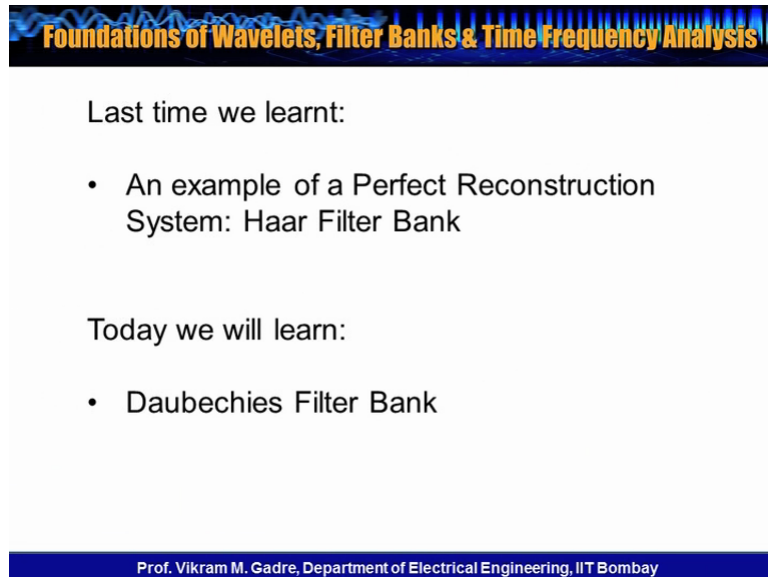


**Foundations of Wavelets, Filter Banks and Time Frequency Analysis.**  
**Professor Vikram M. Gadre.**  
**Department Of Electrical Engineering.**  
**Indian Institute Of Technology Bombay.**  
**Week-4**  
**Lecture -12.3.**  
**Introduction to Daubeshies family of MRA.**

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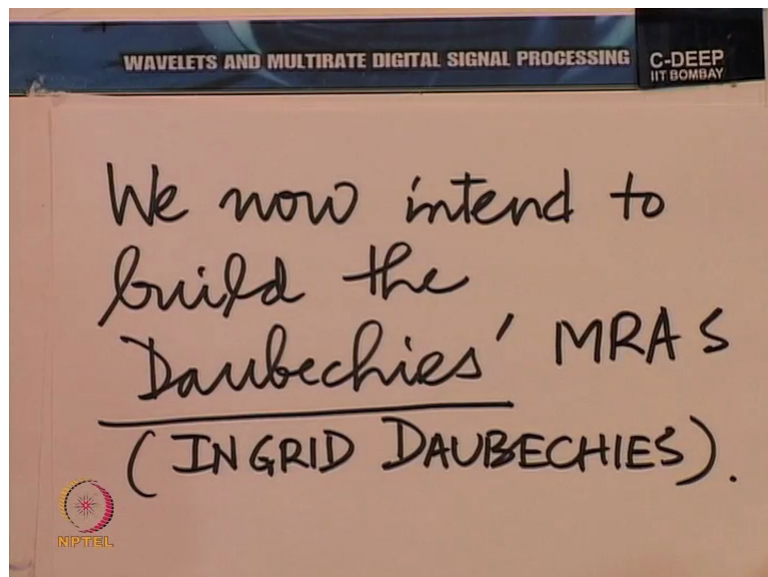
Last time we learnt:

- An example of a Perfect Reconstruction System: Haar Filter Bank

Today we will learn:

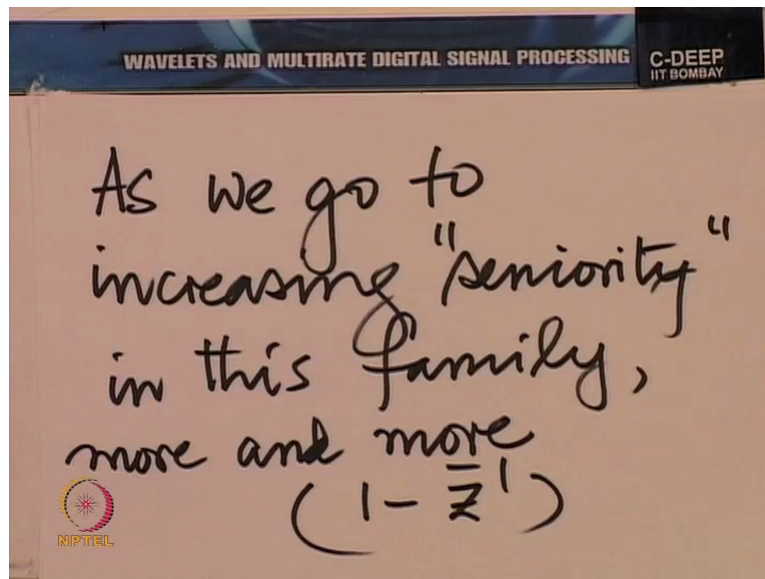
- Daubechies Filter Bank

Prof. Vikram M. Gadre, Department of Electrical Engineering, IIT Bombay



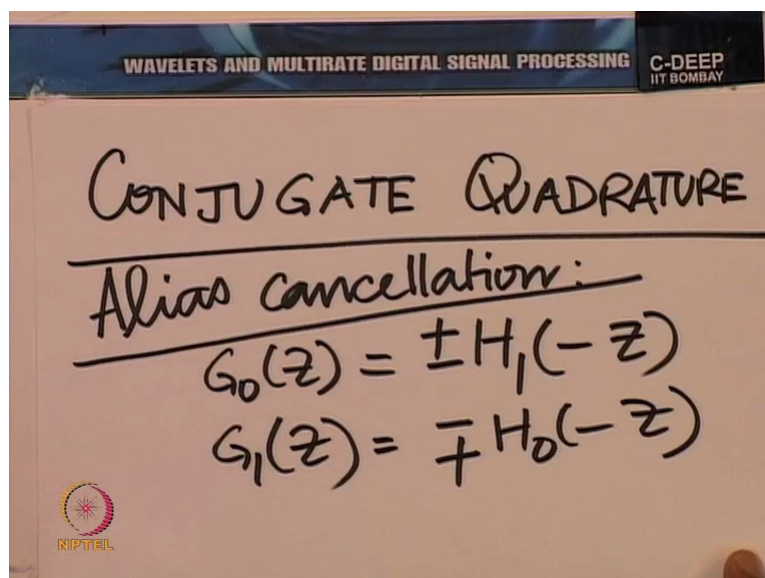
So I will just write that down. We now intend to build what is called the Daubeshies family. You know this name is actually the name of a mathematician scientist, whatever you might want to call her and the full name is this. I believe this is currently pronounced as Dobash but I could be wrong, I think we could just probably say Daubeshies and be content. So anyway we now intend to build the Daubeshies filter bank, the Daubeshies MRAs.

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And the feature of these Daubeshies MRAs is that as we go to the senior members of the family, as we go to increasing seniority there are more and more  $1 - Z$  inverse terms. So on the high pass branch we are effectively cancelling or killing higher and higher order polynomials. So the other way of looking at it is that if you are cancelling or killing them on the high pass branch, they must go to the lowpass branch. So we are retaining more smoothness from the lowpass branch, that is another way of looking at it.

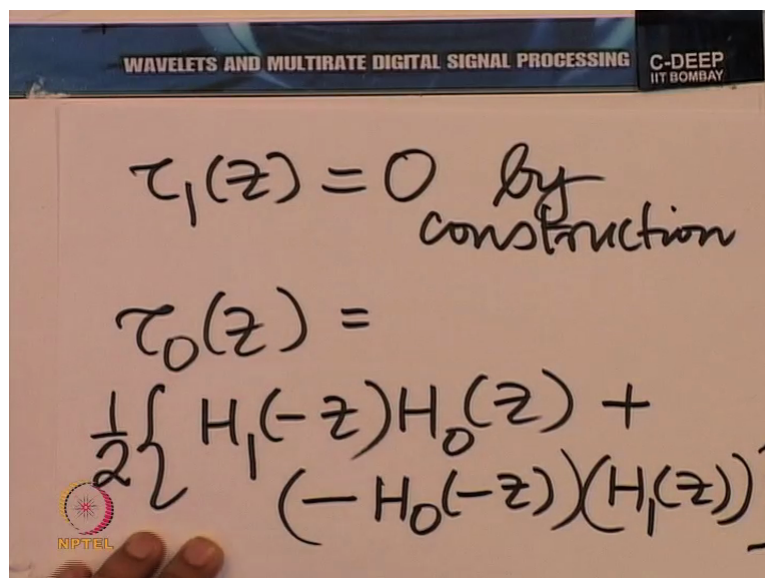
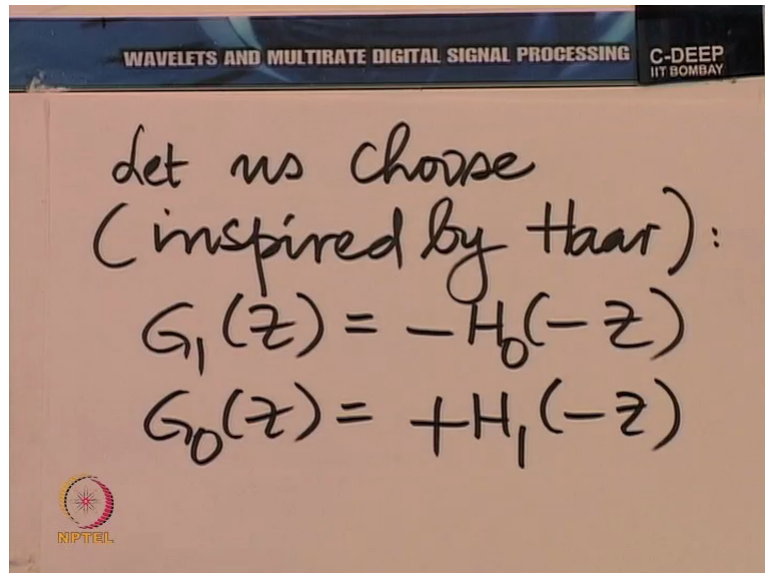
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And addition to doing this, we also want the same kind of analysis and synthesis filter. So all this together leads us to a special class of filter banks which we shall now put down explicitly. And the filter banks are called conjugate quadrature filter banks. So we are looking

at 1 class of what are called conjugate quadrature filter banks. Now the describing equations of these conjugate quadrature filter banks are very simple, we start from the Alias Cancellation condition. The Alias Cancellation condition says  $G_0 Z$  needs to be essentially  $H_1$  of  $-Z$ .

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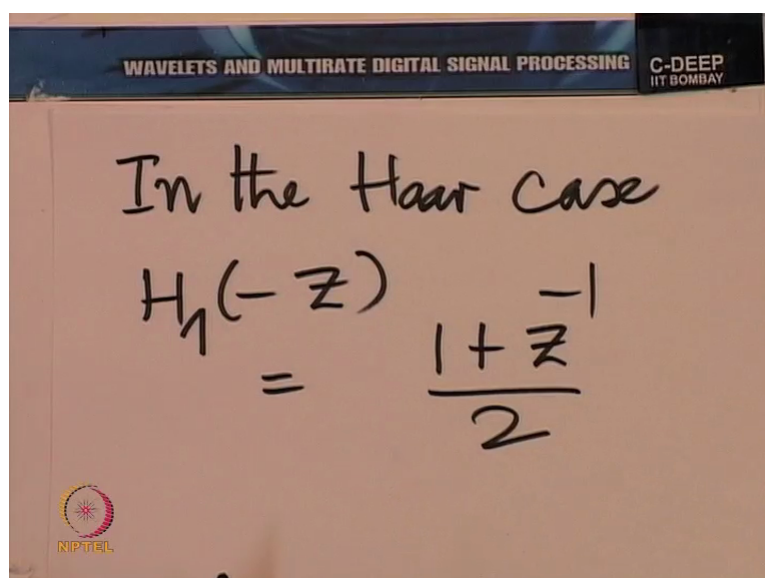
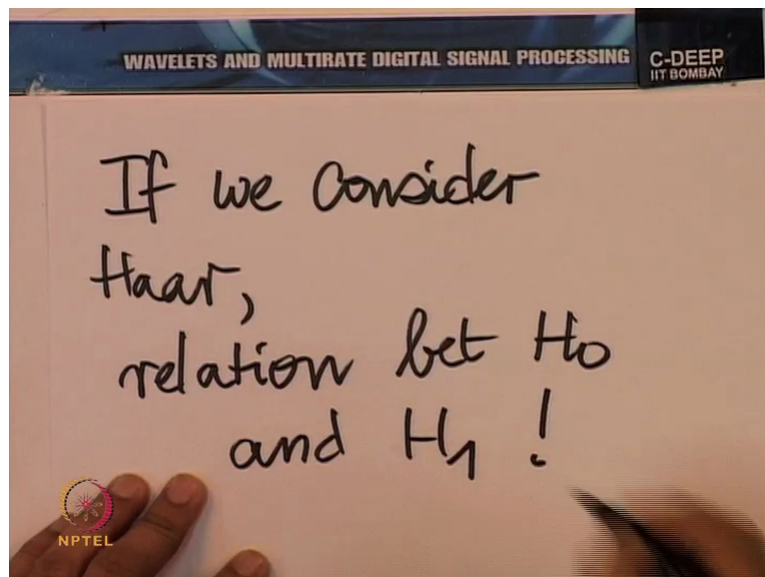


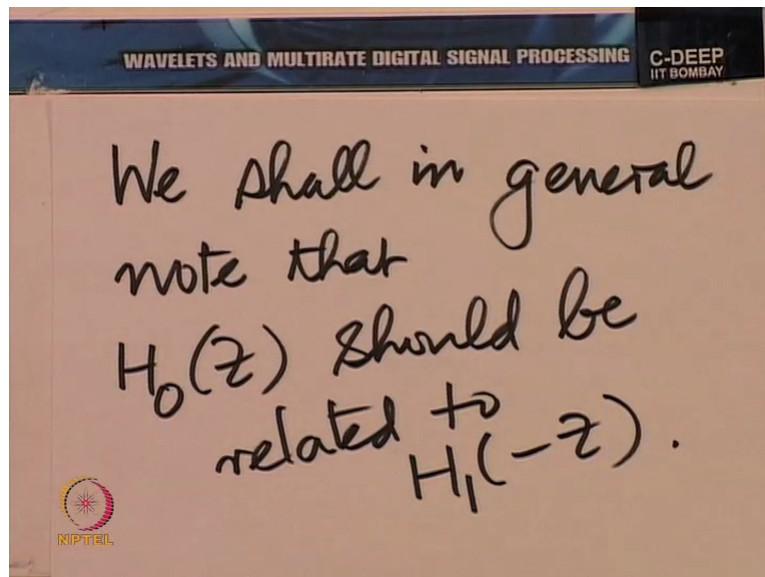
And we have the freedom to put + or - here. Similarly  $G_1 Z$  needs to be correspondingly - or +  $H_0 - Z$ . Now taking inspiration from the Haar, let us take the following choice.  $G_1 Z$  is  $-H_0 - Z$  and therefore  $G_0 Z$  is then +  $H_1$  of  $-Z$ . Now you know we keep away the factor of 2 for the moment, because after all that factor can be absorbed in the  $C_0$  that we have allowed. So with this substitution what do we get? The  $\tau_0 Z$ , of course  $\tau_1 Z$  is identically 0 by construction, but  $\tau_0 Z$  takes the following form then.

Tao  $0Z$  takes the form half  $G0Z$  which is essentially  $H1 - Z$  times  $H0Z + - H0 - Z$  times  $H1Z$ . So we have an interesting situation here. We have this  $H0Z H1 - Z$  product. You know  $H1 Z$  is essentially a high pass filter,  $H1 - Z$  therefore essentially becomes or aspires to become a lowpass filter with a cut-off of  $\pi$  by 2. So here you essentially have a cascade of 2 lowpass filters with a cut-off  $\pi$  by 2 and correspondingly this becomes a cascade of high pass filters with cut of  $\pi$  by 2.

And you are effectively saying that the overall system function with this cascade of lowpass filters of cut of  $\pi$  by 2 and this pair of high pass filters with cut of  $\pi$  by 2 must go towards the perfect reconstruction situation, that is the interpretation of this equation here. So we now focus on  $H1 Z$ .

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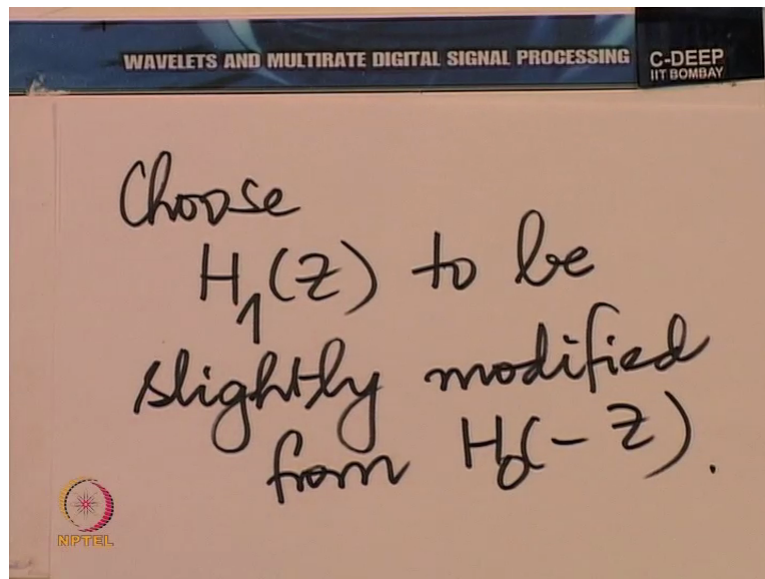


You see if we look at the Haar once again, there is a relation between  $H_0$  and  $H_1$ , in fact what we are trying to say is we of course need to relate the synthesis to the analysis for the purpose of Alias Cancellation. But now we have a perfect reconstruction requirement, so we want this  $Tao\ 0Z$  to essentially go to a delay and a multiplying constant. Now that means you need a relation between  $H_1$  and  $H_0$ . And one simple thing to do is to make  $H_0Z$  related to  $H_1 - Z$ , that is what we have actually done in the Haar. If you look at it carefully in the Haar case,  $H_1 - Z$  is essentially  $1 + Z$  inverse by 2.

So you know  $H_1 - Z$  and  $H_0Z$  are very closely related. Now let us generalise this. In fact the only catch is you know we will later on need to make a little adjustment here. So at this moment even if we simply accept  $H_1 - Z$  to be equal to, in the Haar case this is equal to  $H_0Z$  but we may need to make a little adjustment here. So what we will do is we shall in general note that  $H_1$  or rather  $H_0$  should be related to  $H_1 - Z$ .

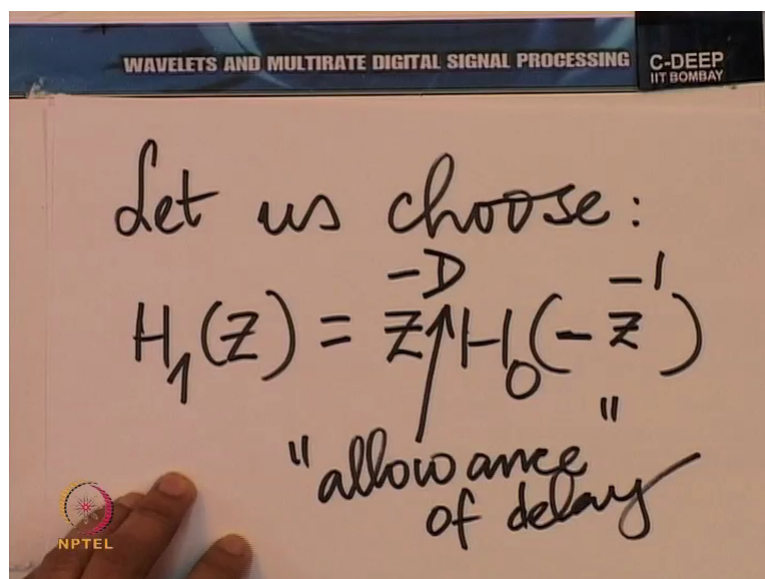


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So in the Haar case they are equal but in general we ask for a relation, very close relationship. The other way of saying it is that if you replace  $Z$  by  $-Z$  in  $H_0$ , you should get the  $H_1$ . So what we are saying is choose  $H_1$  to be derived from or to be slightly modified from  $H_0$  of  $-Z$ . Modified in what way? So in fact you know here again there is a little bit of an issue. What I am now going to do is to put down a choice for  $H_1$  by knowledge or by exposure to the filter banks that I have and justify it later. So you will have to bear with me for a little while, not too long, maybe just about a lecture.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

Haar:

$$z^{-1} H_0(-z^{-1})$$

$$H_0(-z^{-1})(z) = \left(\frac{1-z}{2}\right) z^{-1}$$

$$= \left(\frac{z^{-1}-1}{2}\right)$$

NIPTEIL

WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

Consider

$$H_1(z) = z^{-D} H_0(-z^{-1})$$

NIPTEIL

I shall put down the choice here, I shall partially justify the choice in this lecture and completely justify the choice in the next lecture where we once again look at the whole system in toto. So let us choose  $H_1(z)$  not to be  $H_0(-z)$  but  $H_0(-z^{-1})$ . And we will also allow for the possibility of a  $z$  raised to the power  $D$  here. So we will say  $-D$ , we will allow for this possibility. If we do that, in fact we can verify for the case of Haar, it is very easy.

For the Haar case let us take  $z^{-1}$  times  $H_0(-z^{-1})$  and indeed  $H_0(-z^{-1})$  for the Haar case is essentially  $(1-z)/2$ . And then if I take  $z^{-1}$  times  $H_0(-z^{-1})$ , I multiply this by  $z$  inverse, multiply both sides by  $z^{-1}$ , I have essentially  $(z^{-1}-1)/2$ . So we are doing well. Now let us make this the most general case. So we consider  $H_1(z)$  to be of this form,  $z^{-D}$  times  $H_0(-z^{-1})$ . And write down the  $Tao(z)$  for this.

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WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

$$\tau_1(z) = 0 \text{ by construction}$$

$$\tau_0(z) = \frac{1}{2} \left\{ H_1(-z)H_0(z) + (-H_0(-z))(H_1(z)) \right\}$$

NIPTEIL

WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

$$\tau_0(z) = \frac{1}{2} \left\{ H_0(z) \overset{-D}{(-1)} H_0(z^{-1}) \overset{-D}{z} - H_0(-z) \overset{-D}{z} H_0(-z^{-1}) \right\}$$

NIPTEIL

Tao 0Z would become half then H0 Z H1 of - Z, so H1 of - Z becomes -1 raised to the power - D times H0 Z inverse. Now -, you see you have G0, so again now you have G0 and let me put back the expression for you just for convenience, we want Tao 0Z to be this, H1 - Z times H0Z, for which we have this term H0Z into H1 - Z, all right. So you have Z raised to D there and - H0 - Z times H1 Z. And H1 Z we have accepted to be Z raised to the power - D times H0 - Z inverse.

So we have Z raised to the power - D here. Now what we intend to do in the next lecture is essentially to look at this expression. So we have H0Z, H0Z inverse here, H0 - Z, H0 - Z inverse there, the Z raised to the power - D terms there and of course you have a -1 rest to the power - D here and a -1 here, we need to choose these strategically, it is clear. And then we



also need to put some conditions on  $H_0$  so this indeed becomes a perfect reconstruction situation. We shall complete this in the next lecture and take it further from there. Thank you.