x 1 the periodic convolution of x 1 omega with x 2 omega which is basically 1 over 2 pi over 2 pi x 1 theta x 2 omega minus theta d theta I am sorry this is d theta and this is for any contiguous, this is for any period 2 pi, this is a periodic convolution. This is over this is over any period 2 pi ok.

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The image shows a whiteboard with handwritten mathematical notes. At the top, it says "If $x(n) = \text{Real}$ ". A red arrow points from "Real" to the word "Even Component" written in red. Below this, the equation $x(n) = x_e(n) + x_o(n)$ is written. A red arrow points from $x_o(n)$ to the word "odd Component" written in red. At the bottom, the equation $x(n) \longleftrightarrow A(\omega) + jB(\omega)$ is written in blue.

Now some other properties if $x(n)$ is real let us consider a real signal. Then I can always express a real signal as the sum of even and odd components so this is the even component $x_e(n)$ the even component $x_o(n)$, this is the odd component we have already always we have seen that you can always do this that is $x_e(n)$ and the even component of a signal is $x(n) + x(-n)$ by 2, the odd component is $x(n) - x(-n)$ by 2 we can always do this for any real signal ok.

And now further if I can express the DTFT of $x(n)$ as $A(\omega) + jB(\omega)$. So, in general the DTFT of $x(n)$ is complex you can express this as $A(\omega) + jB(\omega)$.

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The image shows a whiteboard with handwritten mathematical notes. At the top, the signal $x(n)$ is expressed as the sum of its even component $x_e(n)$ and its odd component $x_o(n)$. Below this, the DTFT of $x(n)$ is given as $A(\omega) + jB(\omega)$. The even component $x_e(n)$ is shown to have a DTFT that is the real part of $A(\omega)$, and the odd component $x_o(n)$ is shown to have a DTFT that is purely imaginary, specifically $jB(\omega)$.

$$x(n) = x_e(n) + x_o(n)$$

$= \text{odd component}$

$$x(n) \longleftrightarrow A(\omega) + jB(\omega)$$
$$\Rightarrow x_e(n) \longleftrightarrow \frac{A(\omega)}{2} = \text{Real Part of DTFT } X(\omega)$$
$$x_o(n) \longleftrightarrow \frac{jB(\omega)}{2} = \text{Purely Imaginary}$$

Then it can be shown that and this is something that can be shown that is x_n has the DTFT that corresponds to the real part of the DTFT of x_n that is the even component has the DTFT a ω that is real part of the DTFT x of ω . And similarly as you can expect the odd component has the DTFT that is given by j times b ω ok.

So, the even component has a DTFT that is a times ω that is a real part. And the odd component as there is so the DTFT of the real part even part of the purely real signal you can see is basically real and the DTFT of the odd part odd component of this real signal is purely imaginary you can see that j times B of ω this is purely imaginary ok, this is purely imaginary.

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The image shows a whiteboard with handwritten mathematical derivations. At the top, the equation $X^*(-\Omega) = X(\Omega)$ is written. Below it, the text "FOR ANY REAL SIGNAL $x(n)$ " is written. The next line shows the derivation: $\Rightarrow \frac{(A(\Omega) + jB(\Omega))}{X(\Omega)} = (A(-\Omega) + jB(-\Omega))^*$. The whiteboard interface includes a toolbar at the top and a status bar at the bottom right showing "120 / 133".

$$X^*(-\Omega) = X(\Omega)$$

FOR ANY REAL SIGNAL $x(n)$

$$\Rightarrow \frac{(A(\Omega) + jB(\Omega))}{X(\Omega)} = (A(-\Omega) + jB(-\Omega))^*$$

And now further since we have a real signal recall that for a real signal x of conjugate of minus omega equals x of omega for a real signal x_n where x omega is the DTFT of this is true for any remember this is true for any real signal x_n .

So, this implies that if I take the DTFT a of omega plus j B of omega this is x of omega this must be equal to x of minus omega conjugate so a minus omega j b minus omega conjugate.

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The image shows a whiteboard with handwritten mathematical derivations. The first line shows the equation: $\Rightarrow A(\Omega) + jB(\Omega) = A(-\Omega) - jB(-\Omega)$. Below this, the text "Equating real, imaginary parts," is written. The next line shows the equation: $A(\Omega) = A(-\Omega)$ with the text "Even Function" written below it. The whiteboard interface includes a toolbar at the top and a status bar at the bottom right showing "121 / 133".

$$\Rightarrow A(\Omega) + jB(\Omega) = A(-\Omega) - jB(-\Omega)$$

Equating real, imaginary parts,

$$A(\Omega) = A(-\Omega)$$

Even Function

So, this must be equal to a of minus omega plus j b minus omega conjugate and this implies that A of omega plus j b minus omega equals well a of minus omega minus j b of minus omega. Now equating the real and imaginary parts equating real comma imaginary parts we have A of omega equals A of minus omega which implies A of omega is a even function.

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Even Function

$$B(\Omega) = -B(-\Omega)$$

Odd Function

$$\Rightarrow x(n) = \text{real} + \text{even}$$

$$\Rightarrow x(n) = x_e(n)$$

$$\Rightarrow x_o(n) = 0$$

$$\Rightarrow X(\Omega) = A(\Omega)$$

And further we also have B of omega equals you can see waiting the imaginary parts where B of omega equals minus I am sorry this has to be B of minus omega. So, this is basically B of minus omega ok. So, B of omega equals minus B of minus omega which means this is a this is an odd function.

So, A of omega for a real signal the real part is an even function over a real part of the DTFT and the imaginary part B of omega is an odd function of omega ok. Now, therefore, if x n is real and even now what does this imply this implies two things; one is x n equals real plus even this implies x n equals x e of n comma x o of n the odd component of n is 0. So, this implies x of omega equals A of omega which is basically real and even.

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Handwritten notes on a whiteboard showing the decomposition of a real signal into even and odd components and their corresponding DTFTs.

$$\begin{aligned} \Rightarrow x(n) &= x_e(n) \\ x_o(n) &= 0 \\ \Rightarrow X(\Omega) &= A(\Omega) \\ &\text{Real + Even.} \end{aligned}$$

$$\begin{aligned} x(n) &= \text{real} + \text{odd} \\ \Rightarrow x(n) &= x_o(n) \\ x_e(n) &= 0 \\ \Rightarrow X(\Omega) &= X_o(\Omega) \\ &= jB(\Omega) \end{aligned}$$

So, for a real and even signal the corresponding DTFT is so if the signal is real plus even the corresponding DTFT is real plus also even on the other hand if the signal is real and odd ok. Now look at the other thing if x_n now this implies that x_n equals odd component ok, the even component is 0, if x_n is an odd signal and x_e of n equals 0.

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Handwritten notes on a whiteboard showing the decomposition of a real signal into even and odd components and their corresponding DTFTs, with a note that the result is purely imaginary and odd.

$$\begin{aligned} x(n) &= \text{real} + \text{odd} \\ \Rightarrow x(n) &= x_o(n) \\ x_e(n) &= 0 \\ \Rightarrow X(\Omega) &= X_o(\Omega) \\ &= jB(\Omega) \\ &\text{Purely imaginary} \\ &\text{+ odd.} \end{aligned}$$

And this implies x of ω equals x odd component of ω which is equal to $j b$ of ω which is purely imaginary and odd remember B of ω is odd and this is a

purely imaginary functions. So, this is this is purely imaginary plus odd, so if the signal is real even signal its DTFT is pure is purely real and even.

If $x[n]$ is real and odd signal then its DTFT is purely imaginary and odd ok. So, this is the important thing to keep in mind alright so that basically sums up the properties of the DTFT of real signals real and even real and odd signals alright.

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PARSEVALS RELATION :

$$y(n) = x_1(n) x_2(n)$$

$$Y(\Omega) = \frac{1}{2\pi} \int_{-\pi}^{\pi} x_1(\theta) x_2(\Omega - \theta) d\theta$$

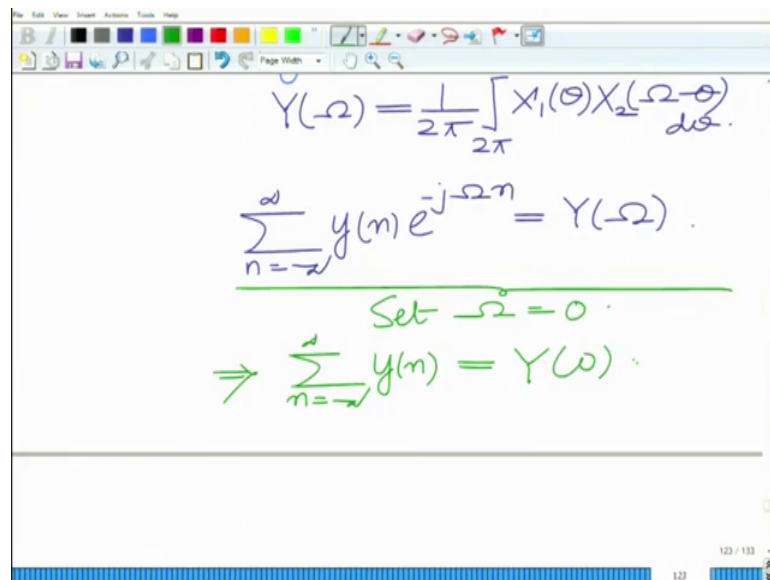
$$\sum_{n=-\infty}^{\infty} y(n) e^{-j\Omega n} = Y(\Omega)$$

And then what we can also do so as to sort of to complete this we can look at the Parseval's relation that is the last topic one of the last properties that we can look at for the DTFT that is the Parseval's relation.

And this is also very simple remember previously we have said let us say we have $y[n]$ equals $x_1[n]$ and that is the multiplication of $x_1[n]$ and $x_2[n]$ then we know that $Y(\Omega)$ is the periodic convolution of $X_1(\Omega)$ and $X_2(\Omega)$. So, that is given as integral $\frac{1}{2\pi} \int_{-\pi}^{\pi} x_1(\theta) x_2(\Omega - \theta) d\theta$ which implies basically that now if I look at summation.

So, now, we have basically what do we have we have basically summation n equals minus infinity to infinity $y[n] e^{-j\Omega n}$ this is equal to $Y(\Omega)$.

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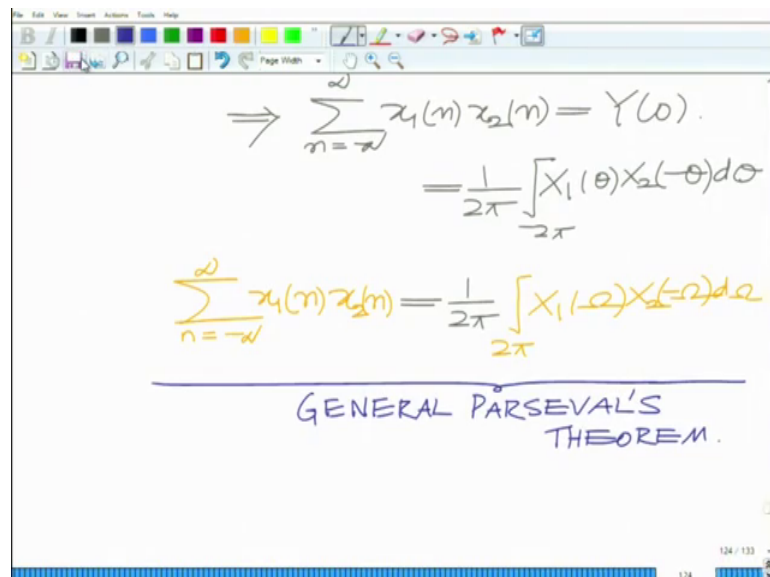

$$Y(\omega) = \frac{1}{2\pi} \int_{-2\pi}^{2\pi} X_1(\theta) X_2(-\omega - \theta) d\theta.$$
$$\sum_{n=-\infty}^{\infty} y(n) e^{-j\omega n} = Y(\omega).$$

Set $\omega = 0$.

$$\Rightarrow \sum_{n=-\infty}^{\infty} y(n) = Y(0).$$

Now if you set omega equal to 0. Now, in this set omega equal to what this implies is summation n equals minus infinity to infinity y n equals y of 0 implies summation n equal to minus infinity to infinity the product x 1 into minus infinity to infinity x 1 n into x 2 of n equals y of 0 and y of 0 is nothing, but set omega equal to 0 on the right.

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$$\Rightarrow \sum_{n=-\infty}^{\infty} x_1(n) x_2(n) = Y(0).$$
$$= \frac{1}{2\pi} \int_{-2\pi}^{2\pi} X_1(\theta) X_2(-\theta) d\theta$$
$$\sum_{n=-\infty}^{\infty} x_1(n) x_2(n) = \frac{1}{2\pi} \int_{-2\pi}^{2\pi} X_1(-\omega) X_2(-\omega) d\omega$$

GENERAL PARSEVAL'S THEOREM.

So, that is integral x 1 of theta x 2 of minus theta d theta and you can also in fact, write that as theta is just an index. So in fact, you can also write that as 1 over 2 pi integral

over 2π ; x_1 of ω x_2 of $-\omega$ $d\omega$ ok. This is summation x_1 of n x_2 of n ok.

So you can always write this thing summation n equal to minus 1 this is nothing, but the correlation between these two discrete time sequences x_1 n x_2 n summation n equal to minus infinity to infinity x_1 n x_2 n that is integral 1 over 2π integral over any 2π x_1 ω into x_2 $-\omega$ $d\omega$ I have simply replaced the integration variable θ by ω ok.

Now if we set x_2 n now in this so this you can think of this as the generalized Parseval's relation. In fact, this is a step ahead of the Parseval's relation it is much more general it talks about two different signals x_1 n and x_2 n . So, you can think of this as a generalized or a general form. So, even think about this as a general form of Parseval's relation.

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The image shows a whiteboard with the following content:

$$\sum_{n=-\infty}^{\infty} x_1(n)x_2(n) = \frac{1}{2\pi} \int_{-\pi}^{\pi} X_1(\omega)X_2(-\omega)d\omega$$

GENERAL PARSEVAL'S
THEOREM.

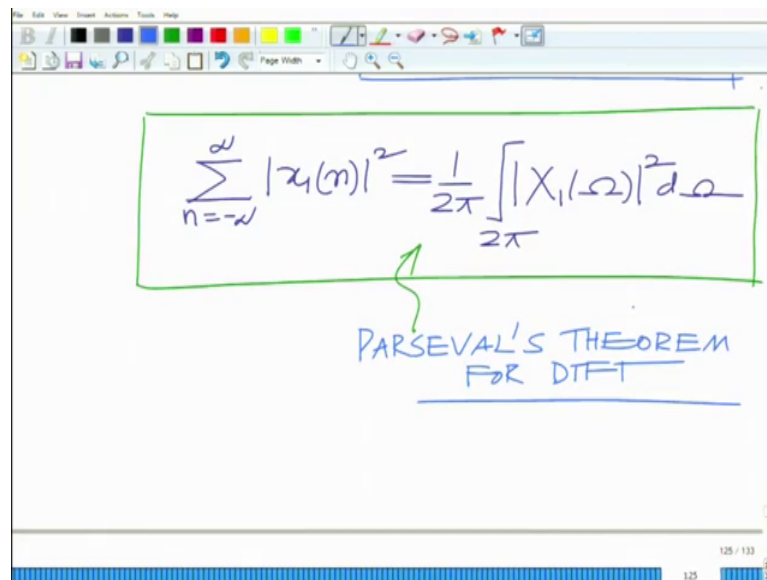
$$x_2(n) = x_1^*(n)$$

$$\Rightarrow X_2(-\omega) = X_1^*(-\omega)$$

$$\Rightarrow X_2(-\omega) = X_1^*(\omega)$$

Now if you said x_2 n equals x_1 conjugate n . Then what you have is x_2 of ω equals x_1 conjugate of $-\omega$ because conjugate sequences is x_1 conjugate of $-\omega$ which means x_2 of $-\omega$ equals x_1 conjugate of ω . So, if you substitute if you substitute x_2 n equals x_1 conjugate on n .

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$$\sum_{n=-\infty}^{\infty} |x_1(n)|^2 = \frac{1}{2\pi} \int_{-\pi}^{\pi} |X_1(\omega)|^2 d\omega$$

PARSEVAL'S THEOREM FOR DTFT

On the left hand side what we have a summation $\sum_{n=-\infty}^{\infty} |x_1(n)|^2$ that is magnitude $|x_1(n)|^2$ whole square that is equal to summation $\frac{1}{2\pi} \int_{-\pi}^{\pi} |X_1(\omega)|^2 d\omega$ which is $\frac{1}{2\pi}$ conjugate of ω that is $\frac{1}{2\pi}$ conjugate of ω which is magnitude $|X_1(\omega)|^2$ into $d\omega$ that is the sum of the energy over 1 period in that this is not the energy.

But you can think of this as the power because you are dividing by 2π so this is the power of the you can think of this as the power of the DTFT ok. So, this is the Parseval's relation for the, this is the Parseval's theorem for DTFT. The summation $\sum_{n=-\infty}^{\infty} |x_1(n)|^2$ equals $\frac{1}{2\pi} \int_{-\pi}^{\pi} |X_1(\omega)|^2 d\omega$ region magnitude $|X_1(\omega)|^2$ of ω whose square $|X_1(\omega)|^2$ alright.

So, basically that completes our discussion of the properties of the DTFT. In the subsequent when the next module will start looking at the DTFT and its properties with relation to the with relation to LTI systems alright. So, we will stop here and continue in the subsequent modules.

Thank you very much.