

Principles of Communication Systems – Part II
Prof. Aditya K. Jagannatham
Department of Electrical Engineering
Indian Institute of Technology, Kanpur

Lecture – 24

**M-ary PSK (Phase Shift Keying) – Part I, Introduction, Transmitted
Waveform, Constellation Diagram**

Hello. Welcome to another module in this massive open online course. So, today in this module we are going to look at a different modulation scheme that is M-ary PSK or M-ary phase shift keying, similar to M-ary QAM which is quadrature amplitude modulation and as M symbols. A M-ary PSK or phase shift keying has M symbols and it is a phase shift keying based digital modulation scheme.

(Refer Slide Time: 00:41)

At high SNR
very small in
comparison to $P_{e,PAM}$

$$P_{e,QAM} = 4 \left(1 - \frac{1}{\sqrt{M}} \right) Q \left(\sqrt{\frac{3E_s}{N_0(M-1)}} \right)$$

Err rate for overall
M-ary QAM.

But, before we proceed further just want to make a minor correction in the overall probability of symbol error derived for M-ary QAM last time there has to be a factor of two. So, this twice as we have seen the problem overall probability of error is twice the probability of error approximately twice the probability of error for the constituent QAM; so for the constituent M-ary PAM. So, this has to be 4, because the probability of error for the constituent PAM is a factor of two. So, this is twice that it will have a factor of 4 this is 4.

(Refer Slide Time: 01:30)

$$P_e = 2\left(1 - \frac{1}{\sqrt{M}}\right)Q\left(\sqrt{\frac{3E_s}{N_s(M-1)}}\right)$$

Average symbol Error rate for each constituent M -ary PAM.

$P_{e,QAM} \approx 2P_e$ as shown below.

Error rate for overall M -ary

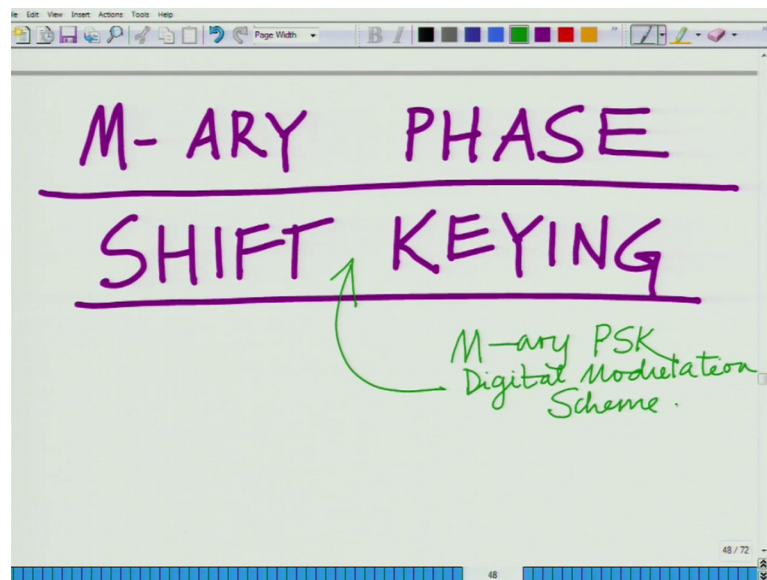
This is a factor of 4 this is twice the probability of probability error of PAM or P_e , QAM approximately twice P_e as shown below.

(Refer Slide Time: 01:57)

QAM - widely used modulation Schemes

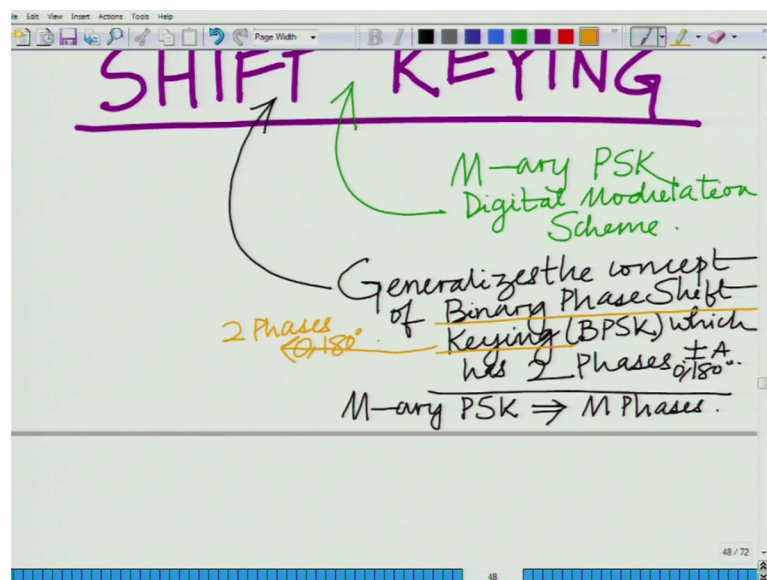
Anyway, let us move on to the modulation scheme under consideration which is yet another modulation scheme a digital modulation scheme which is termed as M -ary phase that is your M -ary PSK.

(Refer Slide Time: 02:06)



This is another digital modulation scheme and again this also generalizes the concept of binary phase shift keying which has 2 phases.

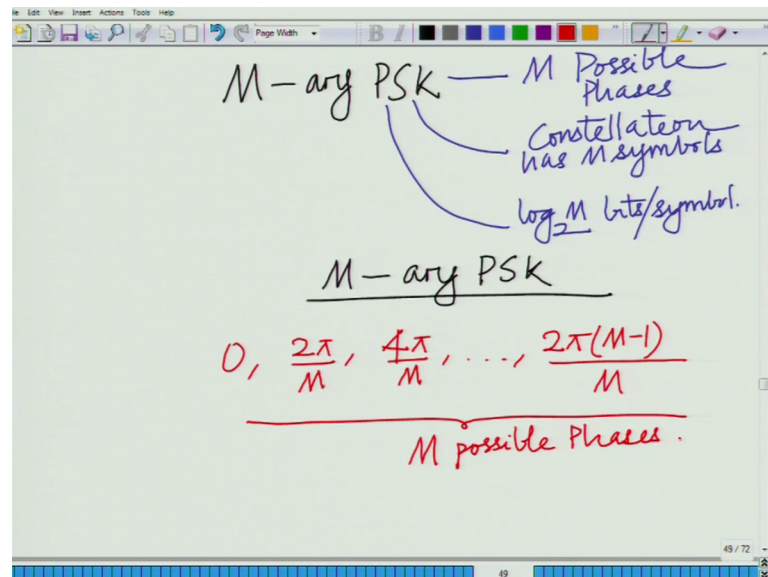
(Refer Slide Time: 03:01)



Generalizes the concept of binary phase shift keying BPSK, binary phase shift keying or BPSK which has 2 phases and it generalizes to M-ary phase shift keying. In M-ary phase shift keying basically there are M phases. So, in binary phase shift keying there are 2 phases plus or minus A that is 0 comma 180 degrees. So, if you look at binary phase shift keying there are only 2 phases which are 0 comma 180 degree.

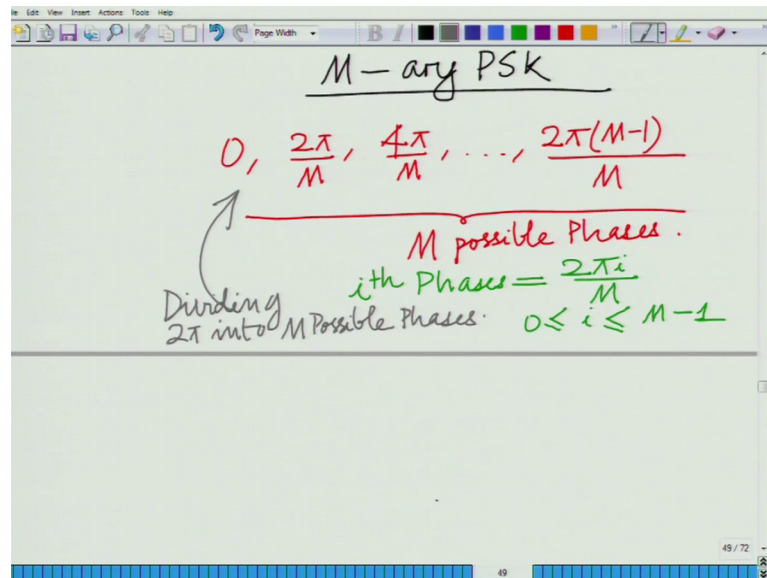
So, M-ary phase shift keying generalizes concept of binary phase shift keying and allows for M phases, that is symbol can have one of M phases and this worth this as M symbols M possible symbols or M possible symbols in the constellation and therefore, once again similar to M-ary QAM one can communicate $\log_2 M$ bits per symbol.

(Refer Slide Time: 04:48)



So, there are M possible phases, M -ary PSK M possible symbols implies the constellation has M symbols implies we have $\log_2 M$ bits per symbol. We can communicate $\log_2 M$ bits per symbol and the M possible phases in M -ary PSK the M possible phases are $0, 2\pi$ by $M, 4\pi$ by M and so on so force until 2π M minus 1 by M . So, these are your M these are the M possible phases. Basically, the M a the i -th phase you can write it as $2\pi i$ by M where $0 \leq i \leq M-1$.

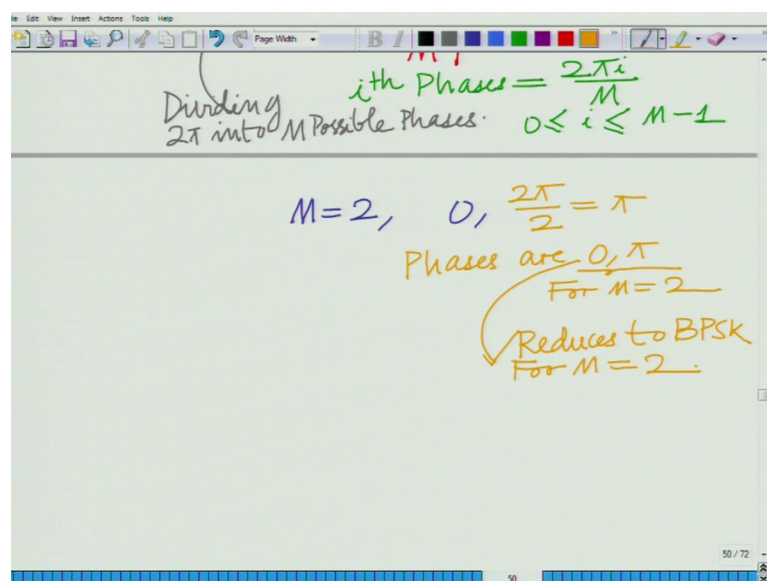
(Refer Slide Time: 06:24)



So, we are taking 2π remember 2π corresponds to 360 degrees and we are dividing it into M possible phases. So, basically what we are doing is dividing 2π into M possible phases dividing 2π into M possible phases.

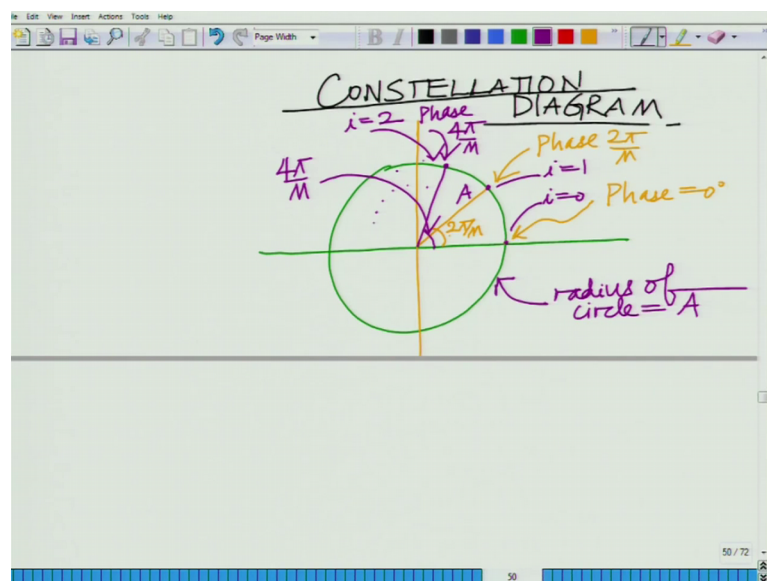
Now, of course, is M equal to 2 again you see if M is equal to 2, for M equal to 2 the phases are 0 comma 2π where i is equal to i divided by M , 2π divided by 2 equal to π .

(Refer Slide Time: 07:18)



So, phases are 0 to 2π for M equal to 2 , which means it reduces to binary phase shift keying. So, if you consider M equal to 2 the phases are again 0 and π that is 0 and 180 degrees. So, for M equal to 2 this reduces to binary phase shift keying and in general you can have any M . So, that is M -ary phase shift keying. So, we are generalizing binary phase shift keying which corresponds to M equal to 2 phases to M -ary phase shift keying which contains one of M possible phases that is symbols can have one of M possible phases that is the whole idea. So, this reduces to M -ary PSK for M is equal to 2 .

(Refer Slide Time: 08:35)

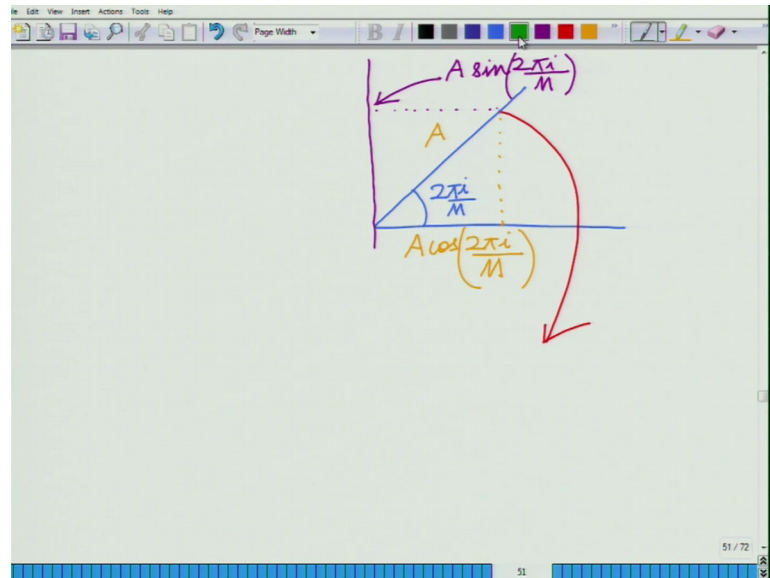


Now, let us look at the constellation diagram as I have said we are taking a phase of 2π we are taking the phase remember phase of 2π corresponds to 360 degree. So, you can represent it using a circle, now the first symbol is at phase 0 second symbol is at phase 2π by M this is let say 2π by M , third symbol is at phase right the phase difference between the symbols. So, third symbol is at phase 4π by M this is phase 4π by M and so on and the radius of the circle.

So, radius of circle equals A ; the radius. So, these constellation point it is a constellation. So, this is another constellation point this is the 0 constellation point corresponding to i equal to 0 constellations. So, this corresponds to i equal to 0 this corresponds to i equal to 1 , this corresponds to i equal to 2 . So, the i -th point where i lies between 0 and M minus 1 is at a phase $2\pi i$ over M , on this circle of radius a , I can take values from 0 to M

minus 1 therefore, there are total of M points each is phase shifted from the previous 1 by 2π by M . So, we are taking the phase total phase of 2π 360 degrees and dividing it into M phases and therefore, now if you look at the i -th point.

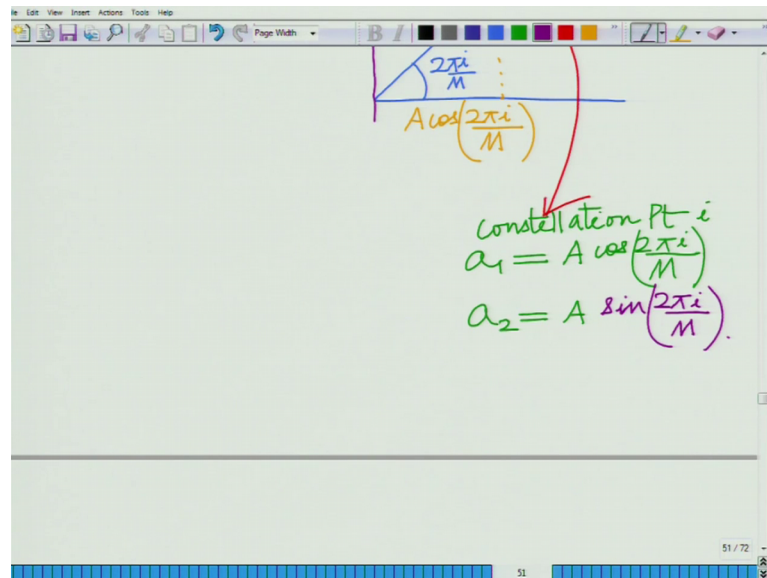
(Refer Slide Time: 11:10)



If you look at the i -th point corresponding to phase. So, if you look at the i -th point $2\pi i$ over M radius is A .

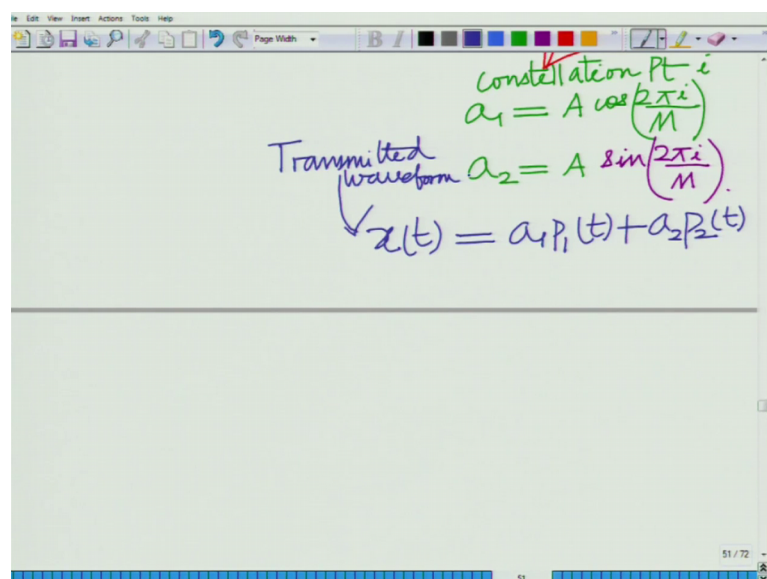
So, if you look at this the x coordinate is going to be $A \cos 2\pi i$ over M , and if you look at the y coordinate that is going to be $A \sin 2\pi i$ by M . So, this can be represented as symbols 2 symbols the 1 is the x coordinate. So, each constellation point.

(Refer Slide Time: 12:05)



So, this is constellation point i. So, this can be represented as a combination of 2 symbols a_1 is the x coordinate which is $A \cos 2\pi i$ divided by M , a_2 equals $A \sin 2\pi i$ divided by M . So, it is a combination of 2 coordinates a_1 and a_2 each correspond to, now you can say the each of these corresponds to basically on each of these is transmitted along, we have 2 symbols right similar to QAM is a combination of 2 symbols, each is transmitted along an orthonormal basis function of the signal space.

(Refer Slide Time: 13:09)



So, the net transmitted waveform $x(t)$ can be $x(t) = 1$. So, this is similar to QAM. So, net waveform.

(Refer Slide Time: 13:37)

The image shows a digital whiteboard with handwritten mathematical expressions. At the top, two sets are defined:

$$a_1 \in \left\{ A \cos\left(\frac{2\pi k}{M}\right), k=0, 1, \dots, M-1 \right\}$$

$$a_2 \in \left\{ A \sin\left(\frac{2\pi k}{M}\right), k=0, 1, \dots, M-1 \right\}$$

Below these, the functions $p_1(t)$ and $p_2(t)$ are written and underlined. An arrow points from the text "orthonormal basis functions for signal space." to these underlined functions.

So, this is your transmitted waveform where a_1 belongs to the set $A \cos$ or let us use the index k $A \cos 2\pi k$ by M . I am simply changing the index to k , k equal to 0 1 up to M minus 1 and a_2 or a_2 element belongs to $A \sin 2\pi k$ by M , k equal to 0 1 up to M minus 1. So, we are transmitting and of course, as we have seen before $p_1(t)$ and $p_2(t)$ these are orthonormal basis functions for the signal space.

These are the orthonormal basis functions for the of the signal space.

(Refer Slide Time: 15:10)

$$a_2 \in \left\{ A \sin\left(\frac{2\pi k}{M}\right), k=0, 1, \dots, M-1 \right\}$$

$P_1(t), P_2(t)$ orthonormal basis functions for signal space.

$$P_1(t) = \sqrt{\frac{2}{T}} \cos(2\pi F_c t)$$

$$P_2(t) = \sqrt{\frac{2}{T}} \sin(2\pi F_c t)$$

$$0 \leq t \leq T$$

$$T = \frac{k}{F_c}$$

For instance we have said that is a $P_1(t)$ equals well again unit norm that is unit energy square root of 2 or $T \cos(2\pi F_c t)$, $P_2(t)$ is square root of 2 or $T \sin(2\pi F_c t)$ where $0 \leq t \leq T$ the symbol duration and T correct T equals k over F_c or basically contains an integer number of cycles so, that is basically what we have seen before also.

(Refer Slide Time: 16:06)

$$P_2(t) = \sqrt{\frac{2}{T}} \sin(2\pi F_c t)$$

$$0 \leq t \leq T$$

$$T = \frac{k}{F_c}$$

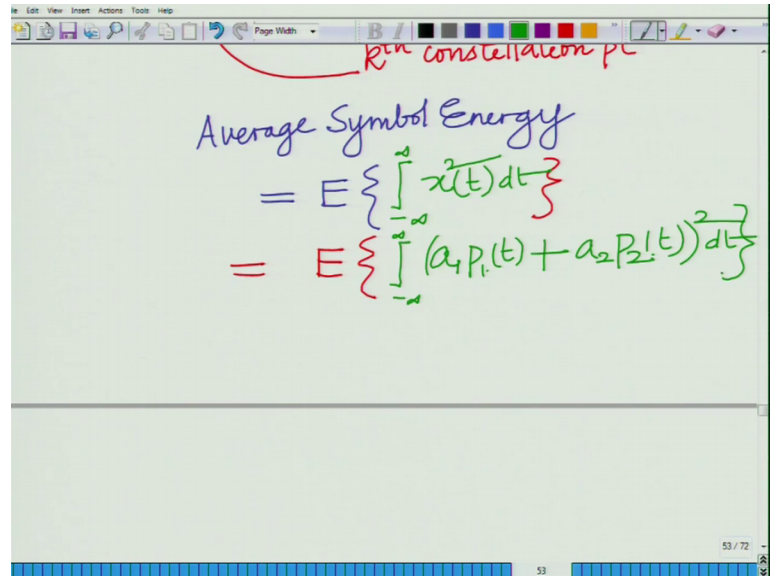
$$S_k = \left(A \cos\left(\frac{2\pi k}{M}\right), A \sin\left(\frac{2\pi k}{M}\right) \right)$$

k^{th} constellation pt

And the k -th symbol remember we have said k -th symbol is A well times $\cos(2\pi k/M)$, $A \sin(2\pi k/M)$. So, the k -th symbol or basically the k -th constellation point is $A \cos(2\pi k/M)$

πk by M , $A \sin 2 \pi k$ by M . So, this is your k -th constellation point this is the k -th constellation point and now if you look at.

(Refer Slide Time: 17:06)



The image shows a digital whiteboard with handwritten mathematical expressions. At the top, a red arrow points from the text "kth constellation pt" to a point on a horizontal axis. Below this, the text "Average Symbol Energy" is written in blue. The derivation follows:
$$= E \left\{ \int_{-\infty}^{\infty} x^2(t) dt \right\}$$

$$= E \left\{ \int_{-\infty}^{\infty} (a_1 p_1(t) + a_2 p_2(t))^2 dt \right\}$$
The whiteboard interface includes a menu bar at the top with options like Edit, View, Insert, Actions, and Tools. A toolbar with various drawing tools is visible below the menu. The bottom of the whiteboard shows a status bar with the page number "53 / 72".

Now, if you look at the average energy average symbol energy that is equal to expected value of, well expected value of you can say average energy of the waveform expected value of x^2 , expected value of or expected or energy of a t^2 or expected value of you can write this as expected value of integral minus infinity to infinity, just be a bit more elaborate minus infinity to infinity $x^2 dt$ which is equal to well let us write it this way minus infinity to infinity, $a_1^2 p_1(t) + a_2^2 p_2(t)$ whole square dt .

(Refer Slide Time: 18:39)

$$= E \left\{ a_1^2 \int_{-\infty}^{\infty} P_1(t) dt + a_2^2 \int_{-\infty}^{\infty} P_2(t) dt + 2a_1a_2 \int_{-\infty}^{\infty} P_1(t)P_2(t) dt \right\}$$

Inner Product
Since pulses are orthogonal.

Well, which is equal to the expected value of minus infinity to infinity, a 1 square P 1 square t d t plus a 2 square integral minus infinity to infinity, P 2 square t d t plus twice a 1 a 2 integral minus infinity to infinity, P 1 t, P 2 t, d t of course this is nothing but energy of pulse E p, energy of pulse E p and this thing is nothing but integral minus infinity to infinity P 2 t, d t this is 0 this is nothing but the inner product of P 1 t, P 2 t pulses are orthogonal since pulses are orthogonal in fact, they are orthonormal.

(Refer Slide Time: 19:57)

$$= E \{ a_1^2 \epsilon_p + a_2^2 \epsilon_p \}$$

$$= E \{ a_1^2 + a_2^2 \} \quad \left\{ \begin{array}{l} \epsilon_p = 1 \\ \text{unit energy pulse} \end{array} \right.$$

$$= E \left\{ \left(A \cos\left(\frac{2\pi k}{N}\right) \right)^2 + \left(A \sin\left(\frac{2\pi k}{N}\right) \right)^2 \right\}$$

So, this is equal to expected value of and you will observe something very interesting over here this equal to expected value of a naught square $E p$, plus a 1 square $E p$, well $E p$ equal to 1.

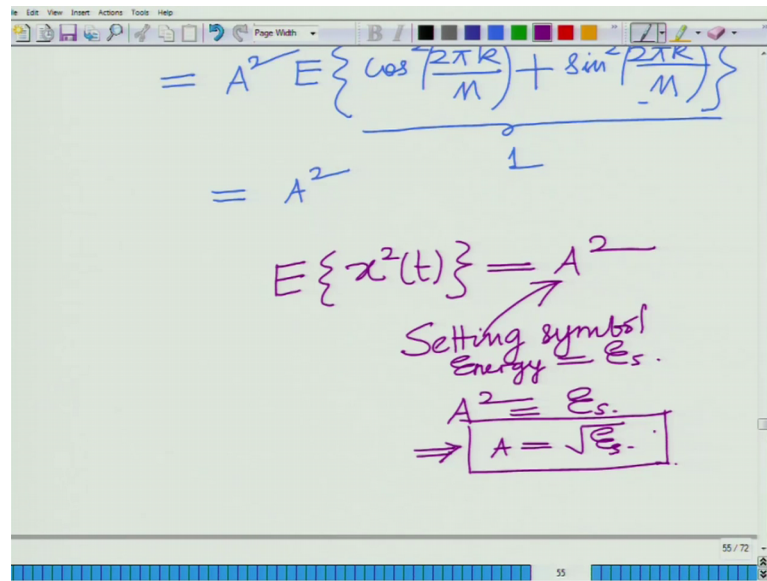
So, this is equal to expected value that energy of the pulse is equal to 1, a naught square plus a 1 square $E p$ equal to 1 that is we are considering a unit energy pulse. Now this is equal to remember a naught I am sorry this is not a naught rather a 1 square plus a 2 square, and remember a 1 is $A \cos 2 \pi k \text{ by } M \text{ whole square}$ plus $A \sin 2 \pi k \text{ by } M \text{ whole square}$ which is equal to of course, $A \text{ square}$ is a constant expected value of cosine square $2 \pi k \text{ by } M$, plus sine square $2 \pi k \text{ by } M$ this is of course 1.

(Refer Slide Time: 21:29)

The image shows a handwritten derivation on a presentation slide. The top line is
$$= E \left\{ \left(A \cos \left(\frac{2\pi k}{M} \right) \right)^2 + \left(A \sin \left(\frac{2\pi k}{M} \right) \right)^2 \right\}$$
 The middle line is
$$= A^2 E \left\{ \cos^2 \left(\frac{2\pi k}{M} \right) + \sin^2 \left(\frac{2\pi k}{M} \right) \right\}$$
 The bottom line is
$$= A^2$$
 with a '1' written below the brace in the previous line, indicating that the expression inside the brace equals 1. The slide has a toolbar at the top and a status bar at the bottom showing '55 / 72'.

So, this is equal to a square.

(Refer Slide Time: 21:59)

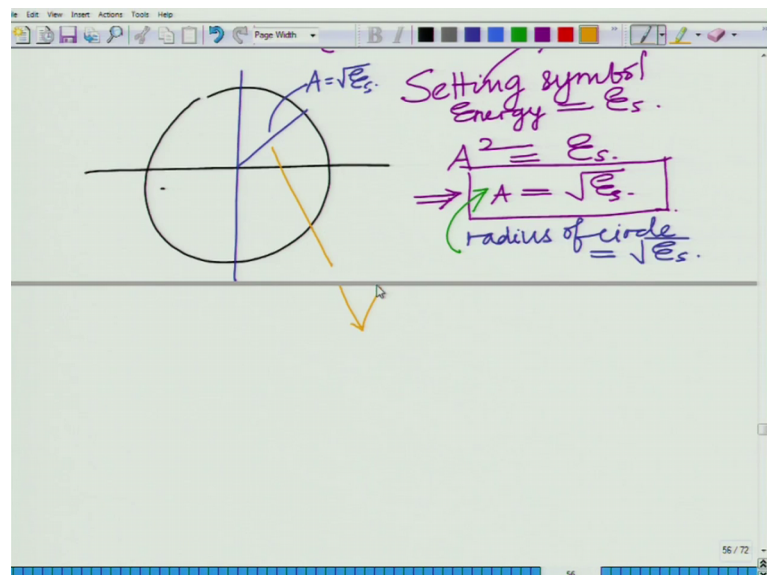

$$= A^2 E \left\{ \cos^2 \left(\frac{2\pi k}{n} \right) + \sin^2 \left(\frac{2\pi k}{n} \right) \right\}$$
$$= A^2$$
$$E \{ x^2(t) \} = A^2$$

Setting symbol Energy = E_s .

$$A^2 = E_s$$
$$\Rightarrow A = \sqrt{E_s}$$

Basically, what we have is if you look at this, the expected value average symbol energy or expected value of x^2 or expected value of x square equals A^2 setting the symbol energy equal to E_s , we have A^2 equal to E_s which implies A is equal to square root of E_s .

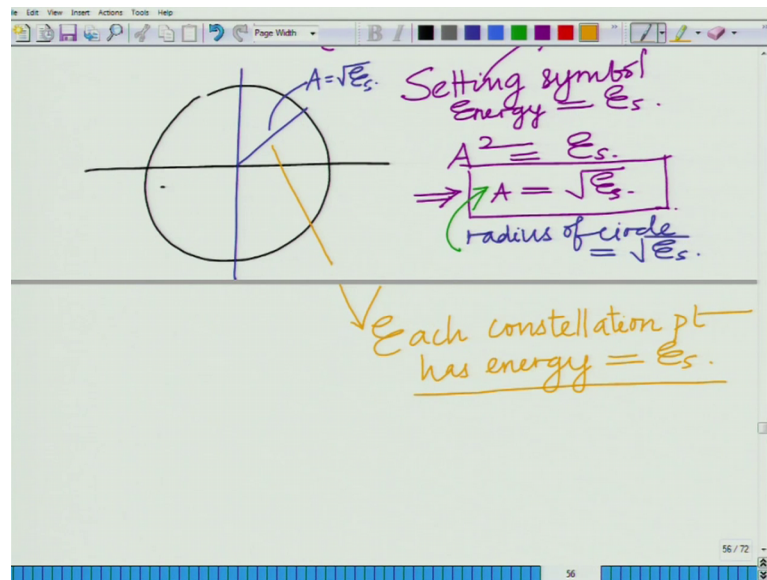
(Refer Slide Time: 22:55)



So, the radius of this circle; the radius of the circle of the constellation: so if you look at the constellation the circle of the constellation correct the constellation is a circle the radius of that circle is square root of E_s . And in fact, all symbols have unlike QAM or

per (Refer Time: 23:06) the average energy is E_s , here not only the average energy, but actually the energy of each symbol if you observe it because the radius is constant the energy of each symbol is E_s . So, that is something interesting about the M-ary. So, if you look at the M-ary.

(Refer Slide Time: 23:51)



So, A is equal to square root of E_s that is radius of circle and you can observe that each point has identical energy. Unlike PAM, each constellation point has not only average energy, but each constellation point actually has energy E_s . So, that is basically your M-ary PSK constellation.

So, M-ary PSK constellation consists of M points arranged around a circle of radius square root of E_s and each constellation point differs from the previous one by phase shift of 2π by M , and the i -th point of the k -th point is given as $A \cos(2\pi k/M)$ or square root of $E_s \cos(2\pi k/M)$ and $A \sin(2\pi k/M)$ or square root of $E_s \sin(2\pi k/M)$. So, we can say since A is square root of E_s with square root of $E_s \cos(2\pi k/M)$ and A is a combination of 2 combination of 2 , basically coordinate that is 1 which is square root of $E_s \cos(2\pi k/M)$ and 2 which is square root of $E_s \sin(2\pi k/M)$.

So, will stop here and look at the receiver design, the detection and also the ensuing probability of error and other aspects in subsequent modules.

Thank you very much.