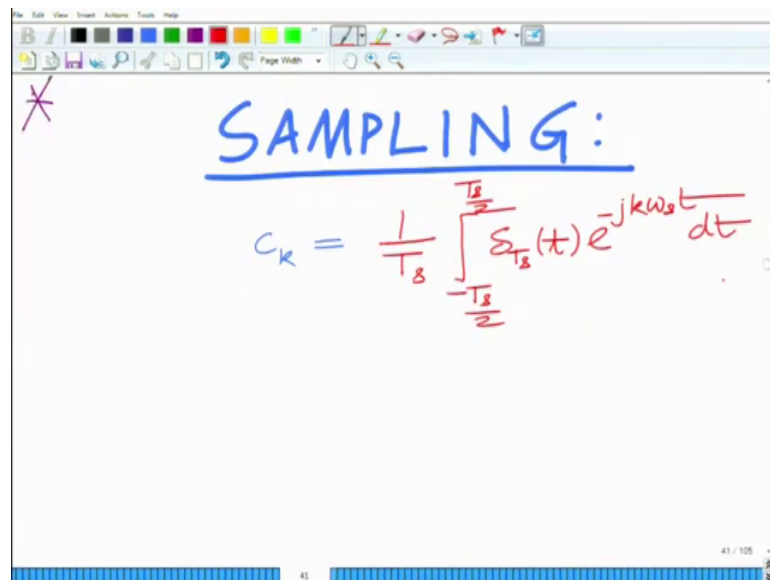


**Principles of Signals and Systems**  
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**Indian Institute of Technology, Kanpur**

**Lecture - 52**  
**Sampling: Spectrum of Sampled Signal, Nyquist Criterion**

Hello. Welcome to another module in this massive open online course. So, we are looking at sampling and considering impulse train based sampling which is the most common form of sampling a continuous and in continuous time analog signal to convert it into a discrete time signal ok.

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SAMPLING:

$$c_k = \frac{1}{T_s} \int_{-\frac{T_s}{2}}^{\frac{T_s}{2}} s_{T_s}(t) e^{-jk\omega_s t} dt$$

So, what we are looking at is we are looking at the principle or the theory behind sampling and try to trying to understand it is various aspects or various properties of sampling using the Fourier transform and what we are using is the impulse train correct to sample the signal.

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The image shows a whiteboard with handwritten mathematical expressions and text. At the top, the equation  $s_{T_s} = \sum_{k=-\infty}^{\infty} \delta(t - kT_s)$  is written. Below it, the text "PERIODIC SIGNAL" is underlined, followed by "Period =  $T_s$ ." and "Fund Freq =  $\frac{2\pi}{T_s} = \omega_s$ ." A blue arrow points from this text to the  $s_{T_s}$  term in the equation above. Below the text, it says "Therefore, we can derive the CEFS". At the bottom of the whiteboard, the equation  $s_{T_s}(t) = \sum_{k=-\infty}^{\infty} C_k e^{jk\omega_s t}$  is written. The whiteboard interface includes a toolbar at the top and a page number "39 / 105" at the bottom right.

$$s_{T_s} = \sum_{k=-\infty}^{\infty} \delta(t - kT_s)$$

PERIODIC SIGNAL  
Period =  $T_s$ .  
Fund Freq =  $\frac{2\pi}{T_s}$   
=  $\omega_s$ .

Therefore, we can derive the CEFS

$$s_{T_s}(t) = \sum_{k=-\infty}^{\infty} C_k e^{jk\omega_s t}$$

And we said the impulse train is a periodic signal  $\delta_{T_s}$ , summation  $\delta(t - kT_s)$   $k$  going from minus infinity to infinity  $T_s$ , this quantity is known as the sampling duration of the sampling interval  $1$  over  $T_s$  which is  $f_s$  is known as the sampling frequency,  $2\pi$  over  $T_s$  that is  $\omega_s$  is the angular frequency of sampling in radians per second ok.

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The image shows a whiteboard with handwritten mathematical expressions. The equation  $s_{T_s}(t) = \sum_{k=-\infty}^{\infty} C_k e^{jk\omega_s t}$  is written. A blue arrow points from the text "CEFS of Impulse Train" below to the  $C_k$  term in the equation. The whiteboard interface includes a toolbar at the top and a page number "40 / 105" at the bottom right.

$$s_{T_s}(t) = \sum_{k=-\infty}^{\infty} C_k e^{jk\omega_s t}$$

CEFS of Impulse Train

And we are trying to find the coefficient  $C_k$  of the complex exponential Fourier series of the impulse train and the  $C_{fo}$  of the impulse train the coefficient can be found as follows,

remember  $C_k$  this coefficient equals well over 1 period  $1$  over  $T_s$  the fundamental period integral over any period.

So, we can choose minus  $T_s$  over  $2$  to  $T_s$  over  $2$  delta  $T_s$   $t$  or delta  $t$  in 1 period it is simply delta  $t$  delta  $T_s$   $t$  let me just write the delta  $T_s$   $t$   $e$  power minus  $j k$  omega  $s t$   $dt$ , this expression for the coefficient  $C_k$  of the complex exponential Fourier series.

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$$C_k = \frac{1}{T_s} \int_{-T_s/2}^{T_s/2} s(t) e^{-jk\omega_s t} dt$$

$$= \frac{1}{T_s} \left[ e^{-jk\omega_s t} \right]_{t=0}^{T_s/2}$$

But in the interval minus  $T_s$  by  $2$   $2$   $T_s$  over  $2$  this simply delta  $t$ , so this is minus  $T_s$  over  $2$  to  $T_s$  over  $2$  delta  $t$   $e$  raised to minus  $j k$  omega  $s t$   $dt$ , which is basically if you look at this  $1$  over  $T_s$   $e$  raise to minus because integral with delta  $t$  is simply  $e$  raise to minus  $j k$  omega  $s t$  evaluated at  $t$  equal to  $0$  and therefore this is simply equal to  $1$  over  $T_s$ .

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The image shows a handwritten derivation in a presentation software. At the top, the expression  $\frac{1}{T_s} e^{-jk\omega_s t} \Big|_{t=0}$  is written. A bracket above the exponent indicates the value  $-\frac{T_s}{2}$ . Below this, the expression is simplified to  $C_k = \frac{1}{T_s}$ , which is enclosed in a purple rectangular box. An arrow points from the text "For all k" to the boxed equation. The software interface includes a toolbar at the top and a status bar at the bottom showing "42 / 105".

So, if you look at this basically all the coefficients  $C_k$  this is for all  $k$  ok. So, all the coefficients  $C_k$  are basically equal to  $1/T_s$  and therefore the complex exponential Fourier series is given as summation  $C_k e^{jk\omega_s t}$  all the coefficients  $C_k = 1/T_s$ . So, this will simply be  $1/T_s$  summation  $e^{jk\omega_s t}$  that is a complex exponential Fourier series. So, your delta  $T_s t$  which is the impulse train with sampling duration  $T_s$  that is simply  $1/T_s$  summation  $k$  equal to minus infinity to infinity  $e^{jk\omega_s t}$ , this is the complex exponential Fourier series of the impulse train.

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The image shows a handwritten equation in a presentation software:  $S_{T_s}(t) = \frac{1}{T_s} \sum_{k=-\infty}^{\infty} e^{jk\omega_s t}$ . Below the equation, it is labeled "CEFS of impulse Train". Underneath, the frequency domain representation is shown as  $e^{jk\omega_s t} \longleftrightarrow 2\pi \delta(\omega - k\omega_s)$ . The software interface includes a toolbar at the top and a status bar at the bottom showing "42 / 105".

This is the complex exponential Fourier series of the impulse train, now taking the Fourier transform what we have is you now you recall that  $e^{jk\omega_s t}$ , this has a Fourier transform  $2\pi \delta(\omega - k\omega_s)$ . In fact,  $2\pi \delta(\omega - k\omega_s)$ .

So, if you take the Fourier transform of the impulse train that will simply be summation  $\frac{1}{T_s} 2\pi \delta(\omega - k\omega_s)$  because the Fourier transform each  $e^{jk\omega_s t}$  is simply  $2\pi \delta(\omega - k\omega_s)$ .

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$$e^{jk\omega_s t} \leftrightarrow 2\pi \delta(\omega - k\omega_s)$$

$$\delta_{T_s} = \frac{1}{T_s} 2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_s)$$

$$\delta_{T_s} = \omega_s \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_s)$$

FT of Impulse Train

So therefore, if you take the Fourier transform of the impulse train, that will simply be  $\frac{1}{T_s} 2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_s)$ ; which is equal to  $2\pi \omega_s \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_s)$ . So, this is basically your Fourier transform of impulse train.

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The image shows a handwritten derivation on a whiteboard. At the top, it says "Impulse Train". The derivation starts with the equation  $x(t) \times \sum_{k=-\infty}^{\infty} \delta(t - kT_s)$ . A purple arrow points from the text "multiplying signal by impulse train" to the multiplication sign. This is followed by the equation  $= \sum_{k=-\infty}^{\infty} x(t) \delta(t - kT_s)$ . Below this, the sampled signal is defined as  $x_{T_s}(t) = \sum_{k=-\infty}^{\infty} x(kT_s) \delta(t - kT_s)$ . A green circle highlights  $x(kT_s)$  in the second equation. A purple arrow points from the text "Spectrum of signal" to the  $x_{T_s}(t)$  term. A purple line underlines the text "SAMPLED SIGNAL".

Now remember what are we doing if you recall what we are doing in the sampling process, what we are doing is a basically we are taking the symbol the signal  $x(t)$  and you are multiplying with the impulse train. So, this is basically multiplying the signal with the impulse train, we are multiplying the signal with the impulse train correct. So, when you multiply 2 signals in the time domain, in the frequency domain right the corresponding Fourier transforms are convolved.

So, it is given by the corresponding spectrum of the sampled signal, is therefore given by the convolution of the Fourier transform of the original signal  $x(t)$  with that of the impulse train that we have just derived ok. So, next step what we want to do is we want to derive the spectrum of the sample signal and now you can now you realize that is fairly simple to derive the spectrum of sampled signal.

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$$x_1(t) \cdot x_2(t) \leftrightarrow \frac{1}{2\pi} X_1(\omega) * X_2(\omega)$$

$$\Rightarrow x_s(t) = x(t) * \delta_{T_s}(t)$$

(sampled signal)

$$\leftrightarrow \frac{1}{2\pi} X(\omega) * \delta_{T_s}(\omega)$$

$$= \frac{1}{2\pi} X(\omega) * \omega_s \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_s)$$

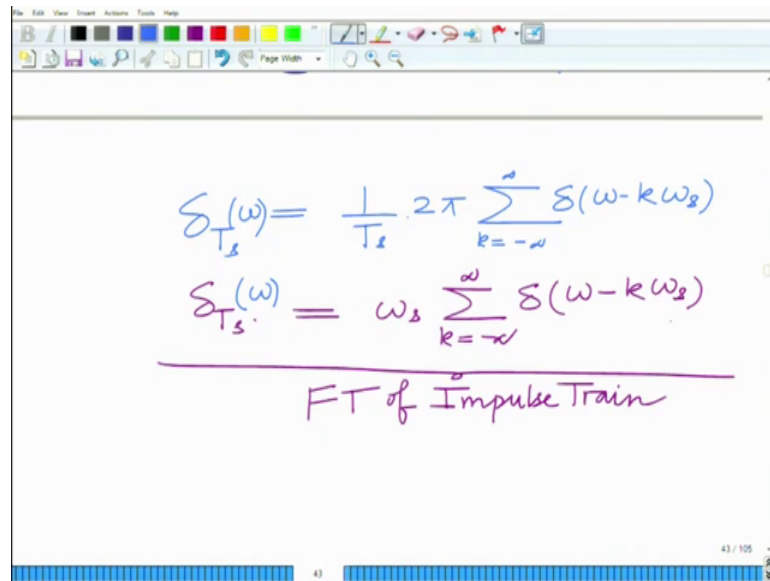

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$$= \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X(\omega) * \delta(\omega - k\omega_s)$$

The spectrum of the sampled signal can be derived as follows. So, we have removing all the property  $x_1(t)$  multiplied by  $x_2(t)$ , the corresponding Fourier transform is  $\frac{1}{2\pi} X_1(\omega) * X_2(\omega)$ . Now, therefore now recall that our sampled signal  $x_s(t)$  is the product of the original signal  $x(t)$  with the impulse train that is  $\delta_{T_s}(t)$  ok.

So, this is your sampled signal, which implies its Fourier transform is going to be  $X(\omega)$  of  $\omega$ . In fact, there has to be a constant factor of  $\frac{1}{2\pi} \frac{1}{2\pi} X(\omega)$  of  $\omega$  convolved with the spectrum  $\delta_{T_s}$ , let us call this  $\delta_{T_s}$  of  $\omega$  or whatever.

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The image shows a handwritten derivation on a whiteboard. The first equation is  $S_{T_s}(\omega) = \frac{1}{T_s} \cdot 2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_s)$ . The second equation is  $S_{T_s}(\omega) = \omega_s \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_s)$ . A horizontal line is drawn below the second equation, and the text "FT of Impulse Train" is written below the line. The whiteboard has a toolbar at the top and a status bar at the bottom showing "43 / 105".

$$S_{T_s}(\omega) = \frac{1}{T_s} \cdot 2\pi \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_s)$$
$$S_{T_s}(\omega) = \omega_s \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_s)$$

FT of Impulse Train

So, convolved with the spectrum of the impulse train that is  $\delta T_s \omega$ , which is basically equal to  $\frac{1}{2\pi} \sum_{k=-\infty}^{\infty} \delta(\omega - k\omega_s)$  and I am sorry  $\delta \omega$  summation  $k$  equal to minus infinity to infinity  $\delta(\omega - k\omega_s)$ ; now which is basically  $\omega_s$  or  $2\pi$  now  $\omega_s$  or  $2\pi$  remember that is  $f_s$  which is  $\frac{1}{T_s}$ .

So, this is  $\frac{1}{T_s}$  taking this since the convolution operator is associative taking the convolution operator inside,  $x_n$  convolved  $x$   $\omega$  convolved with  $\delta(\omega - k\omega_s)$ , which is equal to now  $X$  of  $\omega$  convolved with  $\delta(\omega - k\omega_s)$  is simply when you convolve with shifted impulse by  $k\omega_s$ , it is simply shifting the function that is  $X$  of  $\omega$  shifted to  $\omega_s$  that is we get  $X$  of  $\omega - k\omega_s$ .



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$$= \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X(\omega) * \delta(\omega - k\omega_s)$$

$$X_{T_s}(\omega) = \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X(\omega - k\omega_s)$$

Spectrum of sampled signal.

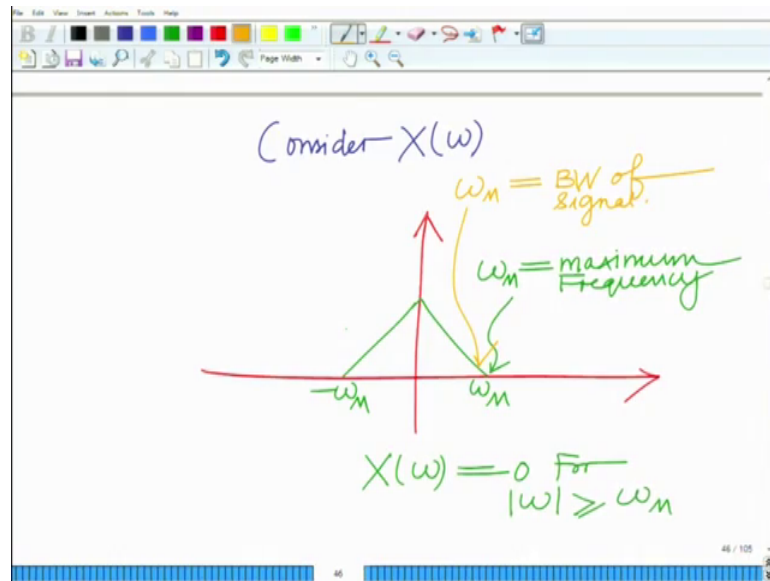
Shifted & Added spectra of original signal.  $x(t)$ .

So, this will be basically  $\frac{1}{T_s}$  summation  $k$  equal to minus infinity to infinity  $X$  of  $\omega$  minus  $k$   $\omega_s$ , this is the spectrum of the this is the spectrum of the sampled signal spectrum of the sampled signal signal sampled by the impulse train.

There is a spectrum of the sampled signal and you can see this basically shifting and adding. So, what you are doing is  $X$  of  $\omega$  you are shifting it by each integral multiple of  $\omega_s$ , that is you are shifting it by  $\omega_s$  that is  $\omega$  minus  $\omega_s$  shifting it by  $2\omega_s$  that is  $X$  of  $\omega$  minus  $2\omega_s$ . Similarly you are shifting it on the negative side that is  $X$  of  $\omega$  plus  $2\omega_s$   $X$  of  $\omega$  plus  $\omega_s$  so on. And you are basically adding all these shifted copies of the original spectrum externally.

So, basically the summation or some shifted and added spectra of the original message signal  $x$  the some of the shifted and so these are replicas right. These are replicas of the original spectrum  $\omega_s$  shifted by every integral multiple of the sampling frequency  $\omega_s$  and then all these shifted replicas are basically added ok.

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Now, consider now let me describe this pictorially describe this process pictorially, consider  $X$  of  $\omega$ ;  $X$  of  $\omega$  given as shown below and we have let us say we have a maximum a spectrum that looks something like this is  $\omega_m$ . So,  $X$  of  $\omega$  equal to 0 for  $\omega$  magnitude of  $\omega$  greater than equal to  $\omega_m$ , that is either  $\omega$  greater than equal to  $\omega_m$  or minus  $\omega$  less than or equal to minus  $\omega_m$  ok.

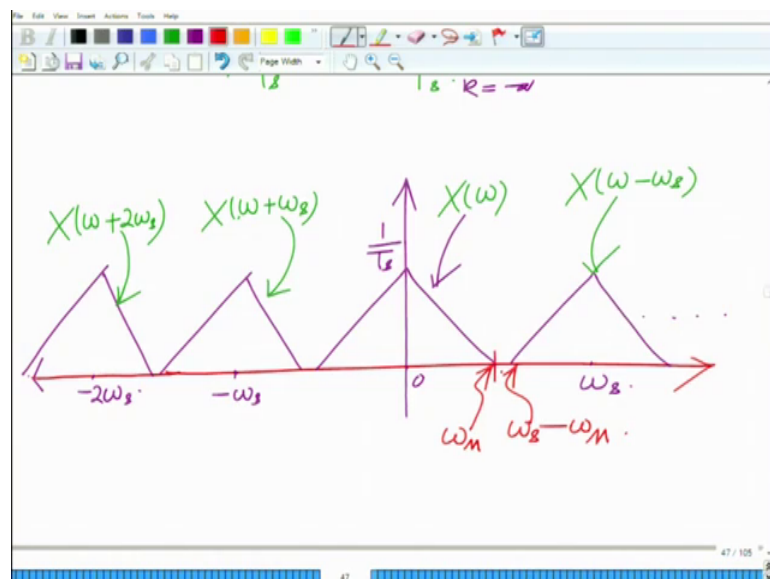
So,  $\omega$  is so the spectrum is 0. So, what we are saying is this  $\omega_m$  this is equal to the maximum frequency. So, this is the maximum frequency of the signal  $\omega$  that is  $X$  of  $\omega$  is 0 when  $\omega$  is greater than  $\omega_m$  as well as when  $\omega$  is less than minus  $\omega_m$ . So,  $X$  of  $\omega$  is restricted to this band to  $\omega_m$  to  $\omega_m$ . So, you can say  $\omega_m$  is the bandwidth of the signal ok, basically  $\omega_m$  is the bandwidth of the signal the signal is 0 outside of this bandwidth.

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The screenshot shows a presentation slide with a white background and a blue border. At the top, there is a toolbar with various icons. Below the toolbar, the handwritten equation is:
$$X_{T_s}(\omega) = \frac{1}{T_s} \sum_{k=-\infty}^{\infty} X(\omega - k\omega_s)$$
 Above the equation, the handwritten text  $|\omega| \geq \omega_m$  is visible. The slide number '47' is at the bottom center, and '47 / 105' is at the bottom right.

Now if you consider this signal which is x of the spectrum of the sampled signal, which is basically scaled by 1 over 2 s let us say the original signal is scaled unity. So, basically this is 1 over T s summation k equals minus infinity to infinity x of omega minus k omega s that as we said consists of shifted replicas of the original Shifted replicas of the original signal.

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So, you have one, you have another that is shifted you have another that is shifted and you have another that is shifted by, you have another that is shifted by 2 time or shifted

to minus 2 omega s. So, this is shifted to omega s this is 0 this is minus omega s minus 2 omega. So, this is your original copy each is scaled by 1 or Ts this is your original copy that is X of omega this is X of omega minus omega s and then you will have several copies over here.

Similarly this is your X of omega plus omega s this is X of omega plus 2 omega s and so on and now, if you look at it something interesting happens if you look at this point, this is omega m and this point of the shifted 1 is omega s minus omega m. Now you look at this gap if omega s minus omega m is greater than or equal to omega m then you will have this gap correct.

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$$\omega_s - \omega_m \geq \omega_m$$

$$\Rightarrow \boxed{\omega_s \geq 2\omega_m}$$

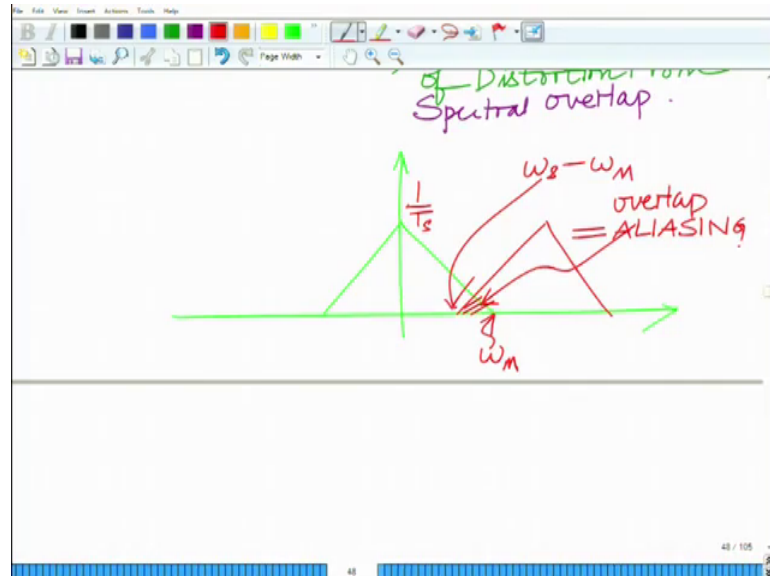
$\Rightarrow$  No interference of Distortion From Spectral overlap.

Which implies that these 2 spectral copies corresponding to X of omega and X of omega minus omega is these 2 do not overlap. As a result they will not interfere with each other; that will happen when there is a gap. And for there to be this gap or this guard band between the original spectrum and its various shifted copies, this condition has to be satisfied that is omega s minus omega the maximum frequency has to be greater than or equal to omega. Which implies basically now the condition that omega s greater than or equal to 2 omega and this quantity omega is a sampling frequency omega m is the maximum frequency.

So, if this is satisfied this implies there is no interference from the spectral overlap, no interference or distortion from spectral overlap. However, if this condition is not satisfied

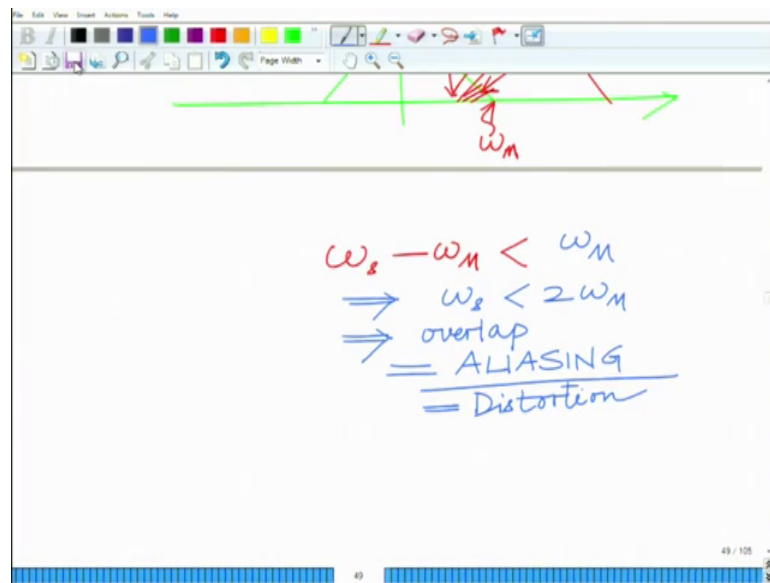
that is if  $\omega_s - \omega_m$  is or  $\omega_m$  is greater than  $\omega_s - \omega_m$ , if this condition is not satisfied and then you can see what happens.

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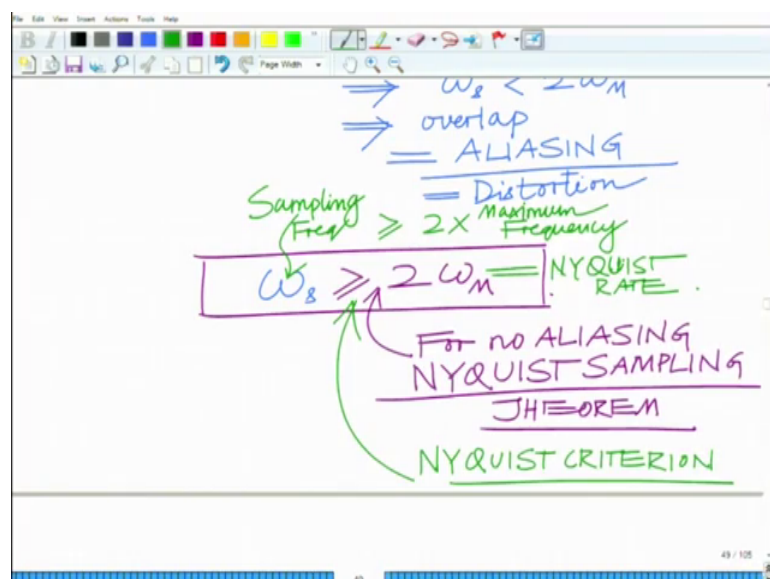
If this condition is not satisfied, then you will have your original copy that is your  $\frac{1}{T_s}$  and then you will have another copy. So, this is your  $\omega_m$  and this is your  $\omega_s - \omega_m$  which is less than  $\omega_m$ . So, these quantities where there is overlap and that causes distortion which is termed as aliasing. So, what happens in this scenario is that  $\omega_s - \omega_m$  is less than  $\omega_m$ , implies  $\omega_s$  is less than twice  $\omega_m$  implies overlap which is basically equal to aliasing this is the distortion.

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So, this basically in this spectral copies in the overlap superimpose over each other that leads to distortion of the original spectrum, then you cannot recover the original message spectrum by any kind of filtering operation. Hence this is termed as aliasing this distortion which is resulting from under sampling, remember you will have no distortion if  $\omega_s$  is greater than or equal to twice  $\omega_M$   $\omega_s$  is less than twice  $\omega_M$  you will have this distortion which is termed as aliasing and this criteria for no aliasing this is termed as the Nyquist sampling theorem or the Nyquist criterion for no aliasing ok, so for low aliasing. So, we need  $\omega_s$  greater than or equal to 2  $\omega_M$ .

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This is the criterion that we need for no distortion and this is done as a Nyquist sampling theorem, Nyquist sampling theorem or the Nyquist criterion this termed a Nyquist sampling theorem or this termed as the Nyquist criterion ok.

So, for no overlap so for no distortion no spectral distortion we need to have this sampling frequency greater than or equal to twice into the maximum frequency of the. So, the sampling frequency has to be greater than or equal to the sampling frequency. So, to avoid aliasing and this is also termed as the Nyquist rate this  $2\omega_m$  this also termed as the Nyquist, the Nyquist rate or basically the minimum rate at which the signal has to be sampled.

So, that there is no distortion this is a very important property 1 of the key properties all right, which one of the key properties which has to be kept in mind which has to be borne in mind; when converting a signal from the analog to the discrete time or the continuous time to the discrete time domain why our sampling. So, if the sampling frequency all right to avoid aliasing the sampling frequency  $\omega_s$  has to satisfy this key criterion.

So, this minimum threshold that  $\omega_s$  must be greater than twice the maximum frequency of this signal, if it is less than twice the maximum frequency then there is going to be distortion this is termed as aliasing. And therefore, this quantity  $2\omega_m$  is termed as the Nyquist the Nyquist rate of sampling a signal. And this is the minimum rate of sampling a signal bad limited signal to avoid aliasing distortion.

So, I will stop here and continue in the subsequent modules.

Thank you very much.