

Statistical Signal Processing
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Lecture – 22
Linear Predictions of Signals 2

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Let us recall

- ❖ Linear prediction formulates the prediction $\hat{Y}(n)$ as

$$\hat{Y}(n) = \sum_{i=1}^M h(i)Y(n-i)$$
- ❖ The WH equations are given by

$$R_y(j) = \sum_{i=1}^M h(i) R_y(j-i) \quad j=1, 2, \dots, M$$
- ❖ In Matrix notation

$$\begin{bmatrix} R_y(0) & R_y(1) & \dots & R_y(M-1) \\ R_y(1) & R_y(0) & \dots & R_y(M-2) \\ \vdots & \vdots & \ddots & \vdots \\ R_y(M-1) & R_y(M-2) & \dots & R_y(0) \end{bmatrix} \begin{bmatrix} h(1) \\ h(2) \\ \vdots \\ h(M) \end{bmatrix} = \begin{bmatrix} R_y(1) \\ R_y(2) \\ \vdots \\ R_y(M) \end{bmatrix}$$
- ❖ The mean square prediction error is given by

Hello students. Welcome to this lecture on Linear Predictions of Signals. Let us recall linear prediction formulates the prediction $\hat{Y}(n)$ as, this is the relationship, $\hat{Y}(n)$ is summation $h(i)Y(n-i)$, i going from 1 to M . Corresponding Weiner Hopf equations are given by this $R_y(j)$ is equal to summation of $h(i)R_y(j-i)$, i going from 1 to M and there are M equations corresponding to j is equal to 1 to M .

In Matrix notation, the Weiner Hopf equations are given by this relationship. Here R_y , this is the autocorrelation matrix. This is the prediction coefficient factor, and this is SCF factor, autocorrelation vector. By inverting this relationship, we can find out the LPC coefficient h_1 , h_2 upto h_m , and similarly the mean square prediction error is given by MMSPE, Minimum Mean Square Prediction Error is equal to $R_y(0)$ minus summation h_i into R_{yi} , i , i going from 1 to M .

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Let us recall

- ❖ The autocorrelation matrix of a WSS process is a symmetric Toeplitz matrix

$$\begin{bmatrix} R_y(0) & R_y(1) & \dots & R_y(M-1) \\ R_y(1) & R_y(0) & \dots & R_y(M-2) \\ \vdots & \vdots & \ddots & \vdots \\ R_y(M-2) & R_y(M-3) & \dots & R_y(1) \\ R_y(M-1) & R_y(M-2) & \dots & R_y(0) \end{bmatrix}$$

- ❖ This lecture will explore the above property to derive the Levinson Durbin algorithm to solve the WH equations for linear prediction

Let us also recall that the autocorrelation matrix of a WSS process is a symmetric Toeplitz matrix. If we see this RY matrix, suppose all elements in this diagonal are equal, all are $R_y(0)$, similarly if I consider this super diagonal, all elements will be $R_y(1)$. Similarly, this sub-diagonal will have all elements $R_y(1)$ like that and also it is symmetric. This element is equal to this, like this element is equal to this, like that. So, this lecture will explore the above property to derive the Levinson Durbin algorithm to solve the WH equations for linear prediction.

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Forward Prediction Problem

- ❖ The linear prediction problem considered so far is a forward prediction problem. For the notational simplicity, we rewrite the prediction equation as

$$\hat{Y}(n) = \sum_{i=1}^M h_M(i) Y(n-i)$$

where the prediction parameters are being denoted by $h_M(i), i=1, \dots, M$.

- ❖ The WH equations in the matrix form is given by

$$\begin{bmatrix} R_y(0) & R_y(1) & \dots & R_y(M-1) \\ R_y(1) & R_y(0) & \dots & R_y(M-2) \\ \vdots & \vdots & \ddots & \vdots \\ R_y(M-1) & R_y(M-2) & \dots & R_y(0) \end{bmatrix} \begin{bmatrix} h_M(1) \\ h_M(2) \\ \vdots \\ h_M(M) \end{bmatrix} = \begin{bmatrix} R_y(1) \\ R_y(2) \\ \vdots \\ R_y(M) \end{bmatrix} \quad (1)$$

$$\text{and MMSE} = R_y(0) - \sum_{i=1}^M h_M(i) R_y(i)$$

We will start with forward prediction problem. The linear prediction problem considered so far is a forward prediction problem. This is obvious from the name itself. For the notational simplicity, we rewrite the prediction equation as $\hat{Y}(n) = \sum_{i=1}^M h_M(i) Y(n-i)$

going from 1 to M, where the prediction parameters are being denoted by h_{Mi} , i going from 1 to M.

Here, we have included this order parameter M as a suffix. The Wiener Hopf equation in the matrix form is given by this. This is the autocorrelation matrix, and this is the filter coefficient vector, first element h_{M1} , h_{M2} like that, and right hand side is the autocorrelation vector R_{y1} , R_{y2} , etc upto R_{yM} , and correspondingly minimum mean square prediction error is given by R_{y0} minus summation h_{Mi} into R_{yi} , i going from 1 to M.

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Backward Prediction Problem

- Given $Y(n), Y(n-1), \dots, Y(n-M+1)$, we want to estimate $Y(n-M)$.
- The backward linear predictor is given by
$$\hat{Y}(n-M) = \sum_{i=1}^M b_M(i) Y(n+1-i)$$
- We have to minimize the mean square prediction error
$$Ee^2(n) = E(Y(n) - \sum_{i=1}^M b_M(i) Y(n+1-i))^2 \quad \text{w.r.t. } b_M(j)s \quad j=1, 2, \dots, M$$
- Applying orthogonality principle.
$$E(Y(n-M) - \sum_{i=1}^M b_M(i) Y(n+1-i)) Y(n+1-j) = 0 \quad j=1, 2, \dots, M.$$

$$\therefore \sum_{i=1}^M b_M(i) R_Y(j-i) = R_Y(M+1-j) \quad j=1, 2, \dots, M$$

We now consider the backward prediction problem. Given Y_n, Y_{n-1} upto Y_{n-M+1} , we want to estimate Y_{n-M} . So, what is the best estimate for Y_{n-M} , given this data Y_n, Y_{n-1} , upto Y_{n-M+1} . The backward linear prediction is given by $\hat{Y}_{n-M}$ that is equal to summation b_{Mi} into Y_{n+1-i} , i going from 1 to M. That is, so this will be equal to b_{M1} into y_{n+1} + b_{M2} y_{n-1} , like that last element will be b_{MM} y_{n-M+1} .

This way we predict Y_{n-M} . We have to minimize the mean square prediction error given by E of e square n , that is equal to E of Y_{n-M} minus summation b_{Mi} into Y_{n+1-i} , i going from 1 to M whole square. So this is the mean square prediction error and this we have to minimize with respect to b_{Mj} , j going from 1 to M. Now we can apply the orthogonality condition. $E(Y_{n-M} - \sum_{i=1}^M b_{Mi} Y_{n+1-i}) Y_{n+1-j} = 0$, j going from 1 to M.

This is the error expression. Into Y_{n+1-j} that must be equal to zero for j is equal to 1, 2, upto M. So taking the expectation now, we will get summation b_{Mi} into R_{Yj-i} , i going from

1 to M must be equal to R_{Yn+1-j} . So E of $Y_n - M$ into Y_{n+1-j} that will give us this. So this set of equation is for j is equal to 1, 2, upto M.

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In the matrix form

$$\begin{bmatrix} R_Y(0) & R_Y(1) & \dots & R_Y(M-1) \\ R_Y(1) & R_Y(0) & \dots & R_Y(M-2) \\ \vdots & \vdots & \ddots & \vdots \\ R_Y(M-1) & R_Y(M-2) & \dots & R_Y(0) \end{bmatrix} \begin{bmatrix} b_M(1) \\ b_M(2) \\ \vdots \\ b_M(M) \end{bmatrix} = \begin{bmatrix} R_Y(M) \\ R_Y(M-1) \\ \vdots \\ R_Y(1) \end{bmatrix} \quad (2)$$

❖ From (1) and (2) we conclude

$$b_M(i) = h_M(M+1-i), \quad i=1,2,\dots,M$$

❖ Thus forward prediction parameters in reverse order will give the backward prediction parameters.

In the matrix form, we get R_{Y0} , R_{Y1} , etc. This is the autocorrelation matrix into that backward prediction filter coefficients b_{M1} , b_{M2} , upto b_{MM} . This is the filter coefficient vector. That must be equal to R_{YM} , and then R_{YM-1} , and upto R_{Y1} . This is the autocorrelation vector, but we observe that this is in reverse order. In the forward prediction, it is in the order, first R_{Y1} , then R_{Y2} , etc.

But here first R_{YM} , then R_{YM-1} , etc. Therefore, from the forward prediction problem 1 and 2 here, we get the relationship between the coefficients b_M of i is equal to h_M of $M+1-i$, where i goes from 1 to M. So, that means the forward prediction parameters in reverse order will give the backward prediction parameters. So this is an important observation, this b_{Mi} 's are nothing but the forward prediction parameters in the reverse order.

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MMSPE for backward prediction

❖ MMSPE

$$\begin{aligned}\varepsilon_M &= E(Y(n-M) - \sum_{i=1}^M b_M(i) Y(n+1-i)) Y(n-M) \\ &= R_Y(0) - \sum_{i=1}^M b_M(i) R_Y(M+1-i) \\ &= R_Y(0) - \sum_{i=1}^M h_M(M+1-i) R_Y(M+1-i)\end{aligned}$$

which is same as the forward prediction error.

❖ Thus

Backward prediction error = Forward Prediction error.

MMSPE, the minimum mean square prediction error that is given by E of $Y_n - M$ minus summation $b_M(i)$ into Y_{n+1-i} , i going from 1 to M , into Y_{n-M} . That is using the orthogonality condition. So, here the remaining terms will contribute zero, because of the orthogonality of the error, this is the error to the data. So, therefore we will have simply mean square prediction error, that is, ε_M , that is the notation is equal to $E(Y_n - M \text{ minus summation } b_M(i) \text{ into } Y_{n+1-i}, i \text{ going from } 1 \text{ to } M, \text{ the whole into } Y_{n-M})$.

And if we substitute, this is equal to $R_Y(0)$ minus summation $b_M(i) R_Y(M+1-i)$, i is equal to 1 to M , and since $b_M(i)$ is equal to $h_M(M+1-i)$, we substitute this $h_M(M+1-i)$ for $b_M(i)$, and we get this expression. If we carefully observe this equation, it is nothing but the forward prediction error, which is same as the forward prediction error. Thus, the mean square backward prediction error is equal to the mean square forward prediction error.

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Levinson Durbin Algorithm

❖ Levinson Durbin algorithm is the most popular technique for determining the LP parameters from a given autocorrelation sequence.

❖ Consider the WH equation for m -th order linear predictor.

$$\begin{bmatrix} R_y(0) & R_y(1) & \dots & R_y(m-1) \\ R_y(1) & R_y(0) & \dots & R_y(m-2) \\ \vdots & \vdots & \ddots & \vdots \\ R_y(m-1) & R_y(m-2) & \dots & R_y(0) \end{bmatrix} \begin{bmatrix} h_m(1) \\ h_m(2) \\ \vdots \\ h_m(m) \end{bmatrix} = \begin{bmatrix} R_y(1) \\ R_y(2) \\ \vdots \\ R_y(m) \end{bmatrix}$$

❖ Writing in the reverse order

$$\begin{bmatrix} R_y(0) & R_y(1) & \dots & R_y(m-1) \\ R_y(1) & R_y(0) & \dots & R_y(m-2) \\ \vdots & \vdots & \ddots & \vdots \\ R_y(m-1) & R_y(m-2) & \dots & R_y(0) \end{bmatrix} \begin{bmatrix} h_m(m) \\ h_m(m-1) \\ \vdots \\ h_m(1) \end{bmatrix} = \begin{bmatrix} R_y(m) \\ R_y(m-1) \\ \vdots \\ R_y(1) \end{bmatrix}$$

Now, we go to Levinson Durbin Algorithm. Levinson Durbin algorithm is the most popular technique for determining the LP parameters from a given autocorrelation data. Consider the Weiner Hopf equations for m -th order linear prediction. Here, this is the autocorrelation matrix, R_y matrix, and this is the prediction coefficient vector $h_m(1)$, $h_m(2)$, up to $h_m(m)$, and the right hand side is the autocorrelation vector, that is given by $R_y(1)$ up to $R_y(m)$.

And if I write the reverse order, this matrix is same, but coefficient is in reverse order. If we write the coefficients in reverse order, and right hand side also accordingly, we have to inverse the order. So first $R_y(m)$ and then $R_y(1)$. So, that way by reversing the order of the coefficient, we get this equation. So, here we consider the m -th order predictor, m -th order prediction.

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Levinson Durbin Algorithm ...

❖ Pre-multiplying both sides in Equation (3) by R_y^{-1} , we get

$$\begin{bmatrix} h_{m+1}(1) \\ h_{m+1}(2) \\ \vdots \\ h_{m+1}(m) \end{bmatrix} + h_{m+1}(m+1) R_y^{-1} \begin{bmatrix} R_y(m) \\ R_y(m-1) \\ \vdots \\ R_y(1) \end{bmatrix} = R_y^{-1} \begin{bmatrix} R_y(1) \\ R_y(2) \\ \vdots \\ R_y(m) \end{bmatrix}$$

$$\therefore \begin{bmatrix} h_{m+1}(1) \\ h_{m+1}(2) \\ \vdots \\ h_{m+1}(m) \end{bmatrix} + h_{m+1}(m+1) \begin{bmatrix} h_m(m) \\ h_m(m-1) \\ \vdots \\ h_m(1) \end{bmatrix} = \begin{bmatrix} h_m(1) \\ h_m(2) \\ \vdots \\ h_m(m) \end{bmatrix}$$

❖ The equations can be rewritten as

$$h_{m+1}(i) = h_m(i) + k_m h_m(m+1-i) \quad i=1,2,\dots,m \quad (5)$$

where $k_m = -h_m(m)$ is called the reflection coefficient or the PARCOR (partial correlation) coefficient.

Then, the $m + 1$ -th order predictor is given by this equation. This is the R_y matrix, this is the h vector, and this is the autocorrelation vector. Here, R_y matrix is of order $m + 1$ by $m + 1$ and this is a vector of dimension of $m + 1$ and similarly R_y is also a vector of dimension $m + 1$. So, our goal is to express this $m + 1$ -th order predictor in terms of m -th order predictor. Let us partition equation 2, these matrices in equation 2 as shown.

So, we will partition the matrices in equation 2, this is the equation 2 like this. Let us partition the matrices in equation 2. We are partitioning by this two lines, so that we have four matrices here, this matrix, this matrix, this matrix, and this matrix. Similarly, two matrices here, and two matrices here. Now, the Weiner Hopf equation can be written in two equations. So first we will multiply this matrix, by this vector, plus this scalar multiply this vector, and then we will get the right hand side.

So, that way this R_y matrix multiplied by this h vector plus h_{m+1} that is this element, into this vector, we will get the right hand side, that is from R_{y1} to R_{ym} , and if I consider the last row, then what we will get is that, this multiplied by this. Summation h of $m + i$ into R_y of $m + 1 - i$, i going from 1 to m , then h_{m+1} into R_{y0} . So last element h_{m+1} into R_y of 0 and that must be equal to R_{ym+1} .

So because of this partition, we get two equations now, equation 3 and equation 4. Now, suppose I pre-multiply this equation by corresponding inverse. Then, I will get this. Similarly, I have to pre-multiply here and pre-multiply here. So, that way pre-multiplying both sides by equations 3 by R_y inverse, we get h of $m+1$, h of $m+2$, up to h of $m + 1$ h_{m+1} of $m+1$ into R_y inverse of this one, R_{ym} , R_{ym-1} up to R_{y1} is equal to R_y inverse multiplied R_{y1} up to R_{ym} .

So, that we have this h_{m+1} vector, plus the scalar h_{m+1} of $m + 1$ into, now this is, if I take R_y inverse, into this vector, I will get the filter coefficient in the reverse order. So, that way it will be h_{mm} , h_{mm-1} , up to h_{m1} and the right hand side will be h_{m1} , h_{m2} , up to h_{mm} . So, what we have observed that, this is $m + 1$ -th order coefficient, and they are related to m -th order coefficients by this matrix equation.

Now, we can write its element like this, h_{mm+1i} is equal to h_{mi} , this is h_{mi} from this vector, plus h_{m+1} into h_{m+1-i} , because it is in reverse order. This is true for i is

equal to 1, 2, upto m. Since this is h_{m+1} of $m+1$, so therefore, this we will write as k_{m+1} h_m of $m+1-1$, because this is $m+1$, here also it will be $m+1$. So, that way we are now able to find $m+1$ -th order LP coefficients in terms of the m -th order LP coefficients and this is true for only i is equal to 1, 2, up to m . This k_m coefficient is called the reflection coefficient or the PARCOR, partial correlation coefficient.

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Levinson Durbin Algorithm ...

❖ From Equation (4), we get

$$\sum_{i=1}^m h_{m+1}(i) R_y(m+1-i) + h_{m+1}(m+1) R_y(0) = R_y(m+1)$$

$$\therefore \sum_{i=1}^m h_{m+1}(i) R_y(m+1-i) - k_{m+1} R_y(0) = R_y(m+1)$$

❖ Substituting for $h_{m+1}(i)$ (5)

$$\sum_{i=1}^m (h_m(i) + k_{m+1} h_m(m+1-i)) R_y(m+1-i) - k_{m+1} R_y(0) = R_y(m+1)$$

$$\sum_{i=1}^m h_m(i) R_y(m+1-i) - k_{m+1} R_y(0) + k_{m+1} \sum_{i=1}^m h_m(m+1-i) R_y(m+1-i) = R_y(m+1)$$

$$k_{m+1} (R_y(0) - \sum_{i=1}^m h_m(m+1-i) R_y(m+1-i)) = -R_y(m+1) + \sum_{i=1}^m h_m(i) R_y(m+1-i)$$

$$k_{m+1} = \frac{-R_y(m+1) + \sum_{i=1}^m h_m(i) R_y(m+1-i)}{\epsilon(m)}$$

$$= \frac{\sum_{i=1}^m h_m(i) R_y(m+1-i)}{\epsilon(m)}$$

where $\text{MMSPE} = \epsilon(m) = R_y(0) - \sum_{i=1}^m h_m(m+1-i) R_y(m+1-i)$.

Now from equation 4, we get summation h of $m+1-i$ into $R_{ym} + 1 - i$, i going from 1 to m , plus h_{m+1} at point $m+1$ into R_{y0} is equal to T_y of $m+1$. This we obtain from this equation 4. So, substituting h of $m+1$ is equal to minus k_{m+1} , we get this relationship. Substituting for h of $m+1-i$ from 5, we have already equation 5, that is the recursive relationship. So, we can write summation $h_{mi} + k_{m+1}$ into h_m at $m+1-i$ into $R_{ym} + 1 - i$, i going from 1 to m minus k_{m+1} into R_y of 0.

That must be equal to R_y of $m+1$ and after simplifying, we will get k_{m+1} into R_{y0} minus summation h_m at point $m+1-i$ into $R_{ym} + 1 - i$, i going from 1 to m , that must be equal to minus R_y of $m+1$ plus summation h_{mi} into $R_{ym} + 1 - i$, i going from 1 to m . So, from this expression, we get k of $m+1$ equals to minus R_y of $m+1$. This is minus R_y of $m+1$ plus summation h_{mi} into R_y $m+1-i$, i going from 1 to m , divided by epsilon m , where this epsilon m is the MMSPE, mean square error given by this expression.

So that way, R_{y0} minus summation $h_{mm} + 1 - i$ into $R_{ym} + 1 - i$, i going from 1 to m . Therefore, k_{m+1} is related with the mean square prediction error by this expression. This is

one important relationship and the reflexion coefficient of order $m+1$ is given by this expression divided by mean square prediction error.

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Levinson Durbin Algorithm ...

$$\varepsilon_{m+1} = R_T(0) - \sum_{i=1}^{m+1} h_{m+1}(m+2-i) R_y(m+2-i)$$

❖ Using the recursion for $h_{m+1}(i)$, we get

$$\varepsilon_{m+1} = \varepsilon_m (1 - k_{m+1}^2)$$

❖ Since MSE is non-negative

$$k_m^2 \leq 1$$

$$\therefore |k_m| \leq 1$$

- If $|k_m| < 1$, the LPC error filter will be minimum-phase, and hence the corresponding synthesis filter will be stable.
- Efficient realization can be achieved in terms of k_m .

Now, the mean square prediction error at order $m+1$ is given by R_y of 0 minus summation h_{m+1} at point $m+2-i$ into R_{ym+2-i} , i going from 1 to $m+1$. Now using the recursion for h of $m+1$, we get this relationship. This mean square prediction also we will be getting recursively. ε_{m+1} is equal to ε_m multiplied by $1 - k_{m+1}^2$, reflexion coefficient square. This will be the MMSPE at order $m+1$.

Therefore, we will be able to determine minimum mean square prediction error also recursively. Now, this quantity is always not negative and since MSE is always non-negative, so from this expression because this is non-negative, this is non-negative, we will get k_m^2 square, that is the reflection coefficient square is always less than equal to 1, implying that reflection coefficient k_m , its magnitude will be less than equal to 1.

If magnitude of k_m less than 1, the LPC error filter will be minimum-phase and hence the corresponding synthesis filter will be also stable. Because here we can ascertain that this quantity is less than 1, so that corresponding LPC error filter will be minimum-phase, because we are elementing in terms of k_m , and the LPC error filter will be minimum-phase and hence corresponding synthesis filter will be also stable.

So, that is one advantage of denoting the modal parameters by the reflection coefficients and the corresponding estimation error. Efficient realization can be achieved in terms of k_m ,

instead of using h_i 's, we can use this k_m , the reflection coefficient to get some efficient implementation of the prediction error filter, that we will discuss later on.

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Levinson Durbin Algorithm ...

❖ k_m represents the direct correlation of the data $Y(n-m)$ on $Y(n)$ when the correlation due to the intermediate data $Y(n-m+1), Y(n-m+2), \dots, Y(n-1)$ is removed. It is defined by

$$k_m = \frac{E e_n^f(n) e_n^b(n)}{R_y(0)}$$

where

$$e_n^f(n) = \text{forward prediction error} = Y(n) - \sum_{i=1}^n h_n(i) Y(n-i)$$

and

$$e_n^b(n) = \text{backward prediction error} = Y(n-m) - \sum_{i=1}^n h_n(m+1-i) Y(n+1-i)$$

❖ For an $AR(M)$ process, $k_m = 0, m > M$

k_m is an very important parameter. It represents the direct correlation of the data Y_{n-m} on Y_n , when the correlation due to the intermediate data Y_{n-m+1} upto Y_{n-1} is removed. If we remove the correlation because of this data, then whatever correlation is present, that is measured by E of forward error into backward error divided by $R_y(0)$. That is the partial correlation coefficient.

So, therefore k_m is equal to $E e_{n-m}^f(n) e_n^b(n)$, where $e_{n-m}^f(n)$ is the forward prediction error is equal to Y_n minus summation $h_{n-m}(i)$ into Y_{n-i} , i going from 1 to n and $e_n^b(n)$ is equal to backward prediction error. That is equal to Y of $n-m$ minus summation $h_{n-m}(m+1-i)$ into Y_{n+1-i} , i going from 1 to m . Now, it can be shown that for an ARM process, this k_m will become zero for m greater than M

So that way it is a test for this partial relation coefficient can be used as a test for the order of ARM process. So that we discussed in the earlier class, that how to determine the order of moving average process, similarly order of AR process can be determined by examining the k_m sequence.

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Steps of the Levinson- Durbin algorithm

- Given order of prediction M and $R_y(m), m = 0, 1, \dots, M$
- Initialization
Take $h_m[0] = -1$ for all m
- $\varepsilon(0) = R_y(0)$
- For $m = 1, \dots, M$

$$k_m = \frac{\sum_{i=0}^{m-1} h_{m-1}(i) R_y(m-i)}{\varepsilon(m-1)}$$

$$h_m(i) = h_{m-1}(i) + k_m h_{m-1}(m-i), i = 1, 2, \dots, m-1$$

$$h_m(m) = -k_m$$

$$\varepsilon_m = \varepsilon_{m-1}(1 - k_m^2)$$

Let us see the steps of the Levinson-Durbin algorithm. Given the order of prediction M and the autocorrelation data for m is equal to 0 to M . We initialize h_m of 0 is equal to minus 1, for all m , then we put Epsilon 0, that is the mean square error of the zero-th order predictor E is equal to $R_y(0)$ and for m equal to 1 upto M , we do these steps. First, we determine k_m by this relationship. So this k_m , reflection coefficient of order M .

These are determined from the $m - 1$ -th order filter coefficient and $m - 1$ order estimation error and then we will update h_m is equal to h of $m - 1$ + k_m into h of $m - 1$ $m - i$. This we will do for i is equal to 1, 2, upto $m - 1$. Then, we will put h_m of m is equal to minus k_m , because here we have determined upto $m - 1$ by this relationship, and since we have determined k_m , h_m of m will be equal to minus k_m .

So m -th order of all the parameters are determined. Then, we will update the mean square prediction error, that is epsilon m will be equal to epsilon $m - 1$ into $1 - k_m^2$. So, that way we will iterate up to M and we will get the prediction coefficients.

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Some salient points

- The reflection parameters k_m and the MMSPEs ε_m completely determine the LPC coefficients. Alternately, given the LP coefficients and the final MMSPE, we can determine k_m and ε_m sequences.
- The algorithm is order recursive. By solving for m -th order linear prediction problem we get all previous order solutions
- If the estimated autocorrelation functions satisfy the properties of an autocorrelation functions, the algorithm will yield stable coefficients.

Some salient points we have to remember, the reflection parameters k_m and the minimum mean square prediction errors, ε_m , completely determine the LPC coefficients. Alternatively, given the LP coefficients and the final minimum mean square prediction error values, we can determine k_m and ε_m for the sequence. This may be required for efficient implementation of the LP error filter.

The algorithm is order recursive, because we first determine, suppose order 1 LP parameters, then determine order 2 LP parameters, like that, therefore the algorithm is order recursive. From the second order LP parameters, we determine the third order LP parameters. By solving the m -th order linear prediction problem, we get all the previous order solutions. So, if we try to solve the m -th order linear prediction problem, we will get the earlier order solutions also, $m-1$ order solutions like that.

If the estimated autocorrelation functions, because all these algorithm we have to use the autocorrelation function, which is actually estimated from data. If the estimated autocorrelation functions satisfy the properties of an autocorrelation functions, that is even symmetry, positive semi-definiteness, etc, the algorithm will yield stable coefficients. That means corresponding construction filter will be also stable. That is one advantage of Levinson Durbin algorithm.

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Summary

- ❖ The forward linear ^{prediction} is given by

$$\hat{Y}(n) = \sum_{i=1}^M h_M(i) Y(n-i)$$

- ❖ The backward linear predictor is formulated as

$$\hat{Y}(n-M) = \sum_{i=1}^M b_M(i) Y(n+1-i)$$

- ❖ Applying orthogonality principle, we get the WH equations

$$\therefore \sum_{i=1}^M b_M(i) R_Y(j-i) = R_Y(M+1-j) \quad j = 1, 2, \dots, M$$

- ❖ It can be shown that

$$b_M(i) = h_M(M+1-i), \quad i = 1, 2, \dots, M$$

- ❖ The backward and the forward PEs are equal

Let us summarize the lecture. The forward linear prediction is given by $\hat{Y}(n)$ is equal to summation $h_M(i)$ into $Y(n-i)$, i going from 1 to M . Similarly, the backward linear predictor is formulated as $\hat{Y}(n-M)$ that is equal to summation $b_M(i)$ into $Y(n+1-i)$, i going from 1 to M . So, these are two prediction problems. Applying orthogonality principle to the backward linear predictor, we get the Weiner Hopf equation as this summation $b_M(i)$ into $R_Y(j-i)$, i going from 1 to M is equal to $R_Y(M+1-j)$, j going from 1 to M .

So from this relationship and earlier our forward prediction relationship, it can be shown that $b_M(i)$ equals $h_M(M+1-i)$, where i is equal to 1, 2 upto M and also we showed that the backward and the forward prediction errors are equal.

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Summary...

- ❖ The Levinson Durbin algorithm exploits the symmetry and Toeplitz nature of the autocorrelation matrix of a WSS process.
- ❖ It computes the LP coefficients recursively from the given ACF data in the following steps:
 - Computation of reflection coefficients k_m
 - Updating $h_m(i)$ s
 - Updating ϵ_m s

The Levinson Durbin algorithm exploits the symmetry and Toeplitz nature of the autocorrelation matrix of a WSS process. It computes the LP coefficients recursively from the given ACF data in the following steps. First step is the computation of the reflection coefficients k_m , then updating the predictor parameters h_m 's, and then updating the mean square prediction errors, ϵ_m . In this way, we can iteratively find out the LP coefficients.

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Try to solve the following problem

Suppose the ACFs of a WSS process are given by

$$R_y(0) = 2.89, R_y(1) = 1.51 \text{ and } R_y(2) = 1.21$$

Find the second-order LP coefficients directly by matrix inversion and applying the Levinson Durbin algorithm.

Try to solve the following problem: Suppose the ACFs of a WSS process is given by R_y of 0 is equal to 2.89, R_y of 1 is equal to 1.51, and R_y of 2 is equal to 1.21. Find the second order LP coefficients directly by matrix inversion and then applying the Levinson Durbin algorithm. Thank you.