

Statistical Signal Processing
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Lecture - 34
Review Assignment 4

Hello students. Welcome to this session on review assignment 4. Let us see the first problem.

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Q.1 Assume the observation model $Y(n) = X(n) + V(n)$ where $V(n)$ is a zero-mean white noise with variance 1 and $X(n)$ has an autocorrelation function $R_X(m) = 0.6^{|m|}$, $m \in \mathbb{Z}$.

(a) Find the transfer function of the causal IIR Wiener filter to estimate $X(n)$.

(b) Find the impulse response of the filter.

(c) Find the mean-square estimation error.

Suppose $H(z)$ is the transfer function of the causal IIR Wiener filter.

$$H(z) = \frac{1}{\sigma_v^2 H(z)} \left[\frac{S_{xy}(z)}{H(z)} \right]_+$$

$$R_X(m) = 0.6^{|m|}, m \in \mathbb{Z}$$

$$S_X(z) = \sum_{m=-\infty}^{\infty} R_X(m) z^{-m} = \sum_{m=-\infty}^{\infty} 0.6^{|m|} z^{-m} = \frac{0.64}{(1-0.6z)(1-0.6z^{-1})}$$

$$S_{xy}(z) = S_X(z)$$

$$S_Y(z) = S_X(z) + 1 = \frac{2 - 0.6(z + z^{-1})}{(1-0.6z)(1-0.6z^{-1})}$$

Assume the observation model $Y_n = X_n + V_n$ where V_n is a zero-mean white noise with variance 1 and X_n has an autocorrelation function R_{Xm} is equal to 0.6 to the power of mod m , m belonging to $\mathbb{Z}$. This problem we considered earlier also. Find the transfer function of the causal IIR Wiener filter to estimate X_n . So, we have to find the transfer function of the causal IIR Wiener filter.

Next, find the impulse response of the filter, and finally find the mean-square estimation error. So, in this problem we have to find out the transfer function, suppose H_z is the transfer function of the causal IIR Wiener filter. We have derived that H_z is equal to 1 by sigma V squared H_z into S_{xyz} divided by $H_c Z$ inverse and the causal part of this. So this is the positive part, causal part. First, we have to find out $H_c Z$.

So what is $H_c Z$. This is the causal filter corresponding to spectral factorization of xyz . So, first we have to find out the causal filter corresponding to factorization xyz . That means, we write suppose, this is our y_n , this is the observed signal and this is V_n where V_n is innovation

sequence with variance σ_v^2 . This one is $H_c Z$, so that way that absorbs the stationary signal is considered as the output of a linear filter with input as the innovation sequence.

So, that we have to find out the expression for $H_c Z$ from the spectral factorization of $S_y(z)$. Now, let us consider the parameters given, so R_x of m is equal to 0.6 to the power mod of m . So m belongs to set of integers. Now, S_x of that power spectral density of that domain, is given by R_x of m into Z to the power minus m , m going from minus infinity to plus infinity. So this we can carry out the algebra I already we have done it m is equal minus infinity to infinity, R_x is equal to 0.6 to the power of mod m and that to the power minus m here.

This we have shown that in earlier problem sheet, we already saw this and this is given by 0.64 divided by $1 - 0.6Z$ into $1 - 0.6 Z$ inverse. So, this is the power spectral density of X_n . Then, also we know that S_{xyz} because of this model signal plus noise model which are independent of each other because of that S_{xyz} will also be equal to S_{xz} that we have already established earlier and $S_{yz} = S_{xz} + 1$, 1 is the power spectral density of the white noise because its variance is 1 and this expression, we got is $2 - 0.6$ into $Z + Z$ inverse divided by this quantity $1 - 0.6 Z$ into $1 - 0.6 Z$ inverse. So, this is the xyz .

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The image shows a handwritten derivation for the spectral factorization of $S_y(z)$. The steps are as follows:

$$\text{Spectral factorization of } S_y(z) = \frac{2 - 0.6(z + z^{-1})}{(1 - 0.6z)(1 - 0.6z^{-1})} = \frac{\sigma_v^2 (1 - \alpha z)(1 - \alpha z^{-1})}{(1 - 0.6z)(1 - 0.6z^{-1})}$$

$$\sigma_v^2 (1 + \alpha) = 2 \quad \text{and} \quad \sigma_v^2 \alpha = 0.6$$

$$\frac{1 + \alpha}{\alpha} = \frac{2}{0.6} = \frac{10}{3}$$

$$\alpha^2 - 10\alpha + 3 = 0 \Rightarrow \alpha = 3 \Rightarrow \alpha = \frac{1}{3}$$

$$\alpha = \frac{1}{3} \quad \sigma_v^2 = \frac{0.6}{\frac{1}{3}} = 1.8$$

$$\therefore S_y(z) = \frac{1.8 (1 - \frac{1}{3}z)(1 - \frac{1}{3}z^{-1})}{(1 - 0.6z)(1 - 0.6z^{-1})}$$

innovation response $H_c(z) = \frac{1 - \frac{1}{3}z}{1 - 0.6z}$

$$H(z) = \text{transfer function of the causal IIR filter} = \frac{S_{xy}(z)}{H_c(z)} = \left[\frac{0.64}{(1 - 0.6z)(1 - 0.6z^{-1})} \right] \left[\frac{0.64}{(1 - 0.6z^{-1})(1 - \frac{1}{3}z)} \right]$$

Now, we have to do spectral factorization of S_{yz} that is equal to $2 - 0.6$ into $Z + Z$ inverse divided by $1 - 0.6Z$ into $1 - 0.6 Z$ inverse. Now, we are interested to factorize this σ_v^2 into $1 - \alpha Z$ into $1 - \alpha Z$ inverse divided by $1 - 0.6Z$ into $1 - 0.6 Z$ inverse. Now, we have to choose this α and σ_v^2 squared in such a way that corresponding filter will be

minimum phase. Spectral factorization whatever this causal part will be that should be minimum phase.

So, we have this expression $2 - 0.6 \text{ into } Z + Z \text{ inverse}$ is equal to $\sigma^2 \text{ into } 1 - \alpha Z \text{ into } 1 - \alpha Z \text{ inverse}$. Now, we can equate this expression and this expression, we will get $\sigma^2 \text{ into } 1 + \alpha^2$ that must be equal to 2 and the other part is $\sigma^2 \text{ into } \alpha$, that must be equal to 0.6. These are the equations; we have to solve to find out σ^2 and α .

Now, dividing this equation by this equation, we will get $1 + \alpha^2$ divided by α is equal to $2 \text{ by } 0.6$ that is equal to $10 \text{ by } 3$. So, this is the relationship we get from this $\alpha^2 - 10\alpha + 3 = 0$, which will give α is equal to 3 and α is equal to $1/3$. For spectral factorization, the filter should be minimum phase, because of that we switch α is equal to $1/3$.

So, that way α is equal to one-third and corresponding σ^2 , from this expression, we can get σ^2 is equal to $0.6 \text{ divided by one-third}$ that is equal to 1.8. Therefore, this numerator part will be now, that is σ^2 will be there 1.8, and then $1 - 1/3 Z \text{ into } 1 - 1/3 Z \text{ inverse}$. So, therefore S_{yz} is equal to $1.8 \text{ into } 1 - 1/3 Z \text{ into } 1 - 1/3 Z \text{ inverse}$ divided by $1 - 0.6 Z \text{ into } 1 - 0.6 Z \text{ inverse}$.

Therefore, this spectral factorization theorem we can apply now, that causal filter will be given by, causal part is this $1 - 1/3 Z \text{ inverse}$ divided by $1 - 0.6 Z \text{ inverse}$. So, that is the causal filter so that this is my innovation sequence with variance σ^2 is equal to 1.8 and it is passed through the system, causal system that is $H_c Z = 1 - 1/3 Z \text{ inverse}$ divided by $1 - 0.6 Z \text{ inverse}$ and output we will get Y_n . So, that way we do the spectral factorization.

Now, we know that H_z that is the transfer function of the causal IIR filter and that is given by $1 \text{ by } \sigma^2 \text{ into } H_c Z \text{ into the causal part of } S_{xyz} \text{ divided by } H_c Z \text{ inverse}$. So that way, now we know $H_c Z$ is given by this N_{xyz} is equal to S_{xz} , so S_{xyz} is equal to S_{xz} and S_{xz} is given by $0.64 \text{ divided by } 1 - 0.6Z \text{ into } 1 - 0.6Z \text{ inverse}$. So, we know S_{xyz} . So therefore $S_{xyz} \text{ divided by } H_c Z \text{ inverse}$ that will be equal to $0.64 \text{ divided by } 1 - 0.6Z \text{ into } 1 - 0.6 Z \text{ inverse}$ divided by $H_c Z \text{ inverse}$, $H_c Z \text{ inverse}$ is $1 - 1/3 \text{ of } Z \text{ divided by } 1 - 0.6Z$.

So, we have to take the causal part of this. This, we can simplify because $1 - 0.6$ and $1 - 0.6$ will get cancelled, so we will have this is equal to 0.64 divided by $1 - 0.6 Z$ inverse into $1 - 1/3Z$. This is the expression for S_{xyz} divided by $H_c Z$ inverse and we have to take the positive part or the causal part of this. So we found the causal part.

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The slide contains the following handwritten derivations:

$$\left[\frac{0.64}{(1-0.6z^{-1})(1-\frac{1}{3}z^{-1})} \right]_+ = \frac{A}{(1-0.6z^{-1})} \quad \text{Causal part}$$

$$A = \frac{0.64}{1-\frac{1}{3}} = 0.8$$

$$H(z) = \frac{1}{\sigma_v^2 H_c(z)} \left[\frac{S_{xyz}(z)}{H_c(z)} \right]_+ = \frac{1}{1.8 \times \frac{1-\frac{1}{3}z^{-1}}{1-0.6z^{-1}}} \times \frac{0.8}{1-0.6z^{-1}}$$

$$H(z) = \frac{\frac{4}{9}}{1-\frac{1}{3}z^{-1}} \quad \text{transfer function of causal Wiener filter}$$

Taking the inverse, impulse response

$$h(n) = \frac{4}{9} \left(\frac{1}{3} \right)^n \quad n \geq 0$$

$$\text{MMSE} = E(x(n) - \sum_{i=0}^{\infty} h(i) y(n-i))^2 = R_{xx}(0) - \sum_{i=0}^{\infty} h(i) R_{xy}(i)$$

$$= 1 - \sum_{i=0}^{\infty} \frac{4}{9} \left(\frac{1}{3} \right)^i 0.6^i$$

$$= 1 - \frac{4}{9} \quad \text{Sum} = 0.444$$

Subtitles/closed captions (c)

Now we have to find out the causal part of 0.64 divided by $1 - 0.6 Z$ inverse into $1 - 1/3$ of Z . So, positive part of this we have to consider causal part of this. Now, this causal part will involve this because this is the causal transfer function, so that way this will be of this form, this will be equal to A by $1 - 0.6 Z$ inverse and anti-causal part will be there. This is causal part, because this part will involve the anti-causal part.

Now, to find out A , we will multiply by $1 - 0.6 Z$ inverse here. This and this will get cancelled and then, we will substitute Z is equal to 0.6 , so that way, A will be given 0.64 divided by here we will put that is equal to $1 - 1/3 \times 0.6$. This is where we will do partial fraction, so using that we get the result and which is equal to 0.8 , because it will be $1 - 0.2$ will be 0.8 , so this will be 0.8 .

So, this part of the causal expression is equal to 0.8 divided by $1 - 0.6Z$ inverse. So, therefore, H_z is the transfer function of the causal IR Wiener filter is given by 1 by σ_v^2 we already out, σ_v^2 by σ_v^2 into $H_c Z$ into S_{xyz} divided by $H_c Z$ inverse that is the causal part of that. So, causal part of that we already found out. This, we have found out. σ_v^2 we have found out. So, that way this will be equal to 1 by 1.8 into 1 divided

HcZ $1 - 1/3$ of Z inverse divided by $1 - 0.6$ Z inverse into causal part of this is given by 0.8 divided by $1 - 0.6$ Z inverse.

So, if you simplify, we will get this as $4/9$, so this part and this part will get cancelled, only we have $1 - 1/3$ Z inverse. This is the transfer function of the causal Wiener filter. This is the transfer function. Now taking the inverse of this, we will get the impulse response which is the causal IR Wiener filter, impulse response $h_n = 4/9$, this is $1/3$ is there into $1/3$ to the power n , $n > 0$. So, this is the impulse response of the causal IIR Wiener filter.

We also have to find out the minimum mean square error MMSE, this is equal to E of X_n minus the filter $h_i y$ of $n - i$, i going from 0 to infinity, because error is orthogonal to data, there is only X_n will be here, and because of that it will be R_x of 0 minus summation h_i into R_{xy} square, i going from 0 to infinity, and if we substitute the values of R_{x0} and $h_i R_{xy}$ of i this will be equal to 1 , R_{x0} is 1 minus summation $4/9$ is there, $1/3$ to the power i , and R_{xyi} , that is the same as R_{xi} that is equal 0.6 to the power i , i going from 0 to infinity.

So, this is the autocorrelation function cross correlation function is same as the autocorrelation function. So, this is an infinite series and carry out this summation, this is an infinite series and common ratio is less than 1 , so this will be simply first term, that is i is equal to 0 , $4/9$ divided by 1 minus common ratio, common ratio is $1/3$ into 0.6 . So that way, this will be $1 - 0.2$, and this is equal to $1 - 5/9$, that is equal to 0.444 . So, this is the minimum mean square error.

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Q.2 (a) Write down expressions for the normal equations and the corresponding MMSPE for the p th-order linear predictor for a signal $y(n)$.

(b) A WSS signal $y(n)$ has an autocorrelation function $R_y(m) = 0.5^{|m|}$, $m \in \mathbb{Z}$. Find the first and second-order linear predictors for $y(n)$.

(a) $\hat{y}(n) = \sum_{i=1}^p h(i) y(n-i)$ *p*th-order linear predictor.
 error $\perp$ data $\Rightarrow E(y(n) - \sum_{i=1}^p h(i) y(n-i)) y(n-j) = 0$ $j = 1, 2, \dots, p$
 $E(y(n) - \sum_{i=1}^p h(i) y(n-i)) y(n-j) = 0$ *Normal equations*
 $\sum_{i=1}^p h(i) R_y(j-i) = R_y(j)$ $j = 1, 2, \dots, p$
 $\text{MMSE} = E(y(n) - \sum_{i=1}^p h(i) y(n-i))^2 = R_y(0) - \sum_{i=1}^p h(i) R_y(i)$

(b) $R_y(m) = 0.5^{|m|}$, $m = 0, \pm 1, \dots$
first-order linear predictor
Normal equation $\sum_{i=1}^1 h(i) R_y(j-i) = R_y(j)$ $j = 1$
 $\Rightarrow h(1) R_y(0) = R_y(1)$
 $\Rightarrow h(1) = \frac{R_y(1)}{R_y(0)} = 0.5$
 $\hat{y}(n) = h_1(1) y(n-1)$
Normal equations $\sum_{i=1}^2 h(i) R_y(j-i) = R_y(j)$ $j = 1, 2$
 $h_1(1) R_y(0) + h_2(1) R_y(1) = R_y(1)$
 $h_1(1) R_y(1) + h_2(2) R_y(0) = R_y(2)$

So this is the first problem. So, we will go to the next problem. One is theoretical, write down the expression for the normal equations and the corresponding minimum mean square prediction error for the p th order linear predictor. Suppose; this is for a signal y_n . So, y_n is this signal, where y_n is equal to summation h_i into y_{n-i} , i going from 1 to p . This is p th order linear predictor.

Now normal equation is obtained from the condition, error is orthogonal to data. So, that way E of $y_n - \text{summation } h_i y_{n-i}$, i going from 1 to p into $y_n - j$ that will be equal to 0, $j = 1, 2$ up to p , p th order linear prediction. So, from this we will get, we can write the expression first, summation h_i into R_y of $j - i$, i going from 1 to p that must be equal to R_y of j . So this normal, $j = 1, 2$ up to p . So, this is the normal equation.

Expression for mean square prediction error, minimum mean square prediction error, MMSPE is equal to E of y_n , that is the data minus prediction $h_i y_{n-i}$, i going from 1 to p and there will be y_n here, this error is orthogonal to the remaining part of the error. So, that way this will be equal to $R_y(0)$ minus summation $h_i R_y(i)$, i going from 1 to p . This is the expression for minimum mean square prediction error.

Part b is R_{ym} is equal to 0.5 to the power mod of m , so that is m is equal to 0, plus minus 1 etc. Now, for first order linear predictor. For that, what we will have, so here is p is 1, so there will be one term, so normal equation will be, there will be only term h_1 into R_y , i going from 1 to p , p is 1, one term only, $R_y(0)$ is equal to $R_y(1)$, imply that h_1 is equal to $R_y(1)$ divided $R_y(0)$, so that will be equal to 0.5 . We know $R_y(1)$ is equal to 0.5 and $R_y(0)$ is equal to 1. So that way it is 0.5 , which we have found out.

Suppose, this is first order prediction that is we will denote here 1. Now, we will go to the second order predictor. Second order prediction is $\hat{y}_n$ is equal to let me call it h_2 of 1 y_{n-1} + h_2 of 2 into y_{n-2} . Now, we will have the normal equations, h_2 of 1 into $R_y(0)$ + h_2 of 2 into $R_y(1)$ that must be equal to $R_y(1)$. This is one equation. Next equation will be h_2 of 1 = $R_y(1)$ + h_2 of 2 $R_y(0)$, that must be equal to $R_y(2)$. So, this is the normal equations for second order case. So, here we can substitute the values.

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$$\begin{aligned}
 h_2(1) R_Y(0) + h_2(2) R_Y(1) &= R_Y(1) \\
 h_2(1) R_Y(1) + h_2(2) R_Y(0) &= R_Y(2) \\
 h_2(1) + h_2(2) \times 0.5 &= 0.5 \\
 0.5 h_2(1) + 1 \times h_2(2) &= 0.25 \\
 \left. \begin{aligned} h_2(2) &= 0 \\ h_2(1) &= 0.5 \end{aligned} \right\} \\
 \text{1st order MMSPE} &= R_Y(0) - h_1(1) R_Y(1) = 1 - 0.5 \times 0.5 = 0.75 \\
 \text{2nd order MMSPE} &= R_Y(0) - h_1(1) R_Y(1) - \underbrace{h_2(1) R_Y(2)}_0 = 0.75
 \end{aligned}$$

So, here we have h_2 of 1 into R_Y of 0 + h_2 of 2 into R_Y of 1 = R_Y of 1 that is first equation. Second equation will be h_2 of 1 R_Y of 1 + h_2 of 2 into R_Y of 0 that is equal to R_Y of 2. So, from this h_2 of 1, R_Y of 0 + h_2 of 2 into R_Y of 1 = 0.5, so that will be equal to R_Y of 1 is 0.5 and from second equation, $0.5 h_2$ of 1 x h_2 of 2 that must be equal to R_Y of 2, that is equal to 0.25. So, these two equations if we solve, we get h_2 of 2 = 0 and h_1 of 1 = 0.5.

So, this is the linear predictor filter coefficients for order 2 and we observe here that h_2 of 2 is equal to zero, because from the data, this is the autocorrelation function of the first order ER process, therefore, if we predict second order terms will be equal to 0. So that is expected. We have to find out the MMSPE also, minimum mean square prediction error, that is for first order case.

Suppose, that will be equal to $R_Y(0)$ minus h_1 of 1 into R_Y of 1 and this is equal to R_Y of 0 and this is 0.5 into 0.5, so that way it will be 0.75 and in the second case, this is first order and this is second order MMSPE is equal to again $R_Y(0) - h_1(1) R_Y(1) - h_2(1) R_Y(2)$. Now, this is 0 and this is same as 0.5 and this also equal to 0.75 only.

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Q.3. The autocorrelation functions of a WSS signal $Y(n)$ at different lags are tabulated as follows:

m	0	1	2	3
$R_Y(m)$	1	0.2353	-0.6059	-0.4071

(a) Apply the Levinson Durbin algorithm to find the first-order and the second-order linear predictors for the process.

(b) Find the reflection coefficient k_3 . Hence comment about the model of the signal.

Levinson Durbin recursion

• For $m = 1, \dots, M$

$$k_m = \frac{\sum_{i=0}^{m-1} h_{m-1}(i) R_Y(m-i)}{\varepsilon(m-1)}$$

$$h_m(i) = h_{m-1}(i) + k_m h_{m-1}(m-i), i = 1, 2, \dots, m-1$$

$$h_m(m) = -k_m$$

$$\varepsilon_m = \varepsilon_{m-1}(1 - k_m^2)$$

Initialization $h_0(0) = -1$ $\varepsilon(0) = R_Y(0)$

$m=1$ $k_1 = \frac{h_0(0) R_Y(1-0)}{\varepsilon(0)} = \frac{-1 \times 0.2353}{1} = -0.2353$

$h_1(1) = -k_1 = 0.2353$

$\varepsilon(1) = \varepsilon(0) \times (1 - 0.2353^2) = 0.9446$

$m=2$ $k_2 = \frac{h_1(0) R_Y(2) + h_1(1) R_Y(1)}{\varepsilon(1)} = \frac{-1 \times (-0.6059) + 0.2353 \times 0.2353}{0.9446} = 0.67$

$h_2(1) = h_1(1) + k_2 h_1(0) = 0.2353 + 0.67 \times (-1) = -0.4347$

$h_2(2) = -k_2 = -0.67$

So, let us see the third equation, the autocorrelation function of a WSS signal Y of n at different lags are tabulated as follows. That lag 0, it is 1, lag 1, it is 0.2353, lag 2, - 0.6059, lag 3, - 0. 4071. So, apply Levinson Durbin algorithm to find the first-order and the second-order linear predictors for the process. Second part is, find the reflection coefficient k_3 , for third order predictor what is the reflection coefficient. Hence comment about the model of the signal.

We know the Levinson-Durbin recursion. This is the recursion. We have to first determine the reflection coefficient that is given by k_m given by summation $h_{m-1} i$ into $R_Y m - i$, i going from 0 to $m-1$, divided by minimum mean square prediction error of the $m - 1$ order. These are the parameters of the $m - 1$ th order, so from that, we will determine the k_m , that is the reflection coefficient of order m .

Then for order m , we will update the filter parameters by this equation, h of m is equal to $h_{m-1} i + k_m h_{m-1} m - i$ and finally h_m is equal to $-k_m$, h_m that is m th order last parameter will be $-k_m$ and mean square prediction error will be updated by this equation. We also have to initialize the algorithm. So, for that initialization that h_m of 0 = - 1, for all m that is one initialization and ε_0 is equal to R_Y of 0.

Because there is no prediction for 0th order predictor and therefore mean square prediction will be simply will be variance that is R_Y0 . Now, we have to apply and find out the Levinson-Durbin algorithm to find out the first order and the second order linear predictors. So,

suppose m is equal to 1 and then K_1 will be equal to from this expression, a sub it is up to $m - 1$, m is equal to 1.

And it will be 1 term will be there and is equal to 0, h_0 of 0 into R_y , $m = 1 - 0$ divided by mean-square error, that is divided by ϵ_0 . So that way, h_0 of 0 = - 1 into R_y of 1 = 0.2353 divided by mean-square error, that is ϵ_0 , so h_0 of 0 = - 1 into R_y of 1 = 0.2353 divided by ϵ_0 , that is same R_y of 0 that is equal to 1, so that is equal to minus 0.2353. So, here only one parameter is there, replacement parameter and that will give us the corresponding linear predictor parameter that is h_1 of 1 will be equal to negative of this minus of K_1 that is equal to 0.2353.

Next, we will go to m is equal to 2 and here we have to update the mean-square error also, before going to m is equal to 2, so, that way ϵ_1 , that is the mean-square prediction error. Minimum mean-square prediction error of the first-order predictor, ϵ_1 is equal to ϵ_0 into $1 - k_1$, we have already found out $1 - 0.2353$ squared, and that will be equal to 0.9446 and now we will go to m is equal to 2, so k_2 we have to determine.

And if I put here, m is equal to 2, so it will be summation from i is equal 0 to 1. Then this term will be h_1 of 0 and h_1 of 1. So, that way first term will be h_1 of 0 into R_y of 2 + h_1 of 1 into R_y of 1 divided by mean-square error ϵ_1 and if we substitute the values, h_1 of 0 that is equal to -1, R_{y2} is given, that is equal to -0.6059 + h_1 of 1, that is equal to 0.2353 into R_y of 1, also is equal to 0.2353 divided by V_1 ; also we found out 0.9446, this will give us k_2 is equal to 0.7.

Therefore, we are interested in second-order parameters. So h_2 of 2 will be equal to minus 0.7. We have to find out h_1 of 1, h_2 of 1 will be equal to, from this that is h_1 of 1 k_2 into h_1 , that is $2 - 1$, that is h_1 of 1. So, this we can again substitute the values and we will get this as 0.4. We know h_1 of 1, k_2 we know, so therefore it will be 0.4. So, we have found out the first order and second order linear predictor.

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$$\begin{aligned} \epsilon(4) &= \epsilon(0)(1 - k_2^2) = 0.4818 \\ k_3 &= \frac{\sum_{i=0}^2 h_2(i) R_Y(2-i)}{\epsilon(4)} = \frac{-1 \times (-0.6059) + 0.4 \times 0.2353 - 0.7 \times 1}{0.4818} = 0 \\ k_3 &= 0 \\ AR(2) \quad , \quad k_m = 0, m > 2 \\ \quad \quad \quad k_m = 0, m > p \end{aligned}$$

We have to find out k_3 . Once, we have these values h_2 of 1, h_2 of 2, we can find out h_2 of 2 is equal to - 0.7 and h_2 of 1 is equal to 0.4, and now we can update the mean-square prediction error, that is the mean-square prediction error, Epsilon 2 is equal to Epsilon 1 minus K_2 squared. So, that will be Epsilon 1, mean square prediction, for a first order predictor we know and K_2 we know. So substitute, this will be equal to 0.4818.

Complete the two recursion in Levinson-Durbin algorithm to find out the first order and second order model. Now, let us go to find out K_3 . K_3 is equal to summation h_2 of i R_Y of $2 - i$, i going from 0 to 2, divided by epsilon 2, minimum mean-square error at iteration 2. Now, we have already values of this, so this will be equal to first one is i is equal to 0, h_2 of 0 is - 1 and R_Y of 2, that is equal to minus 0.6059, then second term will be h_2 of 1, that is 0.4 into R_Y of 1, that is 0.2353 - h_2 of 2 that is - 0.7 into R_Y of $2 - 2$ is 0, R_Y of 0 is 1.

So, this is divided by Epsilon 2, that is the mean-square prediction error at the order 2 and this part, if we carry out the algebra, this will be equal to 0 by epsilon 2, that is equal to 0. So, K_3 is equal to 0. That is 3rd order replacement coefficient is equal to zero. Since K_3 is equal to 0, that optimal linear predictor will be of order 2 only. So, that is our observation that optimal linear predictor will be of order 2 only.

We will not get any benefit by increasing the order and this is why because the corresponding model is AR2 model. So, that is the advantage of this replacement coefficient. Since it is an AR2 model, so after m is equal to 2, K_m will become zero, so this is the observation that since it is an AR2 model, here after m is equal to 2, that is K_3 is equal to 0, therefore it will

be an AR2 model, after m is equal to 2, this replacement coefficient $K_m = 0$. Thus, $K_3 = 0$. So, from this we can say that, we can get the optimal linear predictor of order 2 only, by increasing the order we do not improve the prediction.

Secondly, this $K_3 = 0$ happens when the corresponding model is AR2, so for AR2 model, this replacement parameters will become 0. So, that is a test for AR2 model. So, if it is an AR2 process, it will be equal to 0. Similarly, if it is an ARP process, then K_{p+1} will be equal to 0. So, $K_3 = 0$. It implies that, we do not get any improvement by increasing the linear-prediction model order beyond 2 and this happens because corresponding signal model is AR2 signal, because for AR2 signal, that K_m will be equal to 0 for $m > 2$. Similarly, for any ARP signal, K_m will be equal to 0 for m greater than p . So this is the importance of the replacement coefficients. Thank you.