

Principles of Signals and Systems
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Lecture – 57

**Discrete Time Fourier Transform: Properties of DTFT - Linearity, Time Shifting,
Frequency Shifting, Conjugation, Time-Reversal, Duality**

Hello welcome to another module in this massive open online course.

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$X(z)|_{z=e^{j\omega}} = \frac{1}{1-e^{j\omega}} \neq X(\omega)$

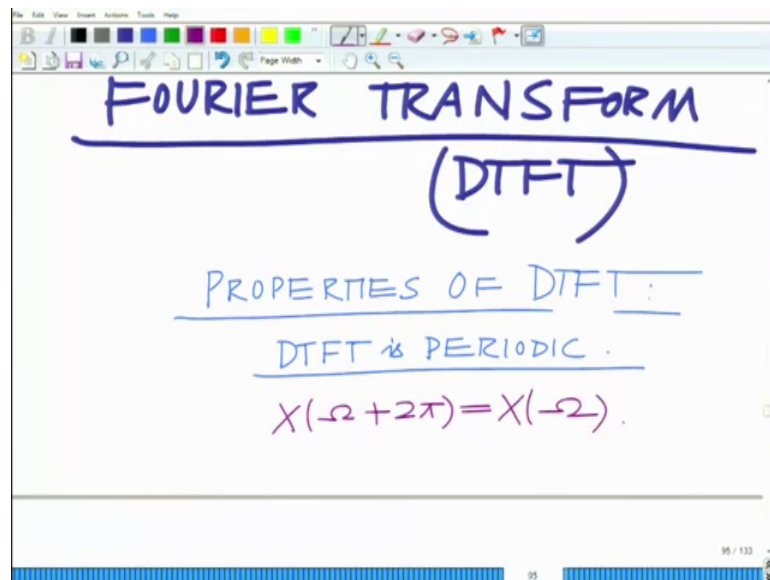
$X(\omega) = \pi \delta(\omega) + \frac{1}{1-e^{j\omega}}$
DTFT of unit-step signal.

DISCRETE TIME

So, we are looking at the Fourier analysis of discrete type signals in particular the Fourier analysis for discrete time aperiodic signals through the discrete time Fourier transform alright. So, we are looking at the discrete time, Fourier transform for discrete time aperiodic signals or what is known as the DFDT the discrete time Fourier transform all right.

So, now, let us look at the properties of the DFDT all right we have looked at the DFDT of a unit step signal.

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So, the properties of let us continue our discussion with the properties of the DFDT of any aperiodic signal x_n . Now, one of the most fundamental properties of the DFDT is that; the DFDT is periodic. So, DFDT is periodic and there in fact, if you can see X of ω plus 2π this is equal to X of ω this can be seen as follows; X of ω equals summation n equals minus infinity to infinity $x_n e^{j\omega n}$.

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$$X(\Omega) = \sum_{n=-\infty}^{\infty} x(n)e^{-j\Omega n}$$
$$X(\Omega + 2\pi) = \sum_{n=-\infty}^{\infty} x(n)e^{-j(\Omega + 2\pi)n}$$
$$= \sum_{n=-\infty}^{\infty} x(n)e^{-j\Omega n} \frac{e^{-j2\pi n}}{1}$$
$$= \sum_{n=-\infty}^{\infty} x(n)e^{-j\Omega n}$$

And X of ω plus 2π this is equal to sum n equal to minus infinity to infinity X of $n e^{j\omega n}$ plus $2\pi n$ which is sum n is equal to minus infinity to infinity X of n

$e^{j\omega n} X[n] e^{-j(\omega + 2\pi)n}$; this is one which is sorry. In fact, this $X[n] e^{-j\omega n} e^{j(\omega + 2\pi)n}$ and this is summation n equals minus infinity to infinity $X[n] e^{-j\omega n}$ which is nothing, but $X(\omega)$ ok.

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$$\begin{aligned}
 X(\Omega + 2\pi) &= \sum_{n=-\infty}^{\infty} x(n) e^{j(\Omega + 2\pi)n} \\
 &= \sum_{n=-\infty}^{\infty} x(n) e^{j\Omega n} \frac{e^{j2\pi n}}{1} \\
 &= \sum_{n=-\infty}^{\infty} x(n) e^{j\Omega n} \\
 &= X(\Omega)
 \end{aligned}$$

$X(\Omega + 2\pi) = X(\Omega)$

So, this is basically again once again this is $X(\omega)$. So, we have $X(\omega + 2\pi)$ equals $X(\omega)$ ok. So, the DFDT is very basically periodic and the period is 2π the fundamental period ok, all right.

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$X(\Omega + 2\pi) = X(\Omega)$

LINEARITY:

$$\begin{aligned}
 x_1(n) &\longleftrightarrow X_1(\Omega) \\
 x_2(n) &\longleftrightarrow X_2(\Omega)
 \end{aligned}$$

$$\begin{aligned}
 &a_1 x_1(n) + a_2 x_2(n) \\
 \xrightarrow{\text{DFT}} &a_1 X_1(\Omega) + a_2 X_2(\Omega)
 \end{aligned}$$

Let us look at some other periodic properties again one of the simplest properties which is obeyed by all the transforms that we've seen before and the DFDT is no exception is the linearity that is if $x_1[n]$ has Fourier transform or DFDT $X_1(\omega)$ and $x_2[n]$ has DFDT $X_2(\omega)$, then $a x_1[n] + b x_2[n]$ where a and b are constants has the DFDT as you can expect $a X_1(\omega) + b X_2(\omega)$. So, this is the linearity that is a linear combination of the signals produces has the DFDT that is the corresponding linear combination of the DFDTs of the corresponding individual signals alright.

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TIME SHIFT :

$$x(n) \leftrightarrow X(\Omega)$$

$$x(n - n_0) \leftrightarrow ?$$

$$\tilde{X}(\Omega) = \sum_{n=-\infty}^{\infty} x(n - n_0) e^{-j\Omega n}$$

$$n - n_0 = m$$

$$\Rightarrow n = m + n_0$$

Time shifting again another important property what happens when you time shift a discrete time signal. So, let us say $x[n]$ has the DFDT $X(\omega)$ what happens when you time shift by n_0 . So, we are asking what is $\tilde{X}(\omega)$ which is summation n equals minus infinity to infinity $x[n - n_0] e^{-j\omega n}$ set $n - n_0 = m$ that implies $n = m + n_0$.

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The image shows a whiteboard with handwritten mathematical derivations. At the top, the substitution $n - n_0 = m$ is written, leading to $n = m + n_0$. Below this, the Fourier transform $\tilde{X}(\omega)$ is expressed as a summation: $\tilde{X}(\omega) = \sum_{m=-\infty}^{\infty} x(m) e^{-j\omega(m+n_0)}$. This is then factored as $e^{-j\omega n_0} \sum_{m=-\infty}^{\infty} x(m) e^{-j\omega m}$. The summation part is identified as $X(\omega)$, resulting in the final equation: $\tilde{X}(\omega) = X(\omega) e^{-j\omega n_0}$.

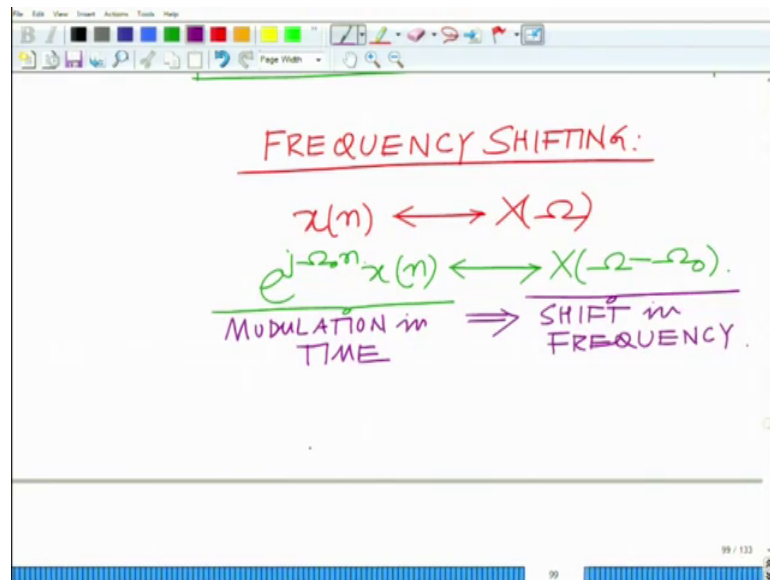
So, this is equal to summation n equals minus infinity to m . In fact, m the index will be m raised to minus j omega m plus n naught which is if you take e raised to minus j omega n naught outside you have summation m equals minus infinity to infinity x of m e raised to minus j omega m ; and this is X of omega and therefore, this is basically equal to X of omega times e raised to minus j omega n naught that is your X tilde omega.

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The image shows a whiteboard with handwritten mathematical derivations. The top part shows the equation $\tilde{X}(\omega) = X(\omega) e^{-j\omega n_0}$ with a label "Modulation in freq" written below it. Below this, a boxed equation shows the Fourier transform pair: $x(m - n_0) \leftrightarrow X(\omega) e^{j\omega n_0}$.

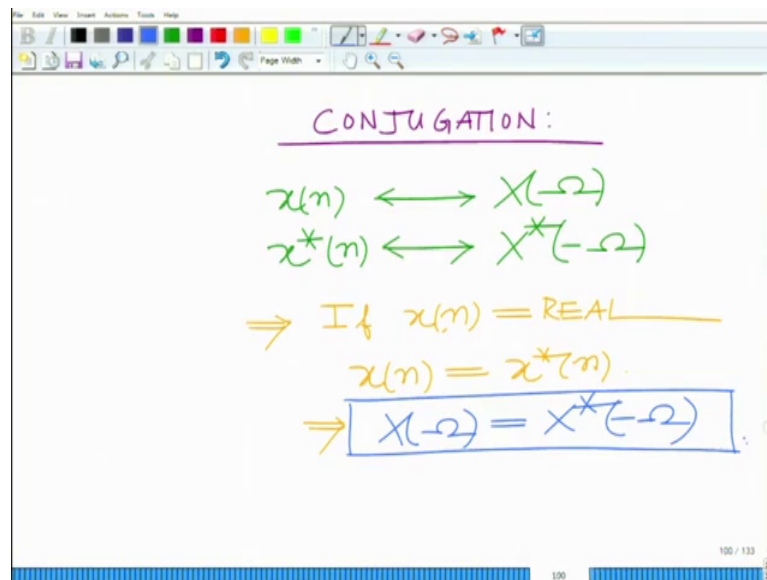
So, when your time shift that is basically a modulation in frequency multiplied by the complex exponential e raised to minus j e raised to minus j $\omega_0 n$. So, x of n minus n_0 as the DFDT X of ω e raised to minus j $\omega_0 n$ ok.

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Similarly, we have frequency shifting now in frequency shifting again one can derive this similarly x n ; let us say x has Fourier transform X of ω then e power e raised to j $\omega_0 n$, that is e raised to j $\omega_0 n$ x n , that is a modulated signal modulation in time leads to a corresponding shift in frequency. So, e raised to j $\omega_0 n$ x n e raised to j $\omega_0 n$ x n , that is a modulation in time because you are multiplying it by a complex exponential of frequency ω_0 in that in the frequency domain leads to a shift in the frequency by ω_0 ok.

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CONJUGATION:

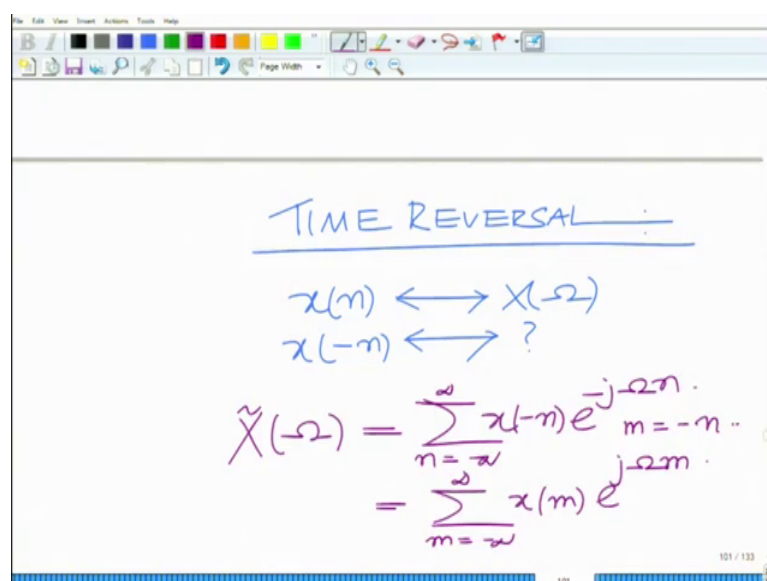
$$x(n) \longleftrightarrow X(\omega)$$
$$x^*(n) \longleftrightarrow X^*(-\omega)$$

\Rightarrow If $x(n) = \text{REAL}$

$$x(n) = x^*(n)$$
$$\Rightarrow X(\omega) = X^*(-\omega)$$

So, this is modulation in time implies a shift in the frequency domain ok. Conjugation when you have $x(n)$ conjugate, that is $x(n)$ has DFDT $X(\omega)$, then $x(n)$ conjugate has DFDT $X^*(-\omega)$ and therefore, this naturally implies if $x(n)$ is real then we have $x(n) = x^*(n)$ and that implies the corresponding DFDTs are equal that is $X(\omega) = X^*(-\omega)$ this is something that we have already seen before; that is for an even that is for a real signal $X(\omega) = X^*(-\omega)$ of ω is basically $x(n)$ conjugate of $-\omega$. So, the magnitude spectrum is an even function of ω and the phase spectrum is an odd function right.

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TIME REVERSAL:

$$x(n) \longleftrightarrow X(\omega)$$
$$x(-n) \longleftrightarrow ?$$
$$\tilde{X}(\omega) = \sum_{n=-\infty}^{\infty} x(-n) e^{j\omega n}$$
$$= \sum_{m=-\infty}^{\infty} x(m) e^{j\omega m}$$

Time reversal again similarly one can look at time reversal, if $x[n]$ has DFDT $X(\omega)$ then $x[-n]$ has DFDT $\tilde{X}(\omega)$ well what is the DFDT that would be $\tilde{X}(\omega)$ equals $\sum_{m=-\infty}^{\infty} x[-n] e^{-j\omega n}$ equals summation setting m equals $-n$ that would be $\sum_{m=-\infty}^{\infty} x[m] e^{-j\omega m}$ where m equals $-n$ and therefore, $\tilde{X}(\omega)$ equals summation m equals $-\infty$ to ∞ $x[m] e^{-j\omega m}$ that is $X(\omega)$.

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The image shows a whiteboard with handwritten mathematical derivations. The top part shows the derivation of the DFT of a time-reversed signal $x[-n]$. It starts with the definition of the DFT of $x[-n]$, which is $\tilde{X}(\omega) = \sum_{m=-\infty}^{\infty} x[-n] e^{-j\omega n}$. Then, it shows that by substituting $m = -n$, the expression becomes $\tilde{X}(\omega) = \sum_{m=-\infty}^{\infty} x[m] e^{-j\omega m}$, which is equal to $X(\omega)$. The bottom part of the whiteboard shows the final result: $\tilde{X}(\omega) = X(\omega)$ and $x[-n] \leftrightarrow X(\omega)$.

$$\tilde{X}(\omega) = \sum_{m=-\infty}^{\infty} x[-n] e^{-j\omega n}$$

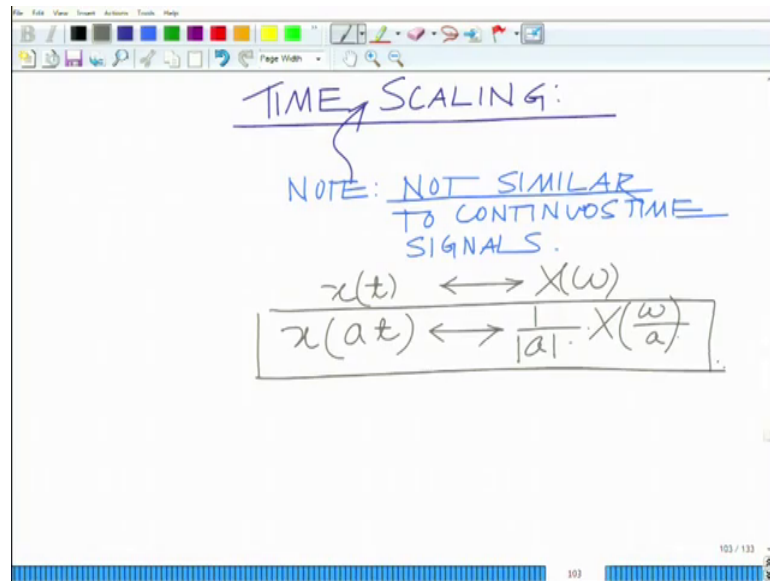
$$\tilde{X}(\omega) = \sum_{m=-\infty}^{\infty} x[m] e^{-j\omega m} = X(\omega)$$

$$\tilde{X}(\omega) = X(\omega)$$

$$x[-n] \leftrightarrow X(\omega)$$

So, we have $\tilde{X}(\omega) = X(\omega)$ which implies $\tilde{X}(\omega) = X(\omega)$ ok. So, basically $x[-n]$ has the Fourier transform as the DFDT $X(\omega)$ of $x[n]$ ok. So, $x[-n]$ has the DFDT $X(\omega)$ alright.

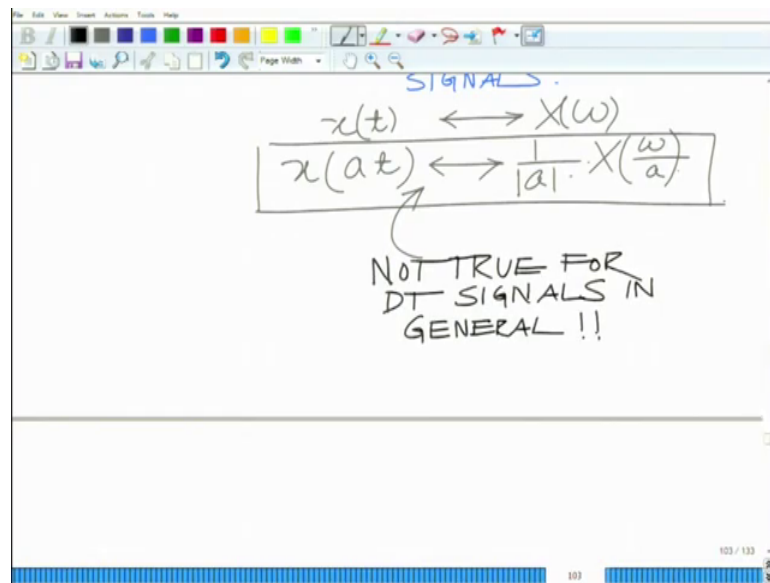
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And another interesting property is that of time scaling; now time scaling is interesting, because it does not directly follow from the, and this is a very interesting property because this is one aspect which does not have a direct which does not directly follow from the continuous time.

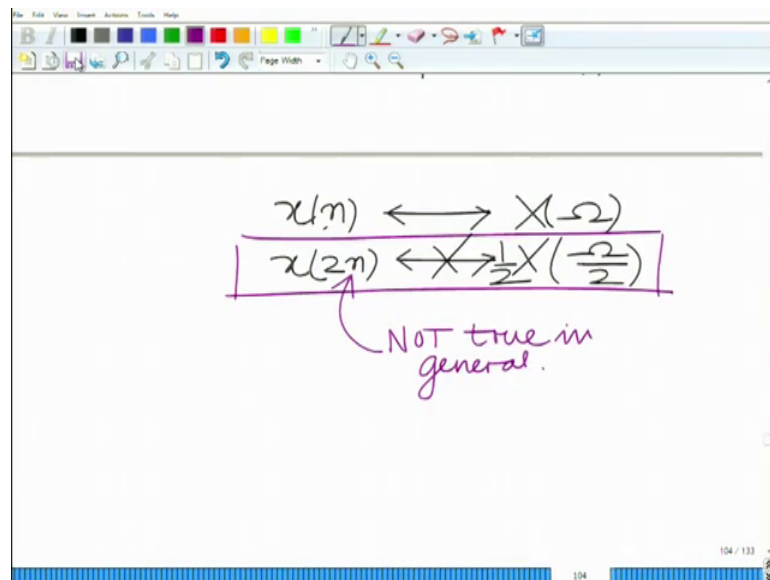
So, one would be mistaken if one would substitute directly the result from the Fourier transform of a continuous time signal and the reason is as follows. So, time scaling caution is important note not while most are similar to the continuous time properties is not similar this is not similar to that for continuous time signals the reason is as follows; that is if you look at a continuous time signal then you have x of $a t$ as the Fourier transform that is if x of t has the Fourier transform X of ω , then x of $a t$ has the Fourier transform one over magnitude of $a X$ of ω over a this is the result for time scaling ok.

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However, this is not true for discrete time signals in general; however, not true for discrete time signals the DT signals; that is if you have x of n has the DFDT X of ω .

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Then x of $2n$ will not have the DFDT half X this is not true; in general this statement is not true this is not true this is not true in general that is x of $2n$ preferring to the sequence that does not have that is I cannot use the formula $\frac{1}{|a|}$ over a $\frac{1}{|a|}$ over magnitude of a X of ω over a that is I cannot get the DFDT by $\frac{1}{|a|}$ over by half X of ω divided by that result is not to true in general.

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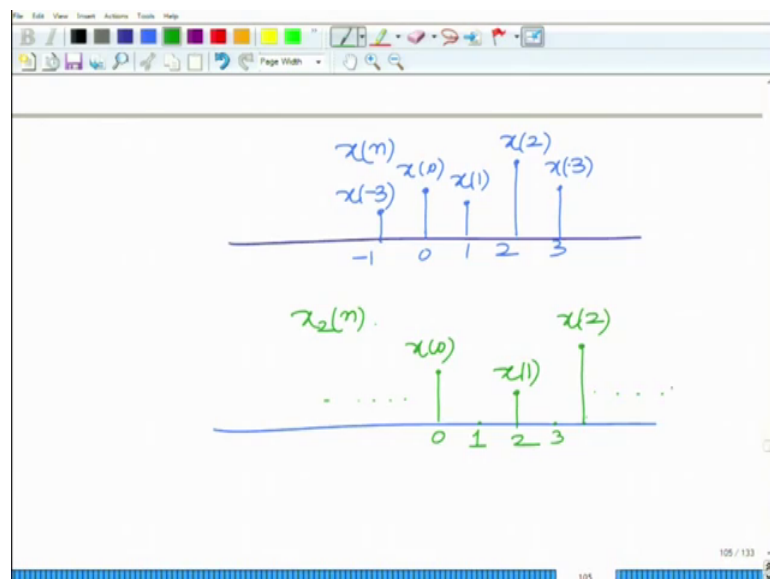
NOT true in general.

$$x_m(n) = \begin{cases} x\left(\frac{n}{m}\right) & \text{if } n = km. \\ 0 & \text{if } n \neq km. \end{cases}$$

$$x_2(n) = \begin{cases} x\left(\frac{n}{2}\right) & \text{if } n = \text{even} \\ 0 & \text{if } n = \text{odd}. \end{cases}$$

However; the following result is with this for the following result is true that is if I define a signal x_m of n equals or any integer $m \times n$ over m if n is a multiple of m , that is if n equals k times m if n is that; is n is divisible by m equals 0 n naught equal to $k m$ for instance for example, x of 2 of n x of m equals x of that is basically equals x of n by 2 ; if n equals even and 0 if n is odd let us consider this for instance ok.

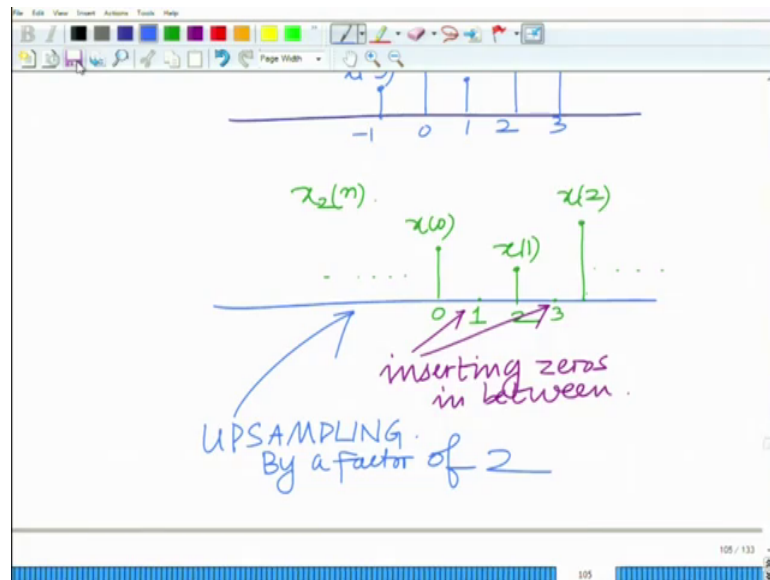
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And for instance this signal can be visualized as follows let us take a simple example; let us say I have this is x of 0 , this is x of 1 , this is x of 2 , x of 3 , this is x of minus 3 , and

then if I consider the signal x_2 of n which is basically; so x_2 of 0 is simply x_0 , x_2 of 1 because 1 is not a multiple of 2 this is 0, x_2 of 2 is x_1 ok. So, x_2 of 2 is x_1 , similarly x_2 of 3 is 0 and x_2 of 4 is x_2 ok. Similarly on the negative axis this is x of n .

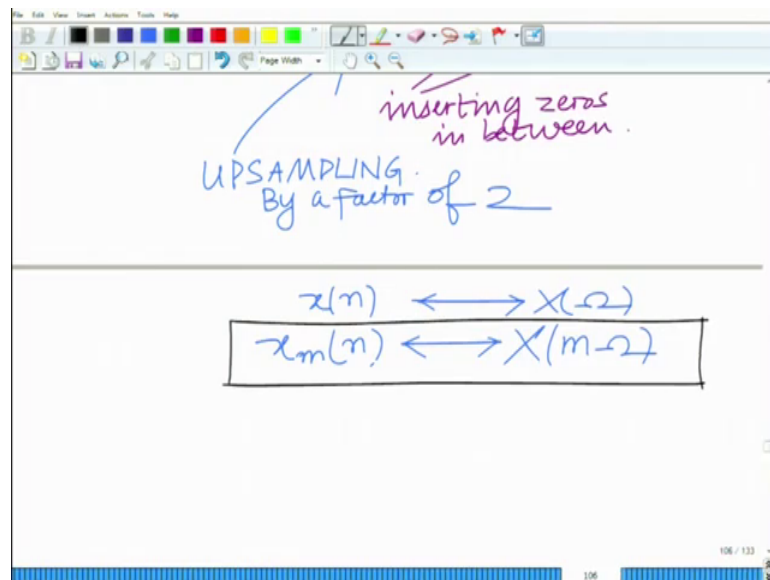
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So, you are interspersing so, you are inserting zeros in between so, first thing realize is basically you are inserting you are inserting zeroes in between and you are spreading the signal out all right. So, x_2 of n , x_2 of 0 is 0, x_2 of 0 is x of 0, x_2 of 1 is 0, x_2 of 2 is x of 1, x_2 of 3 is 0, x_2 of 4 is x of 2 and so, on.

So, we are spreading the signal out and basically you are inserting zeros in between this is known as up sampling. In fact, this is known as up sampling by a factor of 2 ok. So, this operation is basically termed as up sampling; in fact, this is equal to up sampling by a factor of this is up sampling by a factor of 2; there up sampling by a factor of 2 ok.

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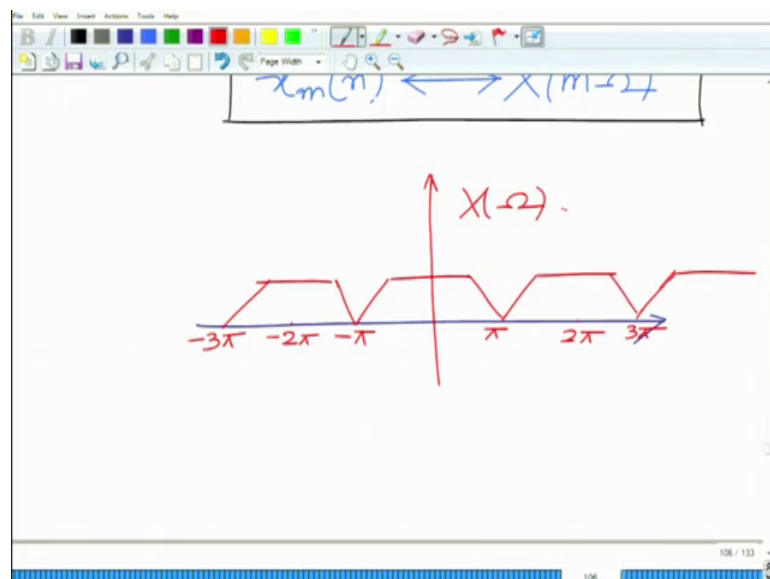
UPSAMPLING
By a factor of 2

inserting zeros
in between.

$$\boxed{x(n) \leftrightarrow X(\Omega)}$$
$$\boxed{x_m(n) \leftrightarrow X(m\Omega)}$$

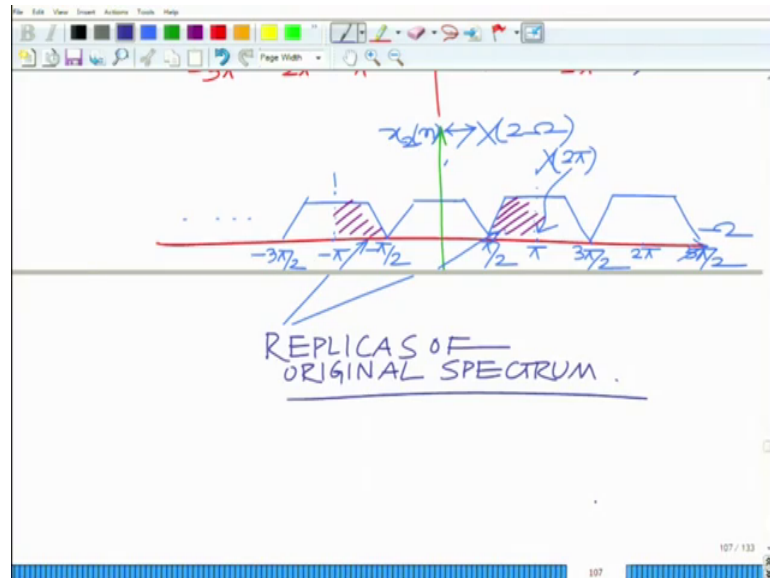
Now even you up sample a signal; that is x of n of n the resulting Fourier transform is x m of ω , that is if x of n has the Fourier transform X of ω , then x m of n has the Fourier transform X m ω that is in the frequency domain it shrinks by a factor of 2. So, we have x of 2 of ω ; that means, with what was at 2π comes 2π ok. So, that can be seen as follows.

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So, X of ω is periodic; let us try to plot a periodic let us plot a periodic DFDT first ok. So, this is let us say. So, this is minus π to π , this is 2π 3π minus 2π minus 3π . So, this is X of ω .

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Now x_2 of n will have the Fourier transform which looks something like this. So, this is like. So, this is your x_2 of ω , which is the DFDT of x_2 of DFDT of x_2 of n . So, that will look something like.

So, you will have basically at; so this will be π by minus π by 2 to π by 2 what is a 2π will come 2π . So, what is that? So, at 2π you will have what is that. So, you have something that looks like π by 2. So, at 3π by 2. So, this is at π 3π by 2 at 2π you will have what is 5π by and so, on and similarly here you will have something that is and so, on ok.

So, this is basically your X of 2ω and remember this is X of basically 2π ok. So, what is at 2π will come to. So, this was spectrum basically shrinks and what happens is as a result of that you can see you have replicas of the same spectrum, if you look at the minus π to π spectrum you will have the replica. So, what this is basically if you can see this it is a replica of the previous copy of the previous signal ok.

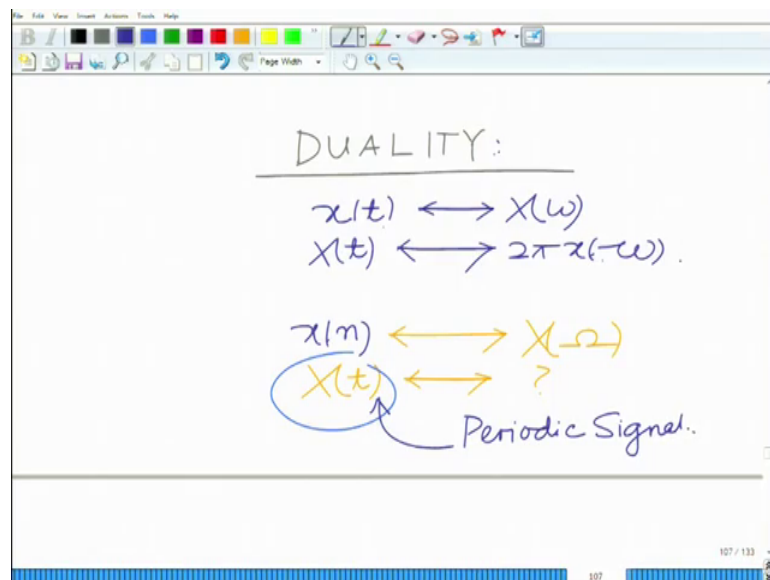
So, you have replicas of the original spectrum n minus π to π and you can see that the number of such replicas is equal to the up sampling factor. So, if you have up sampled by

a factor of 3; then you have 3 replicas 1 is the original and then you have three other replicas if you have up sampled by factor of 4 4 replicas one original and three other replicas. So, so that is the interesting aspect of this that is up sampling all right. Now for the when you down sample that is the opposite of up sample that is you scale down the signal of time domain it is slightly more difficult to derive the Fourier the DFDT of the resulting signal, because you can see that results in a loss of samples.

When you shrink the time domain signal unlike in a continuous time signal, when you shrink the time domain signal, for instance if you shrink it by the factor of 2 take only the even samples then which then it means basically the odd samples of the signal have to be reversed. So, there is if there is a certain loss that is occurring when you are down sampling and therefore, the DFDT is not it is not straightforward to derive the DFDT in such a scenario.

So, basically one has to realize that the result for scaling in the continuous time does not hold; in general for discrete time signal using the DFDT in the context of the DFDT that is the discrete time Fourier transform all right ok.

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Again duality that is another interesting very interesting property what happens to duality; and this is also very interesting, because we see in the continuous time if you have x of t which has a Fourier transform X of ω , then we know that X of capital X of t has the Fourier transform $2\pi x$ of minus ω in the k in the context of a discrete

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$$X(t + 2\pi) = X(t)$$

$$X(-\Omega) = \sum_{n=-\infty}^{\infty} x(n) e^{-j\Omega n}$$

$$X(t) = \sum_{n=-\infty}^{\infty} x(n) e^{-jnt}$$

$$m = -n$$

$$= \sum_{m=-\infty}^{\infty} x(-m) e^{jmt}$$

And now you can see here we have the fundamental frequency omega naught.

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$$T_0 = 2\pi$$

$$\omega_0 = \frac{2\pi}{T_0} = 1$$

$$X(t) = \sum_{m=-\infty}^{\infty} x(t-m) e^{jm\omega_0 t}$$

$$X(t) = \sum_{m=-\infty}^{\infty} x(t-m) e^{jm\omega_0 t}$$

So, we have the time period T_0 equals 2π the fundamental time period, which means the fundamental frequency ω_0 equals 2π by T_0 this is equal to 1. So, I can denote ω_0 by 1 and I can write this as summation m equals minus infinity to infinity $x(t-m)$ $e^{jm\omega_0 t}$ where ω_0 is equal to 1 and; that is the fundamental frequency and now you can see we have the discrete time, we have the complex exponential Fourier

series for X of t . So, I can write it as just going to write it clearly; e raised to j ω m t and what this is; is this is a now you can see this is the complex exponential Fourier series, because you are expressing X t as a linear combination infinite number of complex exponentials at the fundamental frequency ω t and it is harmonics.

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$$X(t) = \sum_{m=-\infty}^{\infty} x(t-m) e^{jm\omega t}$$

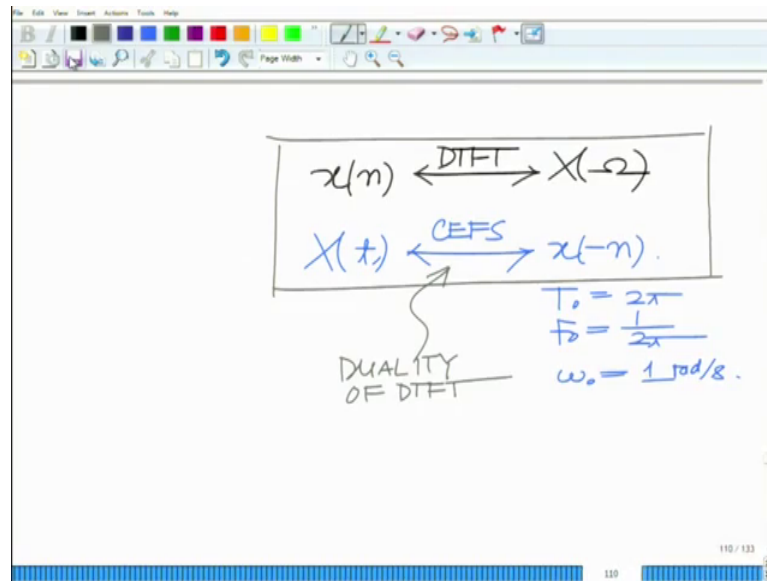
CEFS

$x(t-m)$ = coefficient of $e^{jm\omega t}$ in CEFS.

$$X(t) \xleftrightarrow{\text{CEFS}} x(t-m)$$

So, this is the complex exponential Fourier series and these are the x minus m is the coefficient of e power j m ω t in the complex exponential Fourier series. So, basically you can say X of t has the complex exponential Fourier series given by x of minus m that is the duality result for the DFDT; because the asymmetric nature of the DFDT, when you take the resulting signal in time that is a periodic signal ok.

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And in fact, so the duality result the complete result is x of n has here the top equivalence is in terms of the DFDT and the bottom; that I am going to write now; if you look at the signal in time it is naturally going to be a continuous time signal, because X of ω is a continuous time signal, but it is going to be a periodic signal X of t and this equivalent is in terms of the complex exponential Fourier series x of minus n with the fundamental period t naught equals 2π fundamental frequency linear frequency f naught equals 1 over 2π angular frequency ω naught equals $2\pi f$ not equals 1 radian per second and this is the result this is basically the result for this is basically the result for the duality, duality of the DFDT duality of DFDT alright.

So, basically that completes several properties of the DFDT starting from the simplest property linearity time shifting modulation etcetera and finally, we have looked at up sampling of a discrete time signal and also the duality property for the DFDT that is the duality property for the DFDT of a discrete type aperiodic signal alright. So, we will stop here and look at other aspects in the subsequent modules.

Thank you very much.