

Basic Electronics
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Lecture – 11
RC/RL Circuits in Time Domain (Continued)

Welcome back to Basic Electronics. In the last lecture we got started with the series RC circuit with a periodic piecewise constant input namely a square wave. We will now continue with that example and work out the steady state solution; we will also learn about the meaning of charge conservation for a capacitor and look at its implications for this specific example. So, let us begin.

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RC circuit: example

$0 < t < T_1$ Let $V_C^{(1)}(t) = A e^{-t/\tau} + B$

$V_C^{(1)}(0) = V_1, V_C^{(1)}(\infty) = V_0$
 $\rightarrow B = V_0, A = V_1 - V_0.$

$V_C^{(1)}(t) = -(V_0 - V_1)e^{-t/\tau} + V_0$ (1)

Now use the conditions:
 $V_C^{(1)}(T_1) = V_2, V_C^{(2)}(T_1 + T_2) = V_1.$

$V_2 = -(V_0 - V_1)e^{-T_1/\tau} + V_0$ (3)

$V_1 = V_2 e^{-(T_1+T_2-T_1)/\tau} = V_2 e^{-T_2/\tau}$ (4)

Rewrite with $a \equiv e^{-T_1/\tau}, b \equiv e^{-T_2/\tau}.$

$V_2 = -(V_0 - V_1)a + V_0$ (5)

$V_1 = b V_2$ (6)

Solve to get

$V_1 = b V_0 \frac{1-a}{1-ab}, V_2 = V_0 \frac{1-a}{1-ab}$

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In order to find these constants A and B, we need 2 conditions on V c and phase one; condition number 1 at t equal to 0 this time point here, V c is equal to V 1 what about condition number 2? Let us look at this interval and this interval the input voltage V s is a is equal to V 0, and the capacitor has started charging; now in this phase the circuit is not aware that this transition is going to take place and therefore, its response would be as if this voltage did not change and if that were the case then this V c would finally reach V 0 as t tends to infinity. So, V c at t equal to infinity is equal to V 0, and that gives us the second condition on V c.

So, these are the 2 conditions you have; at t equal to 0 V_c is V_1 , and t equal to infinity V_c is V_0 . Now using these conditions we can find A and B ; B turns out to be V_0 , and A turns out to be $V_1 - V_0$. So, putting this back in this equation we get V_c in phase 1 equal to $V_0 - (V_1 - V_0)e^{-t/\tau}$, note that this number A is positive because V_0 is larger than V_1 .

Now, let us look at the T_2 phase, and let V_c in the T_2 phase denoted by this superscript 2 over here $A' e^{-t/\tau} + B'$. To find these constants A' and B' , we need 2 conditions on V_c in phase 2, one of the conditions is that at t equal to T_1 , V_c is equal to V_2 and the second condition is that as t tends to infinity, this V_c would have gone to 0. So, these are the 2 conditions that we can use and with these conditions we get $B' = 0$, and $A' = V_2 e^{T_1/\tau}$ and when you substitute these back into this equation then you get V_c in the T_2 phase equal to $V_2 e^{-(t - T_1)/\tau}$.

So, these are the 2 equations that describe V_c of t , the first equation describes V_c in this interval and the second equation describes V_c in this interval and notice that our job is not done yet, because we have these unknowns V_1 and V_2 here and we need to figure them out and to find V_1 and V_2 we can use these conditions here, what did they say? V_c in phase 1 at t equal to T_1 is equal to V_2 ; this is V_c in phase 1 and at t equal to T_1 that is this time point here we have V_c equal to V_2 . So, that is the first condition over here. Second condition V_c in the T_2 phase at $T_1 + T_2$ is equal to V_1 this is our V_c in the T_2 phase, and $T_1 + T_2$ is this point here, so what this equation says is that this V_c at t equal to $T_1 + T_2$ is equal to V_1 .

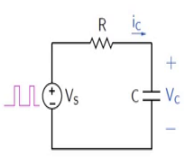
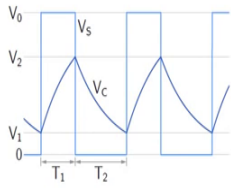
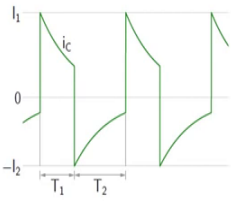
The first condition here gives us this equation 3, let us see how. What is this condition it says that V_c in the T_1 phase at t equal to T_1 should be equal to V_2 , what is V_c in the T_1 phase that is given by this equation 1 here. So, we substitute V_2 over here and we substitute t equal to T_1 and that gives us this equation it is $V_2 = V_0 - (V_1 - V_0)e^{-T_1/\tau}$. The second condition namely V_c in the T_2 phase at $T_1 + T_2$ equal to V_1 , gives us this equation here let us see how. What is V_c in the T_2 phase that is given by equation 2 here. So, what we do is substitute V_1 over here and for t we put $T_1 + T_2$ and then we get $V_1 = V_2 e^{-(T_1 + T_2 - T_1)/\tau}$. This T_1 cancels out and then we get $V_1 = V_2 e^{-T_2/\tau}$.

To simplify the algebra let us define 2 quantities; a equal to e raised to minus T 1 by tau and b equal to e raised to minus T 2 by tau; and using these a and b we can rewrite equations 3 and 4 as follows. Equation 3 can be written as V 2 equal to minus V 0 minus V 1, times e raised to minus T 1 by tau now this quantity is the same as a so therefore, a comes here plus V 0. And equation 4 can be written as this one here V 1 is equal to V 2 times e raised to minus T 2 by tau and this quantity is defined as b. So, therefore, we have B 1 equal to b times V 2.

And now we can solve equations 5 and 6 to obtain V 1 and V 2; and the results are written over here. Now if we know r and c we can calculate tau, tau is equal to r times c and if you know T 1 and T 2 then a and b can be calculated. So, therefore, this entire factor is known and if you know V 0 that is the maximum value of the input voltage, then V 1 and V 2 can be computed. Once we know V 1 and V 2 we have the complete picture for V c in the T 1 phase and V c in the T 2 phase.

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RC circuit: example

$$V_1 = b V_0 \frac{1-a}{1-ab}, \quad V_2 = V_0 \frac{1-a}{1-ab}, \quad \text{with } a = e^{-T_1/\tau}, \quad b = e^{-T_2/\tau}.$$

$$V_C^{(1)}(t) = -(V_0 - V_1)e^{-t/\tau} + V_0, \quad V_C^{(2)}(t) = V_2 e^{-(t-T_1)/\tau}.$$

Current calculation:
 Method 1:
 $i_C(t) = C \frac{dV_C}{dt}$ (home work)
 Method 2:
 Start from scratch!



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Here the results are summarized for the capacitor voltage; we have V 1 and V 2 here in terms of a and b, where a and b are given over here and once we know V 1 and V 2 we can find V c in the T 1 phase using this expression, and V c in the T 2 phase using this expression.

Let us now look at the current calculation i C of t and once again this will have 2 parts i C in the T 1 phase, and i C in the T 2 phase how do we go about this there are 2 ways of

obtaining the current; one we already know the capacitor voltage in the T 1 phase as well as in the T 2 phase, all we need to do is to differentiate that multiply that by C and then we get i_C of t now this is something that you must do as homework we will not do that here.

The second method is to start from scratch pretend that we do not have these results and we start from the beginning and see what we get. And in doing that we will definitely learn something more, let us proceed further now the stop sign has come now, what is it saying? It is saying that there is something to be completed and that is this homework over here. So, you really should stop the video at this point compute i_C in the T 1 phase and also in the T 2 phase by differentiating the capacitor voltage in the T 1 phase and in the T 2 phase and once you have done with that you can get back to the video.

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RC circuit: example

$0 < t < T_1$ Let $i_C^{(1)}(t) = A e^{-t/\tau} + B$
 $i_C^{(1)}(0) = i_1, i_C^{(1)}(\infty) = 0$
 $\rightarrow B = 0, A = i_1$
 $i_C^{(1)}(t) = i_1 e^{-t/\tau}$ (1)

$T_1 < t < T_2$ Let $i_C^{(2)}(t) = A' e^{-t/\tau} + B'$
 $i_C^{(2)}(T_1) = -i_2, i_C^{(2)}(\infty) = 0$
 $\rightarrow B' = 0, A' = -i_2 e^{T_1/\tau}$
 $i_C^{(2)}(t) = -i_2 e^{-(t-T_1)/\tau}$ (2)

Now use the conditions:
 $i_C^{(1)}(T_1) - i_C^{(2)}(T_1) = \Delta = V_0/R,$
 $i_C^{(1)}(0) - i_C^{(2)}(T_1 + T_2) = \Delta = V_0/R.$
 $i_1 e^{-T_1/\tau} - (-i_2) = \Delta$ (3)
 $i_1 - (-i_2 e^{-(T_1+T_2-T_1)/\tau}) = \Delta$ (4)
 $a i_1 + i_2 = \Delta$ (5)
 $i_1 + b i_2 = \Delta$ (6)

Solve to get
 $i_1 = \Delta \frac{1-b}{1-ab}, i_2 = \Delta \frac{1-a}{1-ab}$
 $(a = e^{-T_1/\tau}, b = e^{-T_2/\tau})$

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Let us begin with i_C in the T 1 phase this part, and let i_C be $a e^{-t/\tau} + b$ and that is i_C in the T 1 phase and that is denoted by this superscript 1 over here. Our job is to find A and B and for that we need 2 conditions on the capacitor current. One condition is obvious at t equal to 0 we have i_C equal to i_1 ; what about the second condition? The second condition comes from the fact that if this change did not happen; that means, if this input voltage continued to be b 0 then this i_C would have gone eventually to 0. So, that gives us the second condition namely i_C at infinity equal to 0. So, these are the 2 conditions we can use to find A and B.

So, B turns out to be 0 and A turns out to be I_1 and putting this back in this equation now we get i_C in the T_1 phase equal to $I_1 e^{-t/\tau}$ all right let us now look at the T_2 phase; that means, this part of i_C of t and in that phase let i_C be $A' e^{-t/\tau} + B'$ and here we put a superscript 2 to denote that we are in the T_2 phase.

Now, to find A' and B' we need 2 conditions on i_C in the T_2 phase. The first condition is that i_C at $t = T_1$ is equal to $-I_2$, what about the second conditions? The second condition can be obtained using the fact that this current would have gone to 0 at $t = \infty$, if this change had not happened. So, these are the 2 conditions that we can use namely i_C at T_1 equal to $-I_2$, and i_C at infinity equal to 0 and with these conditions we get V' equal to 0 and A' equal to $-I_2 e^{-T_1/\tau}$, and putting these back in this equation now we get i_C in the T_2 phase equal to $-I_2 e^{-t/T_1 - t/\tau}$ divided by τ .

Our next step is to find these unknowns I_1 and I_2 in equations 1 and 2, and let us see how to do that. Let us look at this picture this is our i_C in the T_1 phase and that is over i_C in the T_2 phase; at $t = T_1$ i_C in the T_1 phase is this value here and at the same time i_C in the T_2 phase is this value over here, which is $-I_2$ and the difference between these 2 is denoted by Δ over here. So, let us first figure out what this Δ should be? At this point $t = T_1$, the input voltage is changing from V_0 to 0 the capacitor voltage is not really changing because V_c at T_1 minus is the same as V_c at T_1 plus and therefore, coming to the circuit now we observe that there is a change in V_s , there is no change in p_c and therefore, this change in V_s must appear across this resistance otherwise KVL would not get satisfied.

V_s is changing by V_0 this quantity here and therefore, the voltage across the resistor must also change by V_0 and what does that mean? That means, there must be a corresponding change in the current i_C here, which is V_0 divided by R and therefore, we conclude that this Δ must be equal to V_0/R , and that is what we write over here; i_C in phase 1 at T_1 that is this value here minus i_C in phase 2 at T_1 this value over here must be equal to V_0/R . And by the same logic this difference must also be V_0/R ; now what is this value? This i_C value is the same as this i_C value and that is i_C in the T_1 phase at $t = 0$, what about this value? That is i_C in the T_2

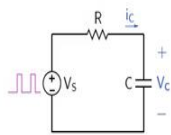
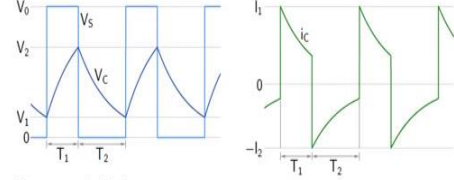
phase at t equal to T_1 plus T_2 and the difference between these 2 values must be equal to V_0 by R and that is what this equation says over here.

The rest is a straightforward what is i_C in the T_1 phase at t equal to T_1 ? This is our equation for i_C in the T_1 phase. So, let us put t equal to T_1 over here and then we get this force term over here. What about i_C in the T_2 phase at t equal to T_1 ? This is over i_C in the T_2 phase, so we put t equal to T_1 here then we get e^{-a} that is 1 and therefore, this i_C at T_1 boils down to $-I_2$ like that, and the difference between these 2 terms must be equal to Δ .

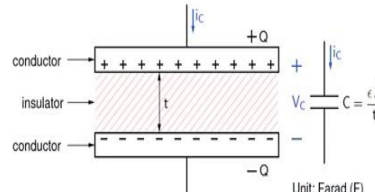
Let us use this second condition now, what is i_C in the T_1 phase at t equal to 0? We put t equal to 0 here and that gives us I_1 the first term here, what about i_C in the T_2 phase? At T_1 plus T_2 we put t equal to T_1 plus T_2 over here, and then we obtain this expression over here and the difference between these 2 must be equal to Δ . Note that, T_1 cancels out over here and we get e^{-a} raised to $-T_2$ by τ . Now using the constants a and b that we defined in the previous slide we can rewrite these equations 3 and 4 as; $a I_1$ plus I_2 equal to Δ that is our equation 3 and I_1 plus b times I_2 equal to Δ that is our equation 4 and these 2 equations can now be solved for I_1 and I_2 . I_1 turns out to be this expression here, Δ times $1 - b$ by $1 - a - b$ and I_2 is Δ times $1 - a$ by $1 - a - b$, where a is e^{-a} raised to $-T_1$ by τ and b is e^{-a} raised to $-T_2$ by τ as we saw in the last slide.

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RC circuit: example

Charge conservation:
 Periodic steady state: All quantities are periodic, i.e.,
 $x(t_0 + T) = x(t_0)$
 Capacitor charge: $Q(t_0 + T) = Q(t_0)$
 $i_C = \frac{dQ}{dt} \rightarrow Q = \int i_C dt$
 $Q(t_0 + T) = Q(t_0) \rightarrow Q(t_0 + T) - Q(t_0) = 0$
 $\rightarrow \int_{t_0}^{t_0+T} i_C dt = 0$
 $\int_0^T i_C dt = 0 \rightarrow \int_0^{T_1} i_C dt + \int_{T_1}^{T_1+T_2} i_C dt = 0$
 $\rightarrow \int_{T_1}^{T_1+T_2} i_C dt = - \int_0^{T_1} i_C dt$



Unit: Farad (F)

$$i_C = \frac{dQ}{dt} = C \frac{dV_C}{dt}$$

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So, now we have obtained expressions for the capacitor voltage V_c in the steady state both in the T_1 and T_2 phases, and also we have got expressions for the capacitor current i_C in the T_1 phase as well as in the T_2 phase. But let us go a little bit beyond that and look at the implications of charge conservation and in the process we will learn a lot more about these waveforms. Here is a schematic diagram showing a capacitor there is 1 metal plate over here with charge plus Q and there is another metal plate here with a charge minus Q . If this is plus Q this has to be always minus Q these 2 metal plates are separated by a dielectric with a thickness t and a dielectric constant ϵ .

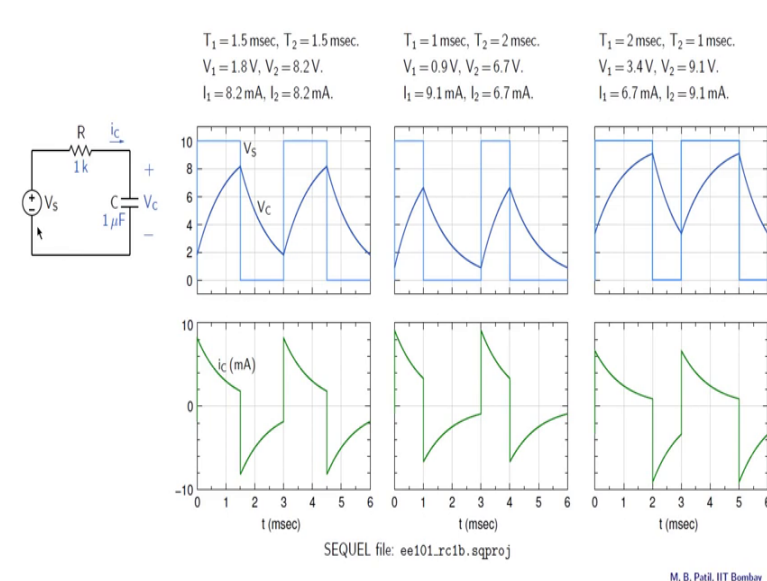
Now, in real life these may not be quite plates, these may be metal foils and the insulator may also be in the form of a dielectric film and this whole thing may be rolled in a cylindrical shape for example, in any case the cross section does look like this and the capacitance is given by $\epsilon A / t$, where ϵ is the dielectric constant of this insulator and t is the thickness A here is the area of the capacitor. Let us now look at the charge here plus Q on this plate and minus Q on this plate; this charge can change only if there is a non 0 current i_C either entering this terminal or leaving this terminal, and the relationship between i_C and Q is given by $i_C = dQ/dt$ and since $Q = C V_c$, we get $i_C = C dV_c/dt$ and this of course, is the equation that we have been using all along.

Let us now look at the implications of this equation in our context and in particular we will look at the implications in the periodic steady state. What is the meaning of periodic steady state? All quantities are periodic that is x at $t_0 + T$ must be equal to x at t_0 ; where x could be a current or a voltage or even the charge on the capacitor. Let us take x equal to Q the capacitor is charged and then we have q at $t_0 + T$ equal to Q at t_0 , this T is the period in our case it is $T_1 + T_2$. Now since $i_C = dQ/dt$ we can write Q as $\int i_C dt$ and let us see what this implies. Now this equation can be rewritten as Q at $t_0 + T$ minus Q at t_0 equal to 0 and because we have this relationship between Q and i_C , this equation here can be rewritten as $\int i_C dt$ from t_0 to $t_0 + T$ equal to 0.

let us take a specific case let us take t_0 equal to 0, this time point here and then this becomes $\int i_C dt$ from 0 to T equal to 0 and now we can split this into 2 integrals one from 0 to T_1 and the other from T_1 to $T_1 + T_2$. So, we can write $\int i_C dt$ from 0 to T_1 plus $\int i_C dt$ from T_1 to $T_1 + T_2$ equal to 0; and this equation

tells us that $\int_{T_1}^{T_1 + T_2} i_C dt$ must be negative of $\int_0^{T_1} i_C dt$. Now let us see what this implies in terms of our graph here, what is $\int_{T_1}^{T_1 + T_2} i_C dt$? It is simply the area under the i_C curve. So, this term here the integral from T_1 to $T_1 + T_2$ is this area here and this is a negative area because our i_C is negative in this region, and this integral from 0 to T_1 is this area and that area of course, is positive because i_C is positive. So, what this equation is saying is that these 2 areas must be equal in magnitude.

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Let us look at some simulation results now, in particular let us take T_1 equal to T_2 1.5 milliseconds each, R equal to 1 k, C equal to 1 micro farad and V_0 equal to 10 volts with these numbers using the formulas that we derived earlier V_1 turns out to be 1.8 volts, V_2 is 8.2 volts, I_1 is 8.2 milliamps and I_2 is also 8.2 milliamps. So, our current the capacitor current is going to vary between plus 8.2 milliamps and minus 8.2 milliamps. The simulation results are shown in these plots this is our input voltage that is the capacitor voltage and that is the capacitor current a milliamps.

Now, V_c is going from 1.8 volts to 8.2 volts and that agrees with these results here and I_1 goes from plus 8.2 milliamps to minus 8.2 milliamps. What about delta this discontinuity? If you recall that is expected to be V_0 divided by R , where V_0 is this voltage ten volts and R is 1 k. So, delta should be 10 volts by 1 k or 10 milliamps; and

this is indeed ten milliamps the same height as this height here and this is also equal to delta that is 10 milliamps.

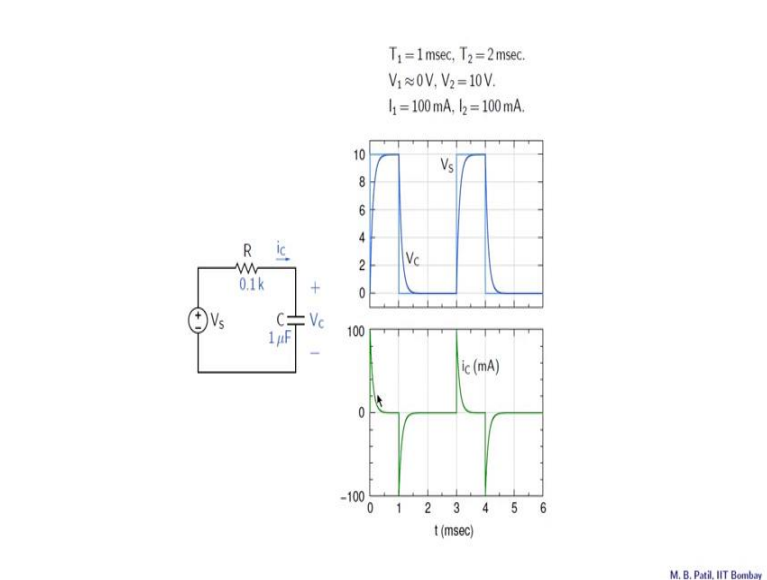
So, all is well and our analysis seems to be correct. Let us now look at the implications of charge conservation, if you recall we expect the area under the i C curve in the T_1 phase and in the T_2 phase to be equal in magnitude, and that does in to be the case this area is equal to this area here. Here is our second case what we have done is we have kept T_1 plus T_2 the same, that is 3 milliseconds, but we have decreased T_1 and increased T_2 . This is our T_1 it is now 1 millisecond, this is over T_2 it is now 2 milliseconds with these changes we get V_1 , V_2 , I_1 , I_2 as given by these numbers here and you can check whether these numbers do actually correspond to the plots which have been obtained by a simulation.

What about charge conservation? We expect this area to be equal to this area here in magnitude, and that does appear to be the case. This is the third case we have once again we have kept T_1 plus T_2 equal to 3 milliseconds, we have increased T_1 to 2 milliseconds and decreased T_2 to 1 millisecond. So, this is our T_1 , 2 milliseconds and this is over T_2 , 1 millisecond and these are the results obtained using our formulas and you should check that these do agree with their simulation results given here. What about charge conservation? This area is expected to be equal in magnitude to this area and that does appear to be the case.

So, there is really quite a lot we can learn from this simple example, here is the sequel circuit file and you are strongly encouraged to run this simulation make sure that you do get these results that we have shown here, and apart from that you can make up your own product choose some different values of R , C , T_1 , T_2 calculate V_1 , V_2 , I_1 , I_2 using our expressions and then turn the simulation to make sure that your values are correct; and in fact you do not need to stop at simulation, you can go to your electronics lab look up this circuit and verify that the results that you get on your oscilloscope do only with your calculations, and in that process you will learn how to use function generator to generate a square wave with different duty cycles, you will learn how to pick components such as 1 k resistor, or 1 micro farad capacitor and you will also learn how to use an oscilloscope.

So, take this as a learning opportunity and it is unlikely that you will regret it.

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Let us consider 1 more case before we leave this example and that is T_1 equal to 1 millisecond, T_2 equal to 2 milliseconds, and we have seen this case in the last slide; but what is different now is the value of r it was 1 k earlier and now it is 0.1 k, and as a result the time constant which was 1 millisecond in our previous example, is now 0.1 k times 1 microfarad which is 0.1 millisecond and therefore, 5 tau is 0.5 millisecond, and that turns out to be smaller than T_1 or T_2 . And what does it mean? That means, we expect the capacitor voltage to reach its final destination 10 volts here, or 0 volts here; and in fact we do not even need to use the formulas that we have derived earlier, because we know that the capacitor is going to get charged all the way to 10 volts during the T_1 phase, and we know that get it is going to discharge all the way to 0 volts in the T_2 phase.

What about the current? The jump in the current which we are called delta is now different because although V_0 is the same as 10, r have changed. So, the new delta would be 10 divided by point 1 k or 100 milliamps and that is what we see over here. So, the jump here or there is 100 milliamps. And as in the voltage waveform we observe that the current also does manage to reach its final destination, that is 0 ampere. Both in the T_1 phase as well as in the T_2 phase.

That brings us to the end of RC circuits with a piecewise constant source. We will find this background very useful in understanding some of the applications to be covered later in this course. See you next time.