

## Fundamentals of Wavelets, Filter Banks and Time Frequency Analysis.

Professor Vikram M. Gadre.

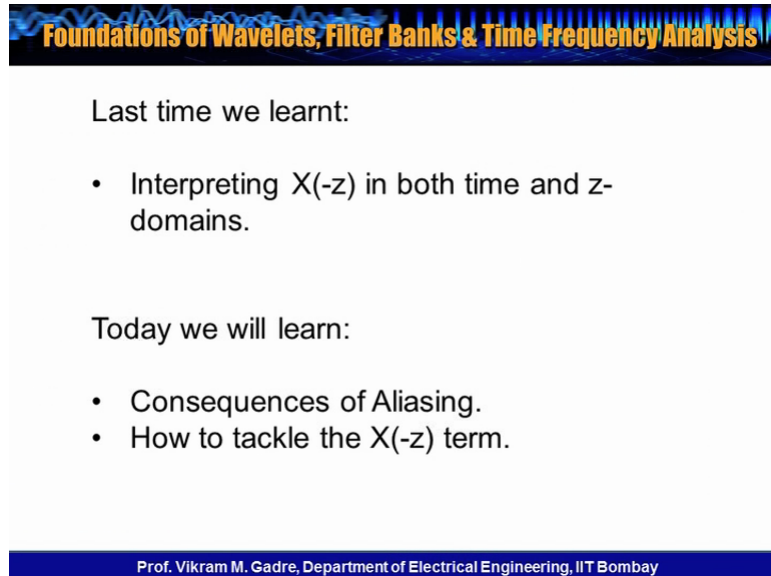
Department Of Electrical Engineering.  
Indian Institute of Technology Bombay.

Week-4.

Lecture-11.3.

Consequences of aliasing and simple approach to avoid it.

(Refer Slide Time: 0:17)



**Foundations of Wavelets, Filter Banks & Time Frequency Analysis**

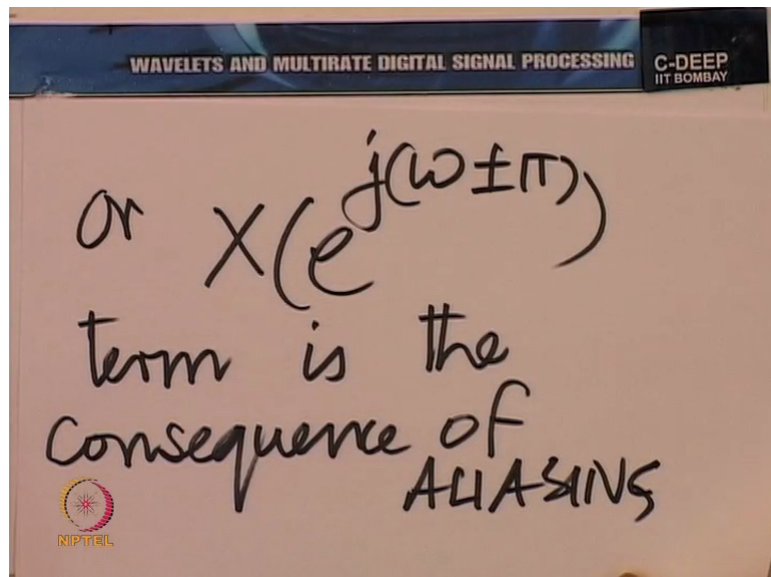
Last time we learnt:

- Interpreting  $X(-z)$  in both time and  $z$ -domains.

Today we will learn:

- Consequences of Aliasing.
- How to tackle the  $X(-z)$  term.

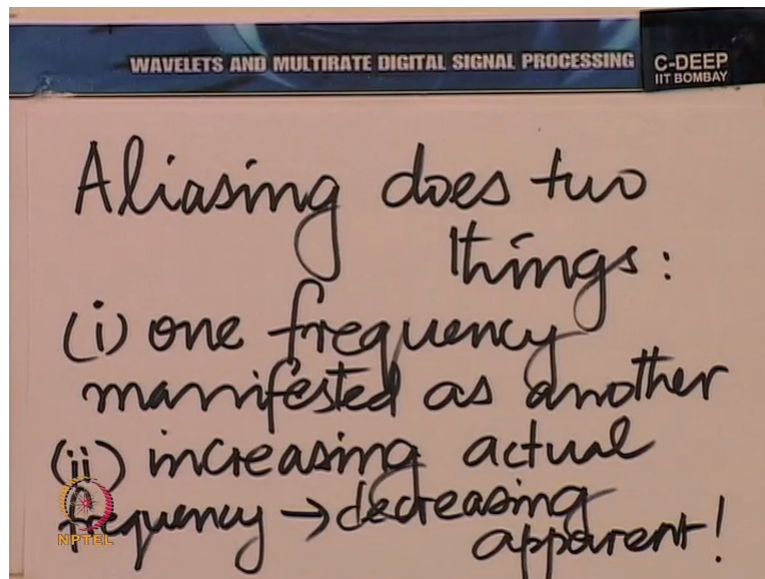
Prof. Vikram M. Gadre, Department of Electrical Engineering, IIT Bombay



WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

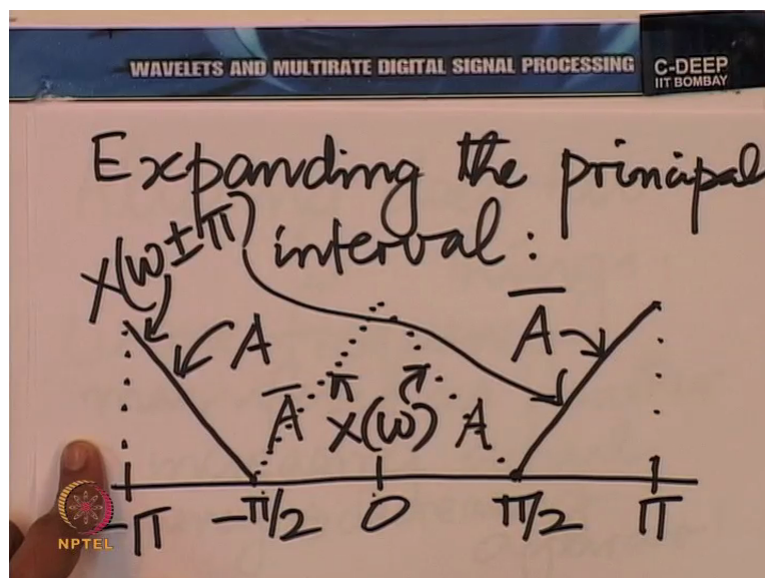
or  $X(e^{j(\omega \pm \pi)})$   
term is the  
consequence of  
ALIASING

NIPTEL



What is the consequence of aliasing, we should put it down very clearly. Aliasing does 2 things, 1 one frequency manifested as another, 2 increasing actual frequency leads to decreasing apparent frequency. So in fact just to emphasise let me put back before you and stress this point.

(Refer Slide Time: 1:33)



If we want  $Y(z)$  to reconstruct  $X(z)$  we must first do away with ALIASING!



$$\frac{G_1(z)}{G_0(z)} = - \frac{H_0(-z)}{H_1(-z)}$$



Very simple choice:

$$G_1(z) = \underline{-} H_0(-z)$$

$$G_0(z) = \overline{+} H_1(-z)$$



As you increase the frequency from 0 to  $\pi$  by  $2$  here, the frequency appears to decrease some  $\pi$  to  $\pi$  by  $2$  there. Now the 1<sup>st</sup> thing we have to do if we want perfect reconstruction is to do away with aliasing. So we put down a condition for aliasing cancellation. If we want  $Y_Z$  to reconstruct  $X_Z$ , we must 1<sup>st</sup> do away with aliasing. And how can we possibly do away with aliasing, that essentially means we want  $G_0 Z H_0 - Z + G_1 Z H_1 - Z$  equal to 0.

And in fact if we wish to explicitly express the synthesis filters in terms of the analysis filters, we could do that as well. Let us rearrange this equation to get  $G_1 Z$  by  $G_0 Z$  is  $- H_0 - Z$  by  $H_1 - Z$ . And of course a very simple choice is  $G_1 Z$  equal to the numerator, let us say  $+ H_0 - Z$  and correspondingly  $G_0 Z$  is  $-$  respectively  $+ H_1 - Z$ . This is a very simple choice of synthesis filter from the analysis filters which can give us alias cancellation.

Now in fact we can even spend a minute in interpreting what we have just said here in this very simple choice. I must of course emphasise, this is a simple choice but definitely not the only choice. In general you should, you should note that there would be a factor cancelled in the numerator and denominator. So more general choice would be as follows.

(Refer Slide Time: 4:22)

WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

More generally:

$$G_1(z) = \pm R(z) H_0(-z)$$

$$G_0(z) = \mp R(z) H_1(-z)$$

$R(z)$ : the factor "cancelled"

WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING

C-DEEP  
IIT BOMBAY

Interpret

$$G_1(z) = +H_0(-z)$$

Ideally:

NIPTEIL

$G_1 Z$  is of the form  $+ -$  some factor, let us say  $R$  of  $Z$  times  $H_0 - Z$ . And  $G_0 Z$  is respectively  $-$  or  $+ RZ H_1 - Z$ . So  $RZ$  is you know in some sense the factor cancelled, you know when you divide  $G_0 Z$  by  $G_1 Z$ , some factor has been cancelled and what is left is  $H_0$  and  $H_1$ , that is the way you should look at it, that is the more general situation. Anyway now let us interpret what we have written here, namely  $G_1 Z$ , this is a very simple choice of  $G_1$  is  $H_0$  of  $- Z$  and  $G_0$  is  $+ \text{ or } -$ , you know you just interchange  $+ \text{ or } -$  here  $H_1$  of  $- Z$ .

So let us take the upper one,  $G_1 Z$  is  $H_0 - Z$ , the other one follows similarly. You see, ideally this is what the frequency response of  $H_0$  should be, it should be an ideal lowpass filter with a cut-off of  $\pi$  by 2. So this is what the frequency response should look like, this is 1, the height here. So what would  $H_0 e^{j\omega}$  raised to the power  $\omega$   $J \Omega$   $+ - \pi$  look like? In other words, what is  $H_0$  of  $- Z$  then?



(Refer Slide Time: 6:37)

WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

$$H_d(-z) \Big|_{z \leftarrow e^{j\omega}} = H_0(e^{j(\omega \pm \pi)})$$

NPTEL

WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

"Zoomed"  $H_0(e^{j\omega})$

NPTEL

WAVELETS AND MULTIRATE DIGITAL SIGNAL PROCESSING C-DEEP IIT BOMBAY

"Zoomed"  $H_0(e^{j(\omega \pm \pi)})$

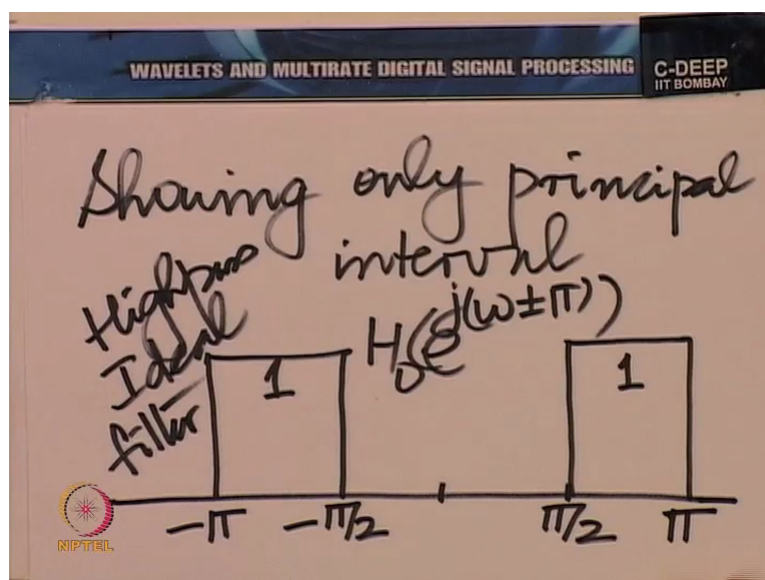
Principal Interval

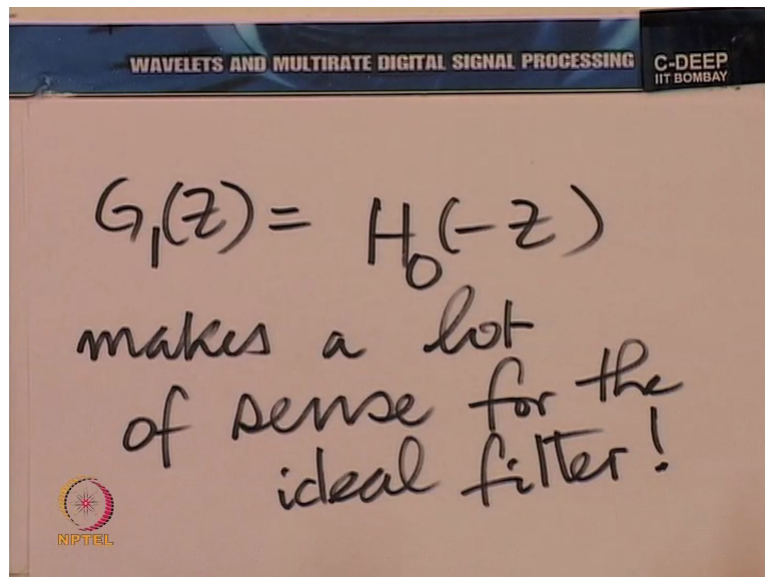
NPTEL

$H_0$  of  $-Z$  with  $Z$  replaced by  $e$  raised to the power  $j\Omega$  is essentially of course as you know  $H_0 e$  raised to the power  $j\Omega + \pi$ . And that is as you can very easily infer looks like this. Again let us use strategy of zooming and then contracting back again. So I have  $\pi$ ,  $-\pi$  there,  $2\pi$ ,  $3\pi$  and so on. This is what  $H_0$  look like, taking note of the periodicity, remember. This is  $-\pi$  by  $2$  there, this is  $\pi$  by  $2$ , this would therefore be  $3\pi$  by  $2$  and so on.

So the zoomed  $H_0 e$  raised to the power  $j\Omega + \pi$  would look like this, it would have an appearance like this and so on here. These pass bands so to speak are now going to lie at the odd multiples of  $\pi$  and you can continue that drawing. Now once again we confine to the principal interval, the principal interval is here. And when we do so and then zoom back to emphasise only the principal interval, what do we get?

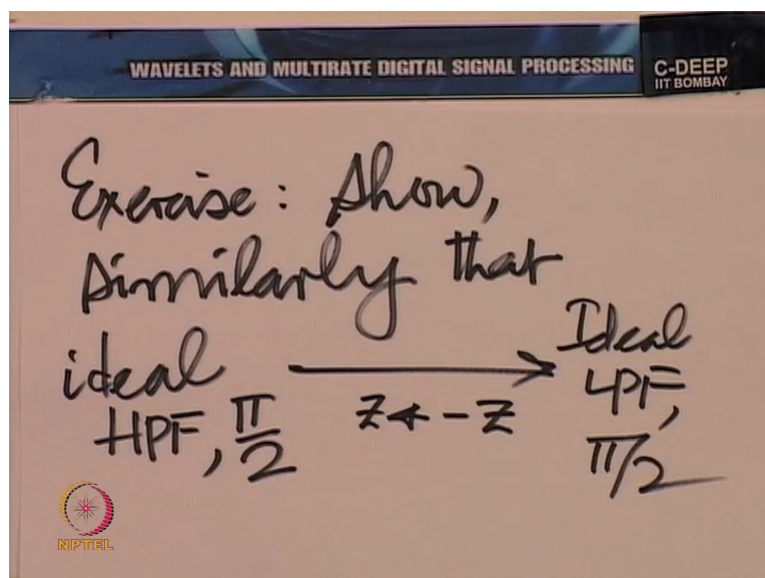
(Refer Slide Time: 9:30)





We get this, lo and behold, from a lowpass filter with a cut-off of  $\pi/2$ , we have a high pass ideal filter of course again with a cut-off of  $\pi/2$ . So this falls into place very well, in fact the equation  $G_1(z) = H_0(-z)$  makes a lot of sense. All that we are saying is that on the synthesis side if you had an ideal lowpass filter with a cut-off of  $\pi/2$  at  $H_0$ , you should put an ideal high pass filter with cut-off of  $\pi/2$  at the point  $G_1$ . And you know if you look back at the other one of the 2, namely  $G_0$ , you would make a similar inference.

(Refer Slide Time: 11:51)



$G_0(z)$  is  $H_1$  of  $-z$  with a  $-$  sign. So if you look at the magnitude sense, again a high pass filter would become a lowpass filter, in fact I leave this to you as an exercise. So exercise, show similarly that an ideal high pass filter HPF with a cut-off of  $\pi/2$  on replacing  $z$  by  $-z$



becomes an ideal lowpass filter with a cut-off of  $\pi/2$ . And that also makes a lot of sense for the other of the 2 requirements for alias cancellation.

So today we have looked at the condition for alias cancellation and we have interpreted the simplest of the conditions. Now if you do not stick to the simple condition and if you have the factor of  $RZ$ , all that we need to do is to say that we are also modifying that filter a little, beyond just this lowpass to high pass and high pass lowpass conversion, that  $RZ$  factor would carry out that modification. So we have taken one out of 2 steps today in building a perfect reconstruction 2 band filter bank.

We have taken the step of alias Cancellation. In the next lecture we shall take the 2<sup>nd</sup> step, namely perfect reconstruction. So we have done away with  $\tau_0 Z$ , I mean  $\tau_1 Z$ , but now we need to see what to do with  $\tau_0 Z$  which we shall do in the next lecture. Thank you.