

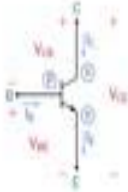
Basic Electronics
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Lecture – 24
Bipolar Junction Transistor (continued)

Welcome back to Basic Electronics. In the last lecture we looked at the Ebers moll model which describes the BJT in all modes of operation we will now use the Ebers moll model to plot the collector current I_C of a BJT as a function of the collector to emitter voltage V_{CE} with the help of the $I-V$ characteristics. We will take another look at the earlier BJT circuit. And figure out the correct solution finally, we will look at a circuit with a PNP transistor. So, let us begin.

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BJT $I-V$ characteristics



- Since BJT is a three-terminal device, its behaviour can be described in many different ways, e.g.,
 - I_C versus V_{CB} for different values of I_B
 - I_C versus V_{CE} for different values of V_{BE}
 - I_C versus V_{CE} for different values of I_B
- The $I-V$ relationship for a BJT is not a single curve but a "family" of curves or "characteristics."
- The I_C-V_{CE} characteristics for different I_B values are useful in understanding amplifier biasing.

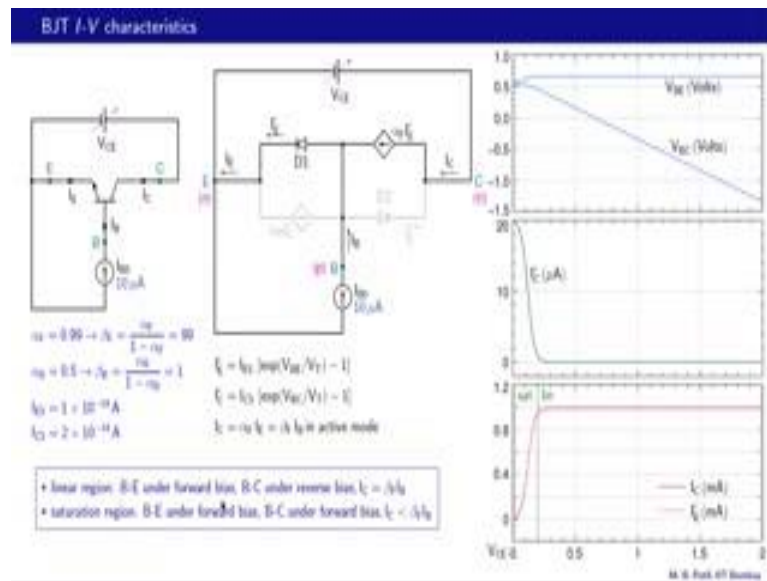
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We now want to discuss the $I-V$ characteristics of a BJT in a graphical form. Now we first note that BJT is quite different from a diode in that it has 3 terminals. For a diode we had only 2 terminals. So, there was exactly one current and one voltage to talk about and therefore, the entire diode behavior could be captured with just one plot the current versus voltage. That is not the case for a BJT. We can look at several different things and here is some of this I_C versus V_{CB} ; so this current I_C versus this voltage V_{CB} for different values of I_E . I_{E1} I_{E2} and so on or we can plot I_C versus V_{CE} I_C as a function of V_{CE} for different values of V_{BE} , V_{BE1} V_{BE2} and so on or we can plot I

C versus V_{CE} , I_C versus V_{CE} for different values of the base current. And of course, there are some other ways of plotting the current as well.

So, the $I-V$ relationship for a BJT is not a single curve, but family of curves or characteristics. And we will look at the I_C versus V_{CE} curves or characteristics for different I_B values because they will find this is very useful in understanding amplifier biasing.

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What we are going to do take an NPN transistor and plot I_C as a function of V_{CE} this voltage drop for various values of I_B . And we will do that by solving the Ebers moll equations. So, the experimental set up would be something like this. Voltage source is connected between collector and emitter this value is V_{CE} . So, this is going to be our x axis variable, the collector current I_C is going to be our y axis variable. So, we are going to plot I_C as a function of V_{CE} . And we will find that I_C versus V_{CE} depends very much on what this base current is. So, therefore, for a given base current let us say 10 micro amps we are going to get one of these curves I_C versus V_{CE} . If I change these 10 micro amps to 20 micro amps or 50 micro amps we are going to get another I_C versus V_{CE} curve and so on.

So therefore, the $I-V$ characteristics of a BJT like we said in the last slide are not one single curve, but a family of curves and that is the complete picture that we want to see. These are the parameters we will consider alpha F the forward alpha as 0.99 that makes

beta equal to alpha by 1.3 minus alpha that is 99. Alpha R the reverse alpha equal to 0.5 that makes the reverse beta equal to alpha R by 1.3 minus alpha R or one and clearly this transistor is very poor in the reverse direction because this beta is so small in the reverse direction. And these are the saturation currents $I_{ES} = 10^{-14}$ ampere and $I_{CS} = 2 \times 10^{-14}$ amperes.

First let us replace the transistor with the Ebers moll model like that and recall that in the active mode we need to only consider I_E prime and αI_E prime, but in general we need to consider all 4 branches. What we want to do is to get I_C as a function of V_{CE} for a fixed value of I_B that is 10 microamperes. What we will do now is to take one specific value of V_{CE} say 1.3 volt and figure out how we can calculate I_C for that V_{CE} with I_B equal to 10 microamperes. So, for convenience let us take the emitter node as the reference node so; that means, V_E is 0 volts and V_C the collector voltage would then be 1.3 volt because we have taken V_{CE} equal to 1.3 volt. And we do not know what V_B is going to be.

So, what we need to do is to solve the Ebers moll equations for V_B and once we know V_B we know this difference V_{BE} we also know this difference V_{BC} and once we know these 2 voltages we can find I_E prime given by that equation, we can find I_C prime and once we know that we can find all other quantities. For example, I_E is going to be I_E prime minus $\alpha_R I_C$ prime and so, on. So, we need to get an equation in terms of V_B and what is that equation that equation is given by I_B equal to 10 microamperes.

What is I_B ? I_B is equal to I_E prime plus I_C prime minus $\alpha_F I_E$ prime minus $\alpha_R I_C$ prime. And I_E prime and I_C prime essentially are functions of V_B because V_{BE} here is V_B minus V_E . V_E is known 0 volts. V_{BC} is equal to V_B minus V_C , V_C is known one volt and therefore, I_E prime and I_C prime can be written in terms of V_B . So, that is the equation we need to solve for V_B , and you can see that it is going to be a non-linear equation because we have exponential terms over here. So, we need to use some iterative method to obtain V_B .

So, by solving this equation that is I_B equal to 10 microamperes, for V_{CE} equal to 1.3 volt we are going to find one value of the collector current. And then we repeat this exercise for other values of V_{CE} . And that is how we are going to plot I_C as a function of V_{CE} . And here are the results that we obtained. This axis is V_{CE} , this is V_{CE} equal

to 0 this is V_{CE} equal to 2 volts. This curve here is V_{BE} ; that means, the voltage across the base emitter diode, this curve is V_{BC} the voltage across the base collector diode, this current is I_C prime, this current here and the solid line here is I_C the terminal current and the dashed line is I_E prime this current here.

Let us look at the various quantities of interest at specific value of V_{CE} , say V_{CE} equal to 1.3 volt and let us start with the junction voltages. Here is V_{BC} and for V_{CE} equal to 1.3 volt V_{BC} is about minus 0.3 volts; that means, the base collector diode is under reverse bias and; that means, I_C prime is very small nearly 0 and therefore, this branch is like an open circuit and so is this branch. And if we look at the I_C prime plot, we see that I_C prime is 0 at V_C equal to 1.3 volt.

Next let us look at V_{BE} the base emitter voltage this plot here and V_{BE} is about 0.6 volt may be between 0.6 and 0.7 and; that means, the base emitter diode is under forward bias and therefore, our transistor is operating in the active mode with the base emitter junction under forward bias and the base collector junction under reverse biased.

Now, since I_C prime is 0. I_C equal to α_F times I_E prime. What is α_F , very close to 1.3 is 0.99 for this transistor. So, I_C is nearly equal to I_E prime and that is what we see over here. I_C and I_E prime are nearly equal. Now since we know that the transistor is operating in the active mode, we also know that I_C can be written as beta times I_B and what is beta is 99. So, I_C is expected to be 990 times I_B or 990 times 10 microamperes, that is 990 microamperes that is nearly 1.3 milliamp and that is what we observe over here I_C is about 1.3 milliamp.

Let us summarize the situation for V_{CE} equal to 1.3 volt. We have about 0.7 volts appearing across D_1 . And the rest of the voltage that is 1.3 volt minus 0.7 volts about 0.3 volts appears across the base collector junction. And since this end is positive compared to this one the base collector diode is not conducting all right. What we will do now is consider another value of V_{CE} namely V_{CE} equal to 2 volts. For this voltage we notice that our base emitter voltage has not really changed, we cannot really tell the difference and the entire difference that is 2 volts minus 1.3 volt equal to 0.7 volt has appeared in V_{BC} .

This was minus 0.3 volts and this is about minus 1.3.3 volts. So, the base collector junction continues to be under reverse biased the transistor continues to be in active

mode I_C prime continues to be 0. And our collector current also remains equal to 1.3 milliamp. So, all that has happened as we go from V_{CE} equal to 1.3 volt to V_{CE} equal to 2 volts is that an additional reverse biased of 1.3 volt has now appeared across this base collector diode. Earlier with V_{CE} equal to 1.3 volt the reverse bias was minus 0.3 volts and with V_{CE} equal to 2 volts it is now minus 1.3.3 volts apart from that there is no real change.

Let us now look at what happens, when we change V_{CE} in the other direction let us say from 1.3 volt to 0.5 volts and let us start looking at the diode voltage is first. Now V_{BE} has not really changed it has remained at about 0.7 volts. V_{BC} has changed from minus 0.3 volts to about 0.2 volts. So, the entire change in V_{CE} has actually got reflected in V_{BC} . And I_C prime continues to be small. Note that V_{BC} has now becomes slightly positive, but it is not large enough to make I_C prime substantially or significantly large and that is why on this scale I_C prime is still looking like 0.

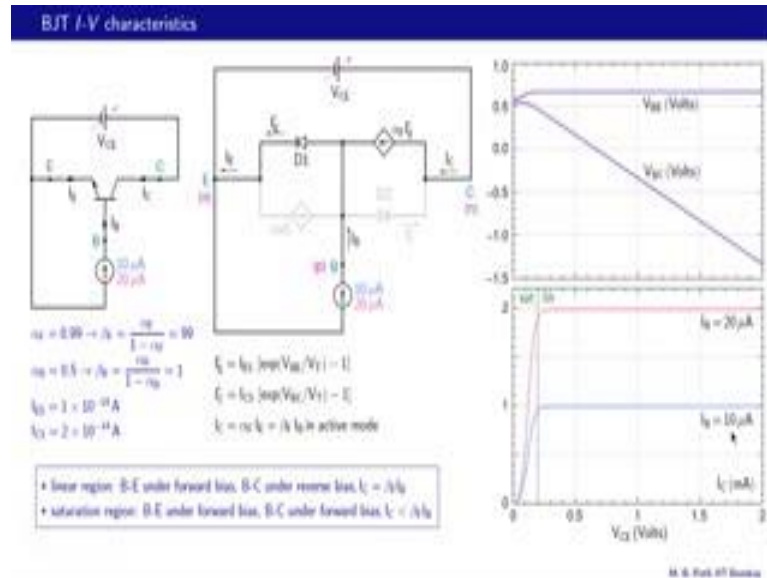
So, the transistor continues to be in the active mode and I_C continues to be about 1.3 milliamp that is beta times I_B . At about 0.2 volts somewhere here the current I_C prime starts raising because in the forward bias across the base collector junction has now increased to make I_C prime significantly large. Apart from that there is also slight reduction in V_{BE} as we see over there and because I_C is $\alpha F I_E$ prime minus I_C prime this current is now decreasing because V_{BE} prime is going down a little bit and this current is increasing there what happens is I_C now starts going down like that.

So, we have 2 distinct regions in BJT $I-V$ characteristics. The right of this slide is the so called linear region which we have been calling as a active region as well and we are already familiar with that. The left side is the saturation region now this region is new to us in this region both the base emitter and base collector junctions are under forward bias and the collector current is less than what it would be in the active region. So, let us summarize these observations. We have a linear region the right side of that line in which the base emitter junction is under forward bias base collector junction is under reverse biased.

And I_C is beta times I_B and in this example it is about 1.3 milliamp. We have the saturation region on the left of this line and in saturation region the base emitter junction is under forward bias and the base collector junction is also under forward bias. And in

this region I_C is less than beta times I_B as we can see over here. So, here is the I_C V_{CE} curve for 10 micro amps again and in the linear region or the active mode the collector current is beta times I_B which is about one milliamp.

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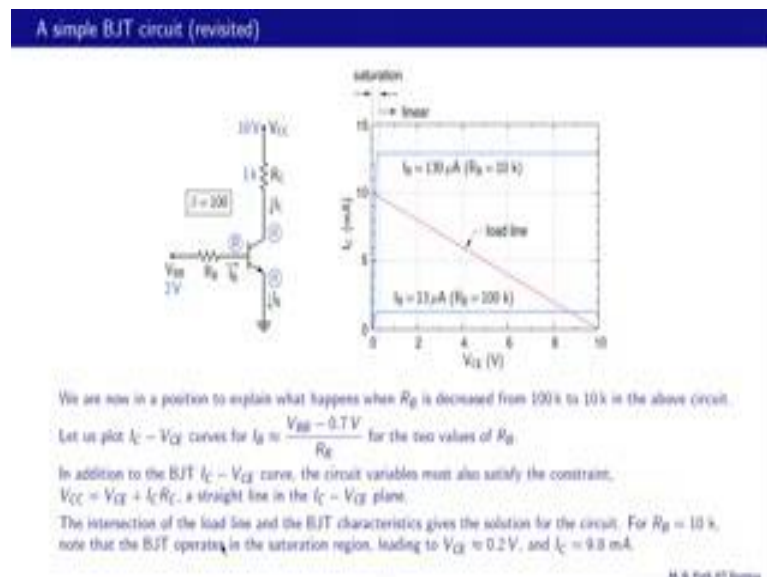
Now, we want to look at the I_C V_{CE} curve when the base current changes from 10 micro amps to 20 micro amps. So, let us look at the results. That is the I_C V_{CE} curve now for 20 micro amps. And we see that it has not really changed much in many aspects except in the linear region I_C has become 2 milliamps. Approximately 2 milliamps approximately 2 milliamps and for 10 micro amps. It was about one milliamp and this of course, is expected because in the linear region we have I_C equal to beta times I_B beta is about hundred. So, 100 times 20 micro amps that is 2 milliamps. So, this part is really expected simply from this equation, and the boundary between the linear and saturation regions has also not really changed the transition is happening at about the same V_{CE} value about 0.2 volts.

So, as we go from I_B equal to 10 micro amps to I_B equal to 20 micro amps there is a large change in I_C from 1.3 milliamp to 2 milliamps that is a factor of 2 and we now want to see how that change gets reflected in the V_{BE} and V_{BC} plots. What is I_C in the linear region I_C is alpha times I_E prime. So, if I_C has changed by a factor of 2 I_E prime would also have changed by a factor of 2. So, let us look at the equation for I_E prime, now I_E prime is $I_{ES} e^{V_{BE}/V_T}$. This one is too small compared

to the first term we will not worry about that and if I_E prime changes by a factor of 2, it calls for a change in this factor e raised to be V_e by V_T by 2.

Now, because this is an exponential relationship. That requires V_{BE} to change only slightly. And that is what we are seeing here essentially. So, the blue curve is for I_B equal to 10 micro amps and the pink one is for I_B equal to 20 micro amps. And we see that for 20 micro amps the base emitter voltage is only slightly larger than the case for 10 micro amps. And because these 2 voltages must add up to V_{CE} that changes is also reflected in V_{BC} . So, overall there is a large change in the collector current plot, but there is hardly any change in the base emitter and base collector voltages. As we change I_B from 10 micro amps to 20 micro amps. And based on this experience we can now predict what is going to happen for some other value of I_B for example, if I_B is made 5 micro amps, instead of 10 I_C is going to be 0.5 milliamp, because we are changing I_B by a factor of 2 I_C will also be changed by a factor of 2 because of this proportionality here and then we will get a curve $I_C - V_{CE}$ curve which looks like this. And we do not really need to do that calculation. Similarly, if I_B is 15 micro amps that are between these 2 values I_C would then look like that. And all of these curves will cross over from linear to saturation regions at about V_{CE} equal to 0.2 volts.

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Let us now revisit this circuit that we have looked at before, and if you remember when R_B was changed from 100 k to 10 k we found that active mode was not possible

anymore, and now having studied the I_C vs V_{CE} characteristics, we would be able to explain what happens. So, we are now in a position to explain what happens when R_B is decreased from 100 k to 10 k. So, what we will do is we will plot I_C versus V_{CE} for 2 values of I_B . 100 μ A and 10 μ A and as we have seen if we take this voltage as approximately 0.7 volts then R_B times I_B is 2 volts minus 0.7. So, that is 1.3. So, I_B is one 0.3 by 100 k or one 0.3 by 10 k, depending on the value of R_B . And let us now look at the I_C vs V_{CE} characteristics for these 2 base currents.

So, here is I_B equal to 13 micro amps, which corresponds to R_B equal to 100 k and what would be collector current here in the linear region that would be 13 micro amps multiplied by beta, that is one 0.3 milliamps and that is exactly what this is. When we change R_B from 100 k to 10 k our base current goes up by a factor of 10, and it now becomes 130 micro amps. And now the collector current in the linear region is 130 micro amps' times 100 or 13 milliamps. So, this collector current is now 13 milliamps.

Now, whatever solution we find for this circuit that is the value of I_C and the value of V_C , must satisfy either this constraint imposed by the transistor if R_B is 100 k or this constraint if R_B is 10 k and note that V_C here this voltage is the same as V_{CE} because the emitter is at 0 volts. Now apart from that there is one more constraint that we need to worry about that and that is imposed by this resistance R_C . So, let us see what is that is. So, in addition to the BJT I_C vs V_{CE} curve the circuit variables must also satisfy the following constraint, and let us see where this is coming from V_{CC} equal to V_{CE} plus $I_C R_C$, V_{CC} must be equal to this voltage plus this voltage.

Now this is just V_{CE} that is this variable here and that voltage drop is $I_C R_C$. Now these equations essentially look like a straight line in the I_C vs V_{CE} plane, and let us plot that straight line now. And there is what it looks like. Its x intercept is 10 volts and that is obtained by putting I_C equal to 0, if I_C is 0 then V_{CE} is the same as the V_{CC} , which is 10 volts and its y intercept is 10 milliamps and if we put V_{CE} equal to 0, now we get I_C equal to V_{CC} by R_C V_{CC} is 10 volts R_C is 1.3 k. So, that is 1.3 milliamps. So, that is the graph of this constraint which is imposed by R_C . The intersection of the load line and the BJT characteristics gives us the solution for the circuit and that is because the solution has to satisfy the load line constraint as well as the constraint imposed by the transistor I_C vs V_{CE} and that constraint is given this curve for R_B equal to 100 k and by that curve for R_B equal to 10 k.

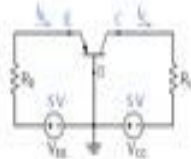
Now, for the first case R_B equal to 100 k that intersection is somewhere here and as we have seen earlier the solution is V_{CE} equal to 8.7 volts and I_C equal to 1.3 milliamps. In the second case that is R_B equal to 10 k we note that the intersection of the load line and the $I_C - V_{CE}$ curve happens to be in the saturation, region not in the linear region. And that is why we found that linear region does not possible when we try to solve this same problem analytically. So, what is the solution now. So, the V_{CE} is between 0 and 0.2 volts we often take this approximately as just 0.2 volts and so therefore, we have a voltage drop of 0.2 volts here this is 10 volts. So, 9.8 volts will drop there. So, 9.8 divided by 1.3 k that is 9.8 milliamps that would be I_C . So, that is the solution V_{CE} of about 0.2 volts and I_C equal to 9.8 milliamps.

So, sometimes these graphical methods are very useful and they help us to visualize things much better than an analytical approach.

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BJT circuit example

Assuming the transistor to be operating in the active region, find R_B and R_C to obtain $I_E = 2 \text{ mA}$, and $V_{BC} = 1 \text{ V}$ ($\alpha = 1$).



$$V_{BE} - V_{BE} + I_E R_E = 0 \rightarrow I_E R_E = 5 - 0.7 \rightarrow R_E = \frac{4.3 \text{ V}}{2 \text{ mA}} = 2.15 \text{ k}$$

$$V_{BC} + I_C R_C - V_{CC} = 0 \rightarrow I_C R_C = V_{CC} - V_{BC}$$

Since $\alpha = 1$, $I_C = I_E \rightarrow I_C R_C = 5 - 1 \rightarrow R_C = \frac{4 \text{ V}}{2 \text{ mA}} = 2 \text{ k}$

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Let us take this example with a PNP transistor. Assuming that the transistor is operating in the active region we want to find R_E and R_C such that I_E this current here diameter current is 2 milliamps and V_{BC} this voltage is 1.3 volt. And we are given that alpha is nearly equal to 1.3. Now to begin with let us first check that this condition V_{BC} equal to 1.3 volt does indeed corresponds to the linear region or the active region. So, we have PNP here. So, this is an n region that is a p region V_{BC} is 1.3 volt. So, V_B is higher than V_C by 1.3 volt so; that means, the n type region is higher than the p type region by 1.3

volt and; that means, reverse biased. So, this junction is indeed under reverse biased and now let us calculate R_E and R_C .

So, in this loop we can write V_{EB} this voltage drop which is going to be 0.7 minus this voltage raise 5 volts plus I_E times R_E equal to 0, and that give us $I_E R_E$ equal to 5 minus 0.7 or 4.3 volts. And since we are given the value of I_E 2 milliamps we can find R_E turns out to be 2.15 k. For the collector resistance the R_C value here we cannot write a loop equation for this loop V_{BC} plus $I_C R_C$ minus V_{CC} equal to 0. And that gives us $I_C R_C$ equal to V_{CC} minus V_{BC} and since alpha is nearly equal to 1.3 I_C and I_E are approximately equal to each other and therefore, instead of I_C here we can use I_E and that gives us R_C . So, I_E times R_C is about 5 minus 1.3 and that gives us R_C is equal to 4 volts by I_E which is 2 milliamps that is 2 k.

In conclusion we have seen how to represent the behavior of a BJT with the help of $I-V$ curves. We have used the $I-V$ characteristics to obtain a graphical solution of the BJT circuit considered earlier we have also taken up a circuit with a PNP transistor and seen that it can be analyzed in much the same manner as an NPN transistor. Having looked at BJT basics we will now start looking at an important topic namely amplification. That is all for now see you later.