

**Principles of Communication- Part I**  
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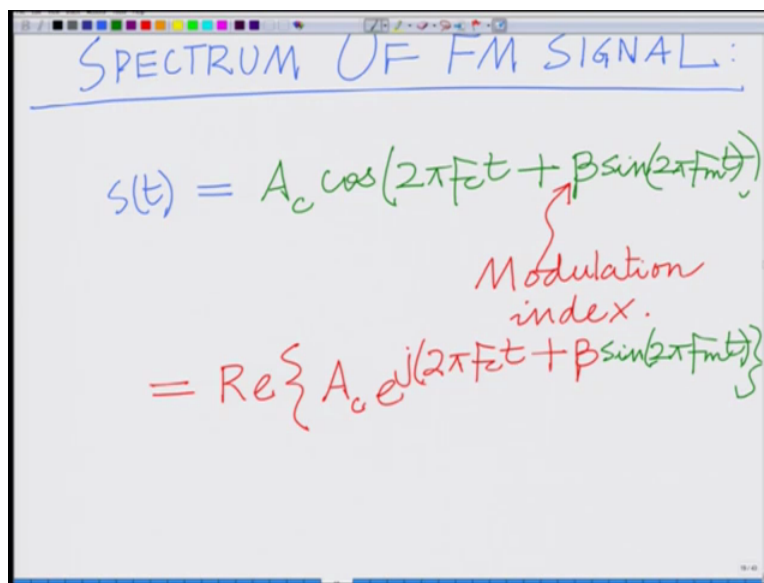
**Module No 6**

**Lecture 32**

**Spectrum of Frequency Modulated (FM) Signals**

Hello welcome to another module in this massive open online course, so this module let us start looking at the spectrum of the FM signal, alright.

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SPECTRUM OF FM SIGNAL:

$$s(t) = A_c \cos(2\pi f_c t + \beta \sin(2\pi f_m t))$$

Modulation index.

$$= \text{Re}\{A_c e^{j(2\pi f_c t + \beta \sin(2\pi f_m t))}\}$$

So we are looking at frequency modulation let us now start looking at the spectrum of a spectrum of an of an Am signal, so we want to look at the spectrum or spectrum of a FM signal. Spectrum of an FM signal, alright. Let us write our FM signal  $s(t)$  equals  $A_c$  let us write this as  $A_c \cosine 2\pi f_c t$  plus  $\beta \sin 2\pi f_m t$  as we have seen where  $\beta$  is the where  $\beta$  is our modulation index, okay. So  $\beta$  is our this is our you have seen this before  $\beta$  is the modulation index. Now what I am going to do I am going to write it as this, I can write this as the real part of  $A_c e$  to the power of  $j 2\pi f_c t$  times  $\beta \sin 2\pi f_m t$  which is equal to well.

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The image shows a whiteboard with handwritten mathematical equations. At the top, it says  $= \text{Re} \{ A_c e^{j(2\pi F_c t + \beta \sin 2\pi F_m t)} \}$ . Below this, it is rewritten as  $= \text{Re} \{ A_c e^{j\beta \sin 2\pi F_m t} e^{j2\pi F_c t} \}$ . A green arrow points from the term  $e^{j\beta \sin 2\pi F_m t}$  to the definition  $x_b(t) = \text{Re} \{ x_b(t) e^{j2\pi F_c t} \}$ . Below this definition, it says "Equivalent complex Baseband Signal of FM Signal".

Now I can expand this as so first I am writing  $A_c \cos 2\pi F_c t + \beta \sin 2\pi F_m t$  as the real part of  $A_c e^{j(2\pi F_c t + \beta \sin 2\pi F_m t)}$ . Now I can split this as  $A_c e^{j2\pi F_c t}$  times  $e^{j\beta \sin 2\pi F_m t}$ .

And we know that whenever I write a signal in this session that is real part of sum signal times  $e^{j2\pi F_c t}$  this becomes this is the complex baseband equivalent signal, okay. So we know that when we have a pass band signal  $x_p(t)$   $x_p(t)$  is the real part of  $x_b(t)$  which is a complex baseband equivalent signal times  $e^{j2\pi F_c t}$ , okay. So this is basically your equivalent complex baseband signal this is the equivalent complex baseband signal this is your equivalent complex baseband signal of the FM signal of the FM signal, okay.

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$$x_p(t) = \text{Re} \{ x_b(t) e^{j2\pi f_m t} \}$$

Equivalent complex Baseband Signal of Fm Signal

$$S_b(t) = A_c e^{j\beta \sin(2\pi f_m t)}$$

Find Spectrum of This signal.

And therefore the complex baseband signal is well, that is your  $S_b(t)$  equals  $A_c e^{j\beta \sin 2\pi f_m t}$ , now what I can do is I can equivalently find the spectrum of the complex baseband so I can find the baseband spectrum then the pass band spectrum is nothing but simply a translation of the baseband spectrum, okay. So I am first first going to find the spectrum of this complex baseband signal and basically going to translate it appropriately because modulation is nothing but translation in the frequency domain, okay.

So first let us start by finding the spectrum of this equivalent complex equivalent baseband FM signal, so we will find the spectrum of this signal we will find the spectrum of this signal which is basically now if you look at this that will be, now first look at this signal  $A_c e^{j\beta \sin 2\pi f_m t}$ .

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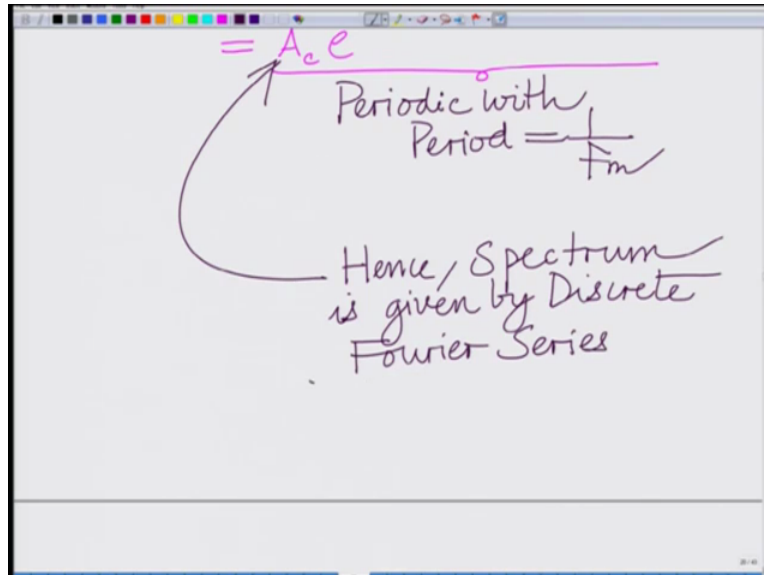
$$\begin{aligned}
 & A_c e^{j\beta \sin(2\pi f_m t + 2\pi k)} \\
 & \text{Periodic - periodic } T = \frac{1}{f_m} \\
 & \downarrow \\
 & A_c e^{j\beta \sin(2\pi f_m(t + \frac{k}{f_m}))} \\
 & = A_c e^{j\beta \sin(2\pi f_m t + 2\pi k)}
 \end{aligned}$$

$$\begin{aligned}
 & \downarrow \\
 & A_c e^{j\beta \sin(2\pi f_m(t + \frac{k}{f_m}))} \\
 & = A_c e^{j\beta \sin(2\pi f_m t + 2\pi k)} \\
 & = A_c e^{j\beta \sin(2\pi f_m t)} \\
 & \text{Periodic with Period } = \frac{1}{f_m}
 \end{aligned}$$

Now realize that this signal is periodic, okay.  $\sin 2\pi f_m t$  is periodic with period well, period is  $T$  equals  $1$  over  $f_m$  naturally  $\sin 2\pi f_m t$  is a periodic signal so  $e$  to the power of  $j\beta \sin 2\pi f_m t$  is also a periodic signal, okay. And with the same period with period  $T$  equal to  $1$  over  $f_m$ , okay. In fact if I look at this I look at at  $kT$  leads later I have  $e$  to the power of  $j\beta \sin 2\pi f_m t$  plus  $2\pi k$  by  $f_m$  that is going to be  $e$  to the (pow)  $A_c e$  to the power of  $e$  to the power of  $j\beta \sin 2\pi f_m t$  plus  $2\pi k$  which is nothing but  $A_c e$  to the power of  $j\beta \sin 2\pi f_m t$  plus  $2\pi k$  is nothing but  $\beta \sin 2\pi f_m t$ , so this is simply  $e$  to the power  $j\beta \sin 2\pi f_m t$ , so this is simply  $e$  to the power  $j\beta \sin 2\pi f_m t$ .

$f_m t$  and therefore this is periodic with period equals with period this is periodic with period equal to 1.

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Therefore now we can find the equivalent spectrum would be now we said that whenever remember that we have a periodic signal the spectrum is given by the Fourier series the discrete Fourier series therefore we have to find the discrete Fourier series hence spectrum will be given by hence spectrum is given by the discrete Fourier series.

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$$S_b(t) = \sum_{n=-\infty}^{\infty} C_n e^{j2\pi n f_m t}$$

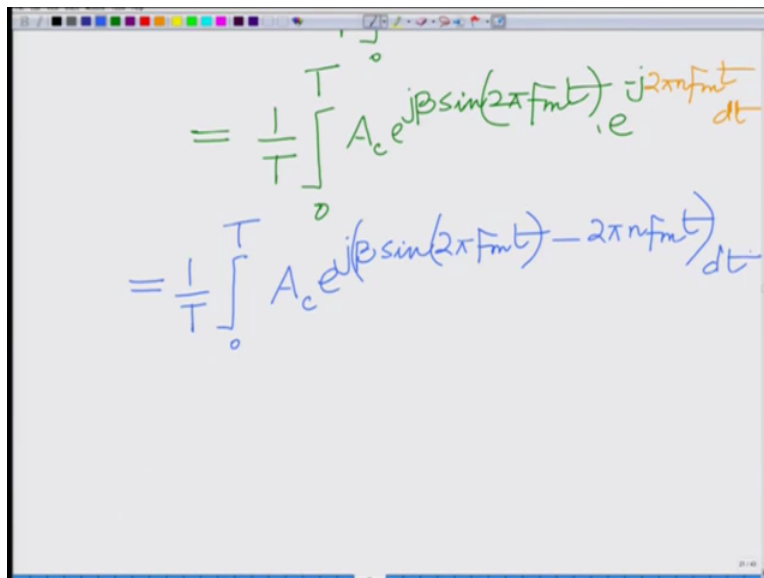
Fundamental Frequency

$$C_n = \frac{1}{T} \int_0^T S_b(t) e^{-j2\pi n f_m t} dt$$

In other words I can write this signal your Sbt which is periodic as summation, remember the discrete Fourier series at the fundamental frequency and the harmonics that is  $C_n e^{j 2\pi n F_m t}$  this  $F_m$  remember we already said  $F_m$  is the fundamental frequency this the fundamental frequency, so I can write it as a as a linear combination in finite summation of harmonics at  $F_m$  and the at  $F_m$  that is frequency  $F_m$  and multiples of  $F_m$ ,  $F_m$  is the fundamental frequency alright. This is the discrete Fourier series this is the representation, alright. (Differ) this is the representation that is the discrete Fourier series representation of your periodic signal, okay.

So now what I have to do is I have to find these coefficients  $C_n$ , these are the coefficients and we also know each  $C_n$  is basically  $1$  over  $T$  where  $T$  is the period times  $0$  to capital  $T$   $A_c e^{j \beta \sin 2\pi F_m t}$  to the power of  $j$  beta sin e to the or for a general baseband signal this is simply your Sbt e to the power of well, minus  $j 2\pi n F_m t$ , okay.

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The image shows a handwritten derivation of the Fourier coefficient  $C_n$  on a digital whiteboard. The derivation consists of two lines of equations. The first line is 
$$= \frac{1}{T} \int_0^T A_c e^{j\beta \sin(2\pi F_m t)} e^{-j 2\pi n F_m t} dt$$
 where the first exponential term is in green and the second is in orange. The second line is 
$$= \frac{1}{T} \int_0^T A_c e^{j(\beta \sin(2\pi F_m t) - 2\pi n F_m t)} dt$$
 where the entire expression is in blue.

The  $n$ th frequency (coef) that is  $C_n$   $n$ th coefficient of the discrete Fourier series which is  $1$  over  $T$   $0$  to  $T$  Sbt  $A_c e^{j \beta \sin 2\pi F_m t}$  times  $e^{-j n F_m t}$  dt is basically  $1$  over  $T$   $0$  to  $T$   $A_c e^{j \beta \sin 2\pi F_m t}$  e to the power of well, let us now use well, I can write this as  $j \beta \sin 2\pi F_m t$  minus  $j \beta$  minus  $F_m t$  into dt.

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$$= \frac{1}{T} \int_0^T A_c e^{j(\beta \sin(2\pi f_m t) - 2\pi n f_m t)} dt$$

$$2\pi f_m t = x$$

$$\frac{dx}{dt} = \frac{dx}{2\pi f_m}$$

$$C_n = \frac{1}{T} \int_0^{2\pi} A_c e^{j(\beta \sin x - nx)} \frac{dx}{2\pi f_m}$$

$$C_n = \frac{1}{T} \int_0^{2\pi} A_c e^{j(\beta \sin x - nx)} \frac{dx}{2\pi f_m}$$

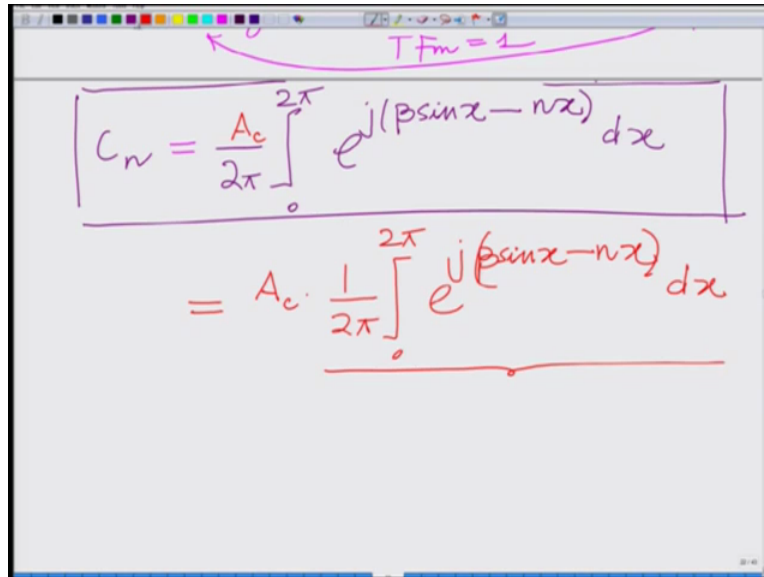
$T f_m = 1$

$$C_n = \frac{1}{2\pi} \int_0^{2\pi} e^{j(\beta \sin x - nx)} dx$$

Now let us employ a substitution I can employ the substitution  $2\pi f_m t$  equal to  $x$  which means that  $dt$  becomes  $dx$  by  $2\pi f_m$  and therefore this integral for  $C_n$  will become  $C_n$  equals well, substitute  $1$  over  $T$  limit becomes  $0$  and well,  $x$  equals  $2\pi f_m t$   $2\pi f_m$  so the upper limit will be  $2\pi f_m$  times capital  $T$  but  $f_m$  equals  $1$  by  $T$ , so this is simply  $2$  by  $2\pi f$   $2\pi A_c e$  to the power of  $j$   $\beta \sin 2\pi f_m t$  will be  $2\pi f_m$  equals  $\sin 2\pi f_m t$  it's  $2\pi f_m t$  is  $x$ , so this is simply  $j \sin \beta x$  minus  $2\pi f_m t$  so is  $x$  so this is simply minus  $nx$  times  $dt$  is  $dx$  by  $2\pi f_m$  of course,  $T$  into  $f_m$  remember  $T$  is  $1$  over  $f_m$  so  $T$  into  $f_m$  equals we have  $T$  into  $f_m$  equals  $1$  and therefore this will

be your well, 1 over T into Fm becomes 1 so this will be 1 or simply one over 2pi 0 to 2pi 0 to 2pi e to the power of j beta sin x minus nx dx that will be your Cn that will be your Cn.

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The image shows a handwritten derivation of the Fourier coefficient  $C_n$  on a digital whiteboard. At the top, a pink arrow points from the text '1 over T into Fm becomes 1' to the expression  $T F_m = 1$ . Below this, the coefficient  $C_n$  is defined as:

$$C_n = \frac{A_c}{2\pi} \int_0^{2\pi} e^{j(\beta \sin x - nx)} dx$$

This equation is enclosed in a purple rectangular box. Below the box, the expression is simplified to:

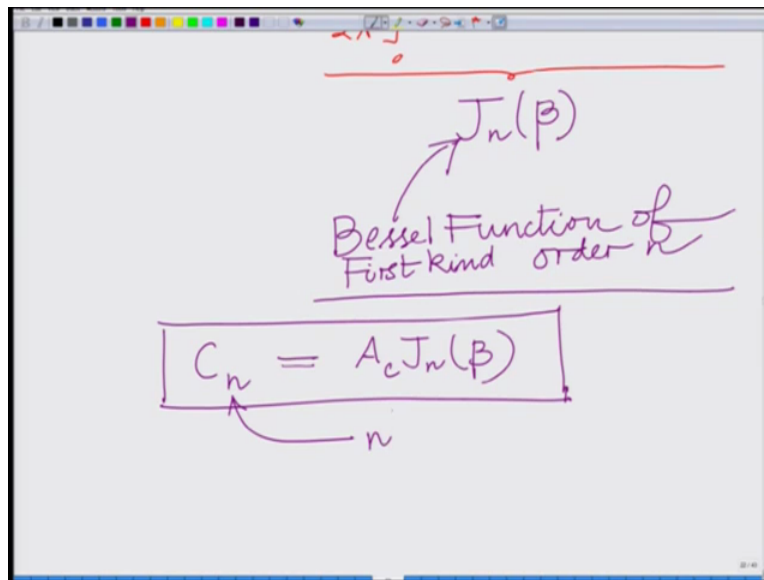
$$= A_c \cdot \frac{1}{2\pi} \int_0^{2\pi} e^{j(\beta \sin x - nx)} dx$$

The simplified equation is underlined in red.

And you can see this is basically your  $C_n$  this is basically in fact I have an  $A_c$  also and this integral now you can see I can write this let me just write this as well,  $A_c$  into 1 over 2pi integral 0 to 2pi e to the power of j beta sin x minus nx now you can see this is a standard integral this is basically the special function this is basically  $J_n$   $J_n$  of beta this is the nth order this is the Bessel function of the first this is the Bessel function of the first guide this is your Bessel function where your this is a standard mathematical function this is a Bessel function of first kind and order n.



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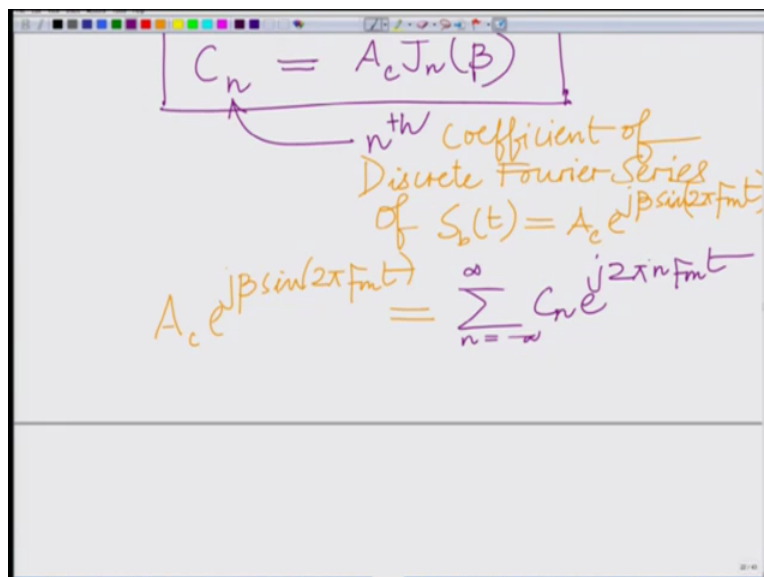


$J_n(\beta)$   
Bessel Function of First Kind order  $n$

$$C_n = A_c J_n(\beta)$$

So this is basically your Bessel function of the first kind and order  $n$  which means your  $C_n$  basically what we have shown is that your  $C_n$  equals  $A_c$  times  $J_n$  beta. So this is what we are able to show that is this is the  $n$ th coefficient of the discrete Fourier series this is remember what is this? This is the  $n$ th coefficient in the discrete Fourier series but not of the FM signal the discrete Fourier series of the complex baseband equivalent signal of the pass band signal, okay.

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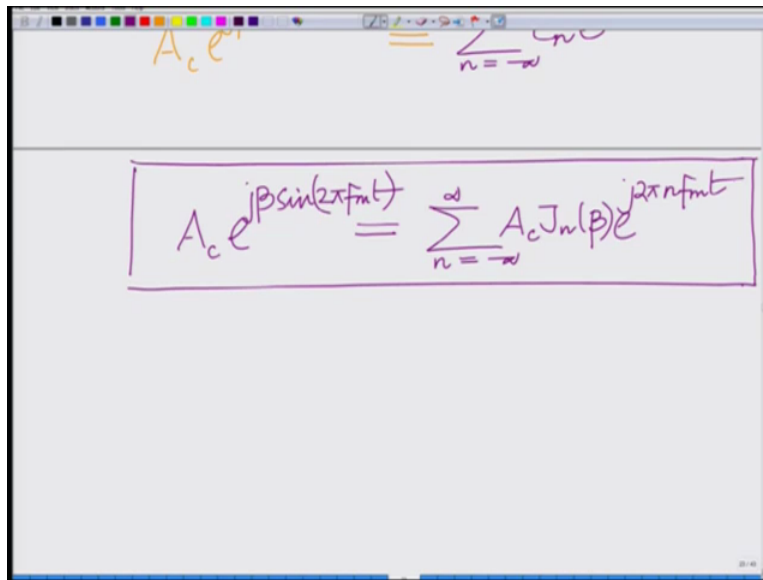

$$C_n = A_c J_n(\beta)$$

$n^{\text{th}}$  coefficient of Discrete Fourier Series of  $S_b(t) = A_c e^{j\beta \sin(2\pi f_m t)}$

$$A_c e^{j\beta \sin(2\pi f_m t)} = \sum_{n=-\infty}^{\infty} C_n e^{j2\pi n f_m t}$$

So this is the  $n$ th coefficient of the discrete Fourier series  $S_b t$  equals  $A_c e$  to the power of  $j \beta \sin 2\pi f_m t$ , okay. So this is the  $n$ th coefficient of the discrete Fourier series. This is the  $n$ th coefficient of the discrete Fourier series, okay. Now therefore I have  $C_n$  equal to  $A_c j^n \beta^n$  now I have therefore I can write the expression or  $j \beta \sin 2\pi f_m t$  this is going to be well, we have already seen this is your summation  $n$  equal to minus infinity to infinity  $C_n e$  raise to power of  $j 2\pi n f_m t$  now substitute for the expression of  $C_n$ .

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$$A_c e^{j\beta \sin(2\pi f_m t)} = \sum_{n=-\infty}^{\infty} C_n e^{j2\pi n f_m t}$$

$$A_c e^{j\beta \sin(2\pi f_m t)} = \sum_{n=-\infty}^{\infty} A_c J_n(\beta) e^{j2\pi n f_m t}$$

So therefore this becomes  $n$  equal to minus infinity to infinity well, that is  $C_n$  is  $A_c J_n \beta^n$   $e$  to the power of  $j 2\pi n f_m t$   $A_c j^n \beta^n$ , this is basically the expression of the DFS of the discrete Fourier series of the complex baseband equivalent signal of the pass band FM signal, okay.

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The image shows a handwritten derivation on a digital whiteboard. At the top, the equation  $A_c e^{j\beta \sin(2\pi f_m t)} = \sum_{n=-\infty}^{\infty} A_c J_n(\beta) e^{j2\pi n f_m t}$  is written in purple. Below it, an arrow points to the expression  $S(t) = \text{Re} \left\{ S_b(t) e^{j2\pi f_c t} \right\}$  in green. This is followed by another green equation:  $= \text{Re} \left\{ \left( \sum_{n=-\infty}^{\infty} A_c J_n(\beta) e^{j2\pi n f_m t} \right) \times e^{j2\pi f_c t} \right\}$ . The term  $e^{j2\pi f_c t}$  is written in blue below the summation.

Now from this we can derive the pass band signal remember pass band signal  $S_t$  is nothing but the real part of this is your  $S_b(t)$  complex equivalent baseband signal, so the real part of  $S_b(t) e^{j2\pi f_c t}$  which is basically the real part of well, if I substitute this  $n$  equal to minus infinity to infinity  $A_c J_n(\beta) e^{j2\pi n f_m t}$  times this whole thing times  $e^{j2\pi f_c t}$  which is equal to...

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The image shows the continuation of the handwritten derivation on a digital whiteboard. It starts with the expression  $= \text{Re} \left\{ \left( \sum_{n=-\infty}^{\infty} A_c J_n(\beta) e^{j2\pi n f_m t} \right) \times e^{j2\pi f_c t} \right\}$  in green. This is followed by a blue equation:  $= \text{Re} \left\{ \sum_{n=-\infty}^{\infty} A_c J_n(\beta) e^{j2\pi (f_c + n f_m) t} \right\}$ . Finally, the result is boxed in red:  $S(t) = \sum_{n=-\infty}^{\infty} A_c J_n(\beta) \cos(2\pi (f_c + n f_m) t)$ .

Now taking  $e$  to the power of the  $2\pi f_c t$  inside this is simply real part of  $n$  equal to minus infinity to infinity  $A_c J_n(\beta) e^{j 2\pi f_c t + n 2\pi f_m t}$  which is now simply basically if you take the real part of this that is simply now you can see real part of  $e^{j 2\pi f_c t + n 2\pi f_m t}$ . Simply cosine  $2\pi f_c t + n 2\pi f_m t$ , so this is simply summation  $n$  equal to minus infinity to infinity  $A_c J_n(\beta) \cos(2\pi f_c t + n 2\pi f_m t)$  that is your  $S(t)$  which is your pass band signal, so this is your pass band signal.

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$$= \text{Re} \left\{ \sum_{n=-\infty}^{\infty} A_c J_n(\beta) e^{j 2\pi (f_c + n f_m) t} \right\}$$


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$$S(t) = \sum_{n=-\infty}^{\infty} A_c J_n(\beta) \cos(2\pi (f_c + n f_m) t)$$


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$$A_c \cos(2\pi f_c t + \beta \sin 2\pi f_m t)$$

And remember the expression for  $S(t)$  is basically  $S(t)$  remember is simply  $A_c \cos(2\pi f_c t + \beta \sin 2\pi f_m t)$ , so this is basically summation  $n$  equal to minus infinity to infinity  $A_c J_n(\beta) \cos(2\pi f_c t + n 2\pi f_m t)$ .

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$$s(t) = \sum_{n=-\infty}^{\infty} A_c J_n(\beta) \cos(2\pi(f_c + n f_m)t)$$

$$A_c \cos(2\pi f_c t + \beta \sin 2\pi f_m t)$$

$$\cos(2\pi f_c t) \leftrightarrow \frac{1}{2} \delta(f - f_c) + \frac{1}{2} \delta(f + f_c)$$

Now we know each cosine, cosine of  $2\pi$  that is we know there is Fourier transform of cosine  $2\pi f_c t$  has Fourier transform impulse of amplitude half at  $f_c$  minus  $f_c$  not plus half  $\delta(f - f_c)$  plus  $f_c$  not therefore each cosine  $2\pi f_c t + \beta \sin 2\pi f_m t$  basically is one impulse if you take the spectrum of that that is half impulse of my amplitude half at  $f_c + n f_m$  and impulse of amplitude half at  $-f_c - n f_m$  therefore now if you look at Start.

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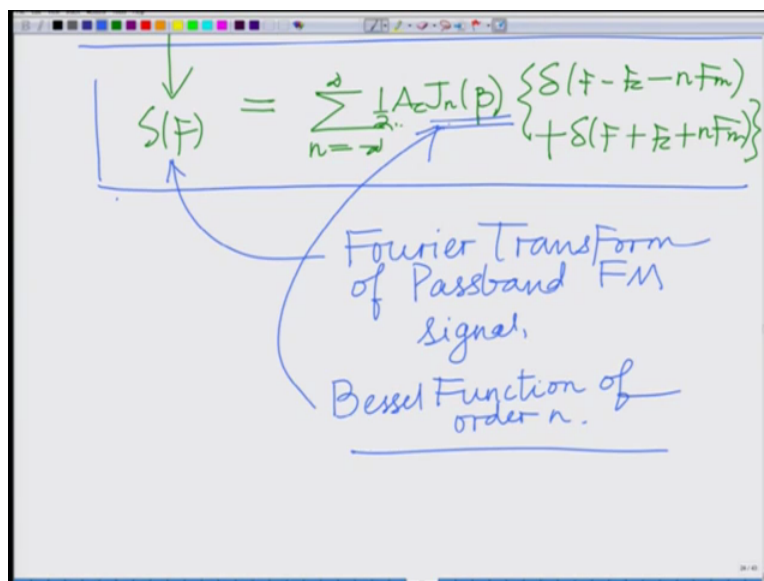
$$s(t) = \sum_{n=-\infty}^{\infty} A_c J_n(\beta) \cos(2\pi(f_c + n f_m)t)$$

$$\frac{1}{2} \delta(f - f_c - n f_m) + \frac{1}{2} \delta(f + f_c + n f_m)$$

$$S(f) = \sum_{n=-\infty}^{\infty} \frac{1}{2} A_c J_n(\beta) \left\{ \delta(f - f_c - n f_m) + \delta(f + f_c + n f_m) \right\}$$

Let me rewrite  $S_t$  again  $S_t$  equals summation  $n$  equals minus infinity to infinity  $A_c J_n(\beta) \cos(2\pi f_c t + 2\pi n f_m t)$ . Cosine  $2\pi f_c t + 2\pi n f_m t$  into  $t$ . Cosine  $2\pi f_c t + 2\pi n f_m t$  into  $t$  and now each of this cosine term has Fourier transform half  $\delta(f - f_c - n f_m)$  that is half  $\delta(f - f_c - n f_m)$  plus half  $\delta(f + f_c + n f_m)$  which implies that the Fourier transform of  $S_t$  therefore now I have simply I have to (trans) substitute the Fourier transform of the cosine that that is basically summation  $A_c J_n(\beta)$  that is half or  $A_c J_n(\beta)$  into well, Delta or half you can say half write here half  $\delta(f - f_c - n f_m)$  plus half  $\delta(f + f_c + n f_m)$ , so this is the Fourier transform of the you are finally variable to derive the Fourier transform of the variable to derive the Fourier transform of the pass band FM signal.

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The image shows a handwritten equation on a whiteboard background. The equation is:

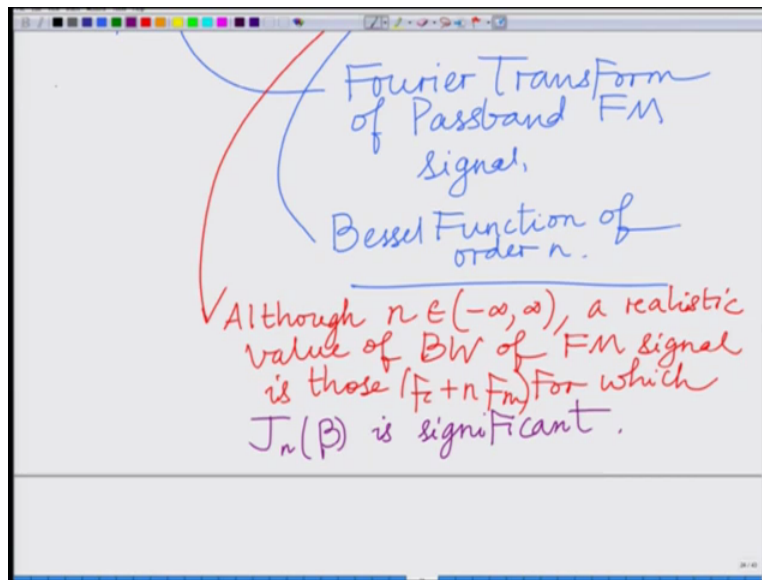
$$S(f) = \sum_{n=-\infty}^{\infty} \frac{A_c}{2} J_n(\beta) \left\{ \delta(f - f_c - n f_m) + \delta(f + f_c + n f_m) \right\}$$

Below the equation, there are two handwritten annotations in blue ink:

- "Fourier Transform of Passband FM signal" with an arrow pointing to the  $S(f)$  term on the left.
- "Bessel Function of order  $n$ " with an arrow pointing to the  $J_n(\beta)$  term in the summation.

What is this? This is the Fourier (trans) the Fourier transform of the of the pass band FM signal and you realize that it depends on this new function that we have seen this is the Bessel function  $J_n(\beta)$  Bessel function of order  $n$  this is the Bessel function of order  $n$ , okay. So this is your Bessel function of order  $n$ , okay. So this is the spectrum of this and naturally therefore it depends on  $J_n(\beta)$  beta naturally of course, this is although is going from  $n$  equal to minus infinity  $n$  equal to infinity obviously since we have to trans in practice we have to limit the bandwidth or spectrum of this transmitted FM signal we have to choose a realistic value of bandwidth. So a realistic value of bandwidth for the signal is the range of frequencies for which this quantity  $J_n(\beta)$  is significant, okay.

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So that is also another thing that can be noted a realistic value of the bandwidth, so although  $J_n$  goes from minus infinity to infinity although  $n$  belongs to well, minus infinity to infinity a realistic value of bandwidth of FM signal is those  $J_n \beta$  for which is those is those frequencies that is  $f_c$  plus  $n f_m$  for which  $J_n \beta$  is significant, right? So we have to choose we have to choose an appropriate range of  $n$  or appropriate range of these frequencies, right? For which  $J_n \beta$  is significant that captures most of the signal power most of the pass band FM signal power, okay. So that gives us an estimate or a measure of a the bandwidth of this FM signal and then once we get the estimate of the bandwidth of this FM signal then we can calculate various other aspects, alright various other quantities, okay.

So that is what we have so what we have now is basically we have considered a pass band FM signal we have derived it's the expression for its spectrum now we have seen that this spectrum literally spans from I mean if you look at it if you look at it analytically or the precise mathematical expression it spans from over the entire frequency band, alright. So realistically we have to come up with a practical limit for the transmit bandwidth or a practical way to characterize a practical and a meaningful how do you put it? A measure to characterize the spectrum of this FM signal and that is given and that is something that we are going to look at in the subsequent modules that is out of this spectrum, how can we extract what is a practically

feasible bandwidth? Or a practically good measure of the bandwidth of this FM system, alright.  
So we will stop here and look at other aspects in the subsequent modules, thank you.