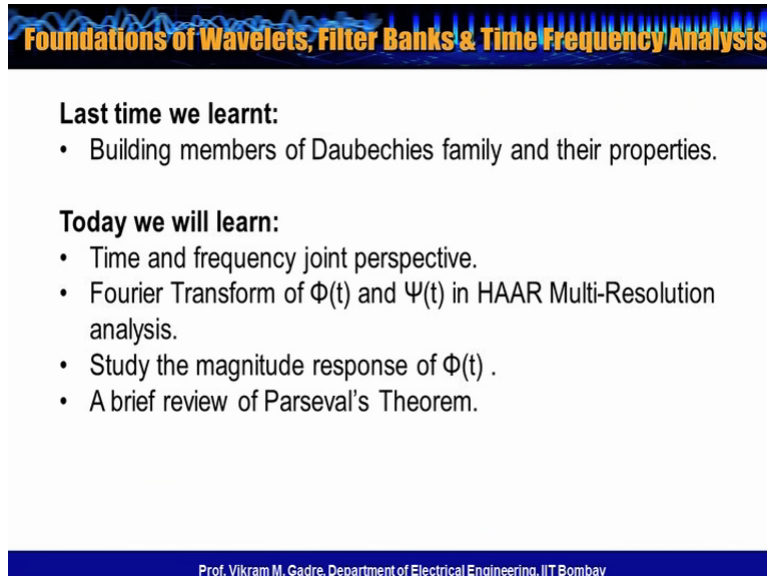


Foundations of Wavelets, Filter Banks and time Frequency Analysis.
Professor Vikram m. Gadre.
Department Of Electrical Engineering.
Indian Institute Of technology Bombay.
Week-6.
Lecture -15.1.
Fourier transform Analysis of Haar Scaling and Wavelet Functions.

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Foundations of Wavelets, Filter Banks & Time Frequency Analysis

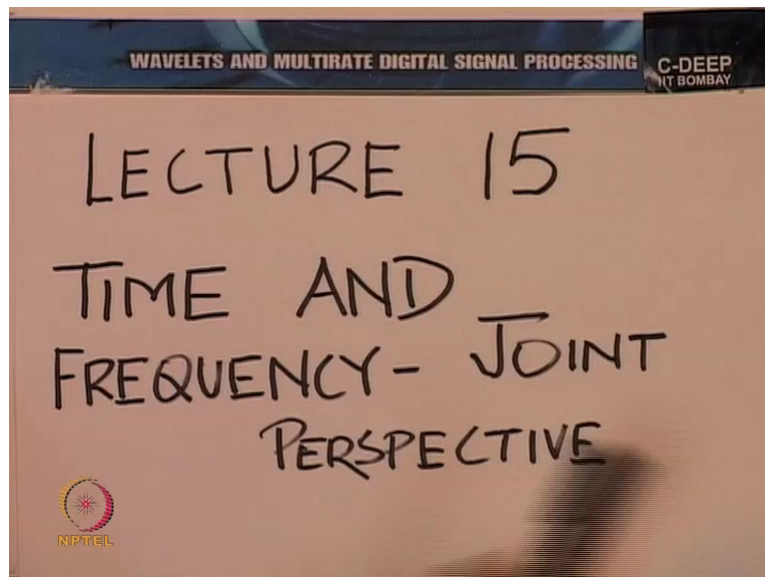
Last time we learnt:

- Building members of Daubechies family and their properties.

Today we will learn:

- Time and frequency joint perspective.
- Fourier Transform of $\Phi(t)$ and $\Psi(t)$ in HAAR Multi-Resolution analysis.
- Study the magnitude response of $\Phi(t)$.
- A brief review of Parseval's Theorem.

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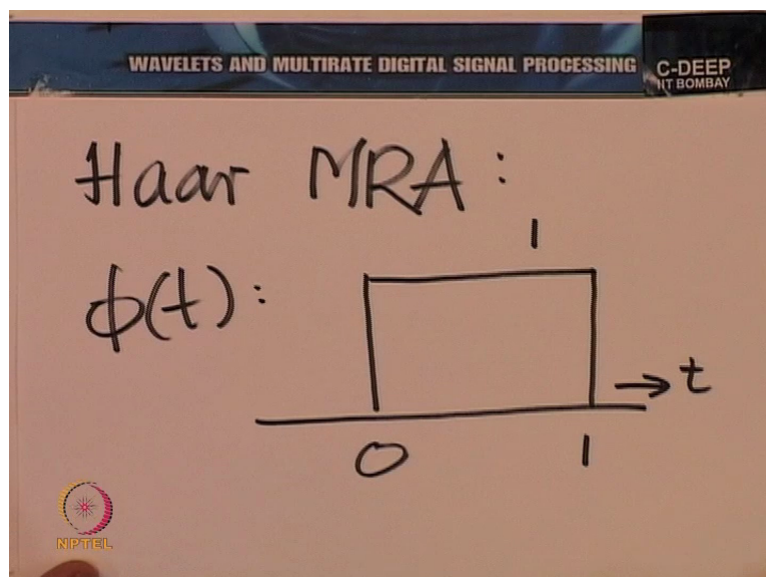
A warm welcome to this lecture on the subject of wavelets and multirate digital signal processing. We had raised a few questions in the previous lecture which we set out to answer in this lecture. the questions pertained to the fact that we were building more than 1 multiresolution analysis. And specifically the question was, what are we looking for when we

go from one multiresolution analysis to the other. Why cannot we be content with the Haar multiresolution analysis? We have been singing the praises of the Haar multiresolution analysis so frequently in some of the previous lecture.

We have been saying that it tells us most of the things that a multiresolution analysis does, why then should we look for others, what is inadequate in the Haar? Well, if you remember, when we began this course, we set out to do something which a basic course on signal theory or signals and Systems does not. namely, look at 2 domains simultaneously. We have been trying to do this for a while. So when we took the Haar, we did at some point talk about the filtering aspects of it. We also talked about the ideal to which we strive in the filter bank and if you recall that ideal had to do with the frequency domain, not the time domain.

So somewhere we have not been giving as much importance to the other domains so to speak as we should. We have been working in the natural domain of time but we have not quite been doing justice to the frequency domain. When we take a Haar multiresolution analysis, what does it do in the frequency domain, we have not quite understood this as yet. And the 1st thing that we would like to do is to answer this question. What does the Haar multiresolution analysis do in the frequency domain and then how do these filter bank that comprise the Haar buildup to their frequency domain behaviour?

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

$$\hat{\Phi}(\Omega) = \int_{-\infty}^{+\infty} \phi(t) e^{-j\Omega t} dt$$

analog radian (angular) frequency

$$= \int_0^1 e^{-j\Omega t} dt$$

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

$$= \frac{e^{-j\Omega t}}{-j\Omega} \Big|_0^1$$

$$= \frac{1 - e^{-j\Omega}}{j\Omega}$$

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So let us set out 1st to answer a more basic question. What is the Fourier transform of the scaling function and the wavelet function in the Haar multiresolution analysis? So we shall call today's lecture a joint perspective, time and frequency. And we shall set out 1st to look at the Fourier transform of $\Phi(t)$ and $\psi(t)$ in Haar, the Haar mRA. Let us begin with the $\Phi(t)$ and let us look at its Fourier transform. Remember how $\Phi(t)$ looked, this was $\Phi(t)$, let us obtain its Fourier transform.

We shall use this notation for the Fourier transform, a cap on top and capital Omega to denote the angular frequency, analog angular frequency, radian or angular frequency. And this is easy to calculate, well I am saying 0 to 1, actually I should write -, I can write - infinity to + infinity but the nonzero part is only between 0 and 1, so it is also all right to write between 0 and 1, only for this specific function. And this is of course equal to integral 0 to 1 e raised to

the power - $J \Omega t$ dt which evaluates to e raised to the power - $J \Omega t$ by - $J \Omega$ from 0 to 1 and that is $1 - e$ raised to the power - $J \Omega$ by $J \Omega$.

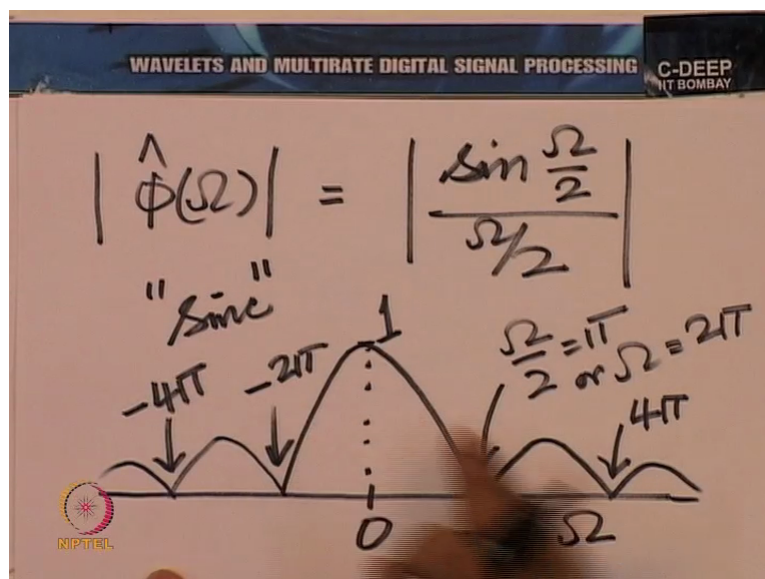
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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

$$= \frac{e^{-j\Omega/2} (e^{j\Omega/2} - e^{-j\Omega/2})}{2 \cdot j\Omega/2}$$

$$= e^{-j\Omega/2} \cdot \left(\frac{\sin \frac{\Omega}{2}}{\Omega/2} \right)$$

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And we can simplify this, we can take e raised to the power - $J \Omega$ by common in the numerator. And that is easy to interpret, in fact if you wish you can make, even make this by $J \Omega$, all right, 2 and then this becomes e raised to the power - $J \Omega$ by 2, if I take the 2J and this together, I get $\sin \Omega$ by 2 there and Ω by 2 here. So this is the Fourier transform. Now let us look at the magnitude of this Fourier transform. So in fact I could straightaway sketch this, it is essentially the magnitude of $\sin \Omega$ by 2 by Ω by 2.

The sketch would look like this, this is a very familiar function to most Electrical engineers, we call it the so-called Sinc function, the Sinc pattern. People have different names for it,

they call it the sampling function, the Sinc function and whatever other names. Anyway, this is a point where Ω by 2 is equal to π for Ω is 2π and this of course is the point where Ω is 4π and this is where it is -2π and this is where it is -4π and so on so forth. At this point the magnitude takes the value 1, in fact recall that the magnitude of the Fourier transform, in fact the value of the Fourier transform at ω equal to 0 is indicative of the area under $\Phi(t)$, so the integral of $\Phi(t)$ over all t .

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

In general
consider $\phi(2^m t - n)$
 $m, n \in \mathbb{Z}$

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

$\phi(t) \xrightarrow{\text{Fourier transform}} \hat{\phi}(\Omega)$

$\phi(\alpha t) \xrightarrow{\quad} \frac{1}{|\alpha|} \hat{\phi}\left(\frac{\Omega}{\alpha}\right)$

$\forall \alpha \in \mathbb{R} - \{0\}$

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Anyways, so much so for the magnitude and now we know what to do when we compress and expand. Let us look in general had the Fourier transform of $\Phi(2t - n)$, so we should interpret that. You know, we are talking about dyadic dilates and translates, so let us consider ϕ , in fact symbol 2 raised to the power m $t - n$, for m and n belonging to the set of integers.

Now, you will recall that there is a very simple result in the Fourier transform which says that if $\Phi(t)$ has the Fourier transform $\hat{\Phi}(\Omega)$ as we do in general, then $\Phi(\alpha t)$ has the Fourier transform $\frac{1}{|\alpha|} \hat{\Phi}(\frac{\Omega}{\alpha})$ for all α belonging to the real numbers other than 0.

This notation says all real numbers except 0, of course α cannot be 0. So for example even if α is negative, we can use this, in particular for example if α is -1, we have a reflection of the Fourier transform as well. So using this we will not take care of the 2 so-called distortions or modification that we have made in $\hat{\Phi}$, the translation and dilation. In fact the translation does not affect the magnitude, the translation only affects the phase or the angle of the Fourier transform. So I can even forget about the translation, I need only look at the 2 raised to the power m term there.

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The slide contains handwritten text and equations. At the top, it says "WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING" and "C-DEEP IIT BOMBAY". The main text reads "Magnitude only:" followed by two equations. The first equation is $\Phi(t) \xrightarrow{\text{Fourier Trans}} |\hat{\Phi}(\Omega)|$. The second equation is $\Phi(2t - n) \xrightarrow{\text{Fourier Trans}} \left| \frac{1}{2} \hat{\Phi}\left(\frac{\Omega}{2}\right) \right|$. There is a small logo for NPTEL in the bottom left corner.

WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
TIT BOMBAY

$$m=1:$$

$$\phi(2t-n) \longrightarrow \left| \frac{1}{2} \hat{\phi}\left(\frac{\Omega}{2}\right) \right|$$

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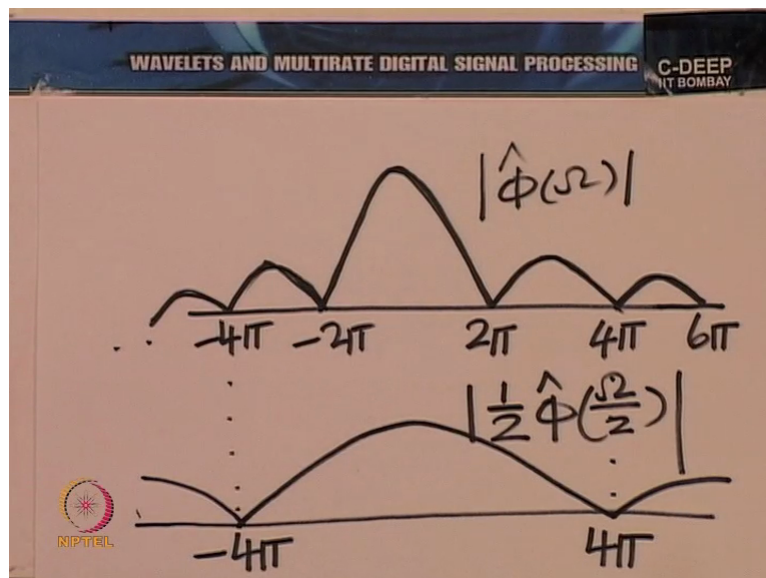
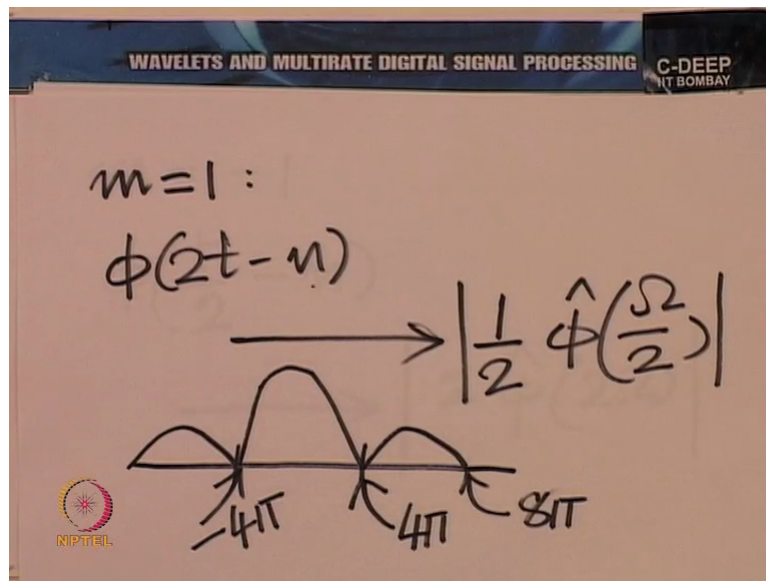
$$m=-1$$

$$\phi\left(\frac{t}{2}-n\right) \longrightarrow \left| 2 \hat{\phi}(2\Omega) \right|$$

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So here if I restrict myself to magnitude only, then $\phi(t)$ has the Fourier transform $\hat{\phi}(\Omega)$ with the magnitude of mod of this, $\phi(2t-n)$ would then have a Fourier transform mod $\hat{\phi}(\Omega)$ divided by 2 raised to the power of m , of course with the constant, the same 2 raised to the power of m but in the denominator. So of course let us take an example, suppose m is equal to 1 and -1 to fix our ideas. And m equal to -1. Now notice that the n is entirely absent, here or here, the n is irrelevant as far as the magnitude goes, the n only contributes to the phase, let us sketch both of these, how would be look.

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So let me take the m equal to 1 case, let me sketch this. So Ω by 2, so if I focus my attention on the main lobe and the principal side lobes that I have here. This was originally 2 pie, now it has become 4 pie, this was originally 4 pie, now it has become 8 pie and so too on this case, this is - 4 pie and so on. So you have stretched by factor of 2, let me put them together, the original Fourier transform and this Fourier transform with m equal to 1 and so on. I am just sketching $|\hat{\phi}(\Omega)|$ and here I sketch $|\frac{1}{2}\hat{\phi}(\frac{\Omega}{2})|$ multiplied by half, of course this should be smooth here, all these are smooth.

So, as expected when we squeeze in time, we have stretched in frequency. And now let us interpret what we are doing when we take the dot product to calculate the coefficients in a V_m . To calculate coefficients in a subspace V_m , what are we effectively doing? We are taking

an inner product, namely the inner product of x with $\phi(2^m t - n)$, remember. Now you know this of course normalisation here. So if you want to work with an orthonormal basis, then it should not quite be $\phi(2^m t - n)$ which is the power of $2^m t - n$, one must normalise it to make it unit normal.

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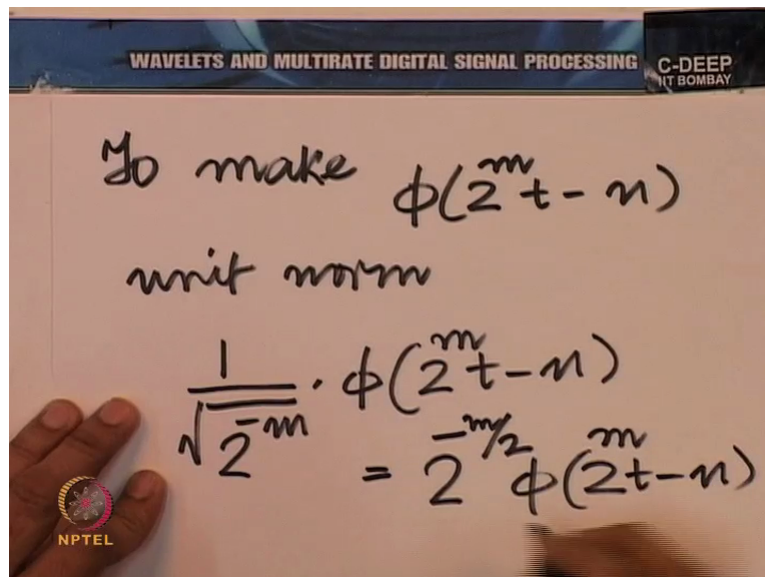
$\phi(2^m t - n)$
 translation does
 not affect $\|\cdot\|_2$
 without loss of
 generality, $n=0$

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
IT BOMBAY

$\|\phi(2^m \cdot)\|_2$
 $= \frac{1}{2^m} \int_0^1 1^2 dt = 2^{-m}$

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So I will take an instance, let us take any arbitrary m and let us look at the norm. So if we consider $\phi(2^m t - n)$ and again the $-n$ does not affect the norm, I am talking about the L_2 norm. And therefore we can as well take without loss of generality, we could take n equal to 0. Let us find out the norm therefore, so I am putting 2 raised to the power of m and then a dot, dot denotes the argument of the function but we are treating the function as an entity, so I do not use the explicit argument here. The norm of this is essentially integral, now you know when you go over $\phi(2^m t - n)$, you are talking about 0 to 2 raised to the power of $-m$ here, 1 square dt and that is obviously 2 raised to the power $-m$.


And therefore if you want to make this unit norm, then you must divide by, this is of course the square of the norm. So if you want to make this unit norm, you must divide the function by the square root of this. We must consider 1 by square root of 2 raised to the power $-m$ times $\phi(2^m t - n)$ and that is easily seen to be 2 raised to the power $-m$ by 2 $\phi(2^m t - n)$. So this is the unit norm, this is now an orthonormal basis. So let us make that note.

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$$\left\{ 2^{-m/2} \phi(2^m t - n) \right\}_{n \in \mathbb{Z}}$$

is an orthonormal
basis for V_m



WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP
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Consider $x(\cdot) \in L_2(\mathbb{R})$

$$\left\langle x(\cdot), 2^{-m/2} \phi(2^m \cdot - n) \right\rangle$$


$2^{-m/2} \phi(2^m t - n)$ for all integer n is an orthonormal basis for V_m , you know what V_m means, V_m is the m th subspace in the ladder of subspaces that leads to $L_2(\mathbb{R})$ as you go rightwards and to the trivial subspace with only the 0 element as you would left words. So it is that m th subspace in the ladder and now we have an orthonormal basis for it. Anyway now let us interpret what happens when we take the dot product of any function in $L_2(\mathbb{R})$ with an element of this orthonormal basis.

So, consider X_t belonging to $L_2(\mathbb{R})$ or if you please X with the argument belonging to $L_2(\mathbb{R})$, as is the correct way to write it. And then consider the inner product of this X with this orthonormal basis elements, $2^{-m/2} \phi(2^m \cdot - n)$. Now we are going to invoke the Parseval's theorem. You will remember that we have

discussed the Parseval's theorem awhile ago in one of the earlier lectures. When I talked about the relationship of functions and vectors, I had mentioned the significance of Parseval's theorem.

There are different ways of stating it, Parseval's theorem on one hand says that the inner product is preserved as we go from time to frequency. Now here if we use angular frequency, a factor of 2π is needed, if we use hertz frequency, that 2π factor is not required. But since we are working with angular frequency, it would be safer to retain that factor of 2π . But that factor of 2π apart, what Parseval's theorem says is that after all when you go from the function to its Fourier transform, in effect you are representing the same function in a different basis.