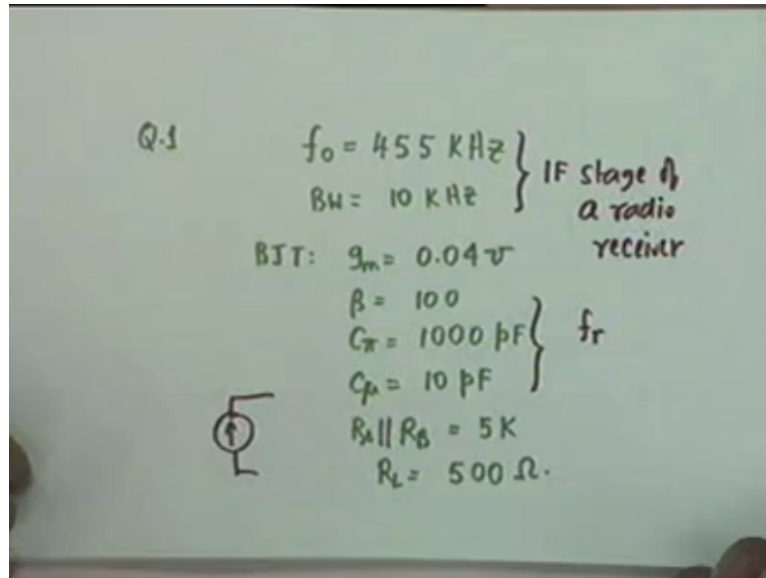


Analog Electronic Circuits.
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Lecture-41.
Problem Session-10 on Tuned Amplifiers.

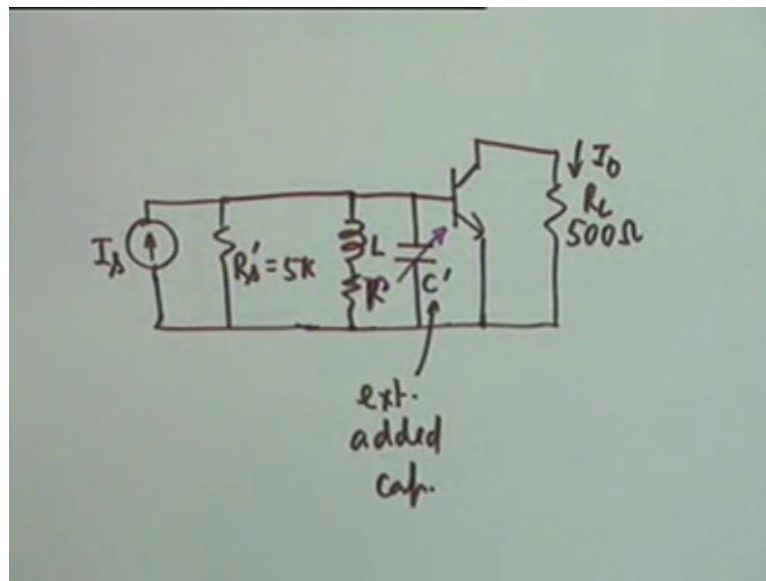
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41st lecture problem session 10 on tuned amplifiers. The 1st problem is to design a single tuned amplifier which has the following specifications. f_0 , the centre frequency is for 45 kilohertz, the bandwidth is 10 kilohertz and as I have mentioned this is a typical IF stage of a radio receiver. The transistor that is to be used has the following specs. g_m is 0.04 mho, β 100, C_{π} 1000 picofarads, C_{μ} 10 picofarads. Now these may be given like this or you may have been given an f_T . If you know f_T then you can calculate C_{π} , provided C_{μ} is given, C_{μ} has to be given.

Then the bias network $r_{sub b}$, the bias that was r_1 parallel r_2 and the source resistance with a current source, the source resistance, they are in parallel, they combined to make 5k. R_L is 500 ohms, in addition as a matter of practical specification it is stated that a coil which is to resonate at 455 kilohertz, at 455 kilohertz coils with Q between 10 and 150 are available, coils between 10 and 150 are available. Which means that you cannot assume the inductor to be perfect, which means the inductor has a series resistance. Is the point clear?

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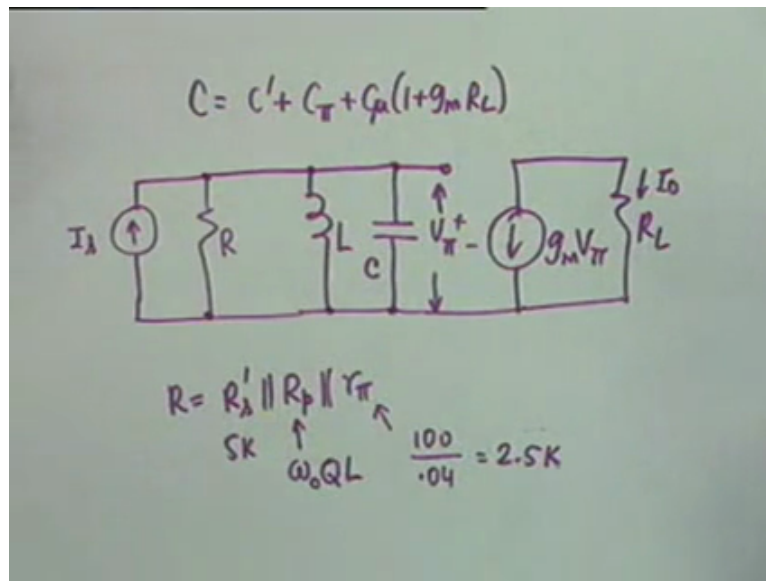


If it is not specified, then you take the inductor to be infinite q , that is its series resistance approximately equal to 0, if it is not specified. If it is specified then you have no alternative. So this is a fairly practical problem, let us see how to solve this problem. The circuit that we plan on, it is almost given, everything is given, it is given that this resistance, we call this r_s prime, okay, r_s parallel r_p , we call this as r_s prime, this is given as 5k, 5k, then you have, you connect the tuned circuit here, it is L in series with resistance r , okay, disease resistance.

We have to use r because q is specified to lie between 10 and 150, that means you have a choice of q . Let us see what q we shall choose. Then there is a capacitance which we call c prime, which is not the total capacitors because the transistor reflects a capacitance, so c prime the externally added capacitance, externally added capacitance. And as we shall see this will dominate, at 455 kilohertz c prime dominates over what is reflected by the transistor. The transistor of course is a common emitter connection, we have not shown the biasing network, that will be taken care of here.

And the load is 500 ohms, this is r_l and this is the current i_0 or i_l and this is the input current i_s , okay. This is the circuit that we plan on. Of course for tuning you have to keep c prime as part of it at least variable, okay. As you know air variable capacitors with the range of few picofarads to what is the maximum, 500 picofarads, actually 478 or something. A few picofarads to 400, a few picofarads to 500 picofarads, this is the air variable capacitor. So if this lies within that, you can use an air variable capacitor and tune it. If it lies outside, if it is more than that, then you can use a fixed capacitor in parallel with an air variable capacitor.

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We will see how that the equivalent circuit now is therefore is, then now what other thing that will come in parallel? One of them is of course r_s prime, okay, we call, would you allow me to change this to small r , the reason is that this equivalent resistance I want to call capital r , okay. Equivalent resistance that comes in parallel with is, allow me to call this small r , then the resistance capital r that will come here would consist of r_s prime which is 5k, then a parallel equivalent of r_l which is r_p . What is the value of r_p ? $\omega_0 Q$ times l , we have derived this.

The series r_l combination has to be converted to a parallel r_l combination and then the equivalent resistance is this, is this correct, $\omega_0 Q l$, yes, it is correct. What is Q ?

$\omega_0 l$ by r .

$\omega_0 l$ by r , okay. Now this parallel, what is the 3rd parameter, r_{π} . And what is r_{π} , β is given as 100 and g_m is 0.04, so this is equal to 2.5 k, okay. This is the total resistance that comes in parallel, then the parallel capacitor, parallel inductance is still l , then we have instead of C' we shall have a C where capital C is equal to C' plus C_{π} plus $C_{\mu}(1 + g_m r_l)$, okay, this is the total capacitors that comes in. And this voltage is V_{π} , then we have the $g_m V_{\pi}$ and r_l , this current i_0 is the output current, all right.

Now the specifications says that the bandwidth, obviously one of the specifications is that parallel, square root of 1 by square root of lC should be equal to ω_0 , whatever is given. But you see we cannot find out l or C directly because C involves r_l and r , that involves, well r does not come into the picture, okay. Yes.

Why do not you consider effective (0)(7:58).

We do not consider because this time, it is usually far away from the resonance frequency. C mu is a small quantity, rl is also a small quantity, so this time constant is very small and therefore it is effective frequency is very large, we do not consider that.

(Refer Slide Time: 8:22)

The image shows handwritten mathematical derivations on a green background. The first equation is $B = 2\pi \times 10^4 = \frac{1}{RC}$. The second equation is $C = \frac{10^{-4}}{2\pi} \left[\frac{1}{5000} + \frac{1}{2500} + \left(\frac{Q}{2\pi \times 455 \times 10^3} \right)^{-1} \right]$. Below this, there is a definition $R_p = Q \omega_0 L = \frac{Q}{\omega_0 C}$. The final equation is $C = \frac{0.95 \times 10^{-8}}{1 - 45.5/Q}$.

Okay, so we take the 1st specification that is bandwidth which is equal to twice pi, look i am multiplying by 2 pi, multiplied by 10 kilos hertz is 10 to the 4, this should be equal to 1 over rc as we have already found out. All right, therefore c, the capacitance that is required is given by 10 to the minus 4 divided by 2 pi multiplied by 1 by r and 1 by r is 1 by 5k +1 by 2.5 k that is 2500 + rp, 1 by rp and 1 by rp is q0.

Which q you will take?

Oh, the series q, okay. Rp is equal to q omega 0l. Now i want to find out c, can i write this as q divided by omega 0c, agreed. Let us do that, i do not know you yet, q has to be chosen, divided by omega 0 is 2 pi times 455 times 10 to the 3c, is that okay? Is this point clear? Q by omega 0, instead of omega 0l i write omega 0 c, instead of omega 0l i write 1 omega 0c because they have to resonate the omega 0.

It is 1 by rp.

Yes, i require 1 by rp because i am making this is equivalent to 1 by r.

(0)(10:15).

Oh, you are right, this should be to the power -1, okay. Now is okay? All right. Now this quantity you see we could have solved for c is the new q, but let us see what are the limitations on q. You will see a very interesting thing. If you find see from here in terms of few after clearing this mess, the result is as follows. 0.95 times 10 to the minus a divided by 1 - 45.5 divided by q, that is what we get. In other words you have to choose a coil whose q is greater than 45.5, is that clear, point clear? And since it is given that you can range between 10 and 150, you choose a convenient value, let us choose q equal to 100.

(Refer Slide Time: 11:18)

Handwritten mathematical derivation on a whiteboard:

$$\text{let } Q = 100$$

$$C = \cancel{0.18 \mu\text{F}} \cdot 0.0173 \mu\text{F}$$

$$C = C' + \frac{1000}{\text{PF}} + 10 \left(1 + 0.04 \times 500 \right)$$

$$\cong C' + 1200 \text{pF}$$

$$C' = \left(\frac{0.0173 - 0.0012}{0.0012} \right) \mu\text{F}$$

$$C' = 0.0161 \mu\text{F}$$

Let us choose a coil with a q of 100, then my c becomes equal to how much? 0.018 microfarads, that is what you get. Pardon me? 0.0181, will someone tell me, i seem to have made a mistake.

It is 1.7 into... yes sir 1.7 into 10 to the power minus 8.

1.7...

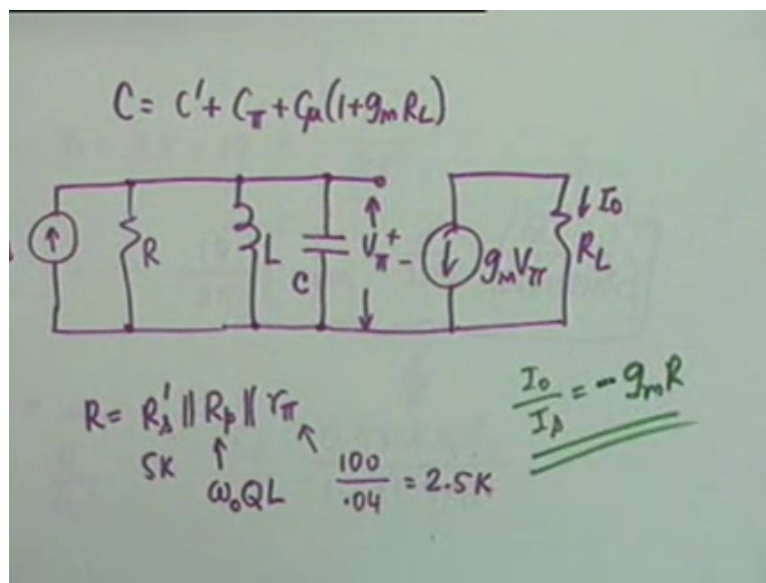
1.743 into 10 to the power -8.

Okay, so 0.0173, is that okay? So many microfarads, agreed. Therefore what is c prime then? We require the external capacitance, you see c prime, c as you know c prime plus c pi which is given at 1000+ c mu which is given as 10 multiplied by 1000 picofarads +10 multiplied by 1+ gm rl, that is 0.04 multiplied by 500 which is approximately c prime +1200 picofarads, agreed. Now therefore c prime is c -1200 picofarads, that is 0.0173 minus let us see how much is this, 0.012, is that okay. 10 to the -2, so 0.1...

It is 0.0012.

0012, do not make a mistake. Now 0.0012, you see that this capacitance is negligible compared to 0.0173. Nevertheless the value is 1610, 0.0161 microfarads, this is c prime. And you see that c prime dominates, c prime dominates in c. This is desirable, why? Because if the transistor is replaced by example, it will have a different c pi and c mu. It should not affect the tuning too badly and nevertheless if the transistor is replaced, that variable component of c prime, that will take care of, you tune it again, that is all. So we have found out c prime, we found out c, what remains to be found out is l.

(Refer Slide Time: 14:37)



$$L = \frac{1}{\omega_0^2 C} = 6.9 \mu H$$

$$R_p = Q \omega_0 L \cong 2 K$$

$$R = 5K \parallel 2.5K \parallel 2K = 910 \Omega$$

$$A_{i0} = -g_m R = -36.4$$

L is $1 / \omega_0^2 c$ and since c is 0.0171, was that, 0.0173 and ω_0 is given, 455 kilohertz, this can be calculated, this comes out as 6.9 micro henry. Now let us also

calculate what is r_p , r_p is $q \omega$ or q divided by ωC_c , if you substitute the value, this comes out approximately as $2k$. Why is r_p needed to be calculated? What did we calculate r_p ?

To see if we require external resistance.

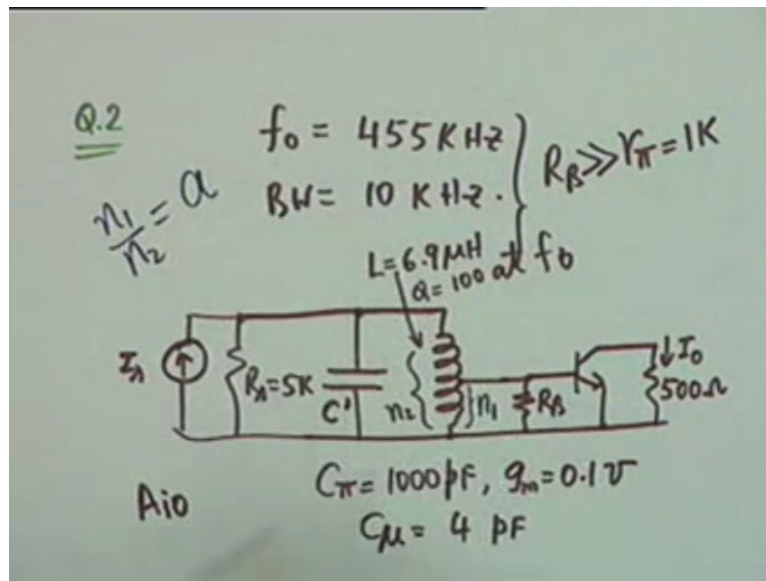
To see if we require external resistance, no. We have already taken care of this, we do not require an external resistance. To find out the mid-band again. Can you tell me what the gain would be? Let us look at the circuit, not midband, the resonant gain, the gain at resonance. Let us look at the surface, what do you think would be the gain at resonance, i_0 by i_s , look at the circuit and tell me? Yes? Pardon me? Would it be a - sign or + sign?

Minus sign.

Minus sign, because i_0 and $g_m v_{pi}$ will not take any. And that resonance you see L and C produce infinite impedance and therefore v_{pi} would be i_s times r , and that would simply be minus $g_m r$, that is the current gain, okay. In order to calculate this we require the value of θ and therefore we require r^2 . And therefore r is the parallel combination of $5k$, $2.5k$ and $2k$ which is equal to 910 ohms in my calculation. And therefore a_{i_0} , the resonant gain is minus $g_m r$ is equal to -36.4 and the design is complete, the design is complete. Any questions before we go to the next one?

Next one would be a little more involved but let me point out that if an inductance, this inductance 6.5 micro henries really a very low inductance. And if an inductance with this value cannot be easily made, maybe it occupies too little space and your standard ferrite core that is available is longer, okay, so it would be beneficial to wind a larger coil, okay. If that is required you do that, you do that and take a tap but the next question is that of designing a tapped inductor or an auto transformer coupled tuned circuit.

(Refer Slide Time: 17:34)



The next question, question 2 is the specifications are the same, almost the same. We will make a slight variation in the transistor, okay. We have, we still have to design a tuned circuit whose f_0 is 455 kilohertz, that is the IF stage of a radio receiver, the bandwidth is still 10 kilohertz and now the circuit is specified, the circuit is specified to be this. Draw the circuit with me. r_s and r_s is found to be 5k, an external capacitance c' , external capacitance c' , then a coil whose inductance is 6.9 microhenry, same as in the previous case, 6.9 microhenry and let say the number of turns here is n_2 .

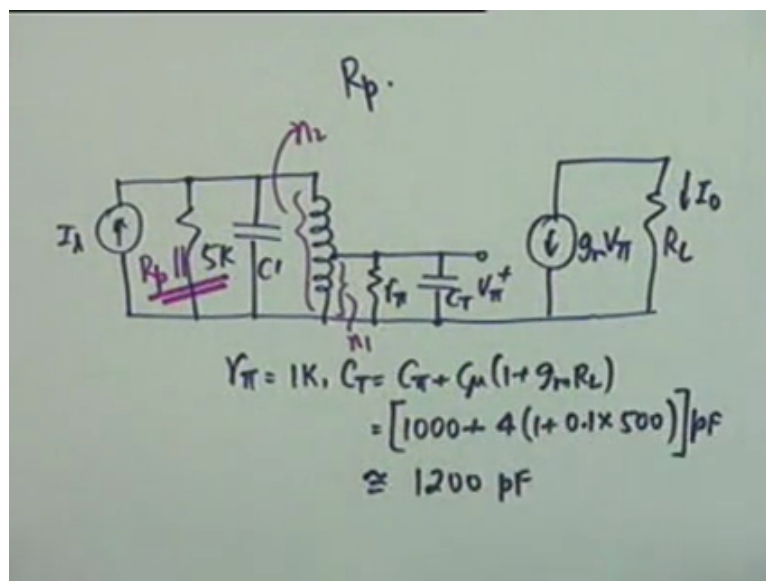
Then you take a tap from here and what comes in parallel with this is the biasing resistance r_b and it is given that r_b is much better than r_{π} and r_{π} is given as 1k, right. So r_b is shown but its effect need not be considered, r_b maybe typically 30 to 40 k and then you have the transistor, emitter, common emitter transistor and the load is the same 500 ohms and this current is i_0 . Here r_{π} is given and the coil Q , coil Q is given as 100 at f_0 , that is at 455 kilohertz, the coil Q is given as 100, c_{π} is given as 1000 picofarads, g_m in order that you cannot use all the data of the previous example, g_m is changed, 0.1 mhos, r_l is given and c_{μ} is also changed to 4 picofarads.

The question is to design this circuit and to find out the mid-band gain A_{i0} , not midband, now it is the resonant gain, centre frequency gain, okay. Now you notice what does the design involve then? All that you have to do now is to find out, let us call this turns ratio as n_1 , all that is required to be done is to find c' , the external capacitance that is needed and that turns ratio n_2 is to n_1 or n_1 is to n_2 , bringing variety into experience, we will call this as a ,

we will call this as the primary and this as a secondary, okay, n_1 by n_2 is equal to a , we will call this either turns ratio.

All that you have to do now is to find out c prime and a , all right, everything else is given. The coil is given to you, you are only allowed to make a tap, now where would you choose the tap, that is the question and what capacitance would be needed. Because the capacitance here would be reflected into the, into the input circuit and therefore c prime that is required shall be different. Let us see how to solve this problem, it is a fairly practical problem and fairly involved problem. The equivalence circuit of this would be the following.

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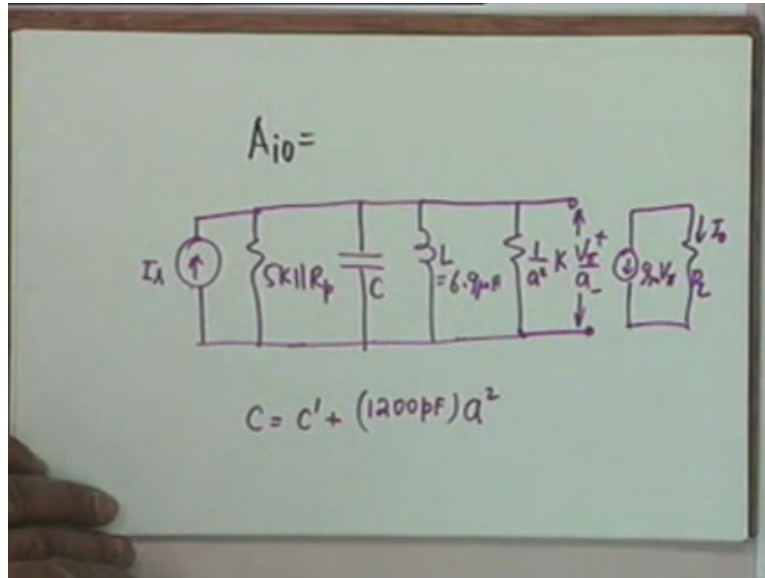
You can draw this very quickly, 5k, then c prime, then the inductance l which is 6.9, then tap, what comes across the tap, 1^{st} is r_{π} and then a c_t , this voltage is v_{π} , where r_{π} , r_{π} is given by 1k and c_t would be c_{π} plus $c_{\mu}(1 + g_m R_L)$, this is equal to 1 + 4 picofarads, 1 + 0.1 multiplied by 500, so it is approximately 1200 picofarads, 1200 picofarads. Then you have $g_m v_{\pi}$ and R_L and this current is i_o , which is the equivalent circuit. And I know everything, pardon me, r_p , oh, the parallel, where would that come, yes, that is a very good point, it should come in parallel with, okay this is, I beg your pardon, the inductance, the inductance is 6.9 micro henry and Q of 100, so this is not a pure inductance.

If it is to be reduced to pure inductance, its equivalent parallel resistance at resonance must be reflected and combined with r_s , very good. So you have to include r_p here, r_p parallel with 5k.

(())(23:32).

No because it is in this, it is above the total and therefore the total inductance has to be here. It is r_{pi} and c_t which will be reflected into the input circuit by, r_{pi} would come as r_{pi} multiplied by, that said, no, no, no, that is wrong, this is n^2 , this as i said i will bring variety into experience so that you are, okay r_{pi} by... c_t would come as a square times c_t , okay.

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So let us look at the input circuit now, equivalent circuit by reflecting r_{pi} including into the input circuit, we have 5k parallel r_p , we have a c now, where c is c' + 1200 picofarads multiplied by a square, is that okay? The capacitance c_t , that would be multiplied by a square, that is how it is reflected. This is c , then you have the inductance l which is 6.9 micro henry as given and the resistance would be reflected as r_{pi} divided by h square, so it is 1 by a square kilos, is that okay? What would this voltage be now?

V_{pi} into a .

Not into a , do not make the mistake again and again, that is why i chose n_1 by n_2 equal to a , it would be v_{pi} by a , then you have $g_m v_{pi} r_{l i0}$. Before i go further, can you tell me what would be the centre frequency gain a_{i0} ? This is not v_{pi} , if it was v_{pi} , then this would have been minus g_m times the equivalent resistance here. This is v_{pi} by a and therefore there should be a factor of multiplication or division?

Multiplication.

So it will be minus $a g_m r$ where capital r is equal to 5k parallel r_p parallel 1 by a square k , agreed. Now let us take the, what is r_p , r_p also we know, 6.9 micro henry, ω_0 we know

and we know q and therefore it would be same as in the previous case, it would be to k. Agreed, the only thing we do not know is 1 by s, what is a, that turns ratio, okay.

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$$R = 5k \parallel 2k \parallel \frac{1}{a^2}k$$

$$\frac{1}{R} = \frac{1}{5000} + \frac{1}{2000} + \frac{a^2}{1000} \quad \checkmark$$

$$= 10^{-4} (7 + 10 a^2)$$

$$\beta = \frac{L}{RC}$$

$\frac{v_i}{v_o}$

$$C = \frac{1}{\omega_0^2 L}$$

$$= \frac{1}{6.9 \times 10^{-6} \times (2\pi \times 457 \times 10^3)^2} \text{ F}$$

$$C' + (200 \text{ pF}) = 0.0173 \text{ MF}$$

$$C' + a^2 \times 0.0012 \text{ MF} = 0.0173 \text{ MF}$$

$$C \cong C' \cong 0.0173 \text{ MF}$$

So the 1st thing we find out is from the bandwidth, from the bandwidth, no, our r is 5k parallel to k parallel 1 by a square k and therefore 1 by r is equal to 1 by 5000, i will require this in the calculation of bandwidth, in the calculation of c, we will see, +1 by 2000 + a square. Divide by thousand or multiplied by thousand?

Divided by.

Okay, wonderful. So many mhos, so many mhos and this comes out as 10 to the - 4, 7 + then a square. So a is the quantity we found out, this can be found out from the bandwidth

specifications. But before we do the bandwidth specification, bandwidth specification as you know is b is equal to 1 over rc , we in radians per second, we require c . Since we know 1 , therefore you can find out c , the required capacitance c is 1 omega 0 square 1 and that is equal to 1 by 6.9 micro henry, okay, 6.9 micro henry multiplied by 2π times 455 times 10 to the 3 whole squared. And this comes out as 0.017 , it would come out exactly as 0.0173 , exactly as in the previous case, is not it, 0.0173 microfarads, so many farads, so many microfarads.

And c as you know, it should come out to be the same, okay and c as you know is c prime plus, did we find that out, 1200 picofarads multiplied by a square. Now that is, you look at now the designer's common sense. C prime plus a squared reply by 1200 picofarads means how many microfarads, 0.0012 microfarads is equal to 0.0173 microfarads. Now the designer comes into effect, a is less than 1 , is not that right, a is $n1$ by $n2$ and therefore this quantity will be very small compared to the right-hand side. And therefore c prime is equal to c prime is approximately equal to 0.173 microfarads, that is approximately equal to c , agreed.

(Refer Slide Time: 30:44)

Handwritten mathematical derivation on a green background:

$$2\pi \times 10^4 = \frac{1}{RC} \quad \checkmark \quad \frac{8RC}{}$$

$$= \frac{1}{0.0173 \times 10^{-6} \cdot 10^{-4} (7 + 10a^2)}$$

$$a^2 \approx 0.4$$

$$a \approx 0.63$$

$$A_{jo} = -a g_m R$$

On the left side of the image, there is a circuit diagram showing a resistor R in series with a capacitor C , and a voltage source V across the capacitor. Below the diagram, the expression $\frac{1}{2\pi RC}$ is written.

Now we can go and find out from the bandwidth specifications turns ratio. The bandwidth specification is 2π times 10 to the minus 4 would be equal to 1 by rc , that is 1 by 0.017 multiplied by 10 to the -6 , that is c , 173 , okay, 1 by c , then 1 by r , 1 by r we have already found out to be 10 to the -4 times $7+10$ a square, right, which has now to be cramped up for finding out the value of a . And a square comes out approximately as 0.4 , so that a is equal to 0.63 , a is equal to 0.63 . Which means that if the original one is 10 turns, this should be 6.3 turns, $n1$ should be 6.3 turns, okay. And the centre frequency gain which was minus $a g_m r$...

Excuse me sir.

Yes?

The unit of $1/RC$ is hertz (31:54).

One over RC is hertz...? Okay, $1/RC$ is radians per second, yes. You see s/RC is a dimensionless quantity and s is in radians per second and therefore RC must be inverse of radians per second, which is okay, do not make a mistake. You see if I have, if I have a simple lowpass filter, whereas the cut-off frequency, $1/2\pi RC$, this is in radians per second.

What is 2π into 10 raised to power minus 4... (32:46).

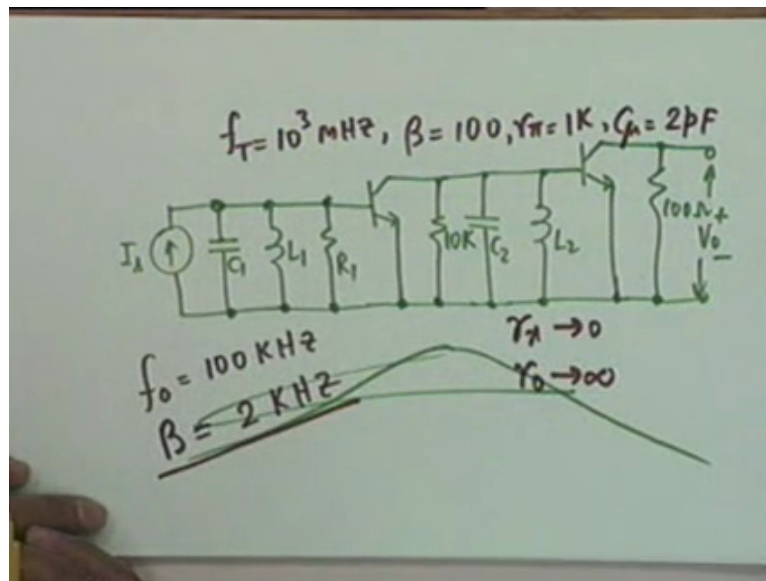
Oh, what mistake did I make?

Minus sign.

It should be plus, I am sorry about that, it should be plus, otherwise everything goes wrong. Okay, the final problem of today, is there any question on this? You have to do this very carefully, what is required to be done must be very clear to you. Your aim is to find c prime and a , these are the 2 quantities and these are interlinked with each other. You see there was a point when, when we were we were absolutely at a loss to understand what should be done. Suppose that we did not ignore this, then we had to have an equation which contains both c prime and a square, only one equation. You could not solve for both but this commonsense that a is less than 1 has helped us to solve the problem.

This is how a designer comes into application of this commonsense. The final problem is that of a synchronously tuned amplifier and we shall assume that we have only 2 stages of synchronous tuning, in other words we have to resonant circuits which resonate at the same frequency and their bandwidths are the same.

(Refer Slide Time: 34:24)



And one of these very simple way of making this is to have 2 transistors, yes, you have for simplicity let us assume that perfect coils are available, if they are not, then you have to convert them into L in parallel with r_p , okay. So let us assume that we have perfect coils like this or coils whose resistance has been taken into account in, we will introduce a resistance r_1 here, okay, in r_1 . Then you have, this is one of the tuned circuits, the other tuned circuits comes in the load or in the collector circuit of this transistor and for some reasons, for biasing and other considerations, this resistance is 10 k.

For biasing this transistor, this resistance is 10 k, then you have a capacitance c_2 , and inductance l_2 , once again we assume that the pile of resistance of l_2 has been observed in 10 k, all right. And then you have a 2nd transistor load is 100 ohms and this is the output, for variety in the experience we assume that the output is the voltage. A question that could be asked at this stage is, well, why do not you take the output from here, from here itself? Why did you take the output from here itself? The condition of the problems says you could not, why?

Pardon me? Because the load is specified, it has very low resistance, 100 ohms, agreed. If you put 100 ohms here, all this tuning, not tuning, the bandwidth, that goes absolutely haywire, the bandwidth becomes very very large. In other words the tuned circuit becomes a flat tuned circuit like this, okay. So you need, you do require a buffer, okay. Now let us see how to solve this problem. The equivalent circuit if we draw, oh the specifications are as follows. Now again to bring variety into experience, again to bring variety into experience,

the specifications are that f_t is given, 10 to the 3 megahertz, okay, f_t is given and the bandwidth, well it is required to resonate at 100 kilohertz and the overall bandwidth b is required to be 2 kilohertz, overall bandwidth is required to be 2 kilohertz, that is it.

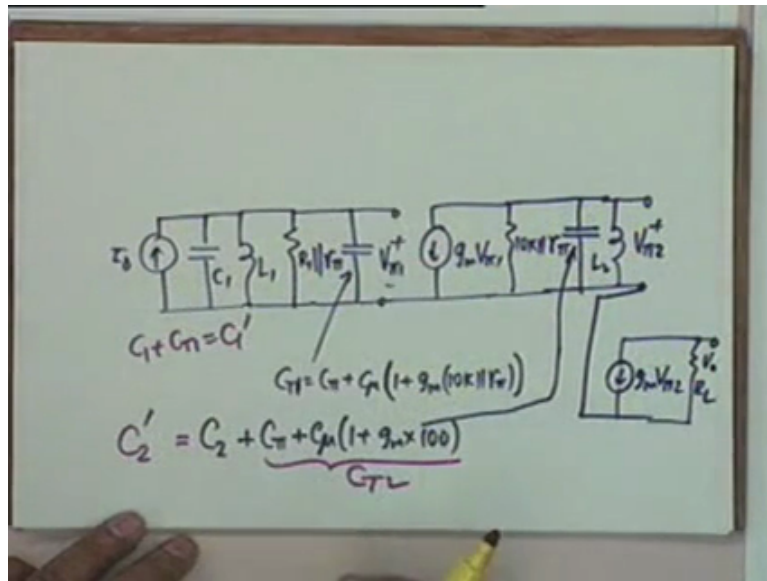
The transistors that are used have f_t equal to 10 to the 3 megahertz, β equal to 100 for both of them, r_{π} is 1k, in other words both have the same g_m , okay, r_{π} is 1k. What else do you need? What else you need? C_{μ} is given as 2 picofarads and it is said that r_x tends to 0 and r_0 tends to infinity, all right. This is the specification of the problem, you have to find out. What are the things to be found out here? This capacitance c_1 , the inductance l_1 and we do not know what this resistance is, we have to find r_1 , then we have to find c_2 , l_2 , that is all?

Gain.

Gain, okay, that of course, that also. Can we write the expression for this, gain at the centre frequency, can we write an expression, write by looking at it? Can you do that? For the last stage without doing the equivalent circuit, for the last stage it is minus g_m times 100, okay. And for the 2nd stage, minus g_m times 10 k parallel r_{π} , 1k, that is right. Then for the 1st stage... yah? For the input circuit. V_{π} over i_s , that is the gain now. So that will be r_1 parallel r_{π} , absolutely wonderful.

We can write this by inspection and therefore if you know g_m , you can find out g_m , all that you need to find out to find the centre frequency, find the centre frequency gain is to find r_1 , that is the only thing that is unknown, all right. And the equations, now let me draw the equivalent circuit, then the equation should be clear.

(Refer Slide Time: 39:35)



What i have is is parallel c1, 11...

Excuse me sir one question.

Yah?

This ai 0 is the ratio between what and what, v 0 by is?

That is right because our output is v0 and input is is. It has a dimension of, look at it, it has a dimension of...? Resistance, this is dimensionless, this is dimensionless, so this has dimension of resistance. It has to be v0 by is, okay. C1, 11, then you have the parallel combination of r1 and r pi, agreed, the 1st transistor. And you have the capacitance as, actually this capacitance, okay, since we have drawn it separately, this is ct1 which is equal to c pi plus c mu 1+ gm times 10 k parallel r pi. Okay. This is v pi 1, then you have gm v pi 1, this resistance should be 10 k parallel r pi, agreed, 10 k parallel r pi, then you have, what would be this resistance?

C2 plus, what shall be reflected? C pi plus c mu + 1 + gm multiplied by 100, that is correct, this is the total capacitance. And then you have l2, of course that has to be there and this voltage is v by 2 and finally, finally we shall have gm v pi 2 and in parallel with rl, this voltage is v0. All right, i could not find place here so i have brought it down, is it okay? Or have i made a mistake? Do not be too sure, i can make a mistake. I shall call this as, what shall i call it, c2 crime let us say , ct2, well, i did not want to use that because i had used this

ct1 here, this is ct 2, okay. And i will call c1 plus ct 1 as c1 prime, that is the total capacitance here. Then i have 1, 2, 3, 4 equations, what are these equations?

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$$\omega_0^2 = \frac{1}{L_1 C_1'} = \frac{1}{L_2 C_2'}$$

$$B_1 = \frac{B \leftarrow 2\pi \times 2 \times 10^3}{\sqrt{2^{\frac{1}{2}} - 1}} = \frac{1}{C_1' (R_1 || r_\pi)}$$

$$= \frac{1}{C_2' (10k || r_\pi)}$$

L_1 L_2
 C_1 C_2
 R_1

Mind you it is synchronous tuning, therefore the equations are omega 0 squared equal to 1 by L1 c1 prime, it should be same as 1 by L2 c2 prime, this gives me how many question, 2? 2. And the other one is the bandwidth equation, now there comes the question. What bandwidth should be used? The overall bandwidth is required to be 2 kilos hertz, so what is the individual bandwidth b1? How do you know?

Synchronously tuned.

Synchronously tuned, so the bandwidth shall shrink. That shrinkage factor you must take into account. It should be the overall bandwidth divided by, obviously the single stage bandwidth has to be larger than the overall bandwidth, so you divide by square root 2 to the power half -1, this you have to find out. And this is given as 2 kilos hertz, so this is 2 pi multiplied by 2 multiplied by 10 to the 3, is a the clear? You might for career, if you use a 2 kilos hertz, then your calculations will go wrong. This should be equal to 1 by, 1 by c1 prime and the total resistance, total resistance is r1 parallel r pi, it should also be equal to 1 by c2 prime multiplied by what?

10 k parallel r pi, how many questions will we get here? 2. So you have 4 equations. What are the things you have to find out? L1, l2, c1, c2, what else and r1, these are the 5 things we have to find out. So to have 4 equations, so one of them you can choose this at your (()) (45:04). You would normally like, you would normally like l1 and l2 to be equal because this

is the problem point. Winding an inductance of a specified value is a problem. So if you have wound it once, then you know exactly how many turns you have to use. So normally you choose l_1 equal to l_2 , that constraints you to 4 equations, okay because you have made one choice, then the rest can be calculated, more on monday.