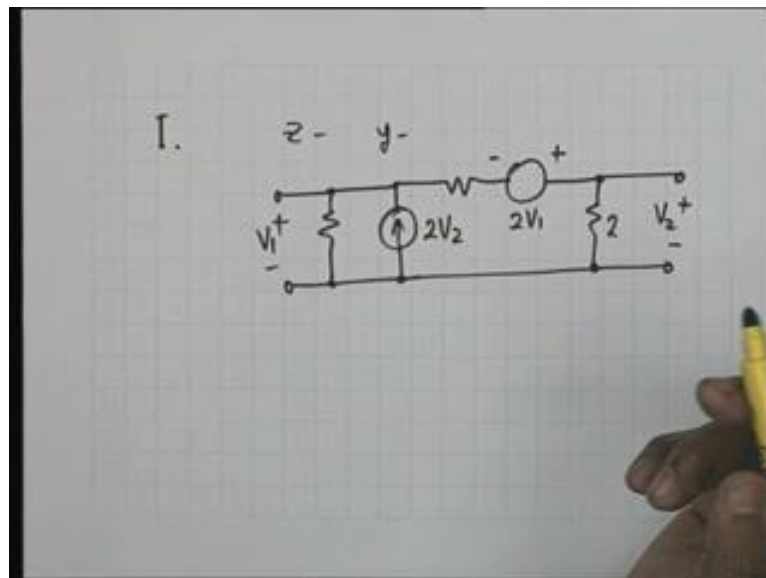


**Circuit Theory**  
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**Lecture - 29**  
**Problem Session 7: Two-port Networks (contd.)**

This is lecture 29 and this is our problem solving session number 7. And we solve problems on 2 port parameters as a finale to 2 port parameters. 2 port parameters we have already discussed in theory and we end up this discussion with a few problems of as i said Non routine nature.

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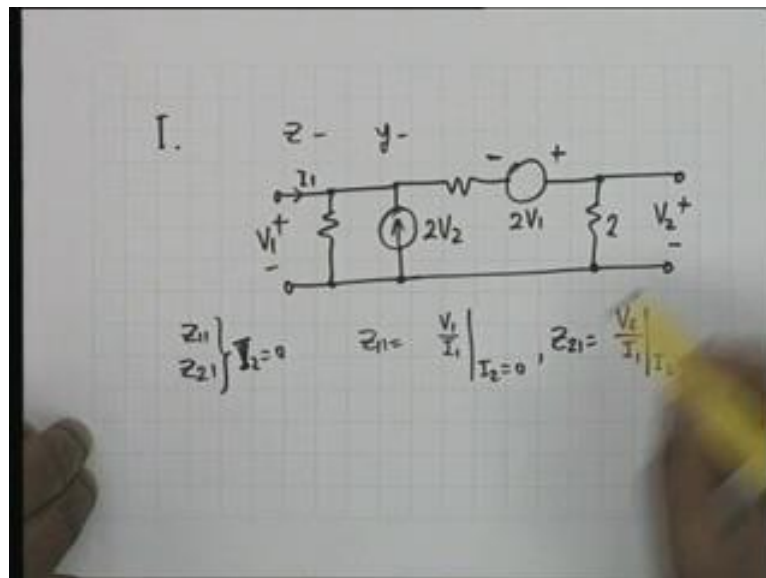


We start with problem number 1 and the problem is to determine the Z and y parameters of the network shown below, which as you see, is not a totally RNC network, it has controlled sources dependent, sources a current generator and a voltage generator with specified polarities. This is 2 ohm and this is  $V_2$  this is  $V_1$ . Just to identify the ports. We have to find out the Z and y parameters of this network. 1 thing that, 1 should be careful about the occurrence of control sources is that if it is voltage for example,  $V_1$  is killed. For example, if we want find  $y_{22}$  then  $V_1$  this will be short circuited. If  $V_2$  is killed then this will also disappear and therefore, it offers simplification.

Another things that 1 should notice is in a similar manner, if  $V_2$  is short circuited if port 2 is short circuited. Then what will happen to this current generator.

It will be opened.  $V_2$  equal to 0 the current is 0. Current generator internal impedance is infinity and therefore, this will be opened. The other thing that one should be careful is, whether the network is reciprocal or not because the occurrence of controlled sources is has the potential of making the network Non-reciprocal and therefore, you should not only calculate  $Z_{1, 2}$  and say  $Z_{1, 2}$  equal to  $Z$  to 1. As you do with reciprocal networks or networks composed of bilateral elements controlled resources we cannot say, If controlled sources occur it is not necessary that the network is not reciprocal, but if controlled sources occur there is a possibility that the network may be Non-reciprocal. In other words, all the 4 parameters have to be computed not 3.

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Now, let us see let us calculate  $Z_{11}$  and  $Z_{21}$  for both of them. The definition says  $V_2$  equal to 0 and  $Z_{11}$  is equal to  $V_1$  by  $I_1$ .

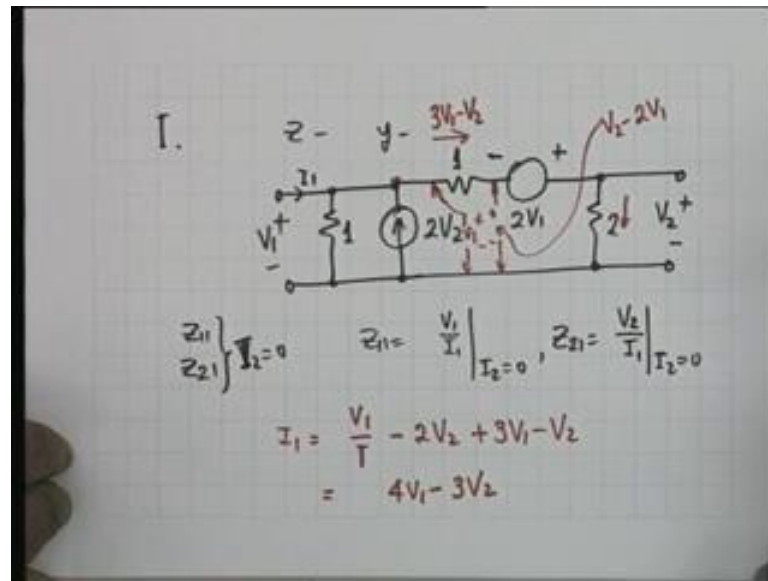
$I_2$  is equal to 0

That is correct.

Thank you.

$V_1$  by  $I_1$ ,  $I_2$  equal to 0 and  $Z_{21}$  is equal to  $V_2$  by  $I_1$  with  $I_2$  equal to 0. So, let us keep  $I_2$  equal to 0 and therefore, these terminals are left open and we have to find out this current  $I_1$  this current  $I_1$  and this voltage  $V_2$ . Then we will be able find out both these parameters.

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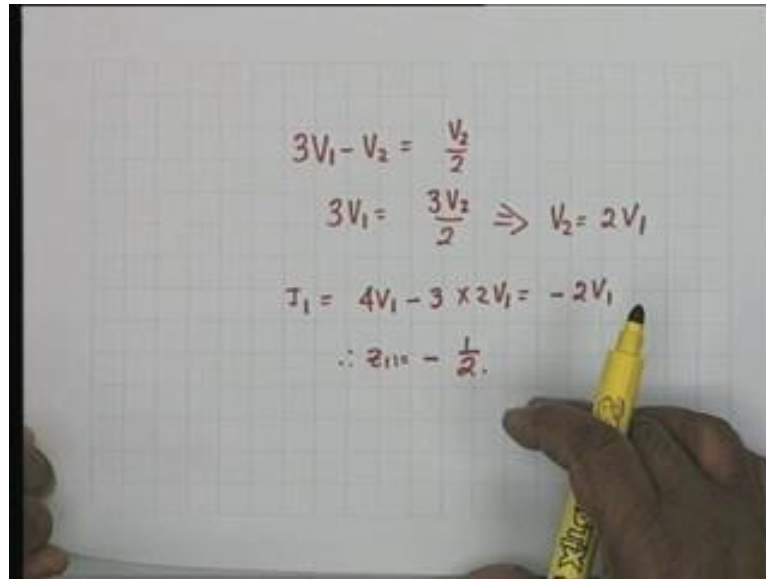
One thing that we notice is that if this voltage now, let me use a different colour. If this voltage is  $V_1$  then this voltage is also  $V_1$ . Now, this resistance is 1 ohm that is given this resistance is 1 ohm and this resistance is also 1 ohm. I did not put the resistance values are here. Let's go systematically. This voltage is  $V_1$  and what is this voltage, this is  $V_2$ .

$V_2$  minus  $2V_1$ .

This is  $V_2$  then there is drop to  $V_1$ . So, this voltage is  $V_2$  minus  $2V_1$  with the same polarity plus minus; therefore, the current through 1 ohm this current shall be equal to this voltage minus this voltage divided by 1 agreed. So,  $V_1$  minus  $V_2$  plus  $2V_1$ . In other words, it is  $3V_1$  minus  $V_2$  agreed and that is solves the problem. Once you are able to identify this current, you see my equation shall be  $I_1$ . If i write KCL at this point  $I_1$  I at this point.  $I_1$  is equal to  $V_1$  by 1 that is this current. Current through this resistance then  $2V_2$  is coming. So, minus  $2V_2$  minus  $2V_2$  then  $3V_1$  minus  $V_2$  is going so  $3V_1$  minus  $V_2$ .

If you simplify this it is simply  $4V_1$  minus  $3V_2$  agreed. This is equal to  $I_1$ . We also notice, while this kind of a simple minded approach requires experience, requires practice while you also notice that this current  $3V_1$  minus  $V_2$  must be the same as this current that is  $V_2$  by 2 agreed.  $V_2$  by 2 it is the same current as this.

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The image shows a hand holding a yellow marker writing on a whiteboard. The equations written are:

$$3V_1 - V_2 = \frac{V_2}{2}$$
$$3V_1 = \frac{3V_2}{2} \Rightarrow V_2 = 2V_1$$
$$I_1 = 4V_1 - 3 \times 2V_1 = -2V_1$$
$$\therefore Z_{11} = -\frac{1}{2}$$

Therefore, the equation of our circuit is  $3V_1 - V_2$  is equal to  $\frac{V_2}{2}$  or  $3V_1$  equal to  $\frac{3V_2}{2}$  which makes  $V_2$  equal to twice  $V_1$  agreed.  $V_2$  equal to twice  $V_1$ , which means if I combine now, with the first equation I get  $I_1$  as equal to  $4V_1 - 3 \times 2V_1$  which is equal to minus  $2V_1$ . Is it have I made a mistake somewhere?

No.

Therefore,  $Z_{11}$  which is the ratio of  $V_1$  to  $I_1$  is minus half is it.  $I_1$  is minus  $2V_1$  and therefore,  $Z_{11}$  is minus half.

It comes as negative that is a negative resistance is realized at port number 1. And happens because of controlled sources which mean, that the circuit is active wherever negative resistance occurs instead of dissipating energy it generates energy and therefore, the circuit is an active circuit.

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$$3V_1 - V_2 = \frac{V_2}{2}$$
$$3V_1 = \frac{3V_2}{2} \Rightarrow V_2 = 2V_1$$
$$I_1 = 4V_1 - 3 \times 2V_1 = -2V_1$$
$$\therefore Z_{11} = -\frac{1}{2}$$
$$Z_{21} = \frac{V_2}{I_1} = -1 \quad \checkmark$$

Number 2: I want to find out  $Z_{21}$  which is equal to  $V_2$  by  $I_1$ .

No.

It is simply  $V_2$  by  $I_1$ .

Yeah pardon me.

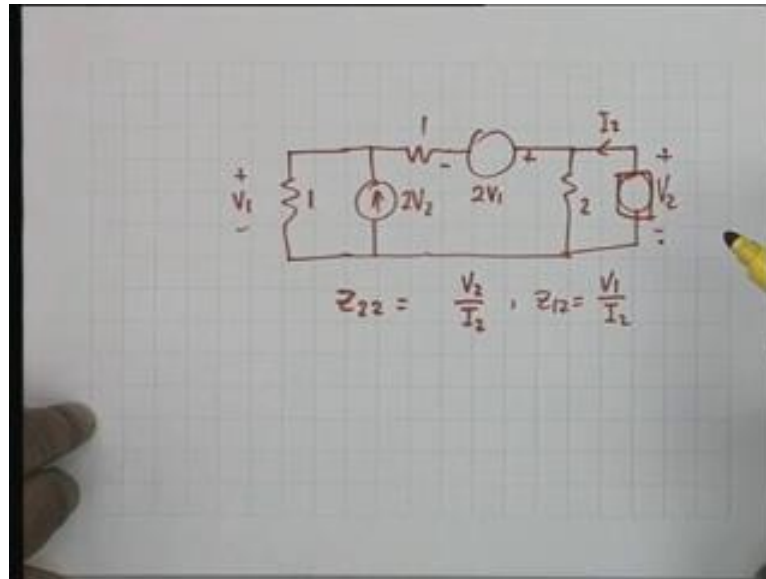
What did you say?

Minus one.

That's right.

Because how do I find this  $V_2$  is  $2V_1$  and  $I_1$  is minus  $2V_1$ . Everything expressed in terms of  $V_1$ . So,  $Z_{21}$  is equal to minus 1. Now, this does not mean that  $Z_{12}$  should also equal be equal to minus one. You have to calculate that, in other words, there is no escape you have to take the network again and put  $I_1$  equal to 0 and the network then becomes.

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1 this is  $V_1$ ,  $I_1$  is 0. No entry of current from port 1, Port one is open circuited then you have a  $2V_2$  and then 1 ohm. The source minus plus  $2V_1$  2 then you have an  $I_2$ . You have to calculate  $Z_{22}$  say you connect the voltage source.  $V_2$  and all you have to find out now is,  $I_1$  is 0; therefore,  $Z_{22}$  is equal to  $V_2$  by  $I_2$  and  $Z_{12}$  is equal to.

Yeah.

$V_1$  divided by  $I_2$ .

Uh it does not matter.

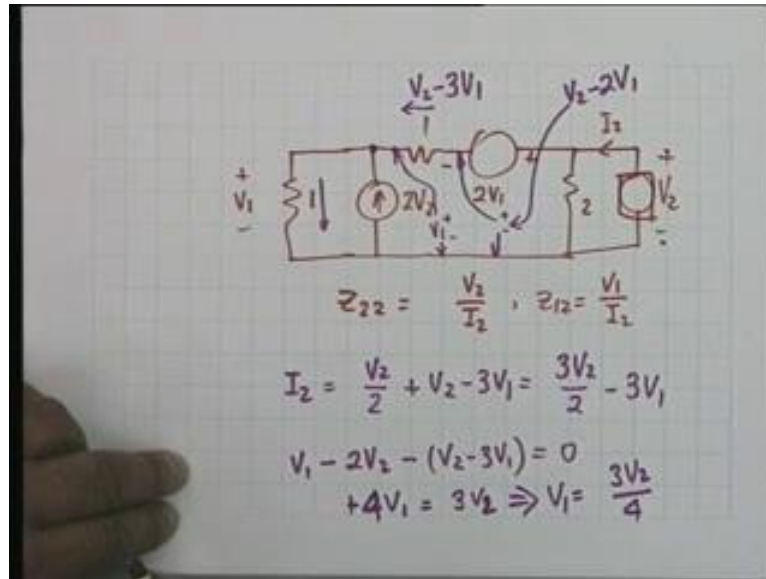
I am not shown, whether it is a voltage source or current source. It could be current source, it could be voltage source. All  $I_1$ , all that matters is this is  $V_2$ , this voltage and the current drawn is  $I_2$  that is all. I have not drawn a well, i have a drawn a voltage source this is the symbol. If i have drawn a current source I would have shown a arrow, but it does not matter you simply call this as source. If necessary you call it a you indicate it by rectangle to indicate that, we do not know what this source.

We do not care.

Not not that we do not know.

We do not care.

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Now, to find  $I_2$ , let us see what these voltages are. This voltage is  $V_1$  and this voltage as we find out earlier is  $V_2$  minus  $2V_1$ ; therefore, this current let us take.

In this direction.

This current would be  $V_2$  minus  $2V_1$  minus  $V_1$ ; that means, it will be  $V_2$  minus.

$3V_1$  all right and then all that you have to do is to write the node equation at this node  $I_2$  would be equal to  $V_2$  by 2.

It is this current plus the current that goes via this branch which is  $V_2$  minus  $3V_1$  that is; equal to  $3V_2$  by 2 minus  $3V_1$  agreed.  $I_2$  equal to  $3V_2$  by 2 minus  $3V_1$  and then you find out the KCL at this node. The current through here, current through this branch is  $V_1$  by 1 then minus  $2V_2$  because it goes it goes up and minus this current comes in minus  $V_2$  minus  $3V_1$  this must be equal to 0.

Which says that Let us say

Minus  $2V_1$  equal  $3V_2$ . So, no I have made a mistake.

This would be plus so  $4V_1$ .

$4V_1$  equal to  $3V_2$ .

Therefore,  $V_1$  equal to  $3V_2$  divided by 4.

Don't allow me to make mistake.

$V_1$  equal to  $3V_2$  by 4 and therefore if I substitute this.

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$$I_2 = \frac{6}{24} \cdot \frac{3V_2}{4} - \frac{9V_2}{4} = -\frac{3V_2}{4}$$
$$Z_{12} = -\frac{4}{3}$$
$$Z_{12} = \frac{V_1}{I_2} = \frac{3V_2/4}{-3V_2/4} = -1$$

Active Reciprocal.

Then I get  $I_2$  equal to  $3V_2$  by 2 minus  $3V_1$ , that is  $9V_2$  by 4. And I can write this as  $6V_2$  by 4; therefore, this is equal to

Minus  $3V_2$  by 4 and therefore,  $Z_{12}$  is equal to minus 4 by 3. This is also negative. Finally  $Z_{12}$  which is equal to  $V_1$ .

Yes.

$V_1$  by  $I_2$ ,  $V_1$  is  $3V_2$  by 4 and  $I_2$  is minus  $3V_2$  by 4. So, this is equal is minus 1.

What was  $V_1$  2?

Also minus 1; therefore, even though  $Z_{12}$  1; therefore, even though controlled sources occur the circuit is reciprocal, but not passive. It is active and reciprocal. Is there any other way of working out this problem?

Pardon me.



What equations?

Mesh and loop equation while that we do not want to do.

Till we are first too

Because; that means, writing a set of simultaneous equation solving for the current sensor. If we do not want, if we can do it by inspection why not?

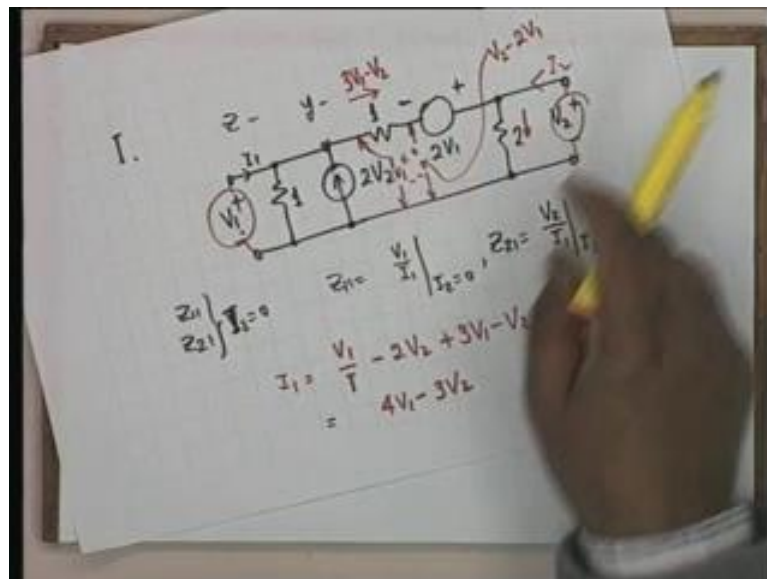
That is

Yes.

What we can do is.

You see after there are only 2 nodes.

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We could connect a voltage source here and write node equations here and here, i could write in terms of,  $I_1$  and  $I_2$  could be expressed in terms of  $V_1$  and  $V_2$ . Then we get the Y parameters and then we could convert them to Z parameters or well no or that is what we will have to do. Otherwise, we will have to write  $V_1$  in terms of  $I_1$  and  $I_2$ . Well that also can be done, you can apply Thevenin's theorem here, make it into a single loop network. There are many ways that this can be done, but let us do in this simple minded manner going back to the roots. Remember going back to the roots, the chances of making a

mistake is reduced considerably. And this problem is not too complicated that we cannot solve it by going back to the roots without any complication.

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$$\begin{aligned}
 & \begin{matrix} y_{11} & y_{12} & y_{21} & y_{22} \end{matrix} \\
 [y] &= \begin{bmatrix} -\frac{1}{2} & -1 \\ -1 & -\frac{4}{3} \end{bmatrix}^{-1} \\
 &= \begin{bmatrix} \frac{-4}{3} & +1 \\ -\frac{1}{3} & -\frac{1}{2} \end{bmatrix} \\
 &= \begin{bmatrix} 4 & -3 \\ -3 & \frac{3}{2} \end{bmatrix}
 \end{aligned}$$

Let's see  $y_{11}$ , the  $y$  parameters  $y_{11}$ ,  $y_{22}$ ,  $y_{12}$  and  $y_{21}$ . To find  $y_{11}$  and  $y_{21}$  this 2.

Pardon me

I can use the transformation also, that is 1 method. Well, the transformation means you will have to find the determinant of the metrics divide the corresponding, it is not too difficult. Let's do that, then maybe we will go back to the roots and see. You see going back to the roots it becomes much simpler. The  $y$  parameters will become much simpler, but suppose we say what was our  $Z_{11}$  minus half and  $Z_{12}$  minus 1 minus 1 then

Minus.

4 by 3

Inverse of this.

So, this is equal to minus 4 by 3, this 1 divided by what is the determinant.

8 by 3 minus 1.

Is it right?

This multiplied by this

2 by 3 minus 1 which is equal to minus one-third agreed. So, this divides by minus one-third then minus 1 divided by minus one-third same comes here.

Here I get minus half

It should be 1 divided by how wonderful.

The negative sign goes off agreed. And this should be minus half divided by.

Minus one-third.

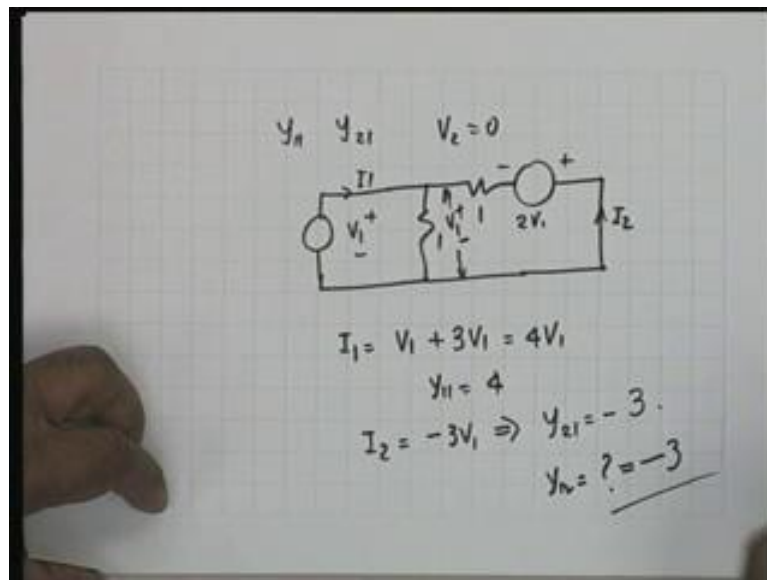
So, the Y parameters finally, become 4.

Minus 3 minus 3 and

3 by 2.

Let us see, if this agree if we make the calculations that is from the roots.

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For  $y_{11}$  and  $y_{21}$ , we have to make  $V_2$  equal to 0. And if we do that, the equation the circuit simply becomes this we have a source  $V_1$ ,  $I_1$  then we have this 1 ohm.  $2V_1$  that currents generator vanishes, because  $V_2$  is made equal to 0. Current generator vanishes and therefore, I have minus plus  $2V_1$  and a short circuit. The current through which is  $I_2$ .

This becomes the circuit and what we have find out is  $I_1$  and  $I_2$ . Now, if you write the node equation at this point  $I_1$  is  $V_1$  by 1 and what is the voltage across, this is  $V_1$ .

What is the voltage across this?

Is it minus 3,1 or plus 3,1?

Plus  $3V_1$  and therefore, the current is  $3V_1$ .

You mustn't make a mistake.

This is  $V_1$  minus plus and this is minus plus and therefore, the drop across this will be  $V_1$  minus minus  $2V_1$  which is equal to  $3V_1$ . Suppose, the polarity was this and this was plus this was minus then  $i$  would simply  $V_1$  minus  $2V_1$ , but no it is  $V_1$  minus minus  $2V_1$ . So,  $4V_1$  is equal to  $I_1$  which means  $y_{11}$  will equal to 4. We have already found that and  $I_2$ ,  $I_2$  is simply,

We have already found this current as  $3V_1$ . So, it is minus  $3V_1$ ; therefore,  $y_{21}$  is equal to minus 3. That's what we have found out. And in a similar manner we can find out  $y_{11}$  and  $Y_{12}$ . Now, if we have found out  $V_{11}$ ,  $V_{12}$ ,  $V_{21}$  and  $V_{22}$ . Do we still have to calculate  $y_{12}$ ?

No.

We argue that we have already shown that the network is reciprocal; therefore,  $y_{12}$  must be equal to minus 3. All that we have to calculate now is  $y_{11}$  and which is also simplicity itself. Because short circuiting the input terminal reduces the voltage source to 0. And there is only a current source all right and it becomes simpler to take care of this situation. We go to problem three problem 3.

We skip one of them and problem 3 says.

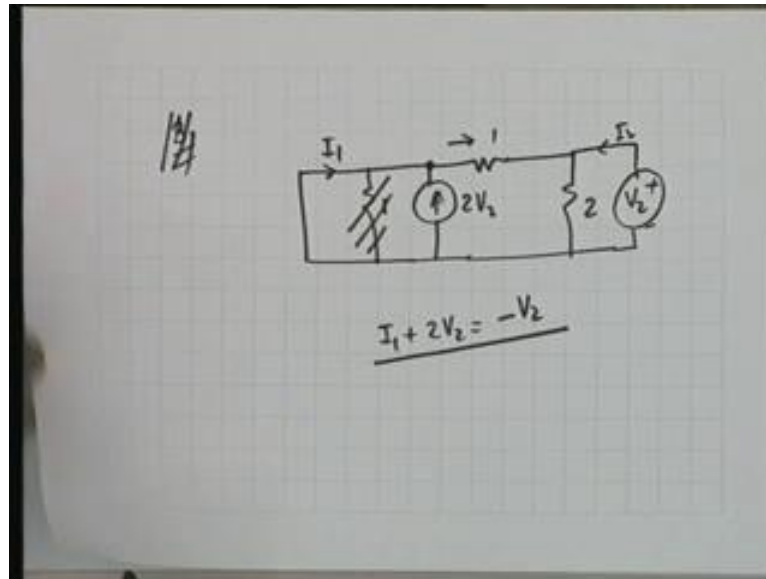
Yes

That is also removed.

That is correct.

Let me draw this network. This is interesting question.

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You see what happens, I am talking of the previous circuit  $V_1$  is made equal to 0. This is the current  $I_1$ , it will be short circuit. 1 then you have  $2V_2$  and then you have 1 ohm and 2 and  $V_2$ . What his question was?  $I_1$  naturally no current shall flow through 1, but this current generator, it cannot be killed. If there is a  $V_2$  there shall be a current  $2V_2$  and therefore, the current that enters here is,  $I_1$  plus  $2V_2$  and that must be even the current that flows here, which is minus  $V_2$  by,

No

$I_1$  plus  $2V_2$  shall be equal to

Minus  $V_2$ , because this point is the same as this point and therefore, this 1 ohm comes across  $V_2$ . And then  $I_2$  shall be equal to this current plus this current.

That's how one find so.

No.

Well we do not care where it flows. All we know is, at this juncture there is a current  $I_1$  coming. This 1 ohm becomes absolutely redundant 1 ohm can become a carrier current, but at this juncture  $2V_2$  comes and  $I_1$  comes. We do not know why not? We do not know, whether the total current will flow through this or not.

No we do not know.

Okay.

Should not it flow through the

That is generated here.

It is a current generator.

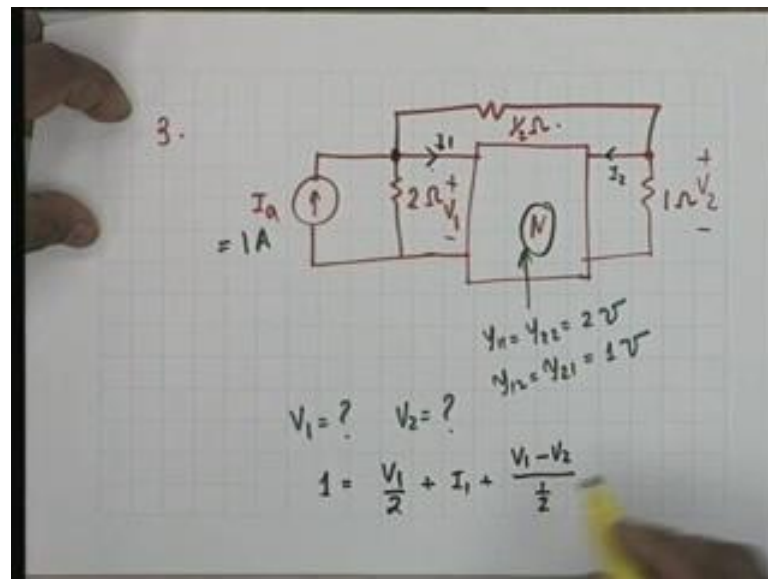
But his question is should not  $I_1$  be equal to minus  $2V_2$ ?

There is an additional current going here, these are not the only 2 currents, that is the third current and we have taken care of by writing

KVM at this particular KCM.

Yes.

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Question 3: Well, each question has to be done carefully, but there is a pot hole here which I has to guard against. Question 3 is, we have a network in which 2 ohm's comes in parallel with a current  $I_a$  and then we have a network N a 2 port N and a 1 ohm resistance, then there is a bridge half ohm. This is the composite network that is there is a 2 port which has which is augmented by 2 ohm's, one ohm here and a bridge of half ohm and this is  $V_2$ . This voltage is  $V_2$ , this voltage is  $V_1$ . What is given is that this parameter

the parameters of this network this i had given. This network is  $y_{11}$  equal to  $Y_{22}$  equal to 2.  $y_{12}$  equal to  $y_{21}$  equal to 1 and  $I_a$  is given as 1 ampere.

The question is to find  $V_1$  and  $V_2$ . This is the question, you understood the question. Now, can you solve it? Let's have suggestions.

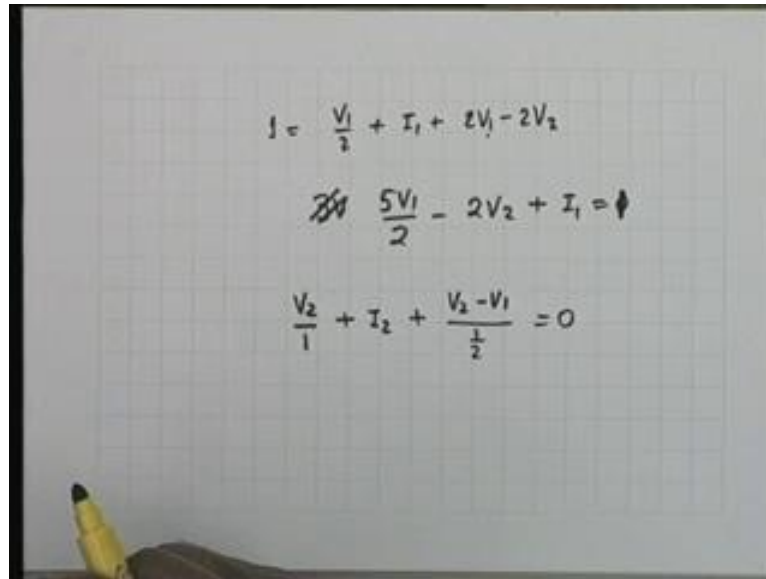
Can we replace this by a pi network?

No, we cannot because the other network, what you are trying to do is to use this as a parallel connection of 2 networks. Parallel connection half two mows 1,2 ohms 1 ohm, but it is not given that this 2 are connected together. So, you cannot make a parallel connection because this could be truly a four terminal network agreed. This is the temptation this is the pot hole. Fortunately in this particular case this also gives correct results and that is the, the point of the question. The question has been framed such that even if you consider this as a parallel connection of 2 true ports. You get the same result as you get by considering N is a truly four terminal network, but if you are happy with the answer. The examiner would be unhappy with the answer, because the method of calculation would be wrong. Now, therefore, what would be the correct method?

Now, go back to the roots. That is, what you do is the following; let's identify some currents and voltages and this is why i said you have to be very careful. I had done it both ways. The wrong way and the right way and they give the same result. Suppose, you have told that this is a short circuit, then there is absolutely no problem. You just find out the y parameters of this 5 pi network. And add to the y parameters of the network N and determine  $I_1$  and  $I_2$ , but no this cannot be done here because this is not specific; therefore, we postulate currents  $I_1$  and  $I_2$ . We postulate currents  $I_1$  and  $I_2$  and I write at this point.

The KCL will be  $I_a$  would be equal to 1 ampere would be equal to  $V_1$  by 2 plus  $I_1$ .  $V_1$  by 2 plus  $I_1$  plus  $V_1$  minus  $V_2$  divided by half. This is 1 of the equations and the other equation is at this node.

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The image shows three equations written on a grid background. The first equation is  $I = \frac{V_1}{2} + I_1 + 2V_1 - 2V_2$ . The second equation is  ~~$\frac{5V_1}{2} - 2V_2 + I_1 = 0$~~ . The third equation is  $\frac{V_2}{1} + I_2 + \frac{V_2 - V_1}{\frac{1}{2}} = 0$ . A yellow highlighter is visible at the bottom left of the grid.

Therefore, this simplification this gives further we give 1 equal to  $V_1$  by 2 plus  $I_1$  plus  $2V_1$  minus  $2V_2$ . Therefore, this is equal to  $2V_1$  and  $V_1$  by 2. So, 3.

$5V_1$  by 2 minus  $2V_2$  plus  $I_1$  equals to 0.

Equal to 1.

The other equation is, if I look at the network  $V_2$  by 21 plus  $I_2$  plus  $V_2$  minus  $V_1$  divided by half this would be equal to 0.

Yeah

Yes we have.

Infact...

I am glad you discovered it. So, what is the way out?

You see writing this equation as assume and he has correctly caught me doing this mistake. You see my  $V_2$  reference is not necessarily same as  $V_1$  reference. And when I say the only problem is here  $V_1$  minus  $V_2$ . If I do that then obviously, the references have to be the same agreed and therefore, this method going back to roots also does not work agreed. So, what is the way out? Shall i discard it at this point? Are you convinced that it should be discarded.

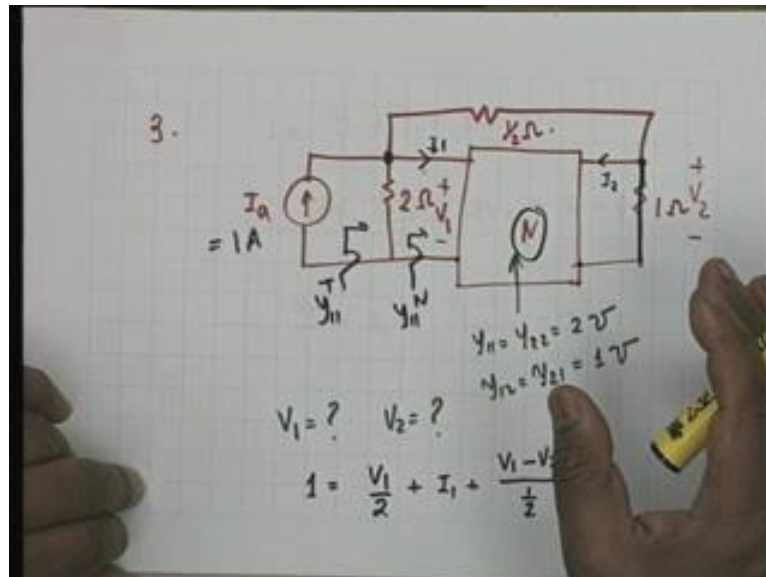


Because doing this simply means that we are assuming that, this 2 are connected together which is not correct? So, what is that we should do now.

What do we do now?

Well, that is called the convenient the true. If there is a problem. Now, we want t counter. We want to confirmed it what is the solution, can we go to find out the  $y_{11}$ ,  $y_{22}$   $y_{12}$  and  $y_{21}$  this total network.

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Can I do that.

How?

For example, if you want to find out  $y_{11}$ . I have to short circuit this. If I short circuit this then what would be,

$y_{11}$ .

Same problem.

So, in other words, you see if I short circuit this let us do it. Short circuit this then my  $y_{11}$  total would be this admittance plus  $Y_{11}$  of the network N half now plus  $y_{11}$  N plus I don't know how this half ohm.

Comes into this circuit because I do not know, what is connected between these 2. And therefore, I have the same problem and I claimed that there is now way

One more variable how do you solve it?

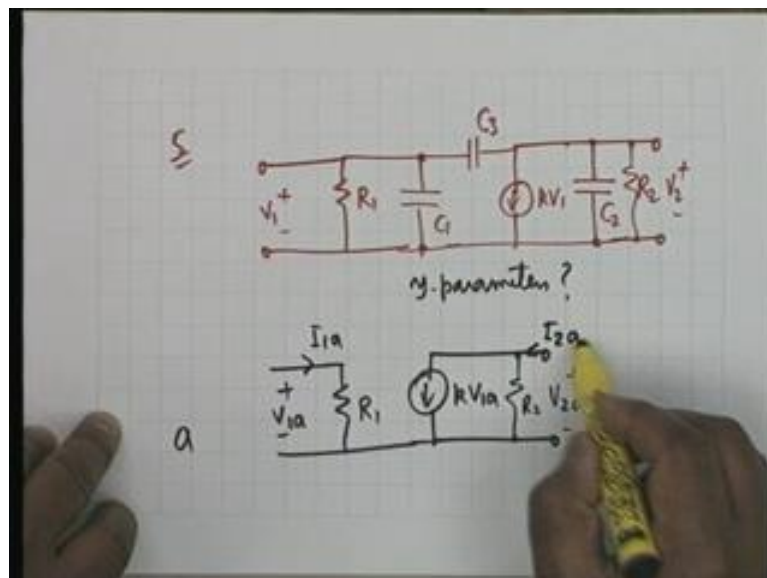
The number of equations will be only 2.

All that you can connect is  $V_1 I_1$ ,  $V_2 I_2$  that is all.

Nothing else. And therefore, this problem is not solvable unless

These 2 points are connected. So, we now we can leave it all right as per suggestion we can go to the next problem. Next problem is question number 5. It is not that I am leaving the tough ones. No I am just taking the alternate problem.

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Determine the  $y$  parameter of the network question number 5 is determine the  $y$  parameters of this network. The network is this  $R_1, V_1$  and then you have a  $C_1$ .

C3 a controlled source KV1 and a capacitor C2.

If you notice carefully.

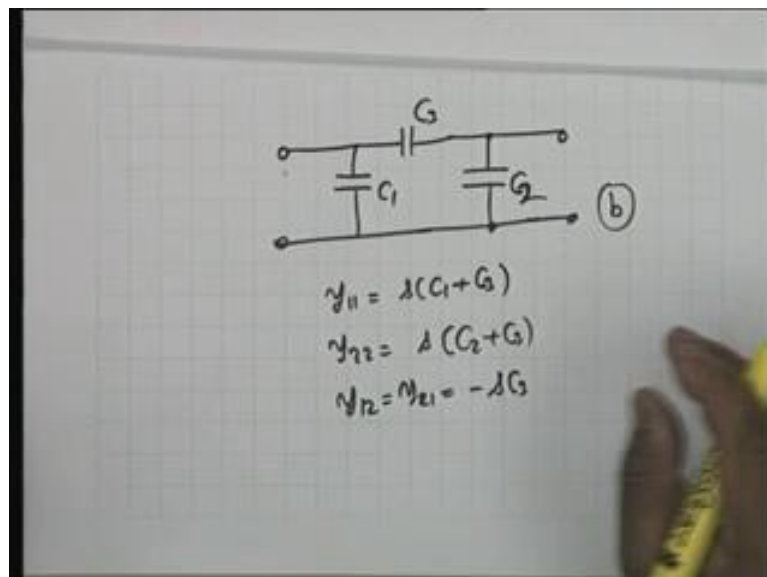
There is an R2 somewhere. Yes there is an R2 here.

If you notice carefully

There is a V2 here port number 2. If you notice carefully this is simply the equivalent circuit of a BJT or FET. Is it not? R1 is simply R phi. C1 is C phi. Are you acquainted at this terminology? C3 is C mu and C2 is C0 output capacitance. R two is R zero that is the collector dynamic resistance and this source kV1.K is simply,

GM and V1 is the voltage across R phi which of course, assumes that Rx is neglected. The base's spreading resistance is neglected. Now, to find the y parameters of this work, the easiest thing to do, easiest thing to do would be to view this as a parallel connection of 2 networks. And you see the how the parallel connection is chosen. One of the networks we choose fortunately there is a straight connection. It is a 3 terminal. So, 1 of the networks that we choose is R1. Let's call this V1 a, the a network and kV1 a that is it. No, let us include R2 also. This is V2 a. I1 a, I2 a this is my a network and the b network

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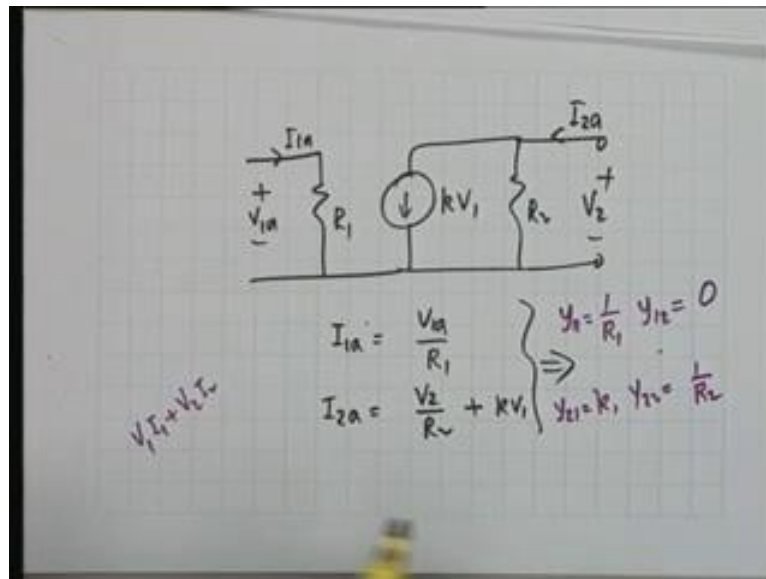


The b network is simply the capacitive network. That is; the b network is C1.

Yeah you have to calculate a parallel impedance.  $R_1$ ,  $C_1$ ,  $R_2$ ,  $C_2$ . Yes we can do that. It can also be done that way. A little bit more calculation here; there is no calculation as you all see. This is a b network, for the b network you can see that  $y_{11}$  is  $sC_1$  plus  $C_3$ . B network (( )) is  $sC_1$  plus  $C_3$ ,  $y_2$  is equal  $s$ .

This is  $C_2$ ,  $sC_2$  plus  $C_3$  and  $y_{12}$ ,  $y_{21}$  is equal to minus  $sC_3$ . This is the y parameter of this network no calculation is by inspection. What about the other network?

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The other network is,  $R_1$ ,  $V_{1a}$ ,  $I_{1a}$ ,  $kV_1$ ,  $R_2$ . This is  $V_2$  and this is  $I_{2a}$  and you can see, that all that I write is  $I_{1a}$  equal to  $V_{1a}$  divided by  $R_1$  and  $I_{2a}$  is equal to  $V_2$  divided by  $R_2$  plus  $kV_1$  and the y parameters are obvious. What is  $y_{11}$ ?

$1/R_1$ ,  $y_{12} = 0$ ,  $y_2$ .

$1/R_2$  and  $y_{22}$  equals to  $1/R_2$ . And because both of them are 3-terminal you can simply, add the y parameters. I did not have to calculate the parallel combination of  $R_1$  and  $C_1$  or  $R_2$  and  $C_2$ . I need not to write I had to write 2-node equation, but...

This is non-reciprocal.

Yes.

BJT is non-reciprocal.

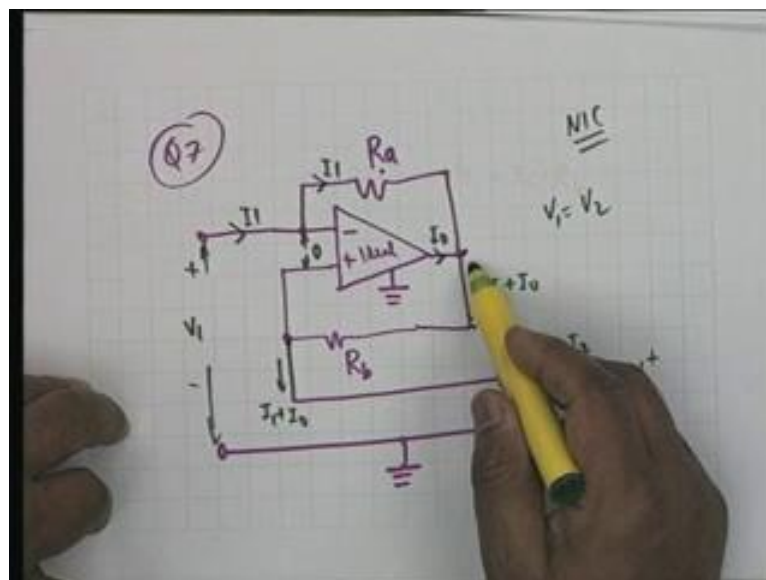
Is it active BJT?

Yes it is active.

The activity; however, is not obvious yet. Activity will be clear if you write  $V_1 I_1$  plus  $V_2 I_2$ . You can see that, it can be less than 0. It is possible to make it less than 0. So, this network is active and can a transistor be passive.

Can a transistor be passive.

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Last question of the day number 7 was there any problem in the fifth-one. Number 7 gives a an op amp circuit, a practical op amp circuit for an NIC negative impedance convertor and the circuit is like this. Let me draw this and the analysis will be extremely simple. This resistance is  $R_{sub b}$ . It is not given in this figure,  $R_{sub b}$  and this is connected of course, the op amp has a ground. This is the input port and from here you have a resistance  $R_a$  connected to the output. The second port is from here, to ground. The question is, this op-amp of course, we have to assume that this is ideal.

You have to show that this behaves as an NIC. The solution is extremely simple. What you have to do is assume a this is the port  $V_1$  and this current is  $I_1$ . We have to show that if port number two if port number 2 is terminated in. Let's say some impedance  $Z_L$  we have to show that  $V_1 I_1$  that is the input impedance is

It can be equal to minus  $Z_L$  or it can be proportional to minus  $Z_L$ . The proportionality constant can be anything. As long as the impedance is the, it is proportional to the negative of the terminating impedance it is an NIC. So, this voltage is  $V_2$  and this current is  $I_2$ .

Proportionality constant someone has it is positive. It will be an NIC quite surprised. Proportionality constant can be negative. If the proportionality constant is negative for an NIC .

Then it is not an NIC. It is a positive impedance (( )) PIC. Now, let us see how to calculate this we invoked the idealness condition of an op-amp. One of the things that is obvious is  $V_1$  is equal to  $V_2$ . Isn't that right because again idealness because the potential difference between this 2points is 0 virtual short not ground. Why did not you say ground? Because neither of this points are grounded. The ground is somewhere else.

This is not ground because this is you r port number 2. So, we do not say virtual ground be careful about this terminology. In this situation it is a virtual short because the ideal op-amp has infinite amount of gain. Once again, the ideal op-amp can just take a current and therefore, the current  $I_1$  must flow here agreed. The current  $I_1$  must flow here, when it comes here the op-amp gives out some current let us say  $I_0$  and therefore, this current would be  $I_1$  plus  $I_0$ . And this current after coming here, nothing can go the op-amp and therefore, this must be equal to  $I_1$  plus  $I_0$  .

So, there are two things that is obvious, that is;  $I_2$  should be equal to minus  $I_1$  plus  $I_0$ . This current and this current and the other thing that is obvious is that this drop. Drop across  $R_a$  should be the same as the drop across  $R_b$ , which means that  $I_1 R_a$  should be equal to  $I_1$  plus  $I_0$  into  $R_b$ .

What means?

Why this should be?

You see this 2points are virtual short and therefore, the drop across  $R_a$  should be equal to drop across  $R_b$ , but the drop across  $R_b$  is  $I_1$  plus  $I_0$  multiplied by  $R_b$  with this terminal positive.

So, there shall be...

A negative sign.

Why should  $I_1$  flow through to because the op-amp is ideal? The input impedance in either port is infinity. If the impedance is infinity it cannot take a current. It cannot in the electronics language sync a current all right.

So, the current is diverted through  $R_a$ .

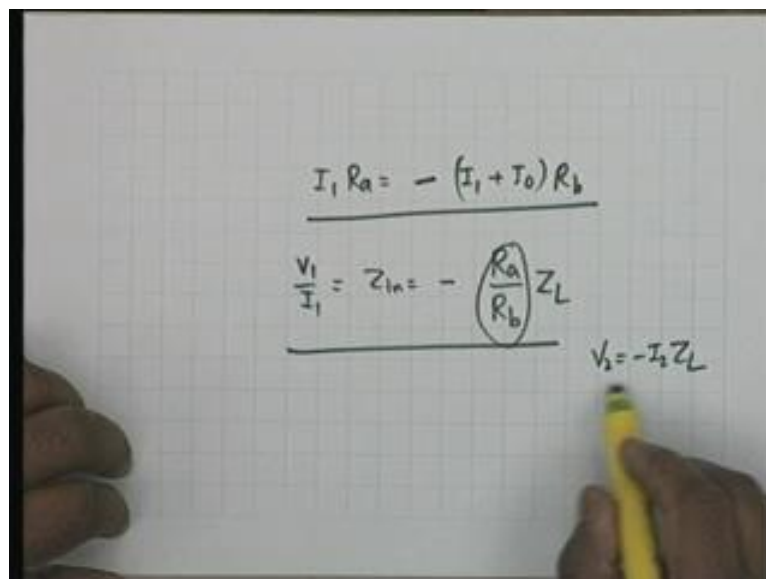
No, because there is no connection. Current has to have a connection, the potential of 2 points may be the same, but there can be no current through that because there is no physical connection.

Any other question?

No,  $I_1$  plus  $I_0$  is flowing like this and  $I_2$  flowing like this. They oppose each other and therefore, one must be the negative of the other.

Why are they approximately short? The op-amp ideal op-amp has a infinite gain. So, to produce a finite output voltage it requires 0 input voltage. 0 input voltage means voltage between this point and this point and therefore, they are short. They must be at the same what, Ideal minded infinite gain.

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The image shows a whiteboard with handwritten equations. The first equation is  $I_1 R_a = - (I_1 + I_0) R_b$ . The second equation is  $\frac{V_1}{I_1} = Z_{in} = - \left( \frac{R_a}{R_b} \right) Z_L$ . To the right of the second equation, there is another equation:  $V_2 = -I_2 Z_L$ . A hand is visible holding a yellow marker, pointing towards the equations.

Now, any other,  $I_1 R_a$  equal to minus  $I_1$  plus  $I_0 R_v$  and therefore,  $V_1$  by  $I_1$  which is equal to  $Z$  input would be equal to minus  $R_a$  divided by  $R_b$  into  $Z_L$ . The 2 simple manipulation of these 2 equations.  $I_1$  is equal to minus  $I_1 I_0$  by  $R_v$  and so on.

What comes have to be plus?

No, where you must not forget that  $V_2$  is equal to minus  $I_2 Z_L$ . So,  $V_1$  by  $I_1$  is  $V_2 I_1$  and  $V_2$  is minus  $I_2 z L$  with minus  $I_2 z L$  by  $I_1$  and we have already found out what  $I_1$  is.  $I_0$  does not come into the picture. It gets cancelled out and you noticed that if this 2 resistances are made equal. There was a negative impedance convertor with a conversion constant of 1. What is the dimension of the conversation constant?

Dimensionless you also noticed that by choosing  $R_a$  and  $R_b$  appropriately 1 can multiply impedance. Not only makes it negative, but then one can increase it or one can decrease. It the way 1 likes and this circuit has lot of uses in active filters and in integrated circuits.

Yeah, this is what it is.

How?

It is very simple, 2 equations  $V_2$  this is  $V_1 V_2$  and  $V_1$  are equal and therefore,  $V_1$  by  $I_1$  is the same as.

Any other question, that is all for the day.