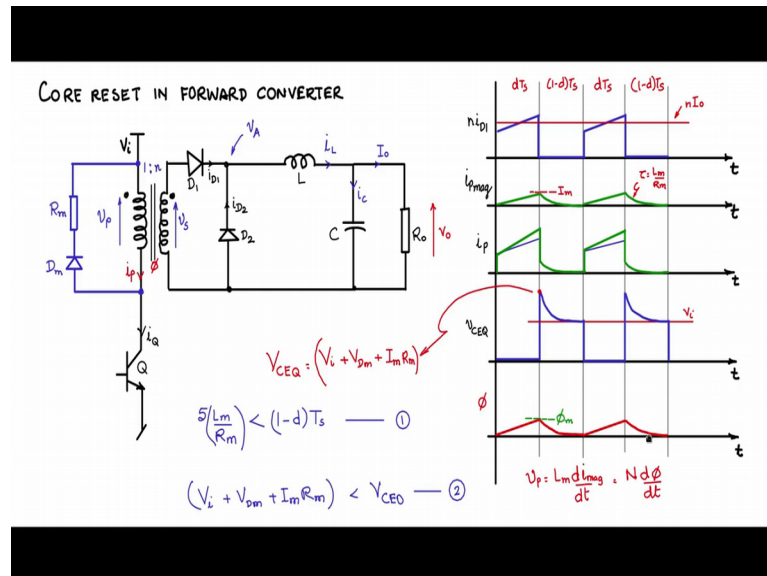


Fundamentals of Power Electronics
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Lecture - 55
Core Reset in Forward Converter

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Till now we discussed on the wave forms on the secondary side of the forward converter which was mostly like the buck converter. Now let us look at the primary side, this switcher and the core reset issue. Let us draw the wave from the related to that and then try to understand that. So, I am going to divide the time again into four parts where you have $d T_s$ $d T_s$ the time during which the switch Q is on $1 - d T_s$ the time during which switch Q is off $d T_s$ again Q is on $1 - d T_s$ Q is off.

So, let us first consider the voltage wave the current waveform through this diode and that is what will actually get reflected on to the primary. So, let us mark here the primary current i_p . Remember that primary current and i_Q are same only during the time when Q is on when you how i_p and i_Q same. The moment Q is off, i_Q becomes 0, but i_p does not become 0 because there is a freewheeling current that will happen in this direction. So, therefore, i_p and i_Q are naught strictly same during $1 - d$ time i_p has some finite current i_Q is 0 during that time.

So, what is i_p ? i_p is the reflected secondary current and at the secondary what is flowing when the transfer Q is on is i_{D1} . So, n times i_{D1} is what is supposed to be reflected on to the primary the load reflected part. So, let us take $n i_{D1}$ which is i_{D1} into the load reflected path flowing through the primary. This is one component. So, we know that the load reflected part during that time inductor current is rising and this is how the current wave shape look like and this we have seen before and if you take the level at this point, you know that that is the n times i_{D1} again load reflected part.

Now, next we should understand that this is a transformer, then there is some current needed to magnetize the transformer and that is the magnetization current. So, a magnetization current also will flow through this primary and that should be added to that to get the total i_p . So, what will be the magnetizing current? I will say $i_{p\ mag}$ when Q is on V_i is applied across the primary winding, it is a constant voltage. So, the magnetizing current from the Faradays law V_i is equal to $L_m di/dt$. The magnetizing current will have a slope of V_i divided by L_m . So, it will start from 0 like that it will have a slope of V_i by L_m , L_m being the magnetizing inductance during $1 - d$ I will explain later.

For now we will leave this part and again it will start from 0. During this time we have to reset the core. We are to reset the core and bring it to 0, so that it will again start back again. In the meantime let us also draw. Before we draw this portion of the magnetizing current let us draw what will be i_p . i_p is the primary current. The primary current will be having two components. One is the load reflected part which is $n i_{D1}$ and other one is the magnetizing part which is $i_{p\ mag}$. These two will add up together to form i_p .

So, this part is magnetizing part. I am actually showing this bit zoomed. It will actually be much smaller as compared to the value of $n i_{D1}$. So, let us say this is the $n i_{D1}$ part which I am writing down here. This part you can add it up here, starts from this value and then gradually it will go up like that same as this $d T_s$ period also.

So, this will be the i_p , this green portion will be the total added up i_p value during the time when Q is on. When Q is off, you will see that i_Q becomes 0, but i_p will start freewheeling in this path. So, it will freewheel in this path because inductor current cannot become 0. There is this resistance R_m and there is a mutual conductance L_m . L

$\tau = L_m / R_m$ will be the time constant. So, at that time constant exponentially i_p will decay. Likewise here also you will see that it will decay in this fashion.

So, now what will be the i_p ? we will in the i_p waveform the same thing will repeat. You will see a drop because the moment Q is off, the reflected component vanishes i_{D1} component vanishes that becomes 0 and only the i_p mag component will come into the picture and that is what we will find in this.

So, the total i_p current we will look like this drawn thick in a thick green line. This will be the i_p current that will be flowing through the primary winding. Let us now see also what is the voltage across V across Q V_{CEQ} . So, let us draw on more access here and see V_{CEQ} during the time when Q is on, you know that this is 0 and during the at the moment when Q is turned off, you will see that there will be a sharp rise. What is that value we will see that and then afterwards post that rise, you will see that this is V_{cc} level the V_{ca} will start decaying down to V_{cc} level. Why does this happen?

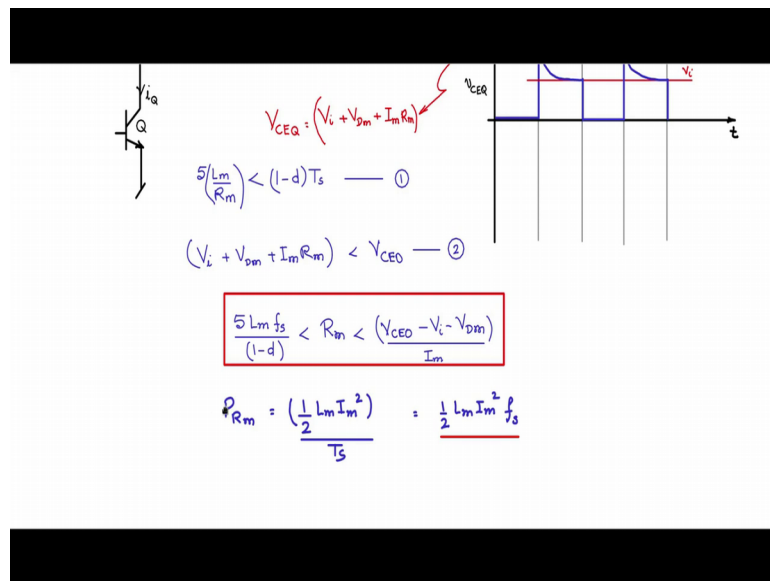
You see that moment this is switched off, this is prevailing in this fashion. So, there is a drop across the diode forward drop across the diode, there is a drop across this R_m i_p into R_m . Current is flowing in this direction i_p into R_m and there is V_i . So, V_i drop across R_m and the diode drop will all be reflecting here when it when this shoots up and then i_p current is decaying as you see here it is decaying exponentially. So, the drop across R_m should also decay exponentially, therefore you see a decay exponentially and once the decay is complete, it will settle at V_i because when once this is off, it is only V_i that will come at this point.

So, this is how the voltage across the V_{JT} will look like and then, it will repeat again the next cycle and decay in this fashion. So, this point I would like to say is the V_{CEQ} which is equal to V_i plus this diode drop V_{Dm} forward diode drop around 0.7 volts plus the current flowing this direction plus minus here you have I_m into R_m if I say this value I_m peek at that point I_m that I_m into R_m will be appearing across. So, this will be the voltage that will come across at this across this transistor Q during the time at the time at the moment when it is being switched off and then further on during the time on minus dT_s , it will start decaying down like this because I_m is decaying down.

Now this decaying time constant is equal to magnetizing inductors L_m / R_m . Now what should be the value of R_m ? So, let us say 5 times time constant 5 times L_m / R_m

m, e is the time constant which says that this would have reached steady state, it would have reached 99.9 percent of the final value. So, by that time this would have core would have reset if I m comes down magnetizing, current comes down to 0, then you can say that the core is reset. So, 5 in 5 times time constant the core would have reset, so that it can take up the next cycle. So, this should be less than this time 1 minus d T s. So, before the off time is over, the core would have been reset. So, this is one relation.

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Another relation is the voltage across Q V_{CEQ} . So, the voltage across that is as we saw here V_i plus V_{Dm} plus $I_m R_m$. This is the peak voltage that the device will see and that it will see at the point of switch off. Now that should be less than V_{CEO} rating of the device is the second relationship. Now these two inequalities using these two inequalities from this you can see that R_m should be greater than $5 L_m$ by $1 - d$ T s. T s I will replace by switching frequency.

So, I will say $5 L_m F s$ by $1 - d$ should be less than R_m . R_m I have taken to the other side and R_m should also be less than. Now if I take this relationship V_{CEO} minus V_i minus V_{Dm} by I_m should be greater than R_m or R_m should be less than that; so, V_{CEO} minus V_i minus V_{Dm} by I_m .

So, this is the value of R_m which will be the upper limit, this is the value of R_m which will be lower limit. R_m should be a value which is in between in this range. So, this is the design equation for R_m selection. Now what is the power rating of R_m ? To see that

when this is switched off, there is a current I_m starts freewheeling through R_m . The energy in the magnetizing inductance at this point is half $L_m I_m^2$. Now that much amount of energy has to dissipate, then only the core will reset. So, that has to dissipate only in R_m because R_m only dissipating element. So, let us say power rating of the R_m can be found out from half $L_m I_m^2$ amount of energy has to be dissipated every cycle, every switching cycle.

So, you can say this is equal half $L_m I_m^2$ into F_s and this is the power that will get dissipated in R_m whatever may be the value of R_m which you are chosen in this range because this energy, the energy here is independent of the value of R_m . So, this is the power rating choice that you have to do for R_m . So, with this you can now say that whole components of this forward converter circuit with core resetting feature can be now implemented in this fashion and also simulated there is one important variable that we will not be able to measure and see on the oscilloscope.

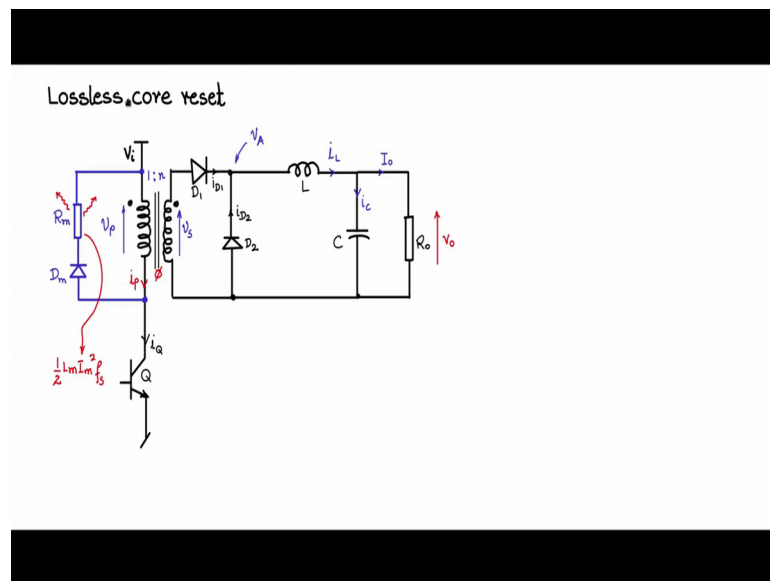
And that is the flux inside this core. We have been talking of core resetting the core in the flux which is to be reset, but we will not be able to see the flux waveform, but indirectly we can know how it looks like. So, let me try to plot an estimate of the flux. See the flux in the core with the symbol ϕ and let us try to plot that one.

Now, we know from Faraday's equation that V_p the voltage across the primary is equal to $L_m \frac{dI_m}{dt}$, the magnetizing inductance dL_m by dt or demagnetizing current by dt it is also equal to $n \frac{d\phi}{dt}$. So, from this basic fundamental equation we see that $L_m \frac{dI_m}{dt}$ is equal to $n \frac{d\phi}{dt}$. L_m is a constant, n is a constant, V_p is fixed for the period of the time and Q is on, therefore $\frac{dI_{mag}}{dt}$ demagnetizing current by dt and $\frac{d\phi}{dt}$ will have the same wave shape.

And therefore, same slopes and therefore, I_{mag} magnetizing current and flux in the core and also have the same wave shape. Therefore, the flux within the core will take the same wave shape as $i_{p\ mag}$ that we have drawn here. So, let us draw that the same linear rising and fall because when V_p is fixed content when Q is on V_p by n V_p by n will be the rate at which the flux will rise and it will fall exponentially because the current I_m is falling exponentially and again the next cycle it will go in a linear fashion to a max value and then, fall exponentially as maximum value is ϕ_m .

So, this is how the flux wave from within the core may look like if you were able to probe it and see it on the oscilloscope, but we are able to get this wave shape indirectly through the Faradays law. So, you see that the flux rises and then the flux is reset. This is the core resetting action. The flux in the core it is the reset to 0 and it is ready for action. The next cycle when it again rises and then during $1 - d$ T s period, the flux is reset again; so, this is core reset in forward converter. We have been discussing about this forward converter with this type of core reset.

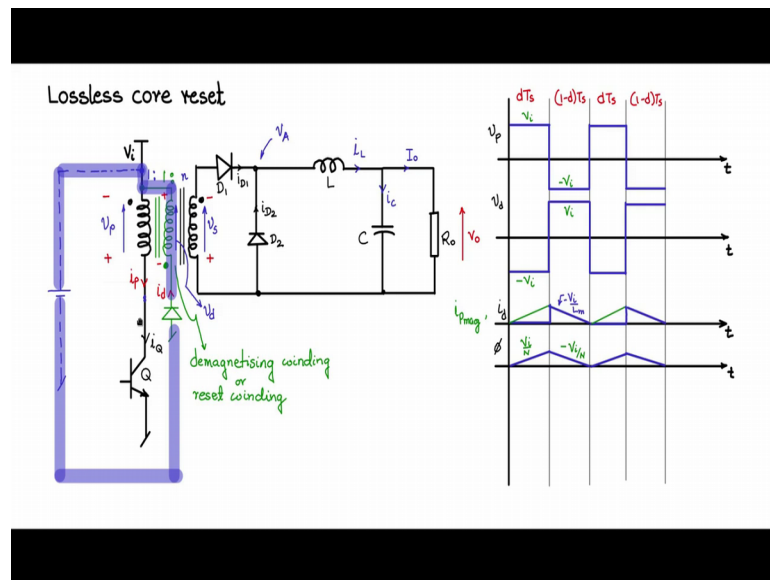
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Now, this is a lossless core reset. The reason being that there is some dissipation occurring in R_m and we calculated that the dissipation that happens in R_m is $\frac{1}{2} L_m I_m^2 F_s$ meaning $\frac{1}{2} L_m I_m^2$ is the energy stored in the magnetization and that $\frac{1}{2} L_m I_m^2$ must be removed every cycle and therefore, $\frac{1}{2} L_m I_m^2 F_s$ is the amount of power dissipation in this R_m . So, this will bring down the overall efficiency of the forward converter.

So, let us discuss another method of core reset where it is lossless. So, lossless core reset, so what we do is first we remove this and then let me make some space.

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And I will introduce one more winding, one more winding which is tightly wound to the primary 1 is to 1. In fact, bicular wound you take the primary and this winding which I will call as demagnetizing winding because it is used for resetting or reset winding you take them together and wind them, so that there is minimum leakage as possible.

And then you attach a diode like this and other end you connected to V_i or V_{cc} and you place the dot polarity in the spanner opposite dot polarity ok, then let me put this as 1 is to 1 and let me push this back here. So, now this is the modified forward converter where you have removed the lossless core reset circuit and you have interposed another winding, no resistance only the winding, but still there is a prevailing diode.

Now, how does this become lossless? Let us see how the core reset action is performed and how it is lossless. So, this winding is normally turn the literature as the demagnetizing winding or simply the reset winding because of the job that it does. Now just like your measuring the voltage V_p here in this fashion V_s here in this fashion will measure the voltage across the demagnetizing winding and let me call that 1 as V_d. So, now it will be appropriate to look at some wave forms to understand the operation of the circuit. Again I am dividing the time into four parts d T s 1 minus d T s d T s and 1 minus d T s two cycles two switching cycles d T s, the time period when Q is on. So, now first let us see the two waveforms V_p and V_d simultaneously V_p and V_d.

Now, during the time when during the timed T_s during the timed T_s Q is on. So, when Q is on the entire V_i is coming across the primary coil with dot end positive, non-dot end negative. So, if that is the case in the other winding you will see the dot end positive non-dot end negative for the reading this fashion and dot end positive non-dot end negative in the secondary and the secondary operation we know it is the regular buck convertor operation. We will not be discussing it now. Let us focus only on these two coils the primary and the demagnetizing coil.

So, during that time V_i comes across V_i comes across V_p for that V_i and as I am measuring in this fashion the dot end being positive, it will appear minus for V_d minus V_i . So, because it is minus V_i , the diode is reversed biased because there is a minus V_i and there is also V_i of the supply. So, minus $2 V_i$ comes across the diode. Diode is reversed biased no current flows through it, but it is also interesting for us to look at what is this diode current i_d .

So, let us also look how one access here for the diode current i_d I would also like to look at the magnetizing part of the i_p . We will come to that later. So, the diode current i_d because the diode is reversed biased, it will be zero current at that moment of time. Now if I take the magnetizing part of i_p remember recall that i_p is consisting of two parts. One is the load reflected part that is the n times i_{D1} and the magnetizing part $i_{p\ mag}$ together that form i_p . So, consider only the magnetizing part $i_{p\ mag}$ which is a kind of a ramp in this fashion which is having a slope of V_i by L_m the slope.

Now during $1 - d$ T_s Q is switched off the moment, Q is switched off. The current here is immediately brought to 0. So, there is a large di by dt negative di by dt and therefore, they will be reversal of potential across the coils. So, let us mark that reversal of potential non-dot end becomes positive dot end is negative, non-naught end is positive; dot end is negative likewise on the secondary side the secondary side now here inductor is freewheeling. This is plus and minus in this fashion this there is a reverse biased. So, the entire secondary is like open circuited. We will not talk about that. We will talk about these two.

Now here if you look at the non-dot end is positive and this will keep this will become as positive as it can pump against this V_i and make this diode forward biased, so that the diode is conducting for freewheeling action. So, if I draw the voltage source V_i portion

also, here you will see that there will be conduction of current in this path like that and comes through this against the source V_i and reaches here from this plus it goes in against the source. So, if the diode is ideal, the drop just a bit more than V_i is what this will be needed to pump current into the pump current for resetting the core. So, there is freewheeling action that will happen.

So, what are the voltage across V_d and V_p ? So, across V_d because it is this a diode clamping plus and minus this is coming directly across V_i , it will be V_i in this measured in this direction nor non-dot end positive. So, in the case of V_p you will see that it is negative minus V_i because this is 1 is to 1 times ratio, this is 1 is to 1 and what happens to i_d see that i_p had charged the flux up to that point and now it will take over from there and discharge during that time with a slope of minus V_i by L_m . This charged with V_i by L_m , this will discharge with minus V_i by L_m .

So, therefore if this is taken x seconds, this should also take x seconds. That is why we say that if this has taken T_s by 2, this would be T_s by 2. You should give the same amount of time for it to reset. Therefore, you cannot have a duty cycle greater than 50 percent in these type of converters, but is not a disadvantage because you have n has a degree of freedom whichever you can adjust the voltage.

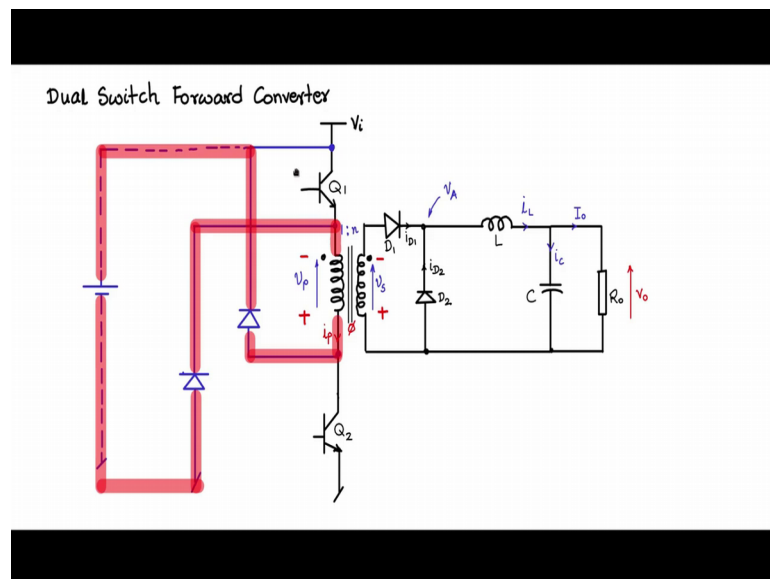
So, next again next cycle next cycle behaves in the same way. You have the repeat and during this portion there is no i_d . There will be i_p mag and then i_d comes down in this fashion. So, i_p mag magnetizing close only during $d T_s$ portions; i_d flows during $1 - d T_s$ portion as the freewheeling path. If you look at the flux within the core, the flux within the core will have the same profile as i_d and i_p mag. It will take the profile of i_p mag during the $d T_s$ period because that is a magnetizing current that is flowing and during the $1 - d T_s$ period. The magnetizing part is flowing in the freewheeling portion of the demagnetizing winding and that will be this part.

So, this is V_i by n , this is also V_i by n negative because it is falling. So, this repeats. So, this will be the nature of the Flux. The flux builds up and the flux resets during $1 - d$ flux builds up, flux resets during $1 - d$ period because the slopes are same. You need to at least that much same amount of time for it to reset therefore, limiting the duty cycle for this type of converter max to 50 percent.

So, this is the way the core reset functions in this. Note two points this windings here the primary winding and the demagnetizing winding should be 1 is to 1 and they are to be tightly woven together, wind them together. That is called bifilar winding, so that the leakage inductance is minimized. The reason is that if there is a leakage here that portion is not covered for mutual inductance for demagnetizing, so that can cause a large spike across the device.

When it is switching off that can also be avoided by having this number circuit which I will discuss later when I am discussing devices. But, I could also provide you with a quick solution provide even the leakage inductance shortly when I am talking of using dual switch or two switches for the forward converter and remove this magnetizing portion, then we can still get lossless core reset and avoid the leakage inductance issues also.

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We can do lossless core reset with dual switch forward converter also. So, let me briefly describe the dual switch forward converter and explain to you how this lossless demagnetizing can be done. So, this is the circuit of a forward converter without any demagnetizing; without any core reset and let us introduce core reset now. So, let I remove the V_{cc} for now and then here I will introduce another switch another BJT this is V_i . Let me call this Q_1 and this is Q_2 ok.

Now, these two switches are switched on simultaneously ok. Now I will add two freewheeling diodes; one diode freewheeling from here to V_i to V_{cc} and another diode from here down to ground in this fashion. I will explain to you how the current flows. Let us also visualize that there is a supply here. So, let us place that supply because we have placed the label. So, let us visualize to understand the current flow.

Now, let us consider the situation dT_s when Q_1 and Q_2 are both on together. So, they are both on together. You will see that this will be this is on. So, V_i this is plus and this is minus, on the secondary side this is plus and minus, secondary side operation we know is buck converter operation, but let us see how the current flows here on the primary. So, it starts from here, goes in this direction, comes like that through the ground and back again to the power supply. So, this is the path of the current flow during the period dT_s and where the voltage across the primary is plus minus dot and positive V_s dot end positive buck operation follows.

Now, what happens when you switch off Q_1 and Q_2 ? So, when you switch off Q_1 and Q_2 there is a cut in the primary current meaning the primary current is cut off to 0, but we cannot allow cut off to 0. There will be a reversal of polarity. So, first these polarities will reverse because of the minus di/dt and there has to be a path for i_p current flow in the same direction in the coil.

So, let us start from here i_p flows here now as Q_2 is off current flow in this direction through this diode and then into the supply. So, I am indicating the supply here goes into the supply and this comes in this direction through this diode up there again and then, down and maintaining the same direction of current flow in the primary only. Now the dot end is negative; non dot end is positive. So, it is demagnetizing.

So, this is this coil is acting like a demagnetizing winding during the $1 - d$ period because the secondary side is open, this diode is off. So, this because the same winding is operating as a demagnetizing winding, there is no leakage inductance problem because everything is included here and primary reflected and you will not have leakage inductance spikes due to due to the leakage inductance spikes on the two switches when they are off. So, this is the path the this is the path where the demagnetizing current will flow and if you will consider diodes as ideal, this whole V_i is coming across the primary plus minus in this direction.

So, you have to allow the same time because again it is 1 is to 1. The magnetizing is happening at V_i by n the demagnetizing is also happening at the rate V_i by n . Therefore, you have to allow the same amount of time for the magnetizing as you are allowed for magnetizing and therefore, this is also limited to 50 max duty cycle. This is not actually a serious limitation because of the n here, you cannot just the value of the n to get whatever required output voltage. So, this is the dual switch forward converter. You have removed the windings, but you have included one extra diode and one extra switch and an extra controls switch means extra get right for that.